

Root-Locus Techniques

chp \approx 08 (Nise).

→ It is plotting system's dynamic characteristics.

Ex: If system with unity feedback is given

by $G(s) = \frac{K}{s(s+2)}$

char. eqⁿ of above system

$$1 + G(s) H(s) = 0$$

$$\therefore 1 + \frac{K}{s(s+2)} = 0$$

$$s^2 + 2s + K = 0$$

$$\frac{C(s)}{R(s)} = \frac{N(s)}{D(s)}$$

if $K=0$

$$s^2 + 2s = 0$$

$$\therefore s=0, s=-2$$

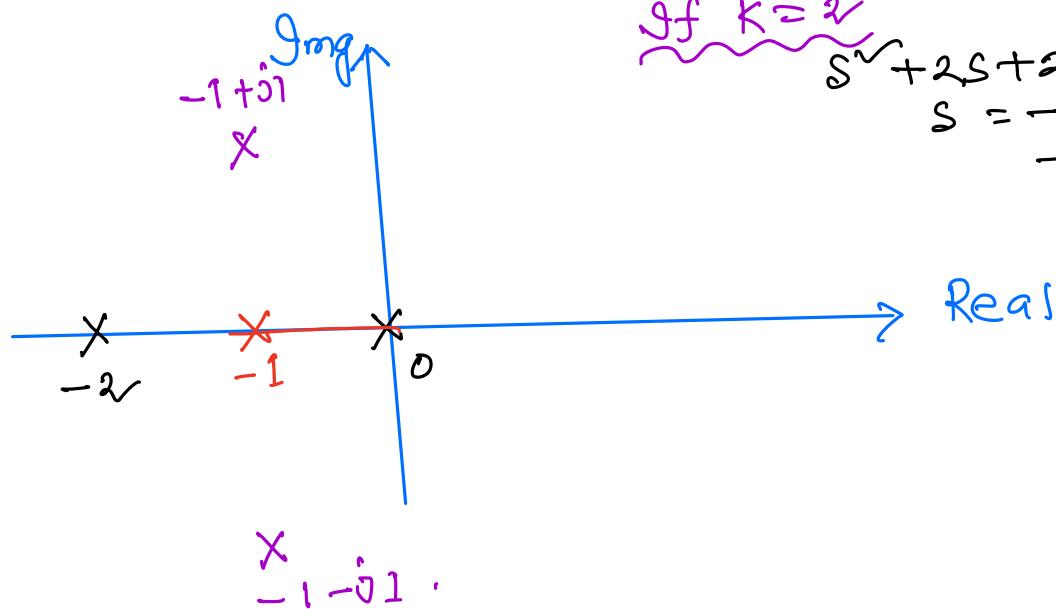
if $K=1$

$$s=-1, -1$$

$$\begin{aligned} s^2 + 2s + 1 &= 0 \\ (s+1)^2 &= 0 \end{aligned}$$

if $K=2$

$$\begin{aligned} s^2 + 2s + 2 &= 0 \\ s &= -1 + j\sqrt{1}, \\ &\quad -1 - j\sqrt{1} \end{aligned}$$



Root locus: It is the representation of the paths of the closed-loop poles as the gain is varied.

Asymptotes:

In the context of **root locus** in control theory, **asymptotes** refer to the straight lines along which the poles of the system tend to move as a system's gain parameter K approaches infinity. These asymptotes describe the direction of the roots of the characteristic equation at extreme values of K , providing insight into how the system behavior changes as the gain increases.

Key Points About Asymptotes in Root Locus:

- Definition:** Asymptotes are lines that represent the direction that the poles (roots of the characteristic equation) move towards as the gain K goes to infinity. The poles on the root locus may diverge to infinity, and these diverging directions are along the asymptotes.

Centroid of Asymptotes:

is the point on the real axis where the asymptotes of the root locus intersect the real axis.

Loci: A locus is the collection of points that share a common geometric property.

Rules of Root locus plot:

Rule - 1: The root locus is always symmetric w.r.t real axis.

$$\frac{P}{Z} = \frac{0}{0}^{-\infty}$$

Rule - 2: Total Loci = $\max \left(\frac{P}{2}, \frac{Z}{2} \right) = 2$

Rule - 3: Total no. of Asymptotes = $\frac{P-Z}{2x+1} = 2-0=2$

Rule - 4: Angle of Asymptotes $\theta = \frac{2x+1}{P-Z} \times 180^\circ$
 $x=0, \theta = \frac{1}{2} \times 180^\circ = 90^\circ$
 $x=1, \theta = 270^\circ$

Rule - 5: Centroid of Asymptotes.

$$P \{ 1+j2, 1-j3 \}, Z \{ 2-j1, 3-j1 \} \Rightarrow \frac{\sum \text{Real } P - \sum \text{Real } Z}{\# \text{ of } P-Z} = \frac{(2+0)-0}{2-0} = \frac{-1}{-1}$$

Rule - 6: Break away point

→ char. eqn $(1+G(S)H(S)=0)$

→ compute $K = \text{polynomial func.}$

→ " $\frac{dK}{ds} = 0$ and find $s = \text{break away point}$

Rule 7° Angle of departure \neq

Rule 8° Intersection to Imag. axis.

→ char. eqⁿ ($1 + G(S)H(S) = 0$)

