

# ROBOT CONTROL : INTRODUCTION

Control problem: determine joint inputs required for the end-effector to execute a desired action

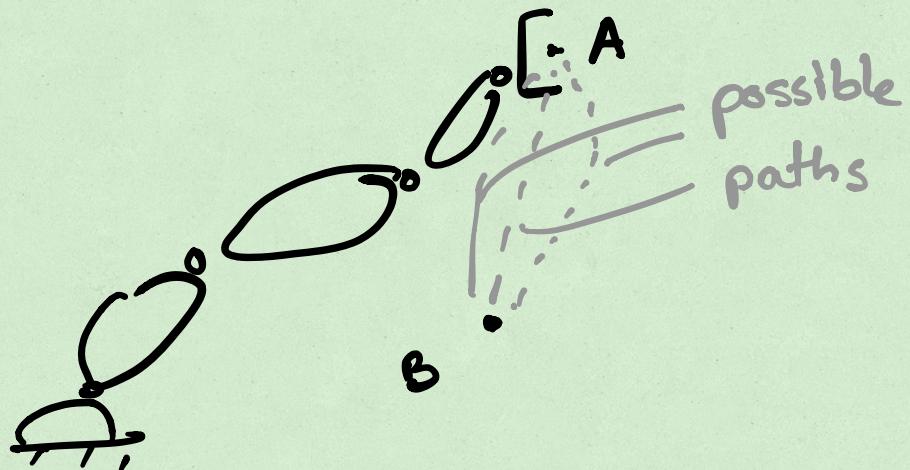
examples of Joint inputs

- motor torques
- voltage to the motors
- joint velocities (say if you use servomotors)

Problems encountered may vary with the mechanical design of the robot (a Cartesian robot vs an elbow robot or use of gears )

## Motion vs force control

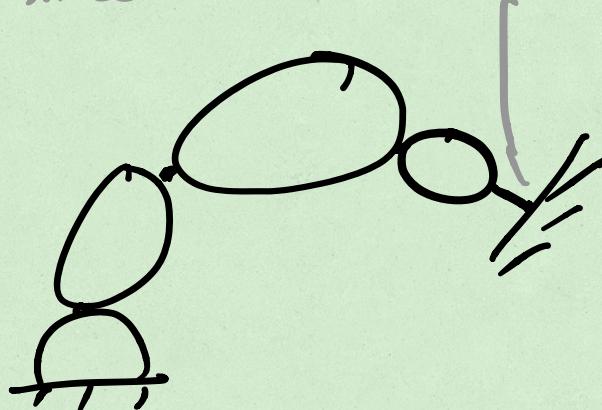
### Position / Motion control



- Point to point (PTP) control : move robot from position A to B. (position here refers to position and orientation)
- Trajectory tracking problem: Move robot along a specified path  $x(t)$

### Force control

end-effector  
in contact with a surface



- cannot move the end-effector 'inside' the surface !
- regulate contact force normal to the surface

## Control design and challenges

Control schemes are based on

feedforward : commands generated at planning stage

feedback : commands generated using  
robot state measures

Control law can be defined either in Joint  
or Cartesian (Task) space.

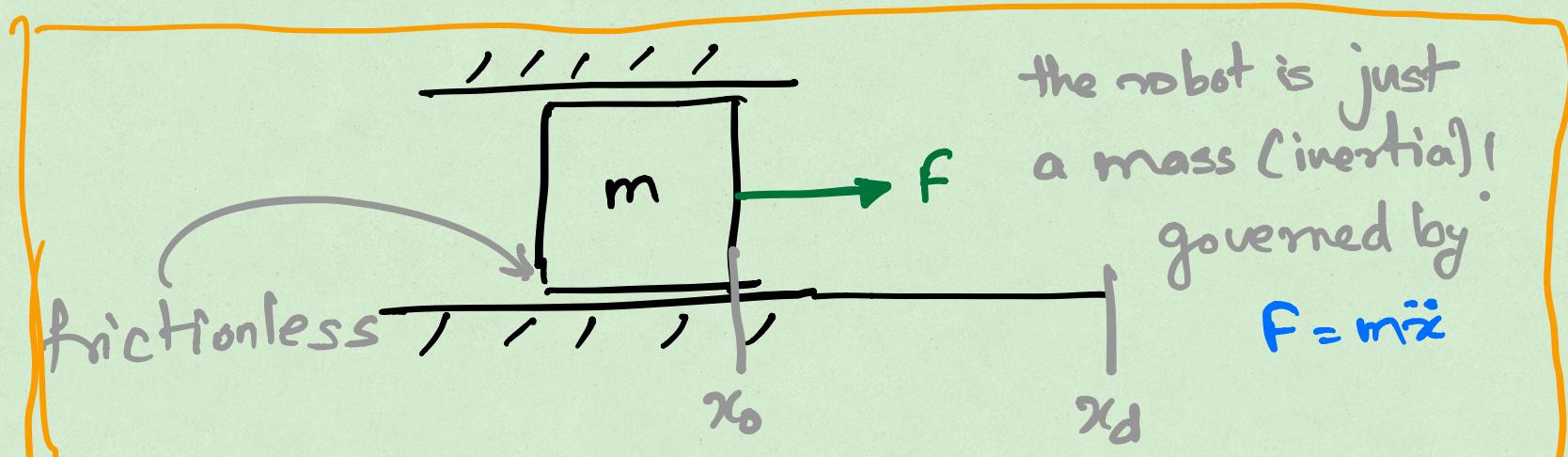
Control action always occurs at joint level  
(actuators driving the robot)

Challenges posed by : - unmodeled effects (friction  
flexibility of transmission)  
• external disturbances  
• errors (initial or due to dist.)  
• uncertain robot parameters etc.

