

## **Data Provided: Physical constants:**

$$\varepsilon_0 = 8.85 \times 10^{-12} \, Fm^{-1}$$
  
 $\mu_0 = 4\pi \times 10^{-7} Hm^{-1}$ 

# DEPARTMENT OF ELECTRONIC AND ELECTRICAL ENGINEERING

Spring Semester 2015-16 (3.0 hours)

**EEE350 Electromagnetic Fields and Devices** 

Answer FIVE QUESTIONS comprising AT LEAST TWO each from part A and part B. No marks will be awarded for solutions to a sixth question, or if you answer more than three questions from parts A or B. Solutions will be considered in the order that they are presented in the answer book. Trial answers will be ignored if they are clearly crossed out. The numbers given after each section of a question indicate the relative weighting of that section.

## Part A

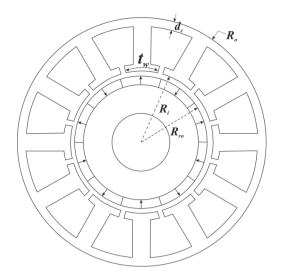
**A1** The major dimensions and parameters of a radial field, 12-slot, 10-pole surface mounted permanent magnet machine shown in Figure A1 are listed in TABLE A1 (*see overleaf*).

By specifying appropriate approximations, answer the following questions.

- a. Assuming the permanent magnets are full arc (180°) and  $\mu_r = 1$ , calculate the magnetic loading B and list possible ways to increase the magnetic loading B. (6)
- **b.** If the allowable peak no-load lamination flux density is 2T, choose a suitable stator tooth width  $(t_w)$  and comment on the result. (4)
- c. For the same peak no-load lamination flux density stated in A1.b, choose a suitable stator yoke thickness  $(d_c)$  and comment on the result. (4)
- **d.** Calculate the electric loading Q if the total Ampere-Turns for the 3-phase windings are NI = 1000A. List the limiting factors for increase in electrical loading Q.
- e. Calculate the electromagnetic torque using the previously obtained electrical loading Q and magnetic loading B. (3)

**(6)** 

**(6)** 



# TABLE A1

Slot number $(N_s)$	12
Pole number (2 <i>p</i> )	10
Stator outer radius $(R_o)$	50 mm
Stator inner radius $(R_i)$	28.5 mm
Active axial length $(L_a)$	50 mm
Air-gap length $(L_g)$	1 mm
Rotor outer radius $(R_{ro})$	27.5 mm
Magnet thickness $(L_m)$	3 mm
Magnet remanence (Br)	1.2 T

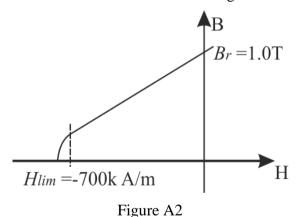
Figure A1

A2 A 4-pole surface mounted permanent magnet machine has 15 stator slots  $(N_s)$  and the following dimensions:

Rotor outside diameter (D) = 30 mm Rotor active length (L) = 80 mm Stator slot area ( $A_s$ ) = 30 mm<sup>2</sup>/slot Magnet pole arc ( $\alpha$ ) = 140° (elec.) Effective airgap ( $l_g$ ) = 1.0 mm

Magnet recoil permeability ( $\mu_r$ ) = 1.1

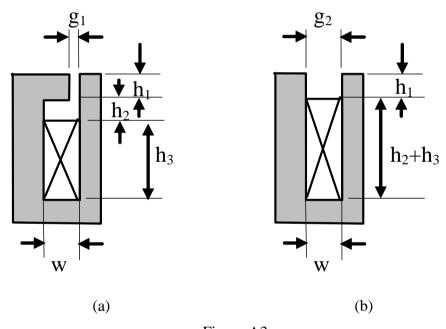
The magnet material has the B-H characteristic shown in Figure A2.