→ construct Routh array

→ find K for marginally stable system

→ place K in Aux. eqⁿ (2nd order eqⁿ) in Routh array

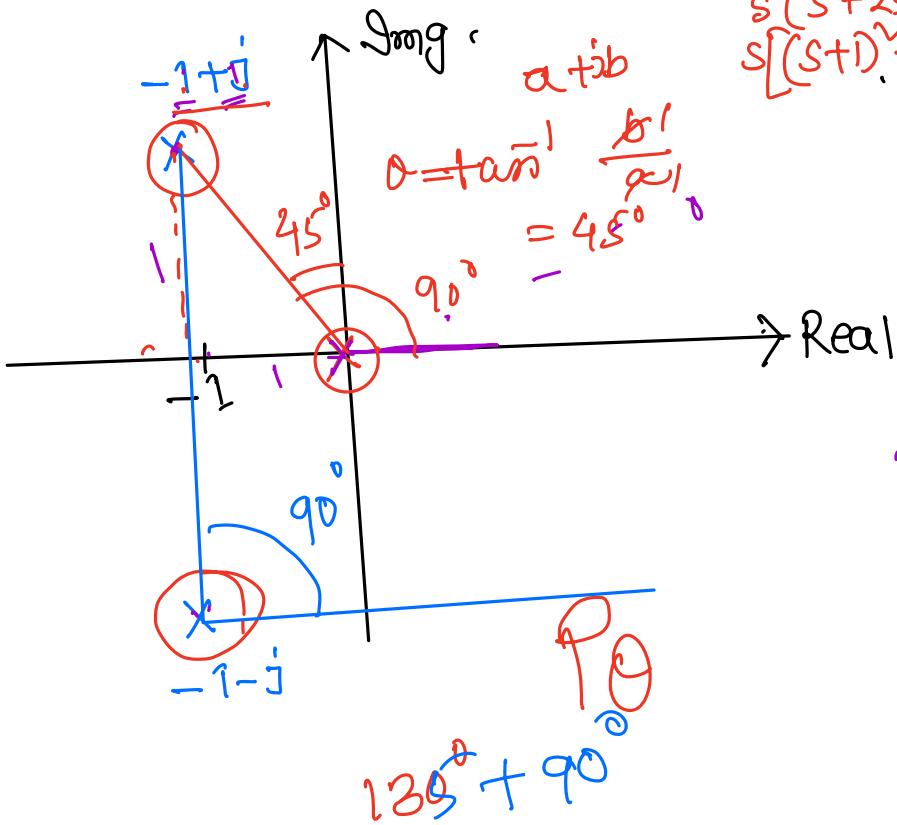
→ $s = \text{Intersection to Imag. axis}$

Rule 7° Angle of departure

Expo 1

$$K(S)H(S) =$$

$$\frac{K}{S(S^2 + 2S + 2)} = \frac{K}{S[(S+1)^2 + 1]}$$



$$\theta_d = 180^\circ - (\sum P_R - \sum Z_R)$$

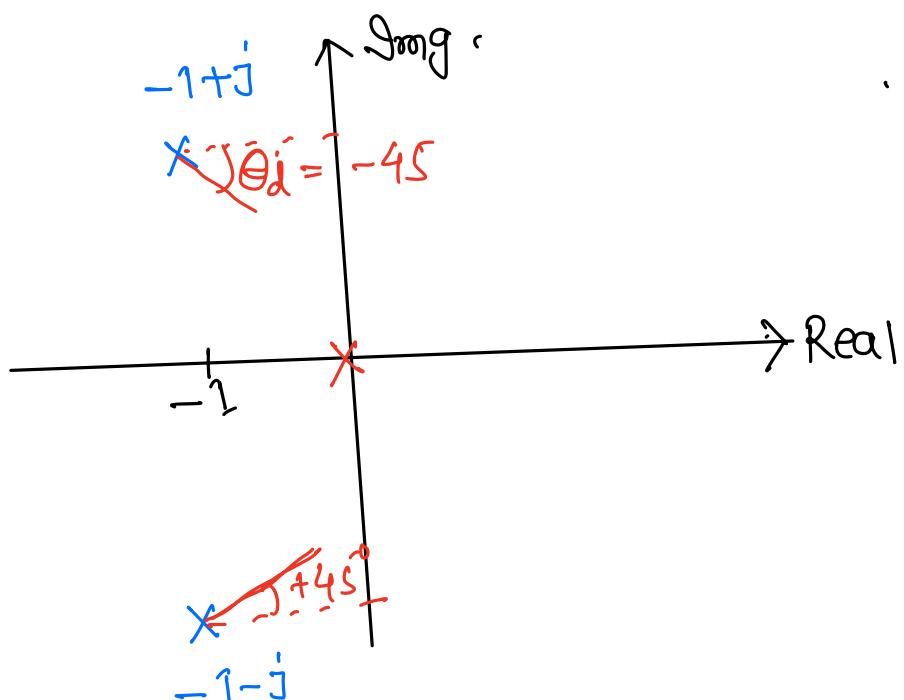
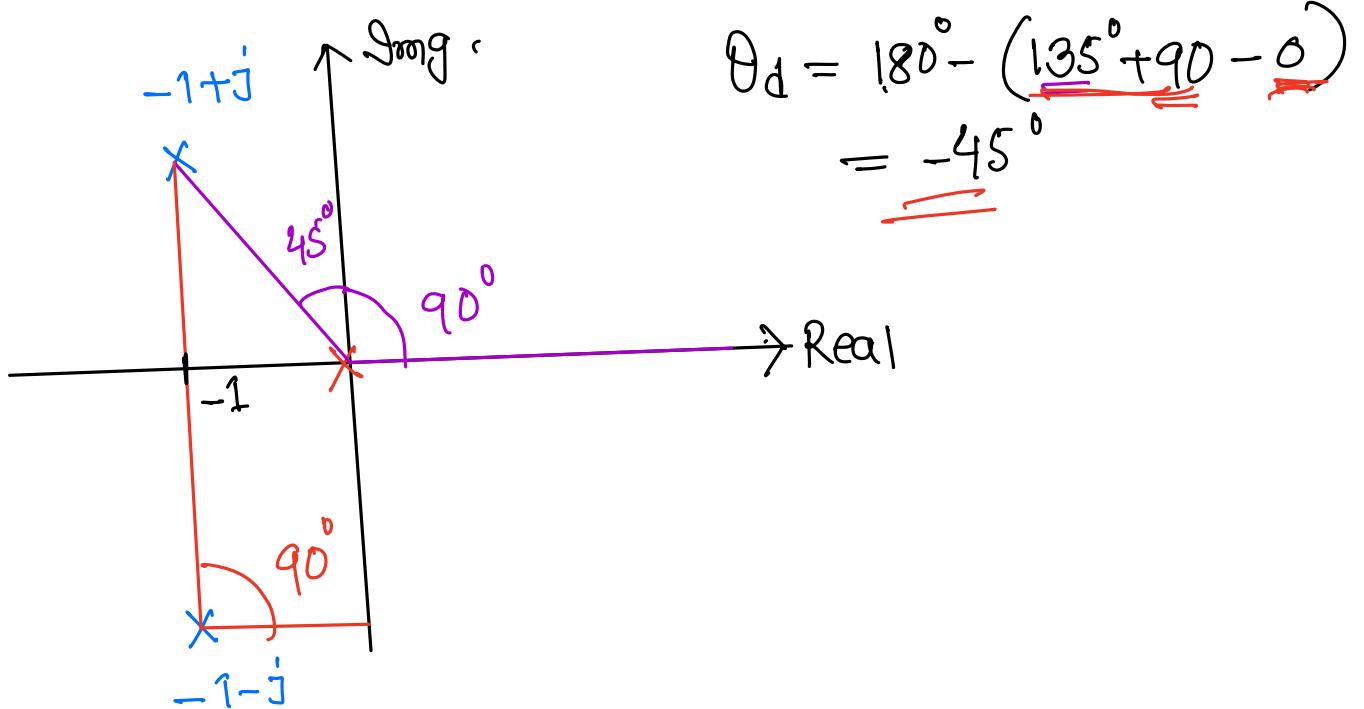
$$S[(S+1)^2 + 1] = 0$$

