

例 3.9: 某反馈控制系统的开环传递函数为

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$$G_0(s) = \frac{K}{s(\frac{1}{6}s+1)(\frac{1}{2}s+1)}$$

要求满足下列性能指标:

(1) 系统为 I 型, 开环增益为 $K_v = 180 \text{ s}^{-1}$;

(2) 剪切频率 $\omega_c \geq 3.5 \text{ rad/s}$;

(3) 相角裕度 $\gamma \geq 40^\circ$ 。

解: $G_0(s) = \frac{180}{s(\frac{1}{6}s+1)(\frac{1}{2}s+1)}$ 转折频率: 2, 6

$$L(\omega) = \begin{cases} 20(\lg \frac{180}{\omega}) & 0 < \omega < 2 \\ 20(\lg \frac{180}{\omega} - \lg \frac{\omega}{2}) & 2 < \omega < 6 \\ 20(\lg \frac{180}{\omega} - \lg \frac{\omega}{2} - \lg \frac{\omega}{6}) & 6 < \omega \end{cases}$$

$$\omega_{c0} = 12.927 \text{ rad/s}$$

$$\angle G_0(j\omega) = -90^\circ - \arctan(\frac{\omega}{2}) - \arctan(\frac{\omega}{6})$$

$$\Rightarrow \varphi_0(j\omega_c) = 180^\circ + \angle G_0(j\omega_c) = -56.31^\circ$$

$$\varphi_0(j\omega_c) = 180^\circ + \angle G_0(j3.5 \text{ rad/s}) = -0.5^\circ, \text{ 相位储备不足}$$

故迟后-超前校正, 设 $\omega_c = 3.6 \text{ rad/s}$

① 超前校正, $G_{c1}(s) = \frac{\alpha T s + 1}{T s + 1}, \alpha > 1, \varphi_m \geq \varphi - \varphi_0(j\omega_c) + \Delta_2 = 40^\circ + 6^\circ = 46^\circ$, 选 $\varphi_m = 55^\circ$

$$\alpha = \frac{1 + \sin \varphi_m}{1 - \sin \varphi_m} = 10.06, T = \frac{1}{\omega_m \alpha} = 0.0876, \alpha T = 0.881$$

$$G_{c1}(s) = \frac{0.881s+1}{0.0876s+1}$$

$$G_1(s) = \frac{180(0.881s+1)}{s(\frac{1}{6}s+1)(\frac{1}{2}s+1)(0.0876s+1)}$$

$$L(\omega) = 20(\lg \frac{180}{\omega} + \lg 0.881\omega - \lg \frac{\omega}{2}) \quad 2 < \omega < 6$$

② 迟后校正, $G_{c2}(s) = \frac{\tau s + 1}{\beta \tau s + 1}, \beta > 1$

$$\beta = |G_1(j\omega_c)| = \frac{180 \cdot 0.881 \cdot 2}{\omega_c} = 88.1$$

$$\frac{1}{\tau} = \frac{1}{10} \omega_c \Rightarrow \tau = \frac{10}{\omega_c} = 2.78, \beta \tau = 244.7$$

$$G_{c2}(s) = \frac{2.78s+1}{244.7s+1}$$

$$G(s) = \frac{180(0.881s+1)(2.78s+1)}{s(\frac{1}{6}s+1)(\frac{1}{2}s+1)(0.0876s+1)(244.7s+1)}$$

$$L(\omega) = 20(\lg \frac{180}{\omega} + \lg 0.881\omega + \lg 2.78\omega - \dots) \quad 2 < \omega < 6$$

$$\Rightarrow \omega_c = 3.60 \text{ rad/s}$$

$$\gamma = 180^\circ + \angle G(j\omega_c) = 47.45^\circ \text{ 满足要求}$$

例 3.10: 某单位反馈系统的不可变部分为传递函数为

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$$G_0(s) = \frac{K}{s(0.1s+1)(0.2s+1)}$$

