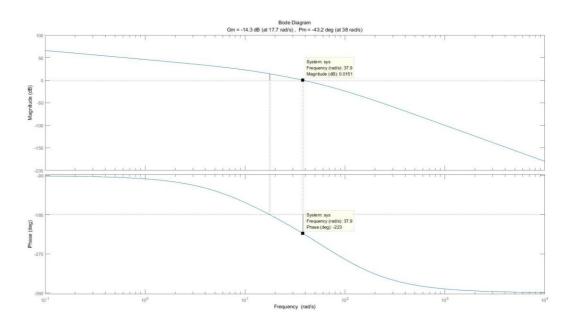
自动控制理论 B

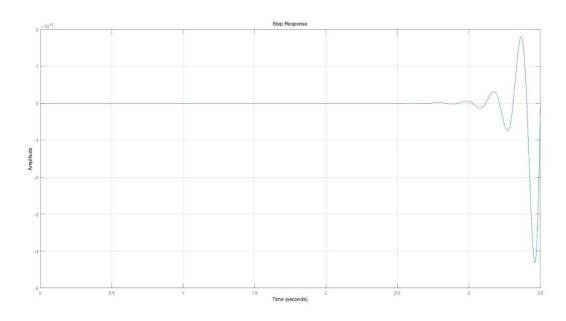
Matlab 仿真实验报告

头验名称	:线性系统的频率校上设计
姓 名	:
学 号	:
班 级	:
撰写日期	: 2023/4/7

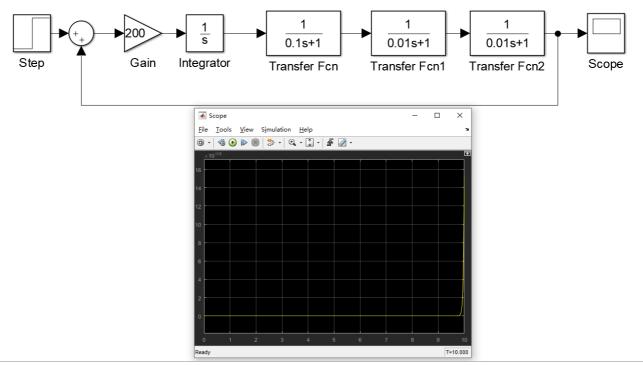
一、 未校正系统的时域指标和频率性能



未校正系统 Bode 图, 剪切频率为 ω_{\circ} =37. 9rad/s, 相角裕度 γ =-43 $^{\circ}$.



未校正系统不稳定,在阶跃输入作用下震荡发散.



要求系统的性能指标为单位阶跃系统的调整时间 t_s <0. 7sec, 单位阶跃响应超调量 σ_p <30%. 由经验公式 σ_p =0. 16+0. 4*(1/sin γ -1), ts= π / ω_c *(2+1. 5*(1/sin γ -1)+2. 5*(1/sin γ -1)²) 得, $\gamma \ge 47$. 79°, $\omega_c \ge 12$. 7rad/s.

二、 迟后-超前校正设计步骤

1. 迟后环节优先的迟后-超前校正设计

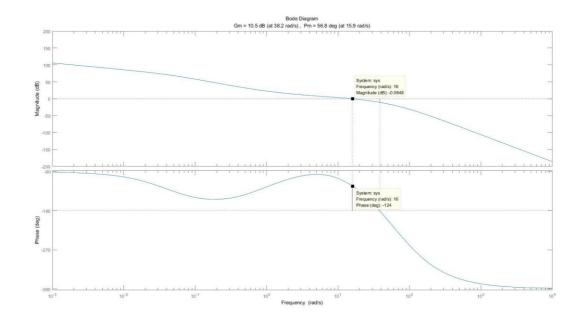
先设计迟后校正环节, 取期望的剪切频率为 ω $_{cl}$ =8rad/s, 使用剪切频率优先迟后校正方法, 得 $_{l}$ =19. 2, 取 $_{\tau}$ =1. 2, 得迟后环节 (1. 2s+1) / (23s+1). 此时系统为

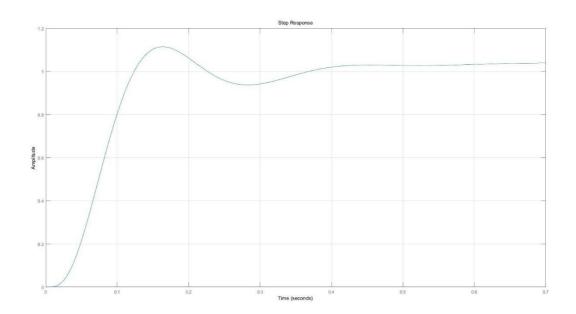
200(1.2s+1)/s(0.1s+1)(0.01s+1)(0.02s+1)(23s+1)

剪切频率为 ω_{\circ} =8rad/s, 相角裕度 γ =31.85°.