$$S=0, (S+1)^2 + 1 = 0$$

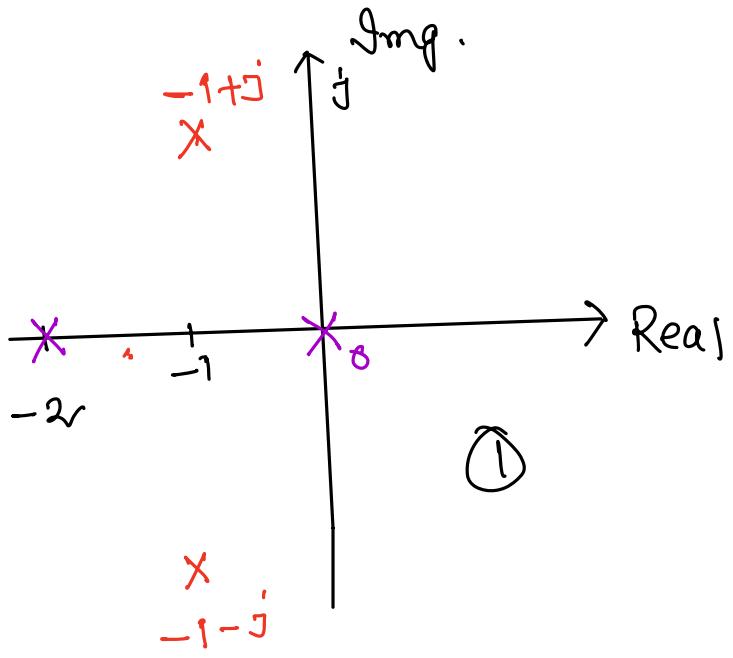
$$(S+1)^2 = -1$$

$$(S+1) = \pm \sqrt{-1} \Rightarrow$$

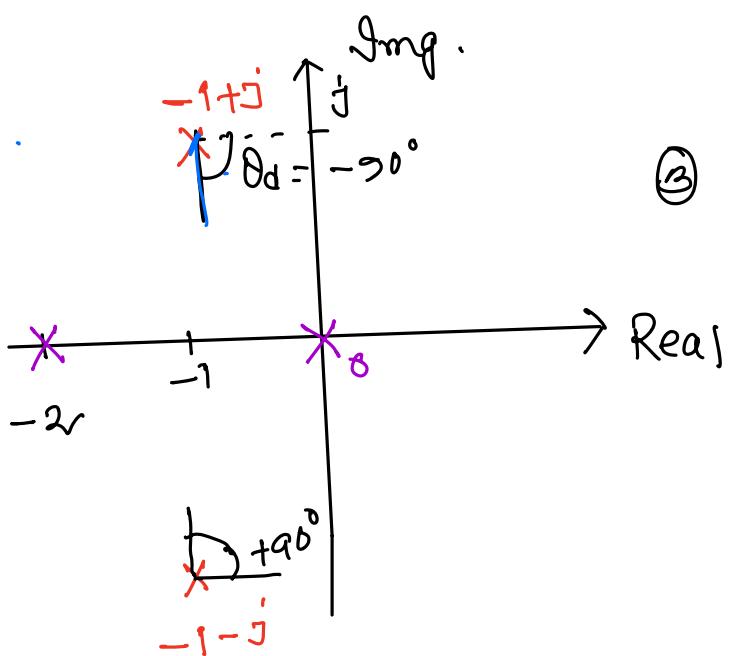
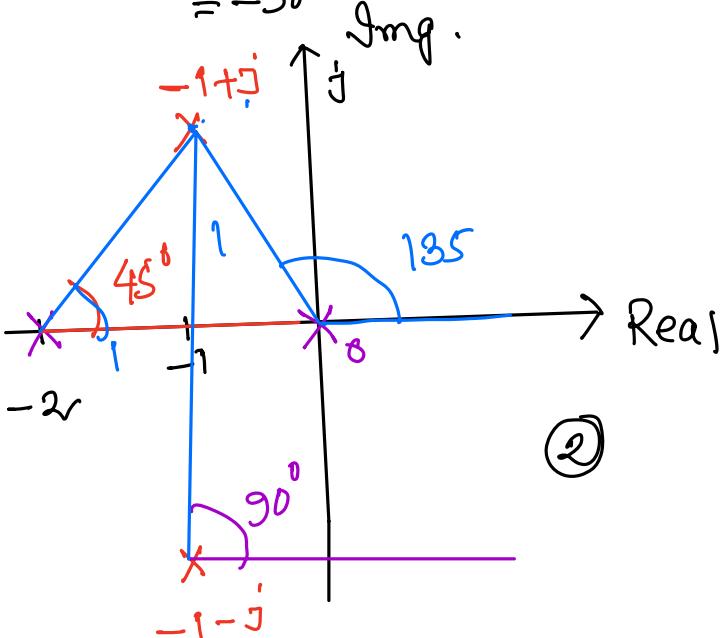
$$S = -1 \pm j$$



$$\text{Exp } \theta_2 \quad K(s) H(s) = \frac{K}{s(s+2)(s^2+2s+1)} = \frac{K}{s(s+2)(s+1)^2}$$