若要求校正后系统满足

- (1) 速度误差系数 $K_v = 30$;
 - (2) 相角裕度 $\gamma \geq 40^\circ$;
 - (3) 增益裕度 $20 \lg K_g \geq 10 \text{dB}$;
 - (4) 剪切频率 $\omega_c \geq 5 \text{rad/s}$
- 试设计串联校正装置。

$$G_0(s) = \frac{30}{s(0.2s+1)(0.1s+1)}$$

$$L(\omega) = \begin{cases} 20 \lg \frac{30}{\omega} & 0 < \omega < 5 \\ 20 \lg \frac{30}{\omega} - \lg 0.2\omega & 5 < \omega < 10 \\ 20 \lg \frac{30}{\omega} - \lg 0.2\omega - \lg 0.1\omega & 10 < \omega \end{cases}$$

$$\omega_{co} = \sqrt[3]{1500} = 11.45 \text{ rad/s}$$

$$\angle G_0(j\omega) = -90^\circ - \arctan 0.1\omega - \arctan 0.2\omega$$

$$\gamma_0 = 180^\circ + \angle G_0(j\omega_{co}) = -25.28^\circ$$

$$\omega_u = 5 \text{ rad/s}, \gamma_0(\omega_u) = 180^\circ + \angle G_0(j\omega_u) = 18.43^\circ$$

需超前-滞后校正,

$$\textcircled{1} \text{超前: } G_{c1}(s) = \frac{\alpha T s + 1}{T s + 1}, \alpha > 1, \text{ 取 } \omega_c = 5 \text{ rad/s}$$

$$\varphi_m \geq \gamma - \gamma_0(\omega_c) + \Delta_2 = 40^\circ - 18.43^\circ + 6^\circ = 27.57^\circ$$

$$\text{选 } \varphi_m = 30^\circ, \alpha = \frac{1 + \sin \varphi_m}{1 - \sin \varphi_m} = 3, T = \frac{1}{\omega_m \alpha} = \frac{1}{5 \times 3} = 0.115$$

$$\alpha T = 0.346$$

$$G_{c1}(s) = \frac{0.346s+1}{0.115s+1}$$

$$G_c(s) = G_{c1}(s) G_0(s) = \frac{30}{s(0.2s+1)(0.1s+1)} \frac{0.346s+1}{0.115s+1}$$

$$\textcircled{2} \text{滞后: } G_{c2}(s) = \frac{\tau s + 1}{\beta \tau s + 1}, \beta > 1, \omega_c = 5 \text{ rad/s}$$

$$\beta = \left| \frac{G_c(j\omega_c)}{G_0(j\omega_c)} \right| = \frac{30}{\omega_c} \frac{0.346\omega_c}{0.2\omega_c} = 10.38$$

$$\zeta = \frac{10}{\omega_c} = 2, \beta \tau = 20.76 \Rightarrow G_{c2}(s) = \frac{\tau s + 1}{20.76s + 1}$$

$$G_c(s) = \frac{30}{s(0.2s+1)(0.1s+1)} \frac{0.346s+1}{0.115s+1} \frac{\tau s+1}{20.76s+1}$$

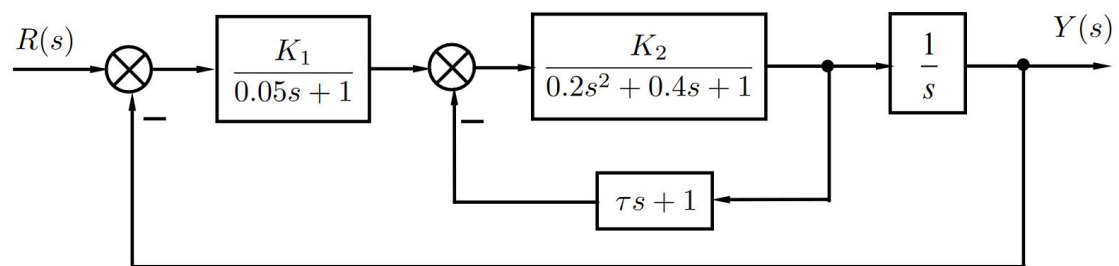
$$L(\omega) = 20 \lg \frac{30}{\omega} + \lg 0.346\omega + \lg 2\omega - \lg 0.2\omega - \lg 20.76\omega \quad 5 < \omega < 10$$

