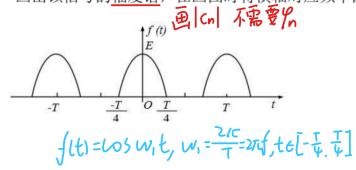
1. 求题图 2-1 所示<u>半波余弦信号的三角傅里叶级数</u>。若 $E=10~\rm{V}$, $f=10~\rm{kHz}$,基于 $|c_n|$ 画出该信号的<u>幅度谱</u>,在画图时将横轴对应频率的单位转为 \rm{Hz} 。(20 分)

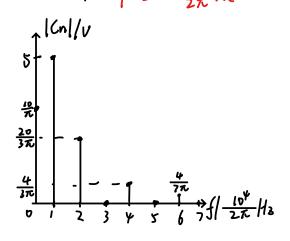


 $\alpha_0 = \frac{1}{T_1} \int_{-\frac{T_1}{2}}^{\frac{T_1}{2}} f(t) dt$ $\alpha_n = \frac{2}{T_1} \int_{-\frac{T_1}{2}}^{\frac{T_1}{2}} f(t) \cos nw.t dt$ $b_n = \frac{2}{T_1} \int_{-\frac{T_1}{2}}^{\frac{T_1}{2}} f(t) \sin nw.t dt$

 $\Delta n = \frac{1}{T} \int_{-T}^{T} f(t) \omega_{3} n w_{1} t dt = \frac{4}{T} \int_{0}^{T} E \omega_{3} w_{1} t \omega_{3} n w_{2} t dt = \frac{4}{T} \int_{0}^{T} \omega_{3} w_{1} t \omega_{3} n w_{2} t dt \\
= \frac{2E}{T} \int_{0}^{T} \omega_{3} \left[(n_{1}) w_{1} t \right] dt + \frac{E}{T} \int_{0}^{T} \omega_{3} \left[(n_{1}) w_{2} t \right] dt = \frac{E}{\pi (n_{1} t)} \sin \left[(n_{1}) \frac{w_{1} T}{4} \right] + \frac{E}{\pi (n_{1} t)} \sin \left[(n_{1}) \frac{w_{1} T}{4} \right] \\
= \frac{E}{\pi (n_{1} t)} \sin \frac{(n_{1} t)}{2} \pi_{1} + \frac{E}{\pi (n_{1} t)} \sin \frac{(n_{1} t)}{2} \pi_{3}$

当n为奇时 (n+1) 当n为行时 (n+1) 当n为行为 (n+1) = $\frac{2E}{\pi(n^2-1)}$ > 只需 $|a_n|$ 而不需 a_n 特别地 $(a_1 = \frac{4}{7}\int_{0}^{\pi} E(\omega)^2w, tdt = \frac{E}{2} + 0 = \frac{E}{2} - 3V$ > 注意特殊情况. 单位换算 $rad/iot_1 \rightarrow \frac{10^4}{27}H_2$

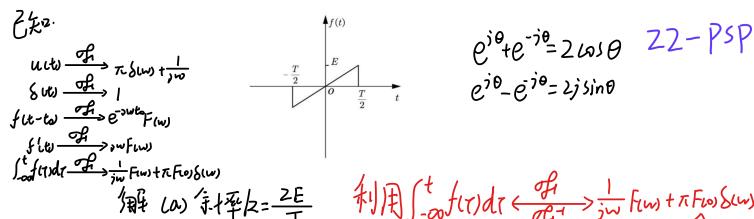
22-PSP



名子上 (元V, N=0 5V, N=1 0, n=2k+1, なe2 20 元(n2-UV, n=2k, be) 2. 求解题图 2-2 (a)、(b) 所示的锯齿脉冲与单周正弦脉冲的傅里叶变换,给出频谱密度

函数的表达式。(20分)

F(w)= F[{(t)]

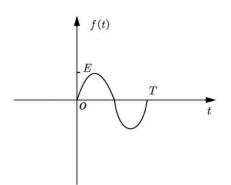


海 (山) 宇本本人= $\frac{2E}{T}$ 和月 $\int_{-\infty}^{t} f(\tau) d\tau$ ($\int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) d\tau$ ($\int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) d\tau$) $\int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) d\tau$ ($\int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) d\tau$) $\int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) d\tau$) $\int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) d\tau$) $\int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) \int_{-\infty}^{t} f(\omega) d\tau$) $\int_{-\infty}^{t} f$

$$\begin{aligned} & 2(\Delta) \vec{x} \dot{x} = \vec{r} \dot{x} \dot{x} \\ & \Rightarrow \mathbf{p}(\mathbf{r}) \\ & \Rightarrow \mathbf{p}(\mathbf{r}) \mathbf{p}$$

2. 求解题图 2-2 (a)、(b) 所示的锯齿脉冲与单周正弦脉冲的傅里叶变换,给出频谱密度

函数的表达式。(20分)