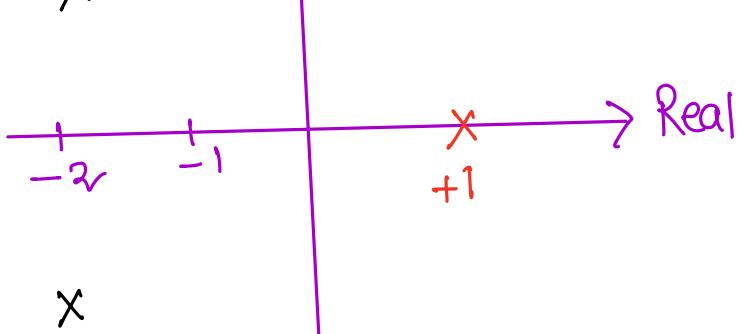
$$\begin{aligned} \theta_d &= 180 - (\sum P_0 - \sum Z_0) \\ &= 180 - (135 + 45 + 90^\circ - 0) \\ &= -90^\circ \end{aligned}$$



$$\begin{aligned} s(s+2)[(s+1)^2 + 1] &= 0 \\ s = 0, s+2 = 0, [(s+1)^2 + 1] &= 0 \\ 0, -2, -1 \pm j \end{aligned}$$

135, 90, 45

$$\text{Exp } \theta_3 \quad K(s) H(s) = \frac{K}{(s-1)(s^2+4s+7)} = \frac{K}{(s-1)[(s+2)^2+3]}$$

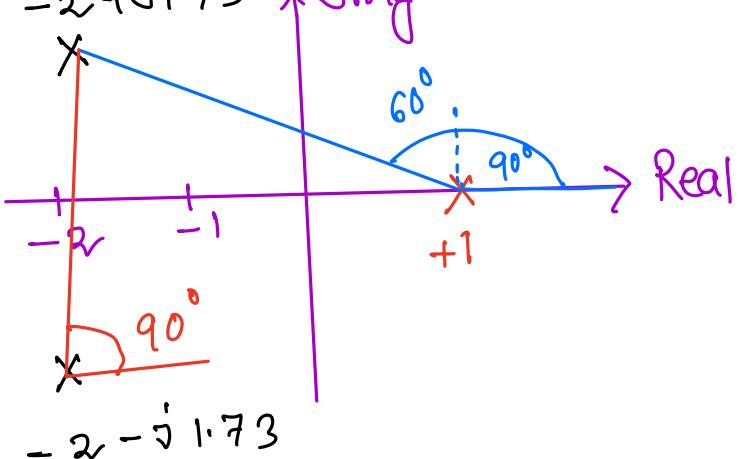


$$s = 1$$

$$s = -2 \pm j1.73$$

$$-2 - j1.73$$

$$-2 + j1.73$$



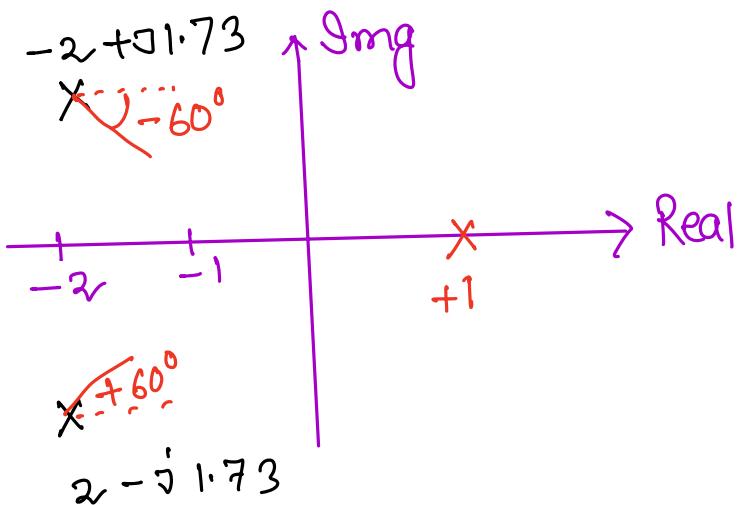
$$\begin{aligned}\theta_d &= 180^\circ - (150^\circ + 90^\circ) \\ &= -60^\circ\end{aligned}$$

$$a + jb$$

$$= \tan^{-1} \frac{b}{a}$$

$$-2 + j1.73$$

$$\cancel{-j-60^\circ}$$



$$\cancel{+60^\circ}$$

$$2 - j1.73$$

Obtain root-Locus plot for the unity feedback system with transfer function.