- a. If the winding has a current density of  $J = 8 \text{ A/mm}^2$  and a packing factor of 0.5, calculate the magnet thickness required to obtain a power output of 600 W at 5000 rpm.
- **b.** List possible ways to improve the open-circuit airgap flux density. (3)
- **c.** Check whether the magnet will be demagnetised at three times the full-load torque.
- **d.** Sketch the difference between reversible and irreversible demagnetizations. (3)
- e. Explain the influence of temperature on the irreversible demagnetization for both Ferrite and NdFeB permanent magnets. (2)

**(4)** 

- A3 a. Derive general expressions for the winding pitch-factor,  $K_p$ , and the distribution-factor,  $K_d$ , for both fundamental and n-th EMF harmonic components. (6)
  - **b.** By employing coil vectors, determine the coil connections that yield a maximum winding factor for a 3-phase, 12-slot, 14-pole permanent magnet machine with double layer, concentrated windings.
  - c. Repeat A3.b for a 3-phase, 12-slot, 14-pole permanent magnet machine with single layer, concentrated windings. (3)
  - **d.** Explain the main advantages and disadvantages of a single layer winding. (4)
  - e. Describe the main differences between concentrated winding and distributed winding and explain their merits and disadvantages. (3)
- **A4 a.** Derive an expression for the slot leakage inductance per-unit length, for a slot shown in Figure A3 (a) with N number of turns, specifying any assumptions which need to be made. Please note full marks will not be given if appropriate assumptions are not specified.



- Figure A3 (6)
- b. For large wind power generators and hydroelectric generators, it is often preferred to employ open slots, as shown in Figure A3(b). Calculate the difference in winding slot-leakage inductances between Figures A3(a) and 3(b). (4)
- c. State the advantages and disadvantages of the slot configurations shown in Figures A3(a) and 3(b). (6)
- **d.** List possible ways to increase winding inductances and state their limitations. (4)

**(7)** 

### Part B

**B1.** a. The variation in the electric flux density in a region of space of permittivity  $\varepsilon_0$  is given by:  $\vec{D} = 3x^2y\vec{u_x} + 4x^2y^2\vec{u_y} + 3y^2\vec{u_z}$  C/m<sup>2</sup>

Calculate the electric field and the divergence of  $\vec{D}$  at the point (1,3,4)m. (4)

- **b.** A single isolated circular conductor of radius  $R_c$  carries a uniform volume charge density q. Starting from Gauss's Law, derive expressions for the variation in the electric field  $\vec{E}$  with distance from the centre of the conductor, r, for:
  - i) The region within the conductor (i.e.  $r < R_c$ )
  - ii) The region outside the conductor (i.e.  $r > R_c$ ) (6)
- c. Figure B1.1 shows a single-phase, two-wire power distribution line, in which the two wires have the same radius  $R_c$  and volume charge densities of  $+q_1$  and  $-q_2$ . By application of the principle of superposition, and assuming  $q_1=q_2$ , derive an expression for the potential difference between the two wires.

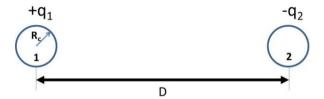


Figure B1.1 Single-phase, two wire power distribution line

d. Calculate the capacitance per unit length for the arrangement shown in Figure B1.1 if  $R_c$ =10mm and D=500mm and that the permittivity of the region outside the conductors is  $\varepsilon_0$ . (3)

EEE350 4 CONTINUED

**(3)** 

**(5)** 

**B2.** a. The electric potential in region of space with permittivity  $\varepsilon_0$  is given by:

$$V = (3yx^3 + 5zx + 7xyz + 12)V$$

Calculate the following at the point (x,y,z) = (3,2,1)m

- i) The electric field strength
- ii) The charge density (3)
- **b.** The variation in electric field strength in a region of space is given by:

$$\vec{E} = 3x^2y \, \overrightarrow{u_x} + 4x^2y^2 \overrightarrow{u_y} \, V/m$$

Calculate the electric potential between the points (2,2)m and (5,5)m (4)

- **c.** A sphere of  $R_s$  carries a uniform volume charge density of q and is surrounded by a region of space of permittivity  $\varepsilon_0$ .
  - i) Starting from Gauss's Law, derive an expression for the electric flux density at a distance *r* from the centre of the charged sphere
  - ii) If the charged sphere has a radius of 10mm and carries a uniform volume charge density of  $1\times10^{-3}$  C/m<sup>3</sup>, calculate the distance from the centre of the sphere at which the magnitude of the electric field strength is 10,000 V/m.
- A 50m long high voltage cable is comprised of a central circular conductor which is surrounded by an insulating layer with a relative permittivity of 6.0 and a minimum electric field breakdown strength of 30 × 10<sup>6</sup>V/m. When this high voltage cable is installed, the 50m length has a capacitance to ground of 0.334μF. Calculate the minimum radius of the central circular conductor if the cable is to sustain a DC voltage of 30kV without breakdown of the insulation.

