

①

Q1 (a) (i) l.f. gain due to resistors.

$$\therefore \text{lf gain} = \frac{R_2}{R_1 + R_2}$$

hf. gain due to capacitors

$$\text{hf gain} = \frac{C_1}{C_1 + C_2}$$

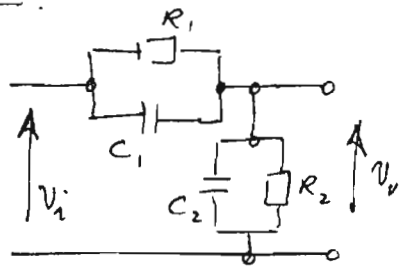
(ii).

$$\frac{V_o}{V_i} = \frac{\frac{R_2 / j\omega C_2}{R_2 + 1/j\omega C_2}}{\frac{R_2 / j\omega C_2}{R_2 + 1/j\omega C_2} + \frac{R_1 / j\omega C_1}{R_1 + 1/j\omega C_1}}$$

$$= \frac{\frac{R_2}{1 + j\omega C_2 R_2}}{\frac{R_2}{1 + j\omega C_2 R_2} + \frac{R_1}{1 + j\omega C_1 R_1}} = \frac{R_2 (1 + j\omega C_1 R_1)}{R_2 (1 + j\omega C_1 R_1) + R_1 (1 + j\omega C_2 R_2)}$$

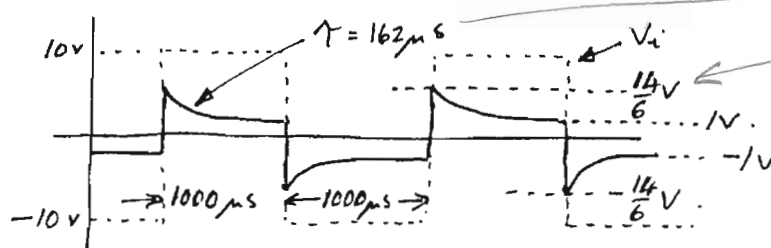
$$= \frac{R_2 (1 + j\omega C_1 R_1)}{R_1 + R_2 + j\omega R_1 R_2 (C_1 + C_2)} = \frac{R_2}{R_1 + R_2} \cdot \frac{1 + j\omega C_1 R_1}{1 + j\omega \frac{R_1 R_2}{R_1 + R_2} (C_1 + C_2)}$$

$$\equiv k \cdot \frac{1 + j f/f_0}{1 + j f/f_1}$$

(iii) if  $R_1 = 9\text{M}\Omega$ ,  $R_2 = 1\text{M}\Omega$ ,  $C_1 = 30\text{pF}$ ,  $C_2 = 150\text{pF} \dots$ 

$$\text{lf gain} = \frac{1}{10} \quad \text{hf gain} = \frac{30}{180} = \frac{1}{6}$$

$$\text{and } \tau = \frac{R_1 R_2}{R_1 + R_2} (C_1 + C_2) = 900\text{k}\Omega \times 180\text{pF} = 162\mu\text{s}$$



(2)

Q1 cont...

(a) (iv) to remove all frequency dependence we need to make:

$$C_1 R_1 = \frac{(C_1 + C_2) R_1 R_2}{R_1 + R_2}$$

$$C_1 \cdot 9 \times 10^6 = (C_1 + 150 \text{ pF}) 9 \times 10^5$$

$$\text{or } 10 C_1 = C_1 + 150 \text{ pF}$$

$$\text{or } 9 C_1 = 150 \text{ pF}$$

$$\text{or } C_1 = \frac{150}{9} \text{ pF} = \underline{\underline{16.7 \text{ pF}}}$$

(b) (i) single op-amp has GBP of 50 MHz.

if gain required is 25, available -3dB BW is  $\frac{50 \text{ MHz}}{25} = 2 \text{ MHz} \rightarrow \text{too small.}$

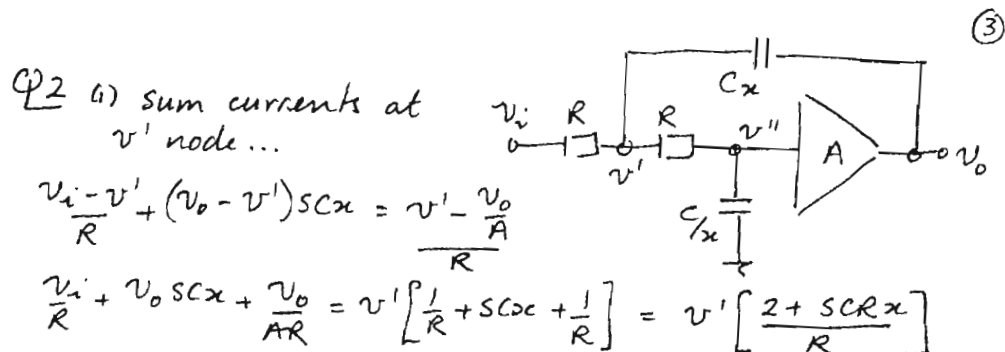
[ or could say, if BW required is 3 MHz, available gain is  $\frac{50 \text{ MHz}}{3 \text{ MHz}} = 16.7$  which is too small ]

(ii) two identical stages so each must have a gain  $= \sqrt{25} = \underline{\underline{5}}$ .