$$G(s) = \frac{K}{s(s+2)}$$

Step : 1 obtain Loci

$$\text{Poles} = 0, -2$$

$$\# \text{ of poles}, P = 2$$

$$\# \text{ of zeros}, Z = 0$$

$$\therefore \text{Total loci} = \max(P, Z) = 2$$

Step : 2 No. of Asymptotes

$$= P - Z = 2 - 0 = 2$$

Step : 3 Angle of Asymptotes

$$\theta = \frac{2x+1}{P-Z} \times 180^\circ, \quad x = 0, 1$$

$$\therefore \theta = 90^\circ, 270^\circ$$

Step : 4 Centroid of Asymptotes

$$G_c = \frac{\sum \text{Real } P - \sum \text{Real } Z}{\# \text{ of } P - Z}$$

$$= \frac{(-2) - 0}{2 - 0} = -1$$

Step : 5: Breakaway point

$$\rightarrow \text{char. eqn}: 1 + G(s)H(s) = 0$$

$$1 + \frac{K}{s(s+2)} \times 1 = 0$$

$$\therefore s^2 + 2s + K = 0$$

$$\therefore K = -s^2 - 2s$$

$\rightarrow \text{Calculate } \frac{dK}{ds} = 0$

$$\therefore \frac{dK}{ds} = - (2s) - 2 = 0$$

$$\therefore s = -1$$

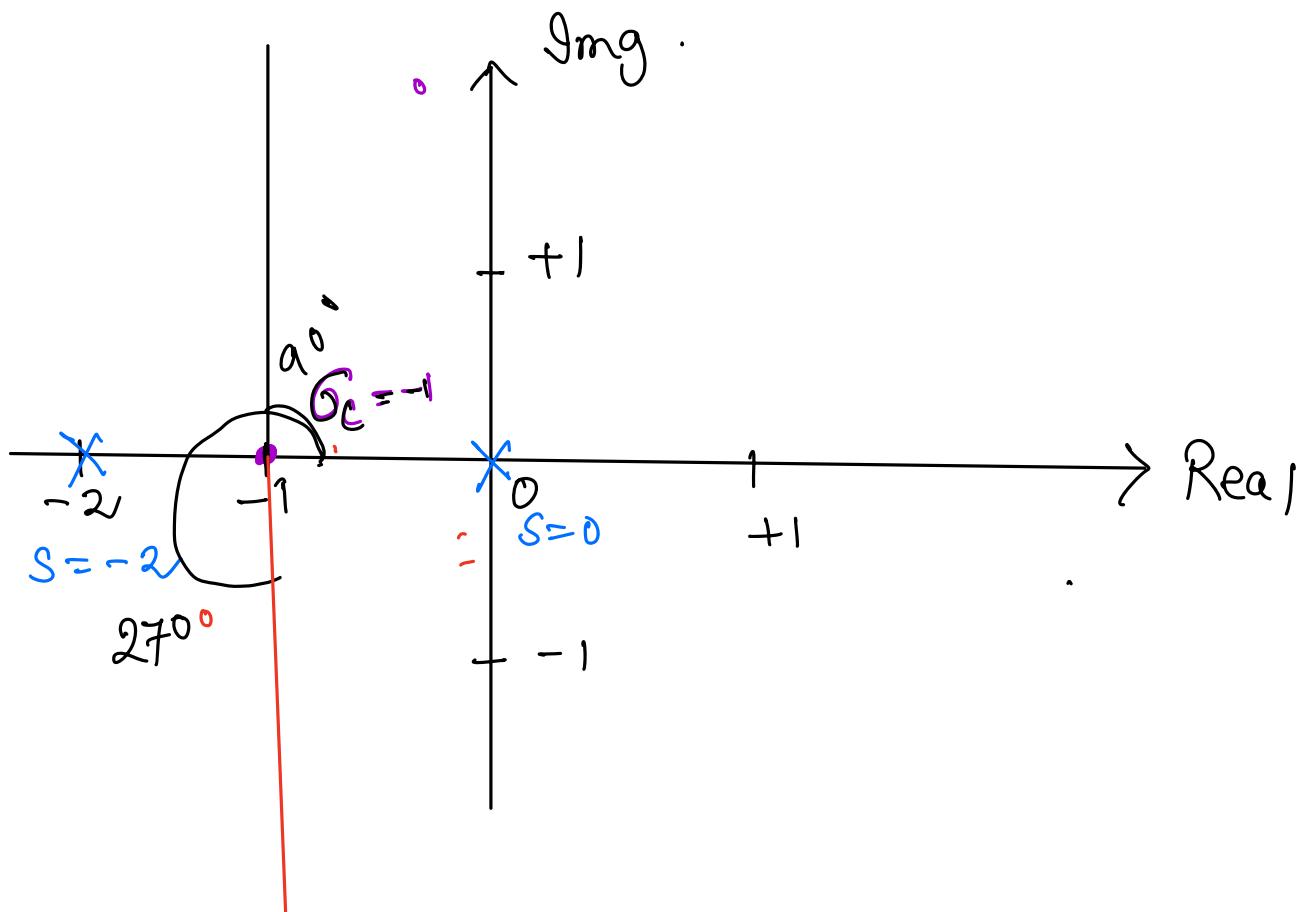
Step: 6: Intersection to Img. axis.

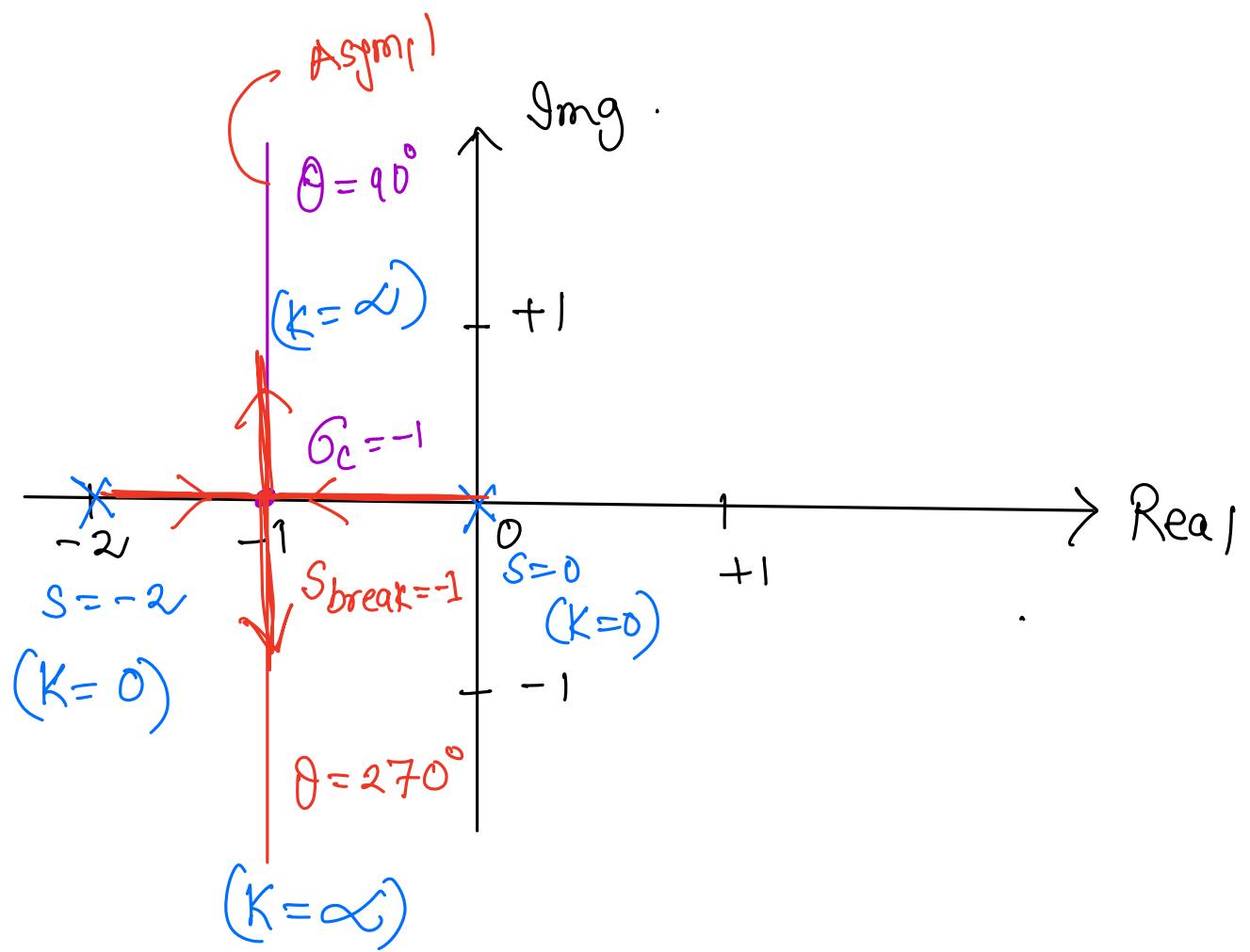
→ char. eqⁿ.
 $1 + G(S)H(s) = 0 \quad \therefore s^2 + 2s + K = 0$

→ Routh array

s^∞	1	K
s^1	2	0
s^0	K	

For stability, $K > 0$.
No K for marginally stable. So, no cross point.





