

Homework #1

By time-hale

1. a) proof: $x * h = \int_{-\infty}^{+\infty} x(\tau) h(t-\tau) d\tau$ and $h * x = \int_{-\infty}^{+\infty} h(\tau) x(t-\tau) d\tau$

let $s = t - \tau \Rightarrow \tau = t - s$ then we have the conclusion that

$$h * x = \int_{-\infty}^{+\infty} h(t-s) x(s) ds = \int_{-\infty}^{+\infty} x(\tau) h(t-\tau) d\tau = x * h$$

b) proof: $x * (h_1 + h_2) = \int_{-\infty}^{+\infty} x(\tau) [h_1(t-\tau) + h_2(t-\tau)] d\tau$

$$= \int_{-\infty}^{+\infty} x(\tau) h_1(t-\tau) d\tau + \int_{-\infty}^{+\infty} x(\tau) h_2(t-\tau) d\tau$$

$$= x * h_1 + x * h_2$$

c) proof: $x * (h_1 * h_2) = \int_{-\infty}^{+\infty} x(\tau) (h_1 * h_2)(t-\tau) d\tau = \int_{-\infty}^{+\infty} x(\tau) \int_{-\infty}^{+\infty} h_1(s) h_2(t-\tau-s) ds d\tau$

$$= \int_{-\infty}^{+\infty} x(\tau) h_1(s) h_2(t-\tau-s) d\tau ds \text{ and } (x * h_1) * h_2 = \int_{-\infty}^{+\infty} (x * h_1)(s) h_2(t-s) ds$$

$$= \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} x(\tau) h_1(s-\tau) h_2(t-s) d\tau ds \text{ let } s-\tau = p \text{ we have } s = \tau + p \text{ so}$$

$$\text{Original} = \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} x(\tau) h_1(p) h_2(t-p-\tau) d\tau dp. \text{ they equal to each other.}$$

2. a) $h[n] = u[n] - u[n-1] = \sum_{k=0}^{\infty} \delta[n-k]$ and $\sum_{k=0}^{\infty} |\delta[n-k]| = 1 < \infty$

so it is stable, and for $h[n]$, all the related components are less than 1.

so it is causal and a FIR system

b) $h[n] = 2^n u[n]$, which is a stable, causal IIR system

c) $h[n] = 2^n u[-n]$, which is a stable, not causal, IIR system

d) $h[n] = n(0.8)^n u[n]$, which is a stable, causal, IIR system

3. a) if signals $x[n], y[n]$ satisfy the equation, we have when $\hat{x}[n] = \lambda x[n]$ and $\hat{y}[n] = \lambda y[n]$

assume: $\lambda y[n] + \lambda a_1 y[n-1] + \dots + \lambda a_N y[n-N] = \lambda b_0 x[n] + \dots + \lambda b_M x[n-M]$

① when $\lambda \neq 0$, both side divided λ , we gain the condition happen in the test.

② when $\lambda = 0$, obviously, it satisfy.

All the above, $\hat{x}[n] = \lambda x[n]$, $\hat{y}[n] = \lambda y[n]$ also satisfy the equation.

- b) if $x_1[n]$, $y_1[n]$ and $x_2[n]$, $y_2[n]$ satisfy the difference equation, we have
 $y_1[n] + a_1 y_1[n-1] + \dots + a_N y_1[n-N] = b_0 x_1[n] + \dots + b_M x_1[n-M]$ ①
 $y_2[n] + a_2 y_2[n-1] + \dots + a_N y_2[n-N] = b_0 x_2[n] + \dots + b_M x_2[n-M]$ ②
 add the equation ① and ②, then let $\hat{x}[n] = (x_1 + x_2)[n]$, $\hat{y}[n] = (y_1 + y_2)[n]$.
 we have $\hat{y}[n] + a_1 \hat{y}[n-1] + \dots + a_N \hat{y}[n-N] = b_0 \hat{x}[n] + \dots + b_M \hat{x}[n-M]$
- c) if $x[n]$, $y[n]$ satisfy the difference equation, then let $k = n-L$, we have
 $y[k] + a_1 y[k-1] + \dots + a_N y[k-N] = b_0 x[k] + \dots + b_M x[k-M]$ for k is any
 and then substitute it. we can verify the $\hat{x}[n] = x[n-L]$, $\hat{y}[n] = y[n-L]$ also satisfy
- d) when $x[n]$ is unit impulse, $x[n] = \delta[n]$. we have $y[n] = y[n-1] + \delta[n]$
 $\Rightarrow y[n] = \sum_{k=0}^n \delta[n-k] + 1$ while when $x[n] = 2\delta[n]$. we have $y[n] = y[n-1] + 2\delta[n]$
 $\Rightarrow y[n] = 2 \sum_{k=0}^n \delta[n-k] + 1$ but it should have the linear property.
 $\therefore x[n] \rightarrow 2y[n] = 2 \sum_{k=0}^n \delta[n-k] + 2 \neq y'[n]$, so the response is not scaled by the same constant.

(e) when $y[n] = \begin{cases} 0, & n < 0 \\ y[n-1] + x[n], & n \geq 0 \end{cases}$ From the above analysis, we can easily verify its linearity and time-invariant

4.a) let $x[n] = \delta[n]$, we have $y[n] = h[n] = 0.5 \delta[n-1] + 0.5 \delta[n] + 0.5 \delta[n+1]$

$$b) H(e^{j\omega}) = \sum_{k=-\infty}^{+\infty} h[k] e^{j\omega k} = \sum_{k=-\infty}^{+\infty} [0.5 \delta[k-1] + 0.5 \delta[k] + 0.5 \delta[k+1]] e^{j\omega k}$$

$$= 0.5 e^{j\omega} + 0.5 e^{-j\omega} + 0.5 = 0.5 \cos \omega + 0.5$$

$$\Rightarrow H(e^{j\omega}) = 0.5 (1 + \cos \omega) = \cos^2 \frac{\omega}{2}$$

c) the type of filter is low-pass.

5 a) let $x(t) = e^{j\omega t}$, $y(t) = H(j\omega) e^{j\omega t}$ so substitute them and we have
 $RC \frac{dy(t)}{dt} + y(t) = x(t) \Rightarrow (RC j\omega + 1) H(j\omega) e^{j\omega t} = e^{j\omega t} \Rightarrow H(j\omega) = \frac{1}{1 + RC j\omega}$

b) $H(j\omega) = \frac{1}{1 + RC^2 \omega^2 (1 - jRC\omega)}$, when $\omega \uparrow$, $|H(j\omega)| \downarrow$ confirm a low-pass