再设计超前校正环节,取期望的剪切频率为 $ω_{c2}$ =16rad/s,使用剪切频率优先超前校正方法,得 α =9. 4, $Φ_{\blacksquare}$ =53. 87°,将频率对准得 T=0. 02,得超前环节 (0. 19s+1) / (0. 02s+1).此时系统为

200 (1. 2s+1) (0. 19s+1)/s (0. 1s+1) (0. 01s+1) (0. 02s+1) (23s+1) (0. 02s+1) 剪切频率为 $ω_s$ =15. 88rad/s, 相角裕度 γ =56. 75°, 满足需求.





2. 超前环节优先的迟后-超前校正设计1

先降低开环增益 K=12, 此时系统为

12/s(0.1s+1)(0.01s+1)(0.02s+1)

剪切频率为 ω_{\circ} =8.8rad/s, 相角裕度 γ =33.64°.

再设计超前校正环节,取期望的剪切频率为 $ω_{c2}$ =16rad/s,使用剪切频率优先超前校正方法,得 α =7. 2, $φ_{\blacksquare}$ =49. 12°,将频率对准得 T=0. 0233,得超前环节 (0. 1677s+1) / (0. 0233s+1). 此时系统为

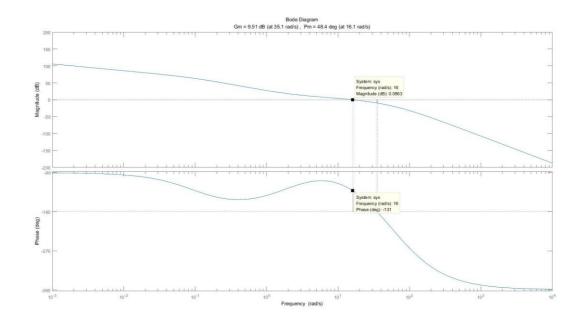
12(0.1677s+1)/s(0.1s+1)(0.01s+1)(0.02s+1)(0.0233s+1)

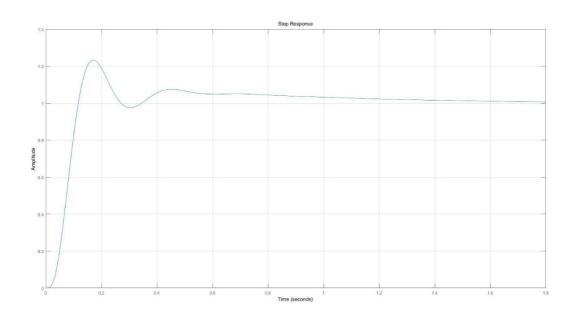
剪切频率为 ω_c =16rad/s, 相角裕度 γ =54.29°.

最后设计迟后校正环节,得 β =16.7, 取 τ =0.6, 得迟后环节 16.7 (0.6s+1)/(10s+1).

200 (0.1677s+1)(0.6s+1)/s(0.1s+1)(0.01s+1)(0.02s+1)(0.0233s+1)(10s+1)

剪切频率为 ω。=16.12rad/s, 相角裕度 γ=48.35°, 满足需求.





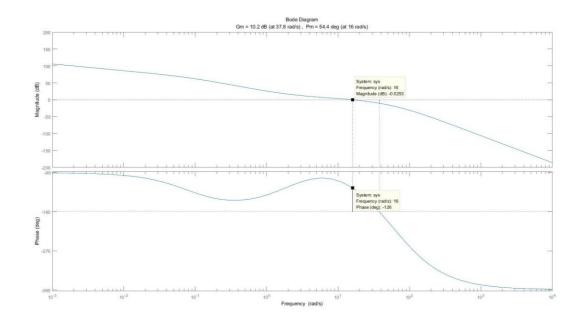
3. 超前环节优先的迟后-超前校正设计 2

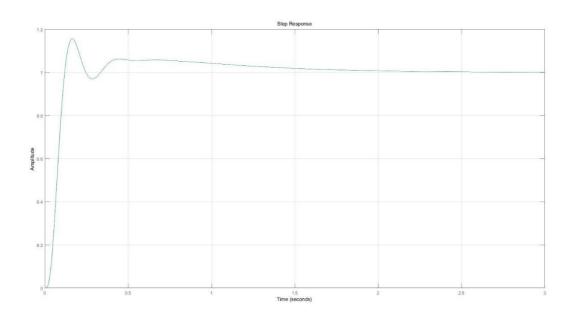
先进行超前校正环节, 选取校正后系统的剪切频率为 ω_{c2} =16rad/s, 未校正系统在此频率的相角储备为 5. 17°, 取超前相角 ϕ_{m} =55°, 得 α =10,将频率对准得 T=0. 02,得超前环节 (0.2s+1)/(0.02s+1).此时系统为

 $200(0.2s+1)/s(0.1s+1)(0.01s+1)(0.02s+1)^{2}$

再进行迟后校正环节, 得 β =20, 取 τ =0. 6, 得迟后环节 (0. 6s+1)/(12s+1). 此时系统为 200 (0. 2s+1) (0. 6s+1)/s (0. 1s+1) (0. 01s+1) (0. 02s+1)²(12s+1)

剪切频率为 ω_c =16rad/s, 相角裕度 γ =54.42°, 满足需求.





三、 期望频率法校正设计步骤

要求系统的性能指标为单位阶跃系统的调整时间 t_s <0. 7sec, 单位阶跃响应超调量 σ_p <30%. 由经验公式 σ_p =0. 16+0. 4*(1/sin γ -1), t_s = π / ω_c *(2+1. 5*(1/sin γ -1)+2. 5*(1/sin γ -1)²) 得, $\gamma \geqslant$ 47. 79°, $\omega_c \geqslant$ 12. 7rad/s, h \geqslant 6. 7, $\omega_2 \leqslant$ 3. 4rad/s, $\omega_3 \geqslant$ 22. 6rad/s.

设计中频段特性, 取 ω_s=16rad/s, ω₂=1.6rad/s, ω₃=50rad/s, 此时 h=31.25.

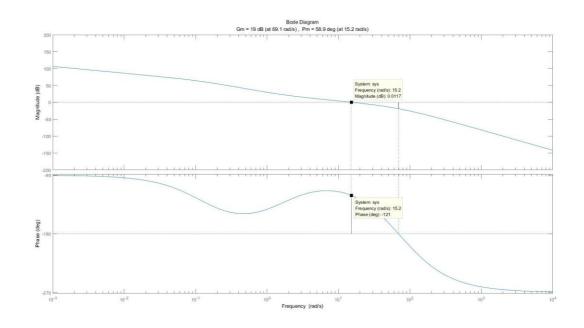
下面求低频段与中频段过渡段下限频率 ω_1 , 20 (1g200-1g ω_c -1g ω_c / ω_1 +1g ω_c / ω_2)=0, 得 ω_1 =0. 128.

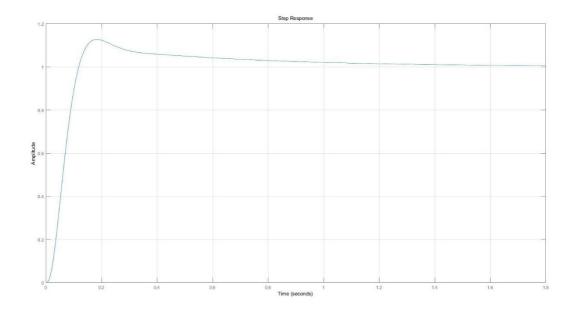
确定高频段与中频段间的过渡特性使校正装置尽可能简单,取ω4=100.

则期望特性系统为

200(s/1.6+1)/s(s/0.128+1)(s/50+1)(s/100+1)

剪切频率为 ω_c =15. 2rad/s, 相角裕度 γ =58. 92°, 满足需求.





四、 校正后系统的时域指标和频率性能

1. 迟后环节优先的迟后-超前校正设计 迟后环节(1. 2s+1)/(23s+1), 超前环节(0. 19s+1)/(0. 02s+1) 200(1. 2s+1)(0. 19s+1)/s(0. 1s+1)(0. 01s+1)(0. 02s+1)(23s+1)(0. 02s+1) 剪切频率为 ωc =15. 88rad/s, 相角裕度 γ =56. 75° .

RiseTime: 0.0730
SettlingTime: NaN
SettlingMin: 0.9062
SettlingMax: 1.1139
Overshoot: 11.3897

Undershoot: 0

Peak: 1.1139 PeakTime: 0.1639

2. 超前环节优先的迟后-超前校正设计1

超前环节(0.1677s+1)/(0.0233s+1),迟后环节(0.6s+1)/(10s+1) 200 (0.1677s+1)(0.6s+1)/s (0.1s+1) (0.01s+1) (0.02s+1) (0.0233s+1) (10s+1) 剪切频率为 ωc =16.12rad/s,相角裕度 γ =48.35°.

RiseTime: 0.0688
SettlingTime: 1.2717
SettlingMin: 0.9013
SettlingMax: 1.2329
Overshoot: 23.2919

Undershoot: 0

Peak: 1.2329
PeakTime: 0.1710

3. 超前环节优先的迟后-超前校正设计 2

超前环节 (0.2s+1)/(0.02s+1),迟后环节 (0.6s+1)/(12s+1) 200 (0.2s+1) (0.6s+1)/s (0.1s+1) (0.01s+1) (0.02s+1) 2 (12s+1) 剪切频率为 ω c =16rad/s, 相角裕度 γ =54.42°

RiseTime: 0.0706
SettlingTime: 1.4515
SettlingMin: 0.9038
SettlingMax: 1.1550
Overshoot: 15.5005
Undershoot: 0

Peak: 1.1550 PeakTime: 0.1649

4. 期望频率法校正设计

校正装置传递函数为(s/1.6+1) (s/10+1)/(s/0.128+1)

200(s/1.6+1)/s(s/0.128+1)(s/50+1)(s/100+1)

剪切频率为ωc =15.2rad/s, 相角裕度 γ =58.92°

RiseTime: 0.0760
SettlingTime: 1.0168
SettlingMin: 0.9025
SettlingMax: 1.1272
Overshoot: 12.7207

Undershoot: 0

Peak: 1.1272 PeakTime: 0.1850