Obtain root-Locus plot for a feedback control system with open loop transfer function.

$$G(s)H(s) = \frac{K(s+2)(s+3)}{(s+1)(s-1)}$$

where K varies from 0 to ∞

SOLN. Z = -2, -3
P = -1, 1

Step 1: No. of loci = max(P, 2) = 2

Step 2: No. of Asymptotes = P - Z = 2 - 2 = 0

Step 3: Break away point

- Char. eqn: $1 + G(s)H(s) = 0$

$$\text{w.r.t. } 1 + \frac{K(s+2)(s+3)}{(s+1)(s-1)} = 0$$

$$\therefore K = -\frac{s^2 + 1}{s^2 + 5s + 6}$$

- find $\frac{dK}{ds} = 0$

$$\Rightarrow \frac{(s^2 + 5s + 6)(2s) - (s^2 - 1)(2s + 5)}{(s^2 + 5s + 6)^2} = 0$$

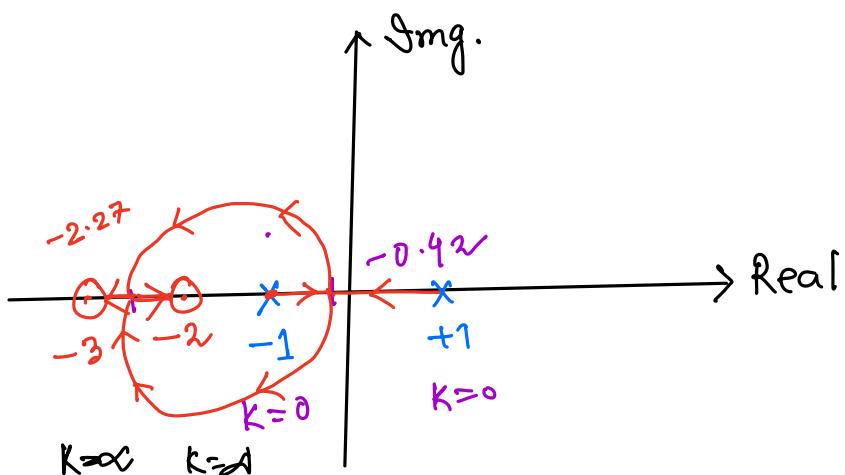
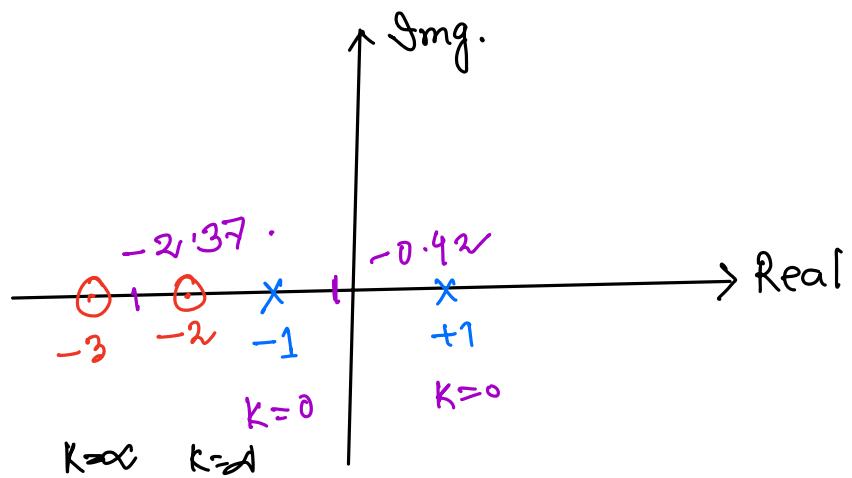
$$\Rightarrow \dots$$

$$\Rightarrow 5s^2 + 14s + 5 = 0$$

$$\Rightarrow s = \frac{-b \pm \sqrt{b^2 - 4ac}}{2a}$$

$$= -0.42, -2.37$$

$$\begin{aligned} \frac{d}{dx} (A/B) \\ = \frac{B \frac{dA}{dx} - A \frac{dB}{dx}}{B^2} \end{aligned}$$

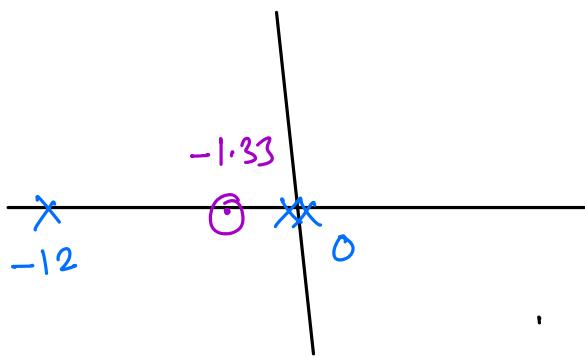


Ex-3 A unity feedback system has an open loop T.F. of $G_l(s) = \frac{k(s+4/3)}{s^2(s+12)}$

Soln: $z_1 = -4/3, P_1, P_2 = 0, P_3 = -12$

Step 1:

$$\begin{aligned} \text{No. of Loci} &= \text{Max}(P, z) \\ &= \text{Max}(3, 1) \\ &= 3 \end{aligned}$$