This means that each amplifier will have a -3dB bandwidth of  $\frac{50 \text{ MHz}}{5} = 10 \text{ MHz}$  ..... we want the -1.5dB frequency of one stage to get the -3dB of the cascade. for one stage  $\frac{V_o}{V_i} = \frac{5}{1 + j f / 10 \text{ MHz}}$  and we want to know when this falls to  $5 \times 10^{-1.5/20}$ . (= 4.207).

$$\frac{25}{1 + f^2 / 10^{14}} = 17.70 \text{ or } \frac{f^2}{10^{14}} = \frac{25}{17.7} - 1 = 0.412$$

$$f_{-3\text{dB for cascade}} = 10^7 \sqrt{0.412} = \underline{\underline{6.42 \text{ MHz}}}$$



or  $v_i + v_o sCRx + \frac{v_o}{A} = v'(2 + sCRx)$

The relation between  $v'$  and  $v''$  is

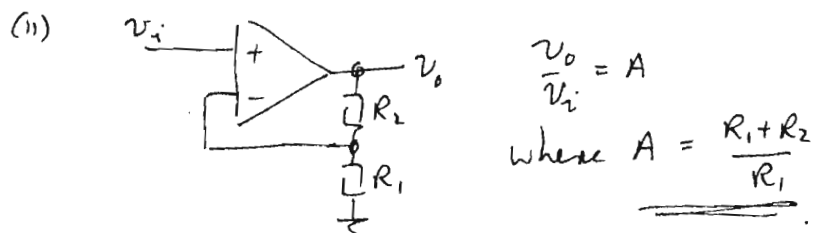
$$v'' = \frac{v_o}{A} = v' \frac{1/j\omega C_x}{R + 1/sC_x} = v' \frac{1}{1 + \frac{sCRx}{x}}$$

$$\therefore v_i + v_o sCRx + \frac{v_o}{A} = \frac{v_o}{A} \left( 1 + \frac{sCRx}{x} \right) (2 + sCRx)$$

$$v_i A = v_o \left[ 2 + sCRx + \frac{2sCRx}{x} + s^2 C^2 R^2 - 1 - sCRx \right]$$

$$= v_o \left[ 1 + sCR \left( \frac{2}{x} + x(1-A) \right) + s^2 C^2 R^2 \right]$$

or  $\frac{v_o}{v_i} = \frac{A}{1 + sCR \left( \frac{2}{x} + x(1-A) \right) + s^2 C^2 R^2}$



(iii) for stability, damping term must be  $+ve$

$$\therefore \frac{2}{x} + x(1-A) > 0 \quad \text{or} \quad x(1-A) > -\frac{2}{x}$$

or  $-A > -\frac{2}{x^2} - 1 \quad \text{or} \quad A < \frac{2}{x^2} + 1$

④

Q2 cont...

(iv) by comparison with standard form..

$$\underline{\underline{\omega_0 = \frac{1}{CR}}} \quad \text{or} \quad \underline{\underline{f_0 = \frac{1}{2\pi CR}}}$$

$$\text{and } \frac{1}{\omega_0 Q} = CR \left( \frac{2}{x} + (1-A)x \right) = \frac{CR}{Q}$$

$$\text{or } \underline{\underline{Q = \frac{1}{(2/x + (1-A)x)}}}$$

(v) 2nd order section

$$\frac{V_o}{V_i} = \frac{1}{1 + s + s^2}$$

1st order section

$$\therefore \left. \begin{array}{l} \omega_0 = 1 \\ Q = 1 \\ \omega_0 = 1 \end{array} \right\} \begin{array}{l} 1 = \text{cut-off} \\ \text{frequency} \\ \text{of overall} \\ \text{filter.} \end{array}$$

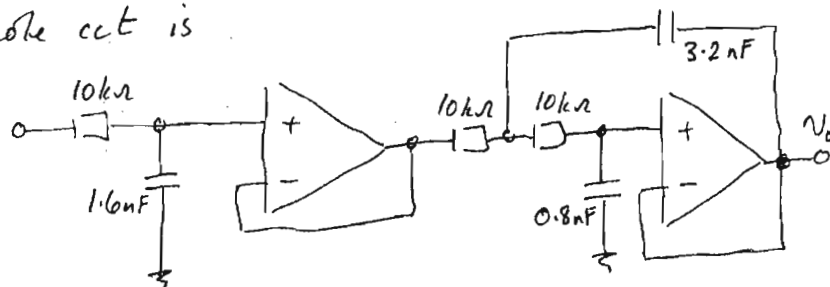
2nd order.