$$\Rightarrow \omega_c = 5.00 \text{ rad/s}$$

$$\gamma = 180^\circ + \angle G_c(j\omega_c) = 43.34^\circ$$

$$\frac{1}{2} \cdot 180^\circ = \angle G_c(j\omega_g) \Rightarrow \omega_g = 10.6 \text{ rad/s}$$

$$20 \lg K_g = -20 \lg |G_c(j\omega_g)| = 14.38 \text{ dB}$$



例 4.7: 控制系统的结构图如4.16所示，欲采用局部反馈来改善系统的性能，要求大闭环系统的闭环主导极点为 $s_{1,2} = -3 \pm j\sqrt{3}$ ，需要确定 K_1 ， K_2 和 τ 的值。

例 6.4: 将如下连续校正装置

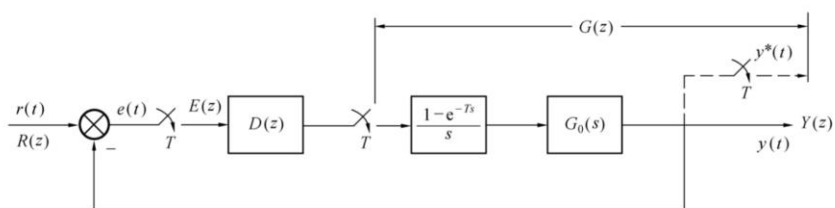
$$D(s) = \frac{1}{s^2 + 0.2s + 1}$$

在采样周期为 $T = 1\text{sec}$ 的情形下，通过双线性变换获得其离散化形式。

例 6.9: 设结构如图 6.10.1 所示单位反馈线性离散系统被控对象及零阶保持的传递函数分别为

$$G_0(s) = \frac{10}{s(s+1)}, \quad H_0(s) = \frac{1-e^{-Ts}}{s}$$

采样周期 $T = 1\text{s}$ ，试设计在控制输入为 $r(t) = t$ 时的最少拍无差系统。



例 8.19: 给定线性定常系统

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$$\dot{x} = \begin{bmatrix} 1 & 1 & 1 \\ 0 & 1 & 0 \\ 1 & 1 & 1 \end{bmatrix} x + \begin{bmatrix} 0 & 1 \\ 1 & 0 \\ 0 & 1 \end{bmatrix} u \quad y = [1 \quad 0 \quad 1]x$$

对该系统进行能控性分解。

解: $A = \begin{bmatrix} 1 & 1 & 1 \\ 0 & 1 & 0 \\ 1 & 1 & 1 \end{bmatrix}$ $B = \begin{bmatrix} 0 & 1 \\ 1 & 0 \\ 0 & 1 \end{bmatrix}$ $AB = \begin{bmatrix} 1 & 2 \\ 1 & 0 \\ 1 & 2 \end{bmatrix}$ $A^2B = \begin{bmatrix} 3 & 4 \\ 1 & 0 \\ 3 & 4 \end{bmatrix}$

$$Q_c = [B \ AB \ A^2B] = \begin{bmatrix} 0 & 1 & 1 & 2 & 3 & 4 \\ 1 & 0 & 1 & 0 & 1 & 0 \\ 0 & 1 & 1 & 2 & 3 & 4 \end{bmatrix} \Rightarrow r = 2 < 3, \text{ 不能控}$$

行满秩要求

取 Q_c 前两列加上 $\begin{pmatrix} 1 \\ 0 \\ 0 \end{pmatrix}$ 有 $P = \begin{pmatrix} 0 & 1 & 1 \\ 0 & 0 & 0 \\ 0 & 1 & 0 \end{pmatrix}$

$$\Rightarrow A_c = P^{-1}AP = \begin{pmatrix} 1 & 0 & 0 \\ 0 & 2 & 0 \\ 0 & 0 & 0 \end{pmatrix} \quad B_c = P^{-1}B = \begin{pmatrix} 1 & 0 \\ 0 & 0 \\ 0 & 0 \end{pmatrix}$$

$$C_c = CP = (0 \ 2 \ 1)$$

3. 两个子系统 Σ_1 和 Σ_2 串联, 如图8.7所示。 Σ_1 和 Σ_2 的系统矩阵、输入矩阵和输出矩阵分别为:

$$\Sigma_1: A_1 = -2, B_1 = 1, C_1 = 1$$

$$\Sigma_2: A_2 = \begin{bmatrix} 0 & 1 \\ -3 & -4 \end{bmatrix}, B_2 = \begin{bmatrix} 0 \\ 1 \end{bmatrix}, C_2 = \begin{bmatrix} 2 & 1 \end{bmatrix}$$