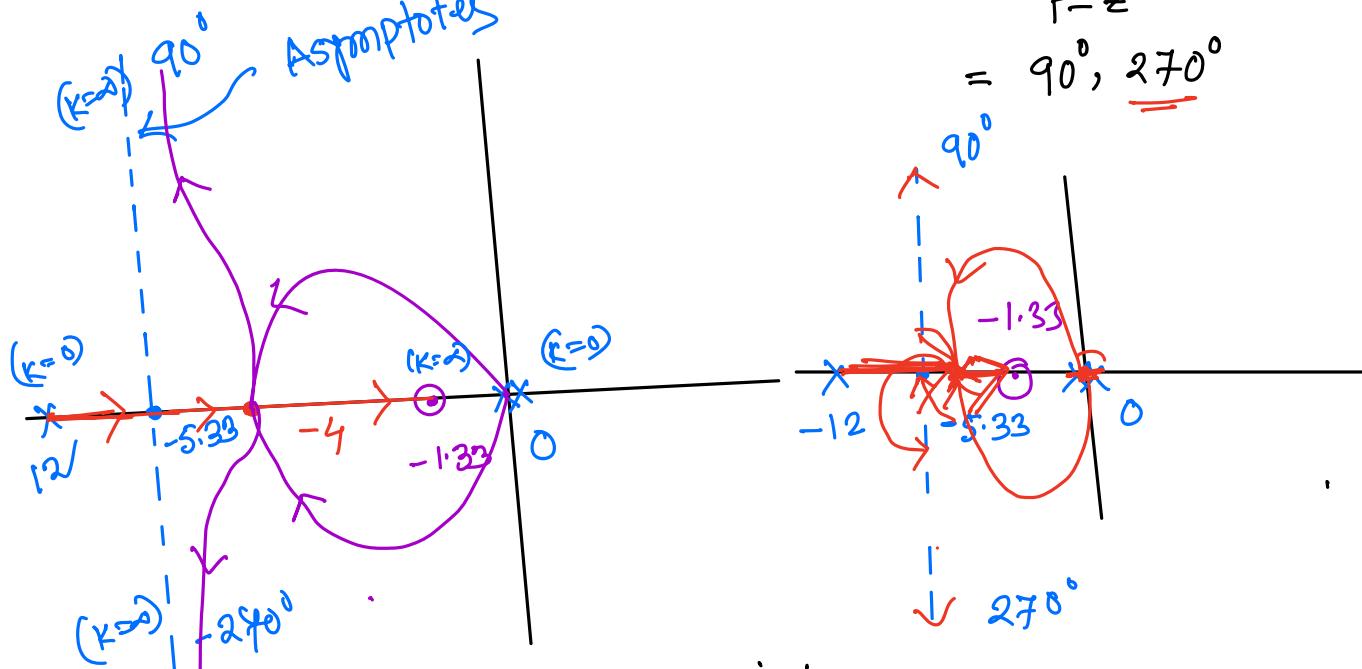
Step 2: No. of Asymptotes = $P - Z = 3 - 1 = 2$

" 3: Centroid " " = $\frac{\sum \text{real of poles} - \sum \text{real of zeros}}{P - Z}$

$$= \frac{(0+0-12) - (-1.33)}{3-1} = -5.33$$

Step 4: Angles of Asymptotes = $\frac{2K+1}{P-Z} \times 180^\circ, K=0, 1, 2, \dots$

$$= 90^\circ, \underline{270^\circ}$$



Step 5: Break away point

$$\text{chr. eqn } 1 + K(s) H(s) = 0$$

$$K = - \frac{s^3 + 12s^2}{s + 4/3}$$

$$\rightarrow \frac{dK}{ds} = 0$$

$$\frac{(s+4/3)(3s^2+24s) - (s^3+12s^2)}{(s+4/3)^2} = 0$$

$$\therefore 2s^3 + 16s^2 + 32s = 0$$

$$\therefore s = 0, -4, -4$$

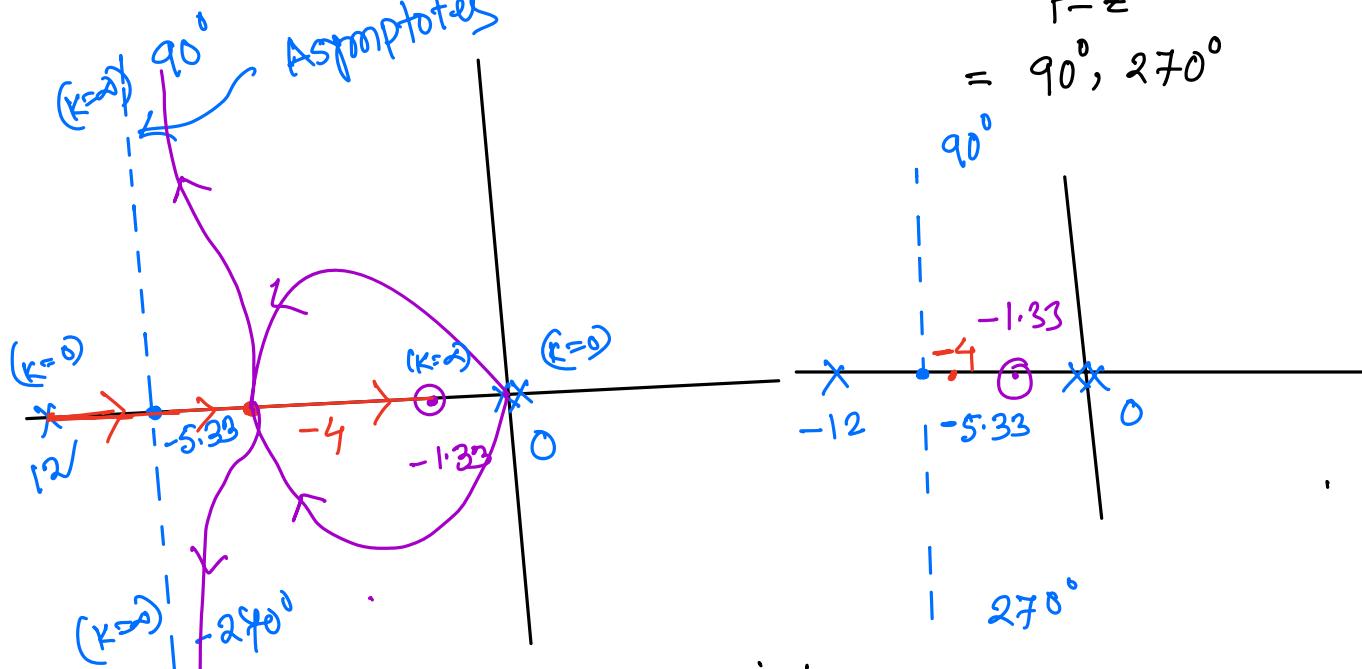
Step 2: No. of Asymptotes = $P - Z = 3 - 1 = 2$