Fun=Fifter] 22-PSP

$$\frac{1}{\sqrt{1}} \cdot \sin \theta = \frac{1}{\sqrt{1}} \left[e^{j\theta} - e^{-j\theta} \right] \qquad e^{j\theta} = \omega_0 \theta + j \sin \theta$$

$$\Rightarrow e^{j2\pi} = 1, e^{-j2\pi} = 1 \qquad de^{j\omega} = \frac{2\pi}{T}$$

$$(b) F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt = E \int_{0}^{\infty} \sin \omega_0 t e^{-j\omega t} dt$$

$$= \frac{E}{2j} \int_{0}^{T} \left[e^{j(\omega_0 - \omega)t} - e^{-j(\omega_0 + \omega)t} \right] dt$$

$$= \frac{E}{2j} \left[\frac{1}{j(\omega_0 - \omega)} e^{j(\omega_0 - \omega)t} \right] + \frac{1}{j(\omega_0 + \omega)} e^{j(\omega_0 + \omega)t} \int_{0}^{T}$$

$$= \frac{E}{2} \left[\frac{1}{w_0 - \omega} \left[e^{j(\omega_0 - \omega)T} \right] + \frac{1}{w_0 + \omega} \left[e^{-j(\omega_0 + \omega)T} \right] \right]$$

$$= \frac{E}{2} \frac{(\omega + |\omega_0|) (e^{-j\omega T} - 1) - (|\omega_0 - \omega|) (e^{-j\omega T} - 1)}{\omega^2 - \omega_0^2}$$

$$= \frac{E}{2} \frac{2\omega_0}{\omega^2 - \omega_0^2} (e^{-j\omega T} - 1) = \frac{\omega_0 E}{\omega^2 - \omega_0^2} (e^{-j\omega T} - 1)$$

$$= \frac{E}{2} \frac{2\omega_0}{\omega^2 - \omega_0^2} (e^{-j\omega T} - 1) = \frac{\omega_0 E}{\omega^2 - \omega_0^2} (e^{-j\omega T} - 1)$$

频域卷积定理

fit) = get) hit)

 $F(w) = \frac{1}{2\pi} (\kappa(w) * H(w))$

72 g(t)=Esin(wot), te1-∞, to0), h(t)=u(t)-u(t-7), f(t)=g(t)h(t)

 $e^{j\omega_0T} = e^{-j\omega_0T} = e^{2\pi} = e^{-2\pi} = [$ $\sin(\omega_0t) \xrightarrow{\sigma_0} j\pi[\delta(\omega_0t) - \delta(\omega_0t)]$

注意到 1—好 2元(w), utl) 好 大(w)+ 1 wo= 2元

$$F(w) = \frac{1}{2\pi} \left[(\kappa(w) * H(w)) \right] = \frac{1}{2\pi} \left[2E\pi \left(S(w+w_0) - S(w-w_0) \right) * \left[(\pi S(w) + \frac{1}{2}w)(1 - e^{-2w}) \right] \right]$$

$$= \frac{2E}{2} \left\{ \left[\pi S(w+w_0) + \frac{1}{2(w+w_0)} \right] \left[1 - e^{-2(w+w_0)T} \right] - \left[\pi S(w-w_0) + \frac{1}{2(w-w_0)} \right] \left[1 - e^{-2(w-w_0)T} \right] \right\}$$

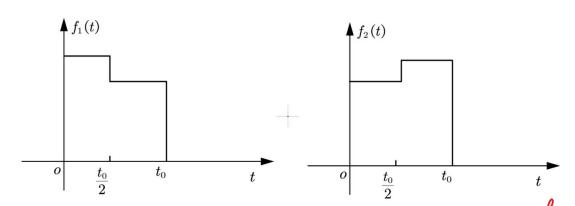
$$= \frac{E}{2} \left[\frac{1 - e^{-2wT}}{w + w_0} - \frac{1}{w - w_0} \right] = \frac{E}{2} \frac{1}{w^2 - w_0^2} \left[-2w_0 + 2w_0 e^{-2wT} \right]$$

$$= \frac{w_0 E}{w^2 - w_0^2} \left(e^{-2wT} - 1 \right) = \frac{-22w_0 E}{w^2 - w_0^2} Sin(\frac{wT}{2}) e^{-2wT}$$

说明可以生物从原点为私的频谱 最后开始王

3. 对题图 2-3 所示波形,若已知 $\mathcal{F}[f_1(t)] = F_1(\omega)$,利用傅里叶变换的性质,求 $f_1(t)$

以 $\frac{t_0}{2}$ 为轴翻转后所得 $f_2(t)$ 的傅里叶变换。(20 分)



解f2t)=f1(-t+6)

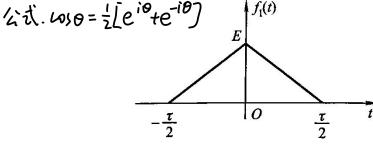
 $f_{2}(t)=f_{1}(-t+t_{0})$ 题图 2-3 f(t) 一, $f_{1}(t)$ 一, $f_{2}(t)$ 一, $f_{3}(t)$ 一, $f_{4}(t)$ 一, $f_{2}(t)$ 一, $f_{3}(t)$ 一, $f_{4}(t)$ 一, $f_{5}(t)$ 一, $f_{6}(t)$ 一, $f_{1}(t)$ 一, $f_{1}(t)$ 一, $f_{2}(t)$ 一, $f_{3}(t)$ 一, $f_{4}(t)$ 一, $f_{5}(t)$ 一, $f_{6}(t)$ — $f_{6}(t)$ $f_{i}(t+t_{0}) \xrightarrow{f_{i}} e^{i\omega t_{0}}$ $f_{i}(t+t_{0}) \xrightarrow{f_{i}} e^{j(-\omega t_{0})}$ $f_{i}(-t+t_{0}) \xrightarrow{f_{i}} e^{j(-\omega t_{0})}$ $f_{i}(-t+t_{0}) \xrightarrow{f_{i}} e^{j(-\omega t_{0})}$ 4. 已知三角脉冲 $f_{1}(t)$ 的傅里叶变换为:

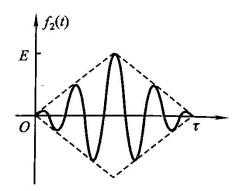
$$F_1(\omega) = \frac{E\tau}{2} \operatorname{Sa}^2\left(\frac{\omega\tau}{4}\right)$$

求 $f_2(t) = f_1\left(t-\frac{\tau}{2}\right)\cos\left(\omega_0 t\right)$ 的傅里叶变换 $F_2(\omega)\circ f_1(t), f_2(t)$ 的波形如题图 2-4 所示。

(20分)

DH% +频移





角子. fit1= = = fit+===)ejwot+=fi(t-==)e-jwot

其中. fi(t-至) of e-jw至 Filw)

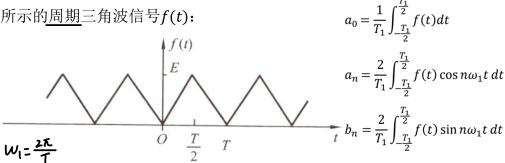
华美符号

^{题图 2-4} 从而·Filw=f[fitt]

 $f_{i}tt^{-\frac{7}{2}}e^{jw_{0}t}\frac{f_{i}}{f_{i}}e^{-j(w-w_{0})\frac{7}{2}}f_{i}(w-w_{0}) = \frac{1}{2}\left[e^{j(w-w_{0})\frac{7}{2}}f_{i}(w-w_{0})+e^{-j(w+w_{0})\frac{7}{2}}f_{i}(w+w_{0})\right]$

 $f_{1}(t-\frac{7}{2})e^{-i\omega t}\frac{f_{1}}{f_{2}}e^{-i(\omega+\omega_{0})\frac{7}{2}}[f_{1}(\omega+\omega_{0})] = \frac{E^{7}}{4}e^{-i\omega \frac{7}{2}}[\int_{a}^{b}(\frac{\omega-\omega_{0}}{4}T)e^{-i\omega_{0}\frac{7}{2}}]$

5. 已知一个如下图所示的周期三角波信号f(t):



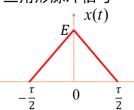
题图 2-5

- (1) 求f(t)的傅里叶级数系数 a_0 、 a_n 、 b_n ,写出完整的傅里叶级数表达式;
- (2)用一个幅值为 E、脉宽为 T 的矩形脉冲为 f(t)加窗取 $t \in [0,T]$ 的部分,记为信号 g(t), 求g(t)的傅里叶变换 $G(\omega)$;

日北京科(3) 对g(t)以等间隔T/10 进行 $_{2}$ 进积,求所得采样信号的频谱 $G_s(\omega)$ 。(20 分)

角体. (1) f(t) 为(思述, bn = 0,
$$Q_0 = \frac{1}{1} \int_{-\frac{1}{2}}^{\frac{1}{2}} f(t) dt = \frac{1}{1} \int_{0}^{\frac{1}{2}} f(t) d$$

三角形脉冲信号



qは相的x(t)幅值×E倍,7=T,向右平移至单位 由傅里啦换的线性与时找稀定理 $G(w) = \frac{E^2T}{2} \int_{0}^{2} (\frac{wT}{4}) e^{-jw\frac{1}{2}}$

(3) 第二章 理想採样心式
$$F_s(\omega) = \frac{1}{T_s} \sum_{n=-\infty}^{\infty} F(\omega - n\omega_s)$$

此处 $T_s = \frac{1}{10} , W_s = \frac{2\pi}{T_s} - \frac{20\pi}{T}$
 $G_{1s}(w) = \frac{1}{T_s} \sum_{n=-\infty}^{\infty} G(w - n\omega_s) = 5E^2 \sum_{n=-\infty}^{\infty} S_a^{\perp} \left(\frac{w - n\omega_s}{4} T \right) e^{-\frac{1}{2}(w - n\omega_s) \frac{1}{2}}$
 $= 5E^2 \sum_{n=-\infty}^{\infty} S_a^{\perp} \left(\frac{wT}{4} - 5nT_s \right) e^{-\frac{1}{2}(\frac{wT}{2} - 10nT_s)} \oplus \int_{n=-\infty}^{\infty} S_a^{\perp} \left(\frac{wT}{4} - 5nT_s \right) e^{-\frac{1}{2}\frac{wT}{2}}$