## MOTION CONTROL : PRELIMINARIES

Basic concepts we will use for manipulators

Consider a single degree of freedom (DOF) robot with a prismatic joint.



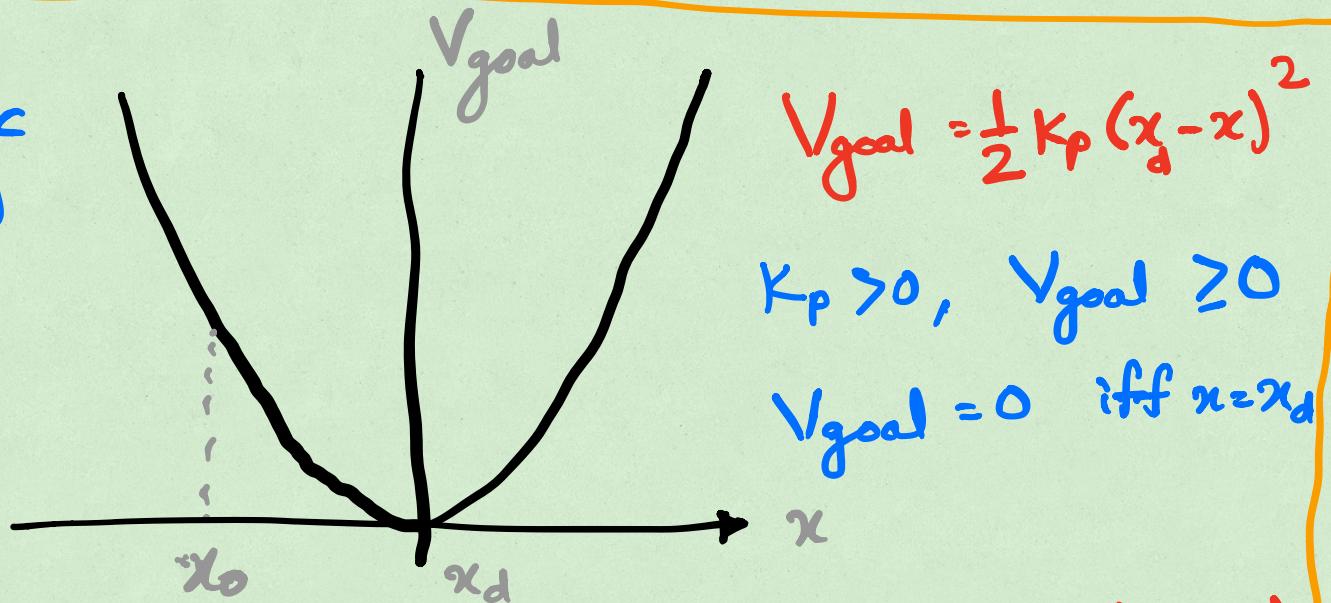
Goal : Find  $f(t)$  to move the robot from the initial position  $x_0$  to a desired (goal) position  $x_d$ .

## Potential Field

Recall: artificial potential fields were useful for path planning

Can we do something similar here?

Quadratic  
artificial  
potential  
field



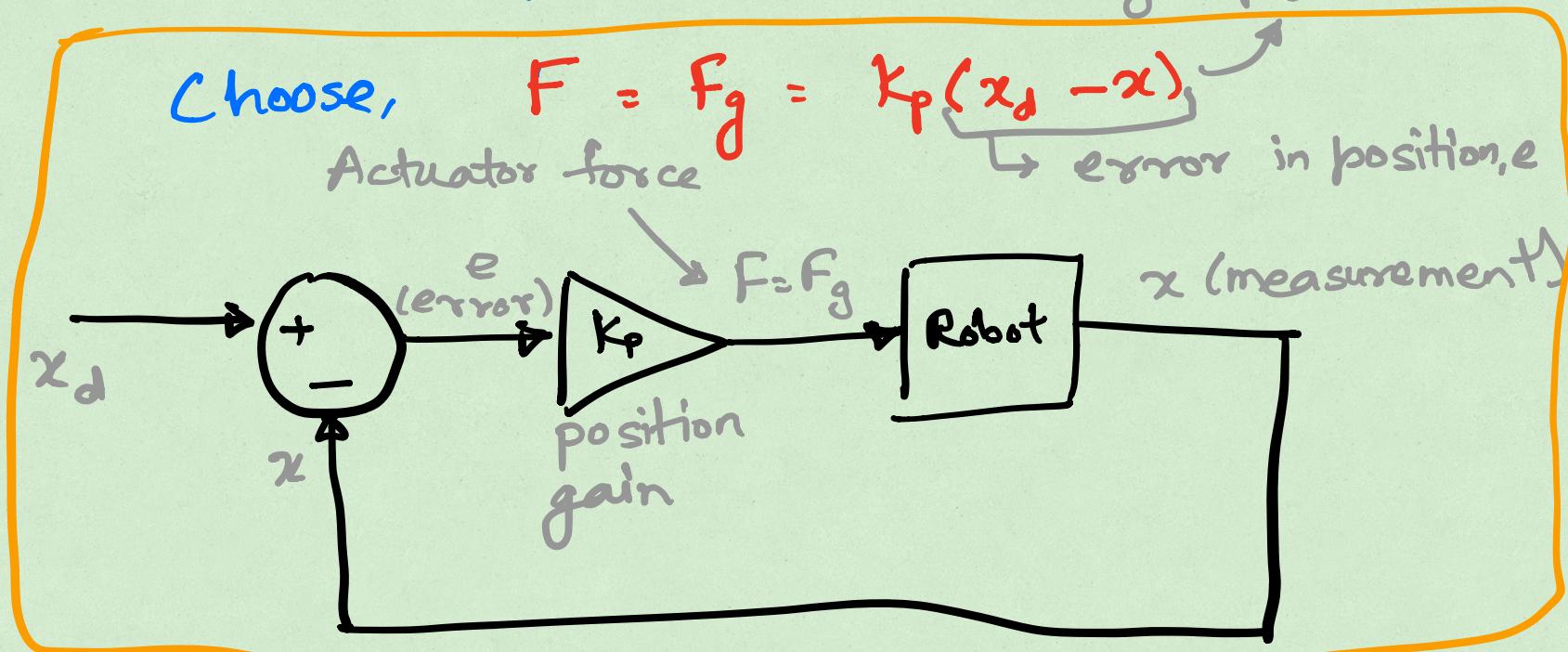
Corresponding force,  $F_g = -\frac{\partial V_{goal}}{\partial x} = K_p(x_d - x)$