(1) 求串联后的状态空间描述; (5分)

(2) 判断 Σ_1 和 Σ_2 串联后的状态能控性和能观性; (5分)

(3) 求串联后的传递函数。(5分)

1. 某系统的状态空间表达式为

$$\begin{cases} \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -2 & -4 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u \\ y = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \end{cases}$$

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试设计一个带全维状态观测器的状态反馈控制系统，使观测器的极点均为 -3 ，闭环系统的极点为 $-5 \pm j5$ ，要求写出观测器方程、状态反馈控制律之表达式，并画出带观测器闭环系统的系统结构图。

解: $A = \begin{pmatrix} 0 & 1 \\ -2 & -4 \end{pmatrix}$ $B = \begin{pmatrix} 0 \\ 1 \end{pmatrix}$ $C = (1 \ 0)$

设状态反馈 $K = (k_1 \ k_2)$ $A+BK = \begin{pmatrix} 0 & 1 \\ k_1-2 & k_2-4 \end{pmatrix}$

$$|\lambda I - (A+BK)| = \begin{vmatrix} \lambda & -1 \\ 2-k_1 & \lambda-k_2+4 \end{vmatrix} = \lambda^2 + (4-k_2)\lambda + (2-k_1)$$

$$\alpha^*(s) = (s+5-j5)(s+5+j5) = s^2 + 10s + 50 \Rightarrow k_1 = -48, k_2 = -6$$

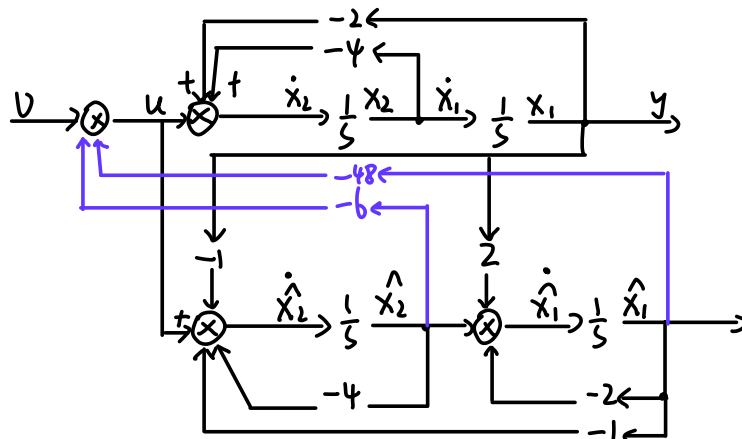
$$K = (-48 \ -6)$$

设 $L = \begin{pmatrix} l_1 \\ l_2 \end{pmatrix}$ $A+LC = \begin{pmatrix} l_1 & 1 \\ l_2-2 & -4 \end{pmatrix}$

$$|\lambda I - (A+LC)| = \begin{vmatrix} \lambda-l_1 & -1 \\ 2-l_2 & \lambda+4 \end{vmatrix} = \lambda^2 + (4-l_1)\lambda + 2-4l_1-l_2$$

$$\beta^*(s) = (s+3)^2 = s^2 + 6s + 9 \Rightarrow l_1 = -2, l_2 = 1 \Rightarrow L = \begin{pmatrix} -2 \\ 1 \end{pmatrix}$$

$$\begin{aligned} \dot{\hat{x}} &= A\hat{x} + Bu + L(C\hat{x} - y) = (A+LC)\hat{x} + Bu - Ly \\ &= \begin{pmatrix} -2 & 1 \\ -1 & -4 \end{pmatrix} \hat{x} + \begin{pmatrix} 0 \\ 1 \end{pmatrix} u + \begin{pmatrix} 2 \\ -1 \end{pmatrix} y \end{aligned}$$



8. 设非线性系统如图10.49所示。试求：

- (1) 两个非线性环节串联后的等效非线性特性；
- (2) 用描述函数法求此系统的自振角频率 ω 和振幅 A 。

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已知： $N_1 = \frac{2K}{\pi} \left[\arcsin \frac{a}{A} + \frac{a}{A} \sqrt{1 - \left(\frac{a}{A}\right)^2} \right], A \geq a$

$N_2 = \frac{4b}{\pi A} \sqrt{1 - \left(\frac{\Delta}{A}\right)^2}, A \geq \Delta$

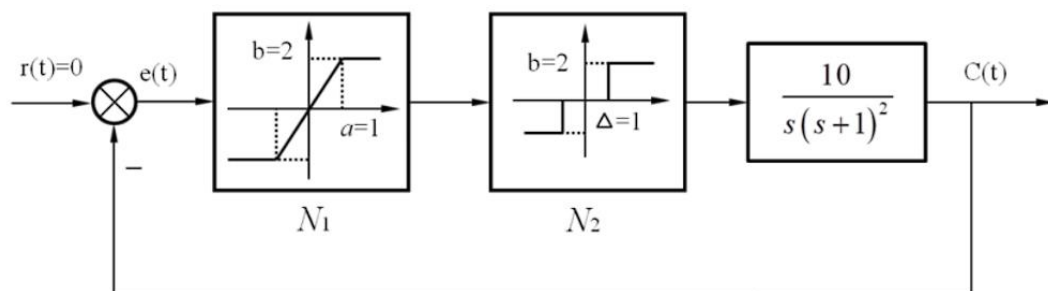
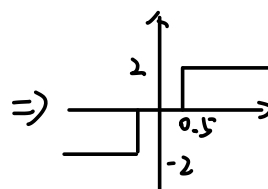


图 10.49

解：① $-0.5 < e < 0.5$ 时， $x = \text{ze} \in (-1, 1), y = 0$

② $e < -0.5$ 时， $x < -1, y = -2$,

③ $e > 0.5$ 时， $x > 1, y = 2$



相当于 $\Delta = 0.5, b = 2$ 的带死区继电,

$$N(A) = \frac{8}{\pi A} \sqrt{1 - \left(\frac{1}{2A}\right)^2} = \frac{16}{\pi} \cdot \frac{1}{2A} \sqrt{1 - \left(\frac{1}{2A}\right)^2} \leq \frac{8}{\pi}, \text{ 当且仅当 } A = \frac{\sqrt{2}}{2} \text{ 等号成立}$$

故 $A \rightarrow 0.5$ 时， $\frac{1}{N(A)} \rightarrow -\infty$, $A \rightarrow \frac{\sqrt{2}}{2}$ 时， $\frac{1}{N(A)} \rightarrow -\frac{\pi}{8}$, $A \rightarrow +\infty$ 时， $\frac{1}{N(A)} \rightarrow -\infty$

线性部分：

$$G(s) = \frac{10}{s(s+1)^2}, G(j\omega) = \frac{10}{-2\omega^2 + j(\omega - \omega^3)} = \frac{10[-2\omega^2 - j(\omega - \omega^3)]}{4\omega^4 + \omega^6 - 2\omega^4 + \omega^2} = \frac{10[-2\omega^2 - j(\omega - \omega^3)]}{\omega^6 + 2\omega^4 + \omega^2}$$

$$\text{Re} = \frac{-20}{\omega^6 + 2\omega^4 + \omega^2}, \text{Im} = \frac{-10(1 - \omega^2)}{\omega^6 + 2\omega^4 + \omega^2}$$

令 $\text{Im} = 0 \Rightarrow \omega = 1 \text{ rad/s}$, $\text{Re} = -5 < -\frac{\pi}{8}$, 即 $G(j\omega)$ 与 $-\frac{1}{N(A)}$ 有交点。

$$-5 = -\frac{1}{\frac{8}{\pi A} \sqrt{1 - \left(\frac{1}{2A}\right)^2}} \Rightarrow A = 12.72, \text{ 由于 } A_1 > \frac{\sqrt{2}}{2}, \text{ 是上面的交点, 是稳定的}$$

会产生自激振荡, 振幅 12.72, 频率 1 rad/s