EEE350 5 TURN OVER

(9)

**(4)** 

**B3.** a. Starting from the vector forms of Faraday's and Ampere's Laws for time varying fields shown below, and listing any assumptions that you make, derive the diffusion equation for a time varying magnetic fields at typical power system frequencies:

$$\nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$$
$$\nabla \times \vec{H} = \vec{J} + \frac{\partial \vec{D}}{\partial t}$$

You may find it useful to make use of the following identity for a general vector  $\vec{F}$ .

$$\nabla \times (\nabla \times \vec{F}) = \nabla(\nabla \cdot \vec{F}) - \nabla^2 \vec{F} \tag{7}$$

b. Figure B3.1 shows a simplified cross-section through a single iron lamination which has a thickness 2b. The lamination is exposed to a sinusoidally time varying magnetic flux which is oriented along the z-axis as shown. The maximum value of the time varying magnetic field strength at each surface is  $H_s$ . Starting from the diffusion equation, show that the x component of current density in the lamination is given by:

$$J_x = \alpha H_s \frac{\sinh(\alpha y)}{\cosh(\alpha b)} e^{j\omega t}$$

where  $\alpha^2 = j\omega\sigma\mu$ 

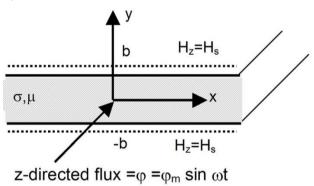


Figure B3.1 Cross-section through an iron lamination of thickness 2b

- c. Calculate the maximum lamination thickness which it would be good design practice to use if the lamination has a *relative* permeability of 500 and an electrical conductivity of  $3\times10^6$  Sm<sup>-1</sup>, and is exposed to a applied magnetic field in the z-direction which has a frequency of 100Hz.
- **B4.** a. The magnetic vector potential in a region of space of magnetic permeability  $\mu_0$  is given by:

$$\vec{A} = 8y^3x \, \overrightarrow{u_x} + 14xz \, \overrightarrow{u_y} + 3xy^3z^2 \, \overrightarrow{u_z} \, Wb/m$$

where  $\overrightarrow{u_x}$ ,  $\overrightarrow{u_y}$  and  $\overrightarrow{u_z}$  are unit vectors in a Cartesian coordinate system

Calculate the magnetic field strength,  $\vec{H}$ , as the point (0.1,0.7,0.15) m. (4)

**b.** The magnetic field strength in a region of space of magnetic permeability  $\mu_0$  is given by:

$$\vec{H} = 6y^2z \overrightarrow{u_x} + 4yx^3 \overrightarrow{u_y} + 2x^2z^3 \overrightarrow{u_z} A/m$$

where  $\overrightarrow{u_x}$ ,  $\overrightarrow{u_y}$  and  $\overrightarrow{u_z}$  are unit vectors in a Cartesian coordinate system.

Calculate the current density at the point (2,5,2) m.

**(4)** 

**(3)** 

- **c.** Figure B4.1 shows the boundary between a region of air and an idealised region of iron with infinite permeability. Assuming that the problem domain is two-dimensional and there is no current present at the interface, then:
  - i) State two magneto-static boundary conditions that could be imposed on this boundary.
  - ii) State a boundary condition that can applied to the magnetic vector potential in Region 1.

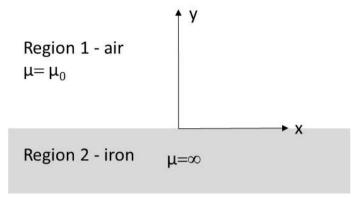


Figure B4.1. Two-dimensional interface between an air and iron region

**d.** A surface current density whose spatial variation with x is defined by the following expression is now introduced at the boundary shown in Figure B4.1:

$$J_z = 100 \cos(x) \text{ A/m}$$

Calculate the magnetic field strength component  $H_x$  in the air region at the boundary at the point x = 1.32m

(3)

**(4)** 

- e. A large cable carrying a DC current of +5000A enters a manufacturing plant. A sensitive piece of measuring equipment in the plant must be subjected to a magnetic field of  $<450\mu T$  in order to function properly.
  - Starting from Ampere's Law derive an expression for the minimum distance that the instrument must be located away from the incoming cable in order to function properly.

What two practical steps could be taken to reduce this distance? (2)

#### GWJ/GJL/ JBW

f.