$$\omega_0 = 2\pi \cdot 10^4 = \frac{1}{CR} = \frac{1}{C \cdot 10k\Omega} \quad \text{or} \quad \underline{\underline{C = 1.6 nF}}$$

$$Q = \frac{x}{2} \quad (\text{since } A=1) \quad \therefore \quad \underline{\underline{x = 2}}$$

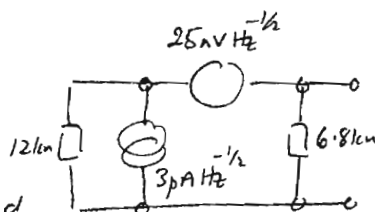
$$\text{1st order... } \omega_0 = \frac{1}{RC} \quad \text{or} \quad \underline{\underline{C = 1.6 nF}}$$

whole ckt is



first order section must come before  
second order section.

Q3 (a) (i) we want the voltage across the  $6.8k\Omega$  resistor due to each source in turn, except for the  $6.8k\Omega$  resistor noise where we want the noise developed across the  $12k\Omega$  resistor.



$$\bar{V}_{on}^2|_{12k\Omega} = 4kT \cdot 12k\Omega \left( \frac{6.8k\Omega}{12k\Omega + 6.8k\Omega} \right)^2 = 198 \times 10^{-18} \times 0.131 = 26.1 \times 10^{-18}$$

$$\bar{V}_{on}^2|_{6.8k\Omega} = 4kT \cdot 6.8k\Omega \left( \frac{12k\Omega}{12k\Omega + 6.8k\Omega} \right)^2 = 113 \times 10^{-18} \times 0.407 = 46.0 \times 10^{-18}$$

$$\bar{V}_{on}^2|_{25nV} = (25nV)^2 \left( \frac{6.8k\Omega}{12k\Omega + 6.8k\Omega} \right)^2 = 625 \times 10^{-18} \times 0.131 = 81.9 \times 10^{-18}$$

$$\bar{V}_{on}^2|_{3pA} = (3pA)^2 \left( \frac{6.8k\Omega \times 12k\Omega}{12k\Omega + 6.8k\Omega} \right)^2 = 1296 \times 10^{-18} \times 0.131 = 169 \times 10^{-18}$$

$$\therefore \bar{V}_{onT}^2 = (26.1 + 46 + 81.9 + 169) \times 10^{-18} = 323 \times 10^{-18} \text{ V}^2/\text{Hz}$$

By inspection  $R_{th}$  is  $12k\Omega // 6k\Omega = 4.34k\Omega$ .

(ii) The noise generated by  $R$  is  $4kTR \text{ V}^2/\text{Hz}$ . If all the noise at the output of the network is assumed to emanate from  $R$  ....

$$323 \times 10^{-18} = 4 \times 1.38 \times 10^{-23} \times T_{eff} \times 4.34k\Omega$$

$$\text{or } T_{eff} = 1348 \text{ K}$$

(b) (i) Signal to Noise ratio is the ratio of signal power to noise power at a specified node in the circuit. It is a measure of signal quality.

$$SNR = \frac{S_k}{N_k} \text{ where } k \text{ is the node of interest.}$$

⑥

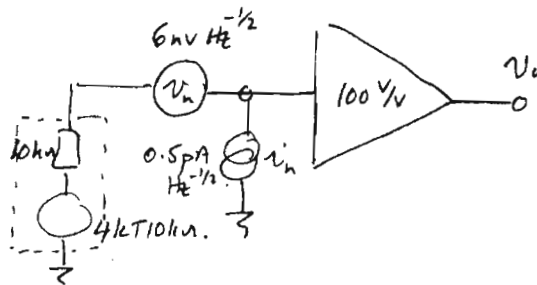
Q3 (b) cont...

Noise factor is defined as the  $S/N$  at a system input divided by  $S/N$  at system output. It is a measure of the degree to which the system degrades the signal.

$$F = \frac{S_i/N_i}{S_o/N_o} = \frac{S_i}{S_o} \times \frac{N_o}{N_i} = \frac{N_o}{A_p N_i}$$

2.

(c) (i)



3.  
(minus 1 per error.)

$$\begin{aligned} (i) \quad \overline{V_o^2} &= A^2 \left[ \overline{i_n^2} R_s^2 + \overline{V_n^2} + 4kT/10k\Omega \right] \\ &= 10^4 \left[ (0.5 \times 10^{-12} \times 10^4)^2 + 36 \times 10^{-18} + 165.6 \times 10^{-18} \right] \\ &= 2.26 \times 10^{-12} \text{ V}^2 \text{ Hz}^{-1} \end{aligned}$$

2.