" 3: Centroid " " = $\frac{\sum \text{real of poles} - \sum \text{real of zeros}}{P - Z}$

$$= \frac{(0+0-12) - (-1.33)}{3-1} = -5.33$$

Step 4: Angles of Asymptotes = $\frac{2K+1}{P-Z} \times 180^\circ, K=0, 1, 2, \dots$

$$= 90^\circ, 270^\circ$$



Step 5: Break away point

$$\text{chr. eqn } 1 + K(s) H(s) = 0$$

$$K = - \frac{s^3 + 12s^2}{s + 4/3}$$

$$\rightarrow \frac{dK}{ds} = 0$$

$$\frac{(s+4/3)(3s^2+24s) - (s^3+12s^2)}{(s+4/3)^2} = 0$$

$$\therefore 2s^3 + 16s^2 + 32s = 0$$

$$\therefore s = 0, -4, -4$$

Expt 4 $G(s) H(s) = \frac{K}{s(s+1)(s+2)(s+3)}$

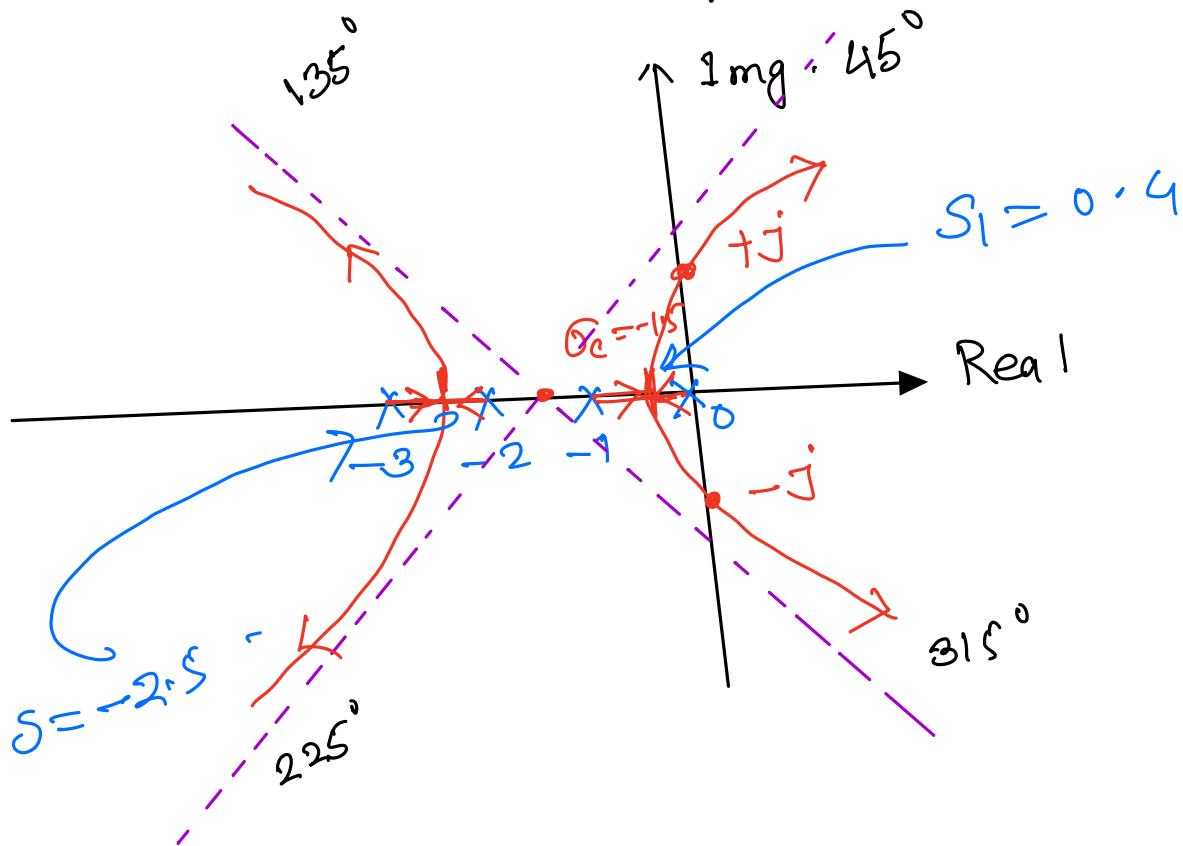
$$\rightarrow P = 0, -1, -2, -3$$

Step 1: No. of Loci = Max (P, 2) = 4

Step 2: " Asy P. = P-Z = 4

Step 3: Centroid = -1.5

Step 4: Angles, $\theta = \frac{(2K+1)180^\circ}{P-Z}$, $K=0, 1, 2, 3$
 $= 45^\circ, 135^\circ, 225^\circ, 315^\circ$



Step 5 Break away:

$$\text{char. eq} : 1 + K(s) G(s) = 0$$

$$\therefore K = \frac{1}{s(s+1)(s+2)(s+3)}$$

$$\frac{dK}{ds} = 0$$

$$\therefore 4s^3 + 18s^2 + 22s + 6$$

$$\therefore s_1 \approx -0.4, \quad s_2 = -2.6$$

Step 6 Intersection \rightarrow long axis.

$$\text{char eq}^n = s^4 + 6s^3 + 12s^2 + 6s + k = 0$$

$$\begin{array}{c|ccc} s^4 & 1 & 11 & k \\ s^3 & 6 & 6 & \\ s^2 & 10 & k & \\ s^1 & \frac{60 - 6k}{10} & 0 & \\ s^0 & k & 0 & \end{array}$$

$$\therefore \frac{60 - 6k}{10} > 0$$

$$\therefore 60 > 6k$$

$$\therefore k < 10$$

\rightarrow Aux eqⁿ

$$10s^2 + k = 0$$

$$s^2 = -k/10$$

If $k \approx 10$,

$$s^2 = -\frac{10}{10} = -1$$

$$\therefore s = \pm j$$