## Initial control scheme (proportional (P) control)

System :  $m\ddot{x} = F$

Based on the potential field,

this is like having a virtual spring between current and goal position!



Actuator force is proportional to the error in robot position !!

## Stability

Kinetic energy of the robot,  $T = \frac{1}{2} m \dot{x}^2$

Using the Lagrangian,

$$\frac{d}{dt} \left( \frac{\partial T}{\partial \dot{x}} \right) - \frac{\partial T}{\partial x} = F = - \frac{\partial V_{goal}}{\partial x}$$

$$\text{or, } \frac{d}{dt} \left( \frac{\partial T}{\partial \dot{x}} \right) - \frac{\partial (T - V_{goal})}{\partial x} = 0$$

L  
no non-conservative forces

$V_{goal}$  is like potential energy associated with our closed loop system!!

Energy is conserved  $\Rightarrow$  system is stable

## What about performance?

$$m\ddot{x} = F = F_g = k_p(x_d - x)$$

if initial position,  $x_0 = x_d \rightarrow F=0 \Rightarrow \ddot{x}=0$

if  $x_0 \neq x_d$ , initial potential energy,

$$V = \frac{1}{2} k_p (x_d - x_0)^2$$

But, energy is conserved ! (potential energy is converted to Kinetic and viceversa)

robot doesn't stop at  $x_d$  ( $V=0$ ) !!!

Robot motion is oscillatory about  $x=x_d$ ,  
similar to that of a simple pendulum or  
mass-spring system.

## Velocity feedback (proportional-derivative (PD)) control

Consider a modified control law,

$$F = F_g + F_v = K_p(x_d - x) + F_v$$

where,  $F_v$  should be a force that helps bring the robot to rest at  $x = x_d$

then,

$$\frac{d}{dt} \left( \frac{\partial T}{\partial \dot{x}} \right) - \frac{\partial (T - V_{goal})}{\partial x} = F_v$$

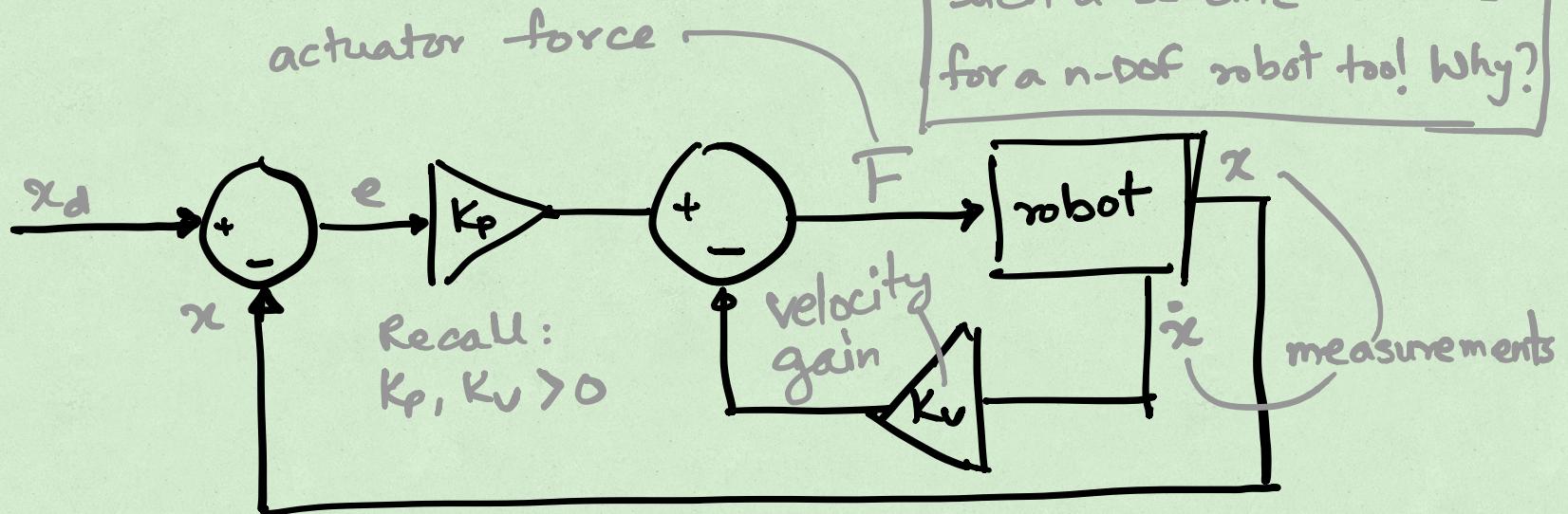
Recall: rate of change of energy,  $\dot{H} = F_v^T \dot{x}$

to bring the robot to rest at  $x_d$ ,  $\dot{H} < 0$ , if  $\dot{x} \neq 0$

a simple choice is  $F_v = -K_v \dot{x}$ ,  $K_v > 0$

overall,  $F = K_p(x_d - x) - K_v \dot{x}$  like a damper

## PD control (continued)



PD controller ensures that the robot reaches, and comes to rest at  $x = x_d$ , eventually.

How does choice of  $K_p, K_v$  affect performance (in terms of oscillations, time to settle at goal position, effect of disturbances etc.)?

## PD control (performance)

The PD controller imitates a mass-spring-damper system

$$m\ddot{y} + b\dot{y} + Ky = 0$$

damping stiffness

This is same as the PD controlled system with the end of the spring fixed at 0.

You should be able to verify that this can be obtained from the earlier one with a simple change of variables,  $y = x - x_d$

Recall that when  $b=0$  (no damping), motion is oscillatory with a constant amplitude

## Second-order system

$$m\ddot{x} + b\dot{x} + Kx = 0$$

$$\text{or, } \ddot{x} + \frac{b}{m}\dot{x} + \frac{K}{m}x =$$

Define,  $\omega_n^2 = \frac{K}{m}$  ;  $\xi = \frac{b}{2\omega_n m} = \frac{b}{2\sqrt{Km}}$

$\omega_n$ : natural frequency      damping ratio

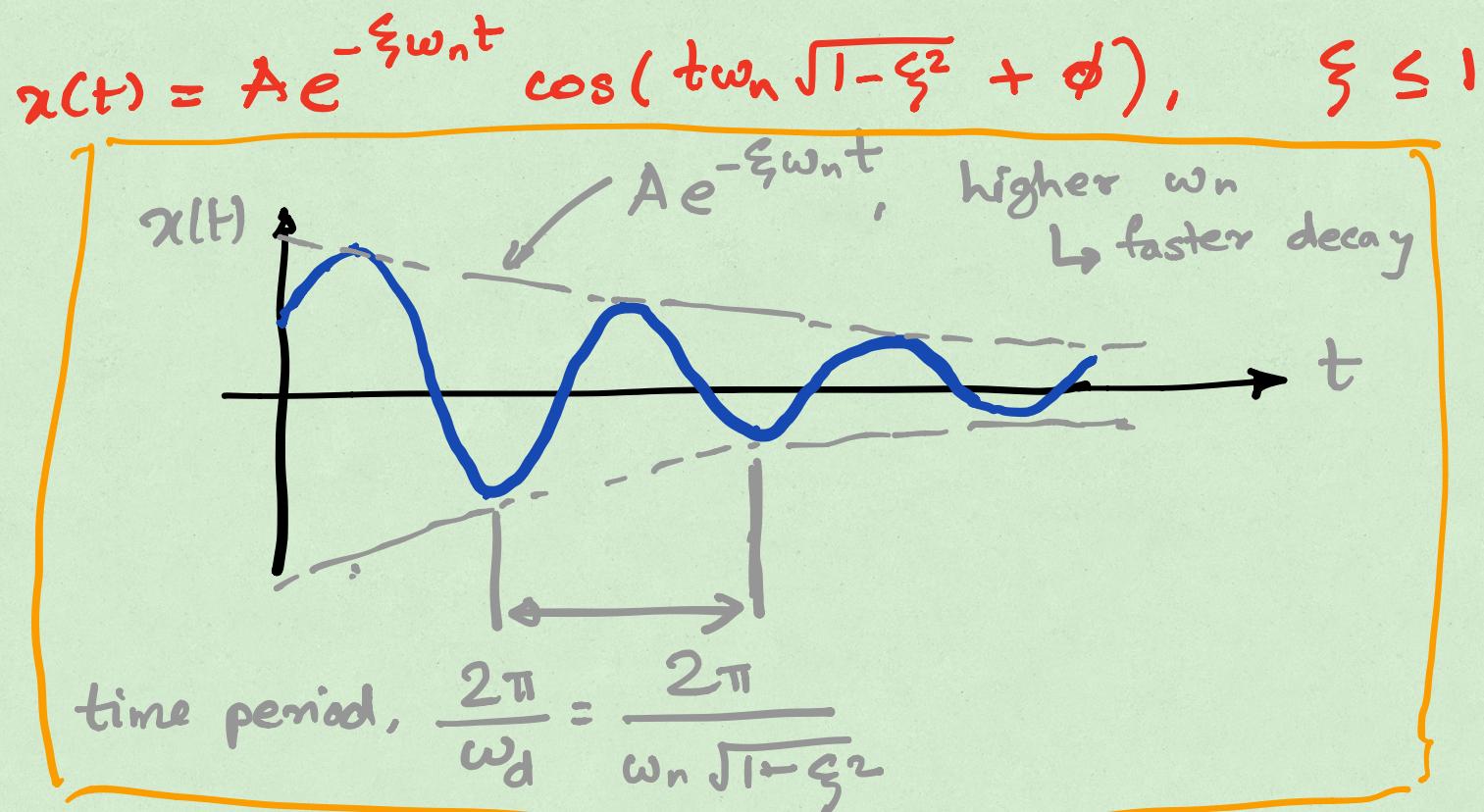
The solution to this system is given by,

$$x(t) = A e^{-\xi \omega_n t} \cos(\omega_d \sqrt{1-\xi^2} t + \phi), \quad \xi \leq 1$$

amplitude decays exponentially

$\omega_d$  : damped natural frequency

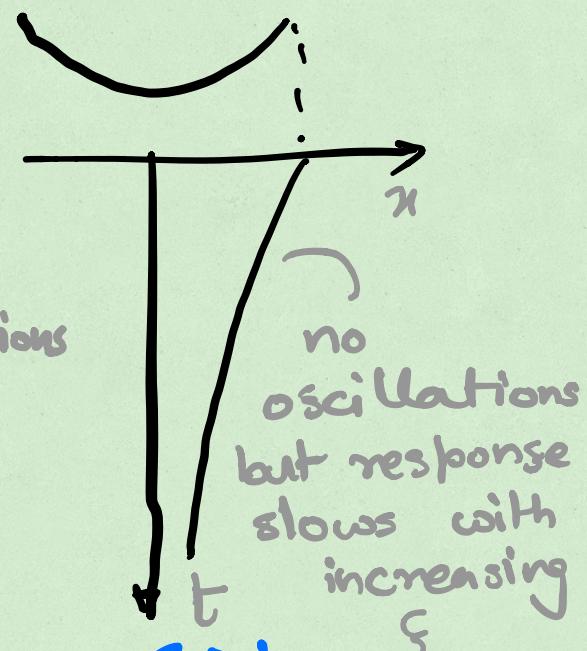
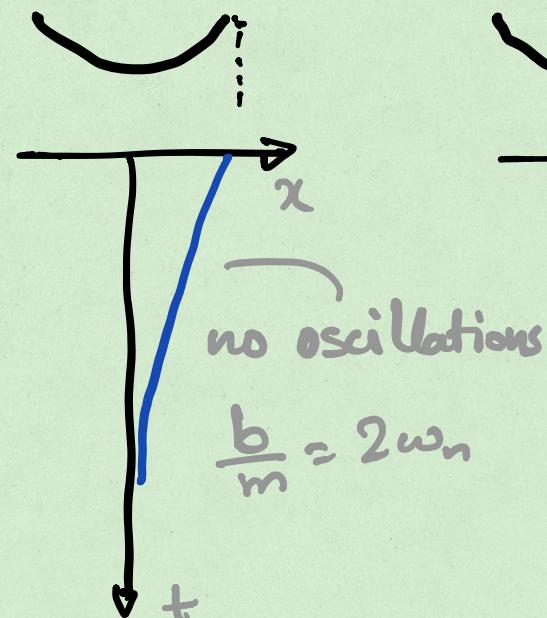
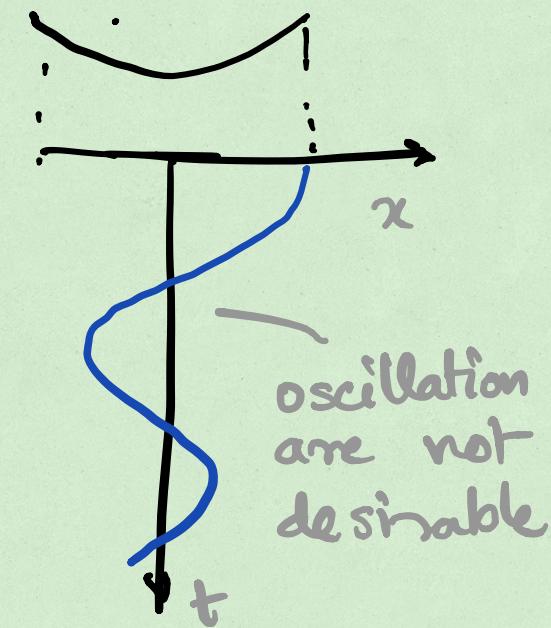
## Second-order system (time-response)



For  $\xi > 1$ , the response is exponential with no oscillations.

## Second order system (time response)

Increasing damping



We want to choose  $K_p$  &  $K_v$  to have a drfully damped system

## PD control (choice of $K_p$ , $K_v$ )

$$m\ddot{x} = F = K_p(x_d - x) - K_v \dot{x}$$

$$\text{or, } m\ddot{x} + K_v \dot{x} + K_p(x - x_d) = 0$$

↑  
velocity gain      ↗  
position gain

$$\text{or, } \ddot{x} + \frac{K_v}{m} \dot{x} + \frac{K_p}{m} (x - x_d) = 0$$

comparing  $\ddot{x} + 2\xi\omega \dot{x} + \omega^2 (x - x_d) = 0$       ↴  
second order  
mass spring  
damper!

closed loop  
damping ratio,  $\xi = \frac{K_v}{2\sqrt{K_p m}}$

closed-loop natural  
frequency,  $\omega = \sqrt{\frac{K_p}{m}}$

## Selection of $K_p$ and $K_v$

- Typically, we want the robot behavior to be critically damped (reaches goal as fast as it can with no oscillations)



$$\xi = \frac{K_v}{2\sqrt{K_p m}} = 1$$

- Choice of  $\omega$  determines  $K_p$ !

- we want  $\omega$  as high as possible (faster response)

- high  $K_p$  is also good to limit deviations due to disturbances!

## Selection of $K_p$ & $K_v$ (continued . . .)

Hence,

$$K_p = m\omega^2$$

$$K_v = m(2\xi\omega)$$

$\omega$  ; typically limited by unmodeled dynamics — from flexibility in the robot structure (due to couplings, transmissions etc.), communication delays in the control loop etc.

In practice, start with low  $\omega$  and increase till the response is free of vibrations.

## Control partitioning

Both  $K_p$  and  $K_v$  are scaled by mass  $m$ !

### Unit mass system

$$K_p' = \omega^2$$

$$K_v' = 2\xi\omega$$

### $m$ - mass system

$$K_p = m K_p'$$

$$K_v = m K_v'$$

1. Design using the unit-mass system.
2. Scale by mass  $m$  to keep the same  $\omega$

This works even if  $m$  changes with  $q$ ,  
as in case of our robot manipulators!

## Control Partitioning

$$m\ddot{x} = F \rightarrow m(1.\ddot{x}) = mf'$$

$$f' = -K_V'\dot{x} + K_P'(x_d - x)$$

$$\begin{aligned} F &= m(-K_V'\dot{x} + K_P'(x_d - x)) \quad (K_P = mK_P') \\ &= mf' \quad K_V = mK_V' \end{aligned}$$



$$m\ddot{x} = F = mf'$$

or,  $\ddot{x} = f'$

or,  $\ddot{x} + K_V'\dot{x} + K_P'(x - x_d) = 0$

same dynamic behavior as the unit mass system.

Other characteristics may be different.

## Handling Non-Linearities

$$m\ddot{x} + \underbrace{b(x, \dot{x})}_{\text{non-linear terms}} = F$$

try to model  
and compensate

(could be due to friction, gravity etc.)

## Control partitioning:

$$F = \alpha f' + \beta$$

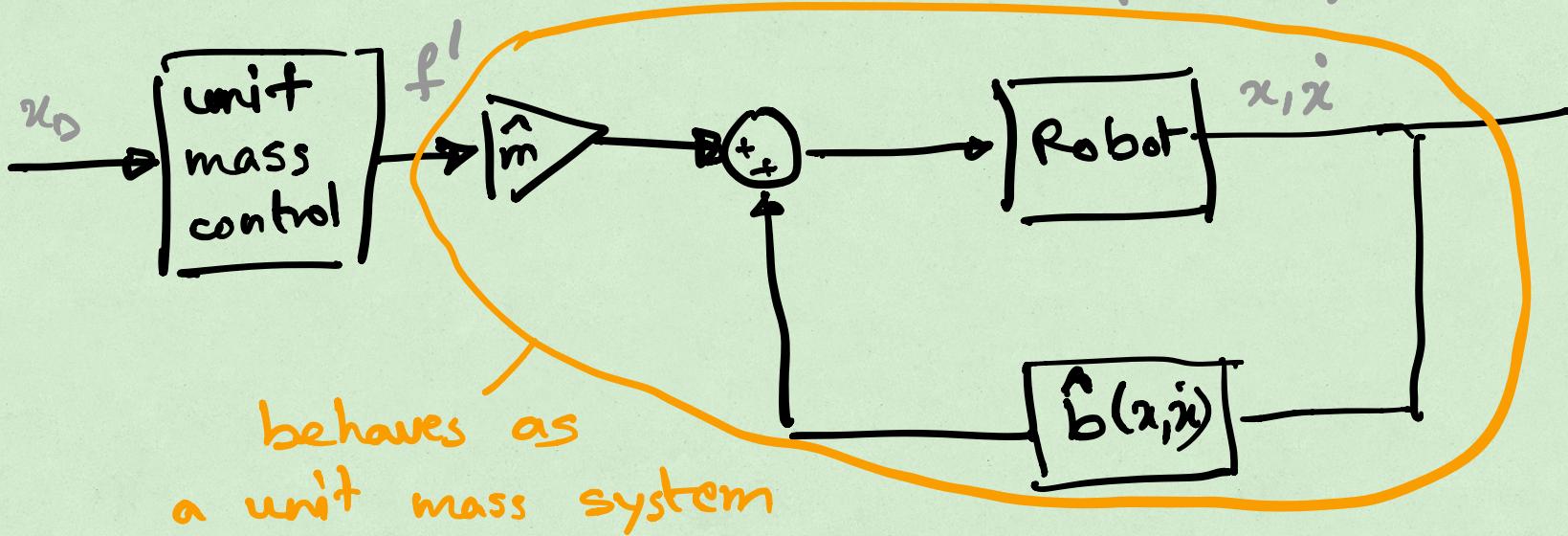
where,  $\alpha = \hat{m}$ ,  $\beta = \hat{b}(x, \dot{x})$   
 $\hat{m}$  our estimate of m       $\hat{b}$  our estimate of b

## Handling non-linearities (continued . . .)

$$m\ddot{x} + b(x, \dot{x}) = \hat{m}f' + \hat{b}(x, \dot{x})$$

With perfect estimates and measurements, this behaves as a unit mass system!

$\ddot{x} = f'$  (in practice, this will be close but not perfect)



## Motion control : Trajectory tracking

How can we modify the controller to follow a path?

i.e. we want to design  $f'$  (we have already converted the system to behave as a unit mass system) to follow a path defined by  $\ddot{x}_d(t)$ ,  $\dot{x}_d(t)$  and  $x_d(t)$

move the free end of the spring and damper (of the earlier PD controller) along the path!

feed forward  $f' = \underbrace{\ddot{x}_d}_{\text{to track the acceleration}} - K_v(\dot{x} - \dot{x}_d) - K_p(x - x_d)$

## Motion control (continued)

Closed loop system:

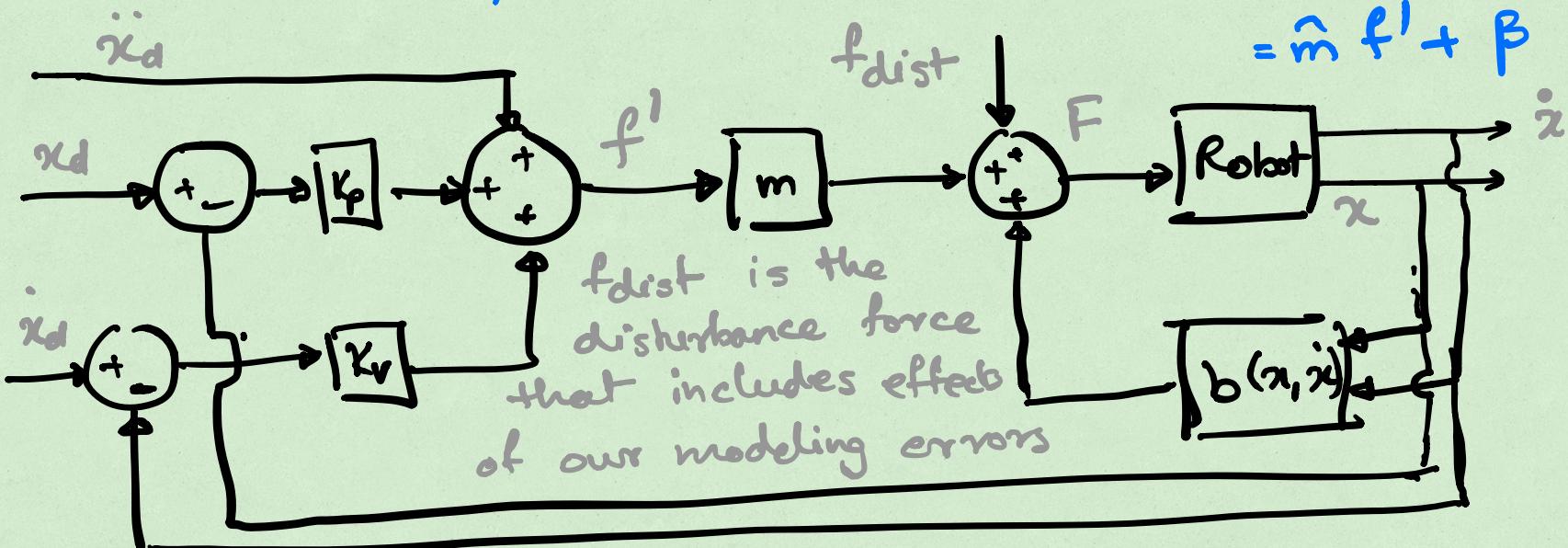
$$(\ddot{x} - \ddot{x}_d) + K_V (\dot{x} - \dot{x}_d) + K_P (x - x_d) = 0$$

linear  
second or,  
order system

$$\ddot{e} + K_V \dot{e} + K_P e = 0$$

$e = x - x_d$   
is the error

For the full system,  $m\ddot{x} + b(x, \dot{x}) = F$



## Steady-state error

$$m\ddot{x} + b(x, \dot{x}) = F + f_{dist}$$

control:  $F = mf' + b(x, \dot{x})$  any modeling errors are in  $f_{dist}$

Assume:  $f_{dist}$  is bounded  $|f_{dist}| < a$   
at all times

$$\text{Closed-loop: } \ddot{e} + K'_p \dot{e} + K_p e = \frac{f_{dist}}{m}$$

Steady state error: error when  $\ddot{e} = \dot{e} = 0$

(how far we are from rest position)

if  $\ddot{e} = \dot{e} = 0$ ,  $e = \frac{f_{dist}}{m K'_p}$  recall  $K_p'$  determines  $\omega$

if  $m$  varies with position  
so does steady state error!

Large  $K_p \Rightarrow$  small errors

## PID control

Add an integral action to address steady-state error.

System:  $m\ddot{x} + b(x, \dot{x}) = F + f_{dist}$

Control:  $f' = \ddot{x}_d - K_v(\dot{x} - \dot{x}_d) - K_p(x - x_d) - K_i \int (x - x_d) dt$

|  
builds up over  
time and compensates

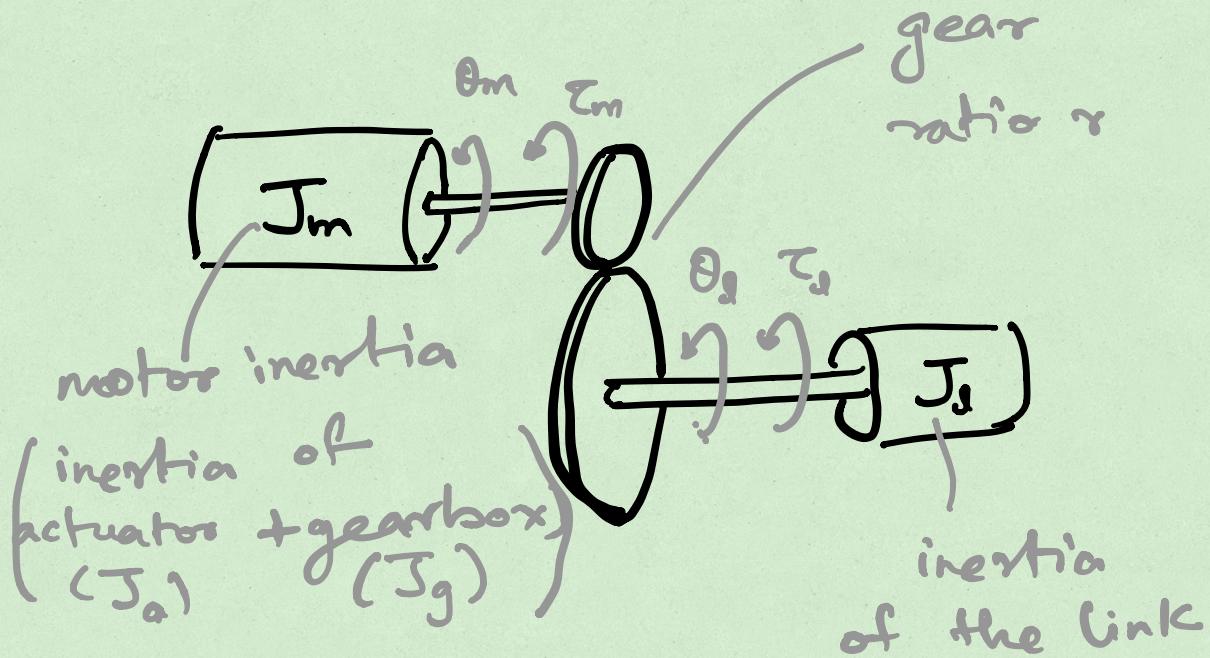
Closed-loop system:

$$\ddot{e} + K_v \dot{e} + K_p e + K_i \int e dt = \frac{f_{dist}}{m}$$

or  $\ddot{e} + K_v \dot{e} + K_p e + K_i e = 0 \rightarrow e = 0$

In practice, integral action is usually used only when close to goal.

## Gear reduction



Why use gears?

- small  $\gamma \rightarrow$  need large motors to get required torques  
↳ larger motors, increases weight!

$$\theta_m = \gamma \theta_g$$

$$\dot{\theta}_m = \gamma \ddot{\theta}_g$$

$$\tau_s = \gamma \tau_m$$

$\gamma \sim 25 - 300$   
for most industrial robots

- Most motors work efficiently at higher speeds than needed for robot

## Gear reduction (continued..)

How do gears change the dynamics?

$$J_m \ddot{\theta}_m + b_m \dot{\theta}_m = \tau_m - \frac{\tau_g}{\gamma}$$
$$= \tau_m - \frac{1}{\gamma} (J_g \ddot{\theta}_g + b_g \dot{\theta}_g)$$

or,

$$\left( J_m + \frac{J_g}{\gamma^2} \right) \ddot{\theta}_m + \left( b_m + \frac{b_g}{\gamma^2} \right) \dot{\theta}_m = \tau_m$$

effective inertia

effective damping

or, on  
link side

$$\left( J_d + \gamma^2 J_m \right) \ddot{\theta}_g + \left( b_d + \gamma^2 b_m \right) \dot{\theta}_g = \tau_g$$

these can be very large even for small  $J_m, b_m$ !

Problems with gears: increased inertia, friction  
and backlash

## Actuator dynamics

- To control using voltage commands to the actuator
- For a permanent magnet DC (PMDC) motor, (other actuators will have different dynamic behavior)

relationship between current, voltage and motor torque is governed by

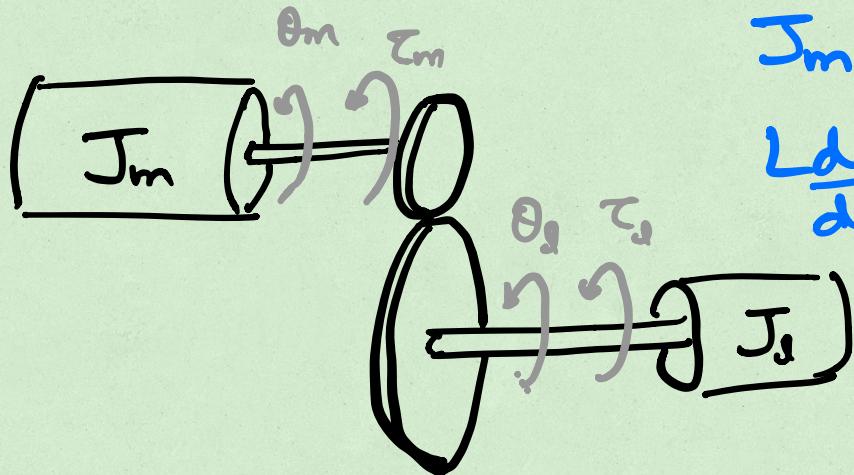
$$L \frac{di}{dt} + Ri = V - V_b$$

inductance of armature       $\xrightarrow{\text{resistance}}$  resistance  
of armature       $\xrightarrow{\text{applied voltage}}$  applied voltage

$$V_b = K_b \dot{\theta}_m$$
$$\tau_m = K_m i$$

motor constants

## Actuator dynamics (combined model)



$$J_m \ddot{\theta}_m + B_m \dot{\theta}_m = K_m i - \frac{T_d}{\tau}$$

$$L \frac{di}{dt} + R i = V - K_b \dot{\theta}_m$$

governed by  
these two coupled  
equations

However, motors respond much faster than the robot.  
This can be used to simplify and model the dynamics  
using a single equation (ignore  $di/dt$ )

$$J_m \ddot{\theta}_m + \left( B_m + \frac{K_m K_b}{R} \right) \dot{\theta}_m = \frac{K_m V}{R} - \frac{T_d}{R}$$