Total voltage over a bandwidth of 10kHz

$$\begin{aligned} \overline{V_{ot}^2} &= 2.26 \times 10^{-12} \times 10 \text{ kHz} \\ &= 2.26 \times 10^{-8} \text{ volt}^2 \end{aligned}$$

$$\therefore \text{rms reading} = \underline{150 \mu\text{V}}$$

2.

(7)

Q4 (i) Power supplied = Power to load + Power dissipated

$$\text{or } P_D = P_S - P_L$$

$$P_D = \int_0^T V_S I_S dt - \frac{V_{Lrms}^2}{R_L} = 2V_{CC} I_{AVE} - \frac{V_P^2}{3R_L}$$

$$= \frac{2V_{CC} V_P}{4R_L} - \frac{V_P^2}{3R_L} \quad \begin{matrix} \uparrow \\ \text{because there} \\ \text{are 2 supplies.} \end{matrix}$$

$$\frac{dP_D}{dV_P} = \frac{2V_{CC}}{4R_L} - \frac{2V_P}{3R_L} = 0 \text{ for a maximum}$$

$$\text{or } V_P = 3V_{CC}/4$$

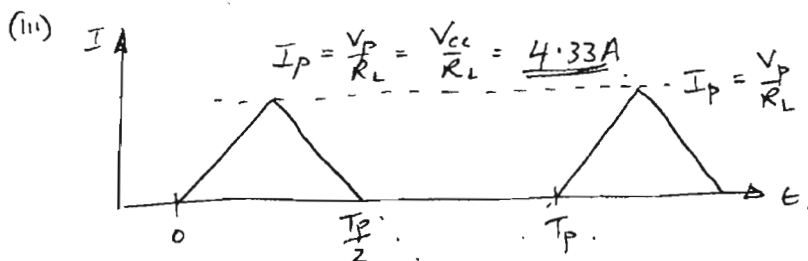
$$\therefore P_D = \frac{2V_{CC} \cdot \frac{3V_{CC}}{4}}{4R_L} - \frac{\left(\frac{3V_{CC}}{4}\right)^2}{3R_L} = \frac{6V_{CC}^2}{16R_L} - \frac{3V_{CC}^2}{16R_L}$$

$$= \underline{\underline{\frac{3V_{CC}^2}{16R_L}}} \quad \text{and } P_L = \frac{V_P^2}{3R_L} = \frac{\left(\frac{3V_{CC}}{4}\right)^2}{3R_L} = \underline{\underline{\frac{3V_{CC}^2}{16R_L}}}$$

(ii) max load power occurs when  $V_P$  is max...ie when  $V_P = V_{CC}$ 

$$\therefore 50W = \frac{V_{CC}^2}{3.8}$$

$$\text{or } V_{CC} = \sqrt{1200} = \underline{\underline{\pm 34.6V}}$$

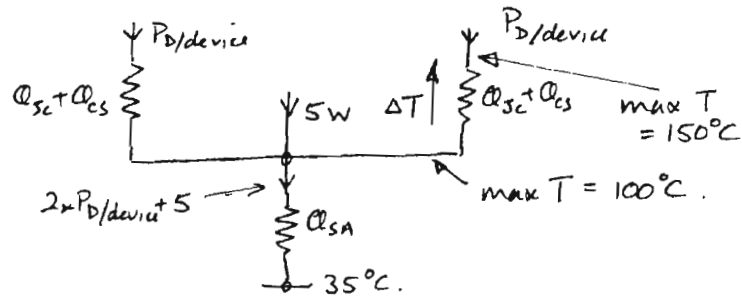


shape = 1  
times = 1  
 $I_P = 1$

⑧

Q4 cont...

(iv)



$$P_{DM} = \frac{3V_{CC}^2}{16R_L} = \frac{3 \times 1200}{16 \times 8} = 28W \text{ in total} \\ = 14W \text{ per device.}$$

$$\therefore \Delta T = 14 \times (\theta_{JC} + \theta_{CS}) = 14 \times 2.75 = 38.7^\circ$$

$\therefore$  with heatsink at 100,  $T_J$  will be 138.7  $\rightarrow$  safe.  
so heatsink temp is the limiting factor.

$$\therefore \frac{100 - 35}{\theta_{SA}} = 2 \times 14 + 5 = 33$$

$$\text{or } \theta_{SA} = 65/33 = 1.97^\circ\text{C/W}$$

(v) during rising triangular half cycles.

$$I = C \frac{dv}{dt} = 33 \times 10^{-6} \times \frac{60}{0.5\text{ms}} \\ = 3.96A.$$

Current waveform is:

