

# TTK4225 - Systems Theory, Autumn 2020

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A complete example, from the beginning to the end

# Roadmap

- Lotka-Volterra, but this time from the beginning to the end

# Lotka-Volterra

- $y_{\text{prey}} := \text{prey}$
- $y_{\text{pred}} := \text{predator}$

$$\begin{cases} \dot{y}_{\text{prey}} &= \alpha y_{\text{prey}} - \beta y_{\text{prey}} y_{\text{pred}} \\ \dot{y}_{\text{pred}} &= -\gamma y_{\text{pred}} + \delta y_{\text{prey}} y_{\text{pred}} \end{cases}$$

`GitHub/TTK4225/trunk/Jupyter/Lotka-Volterra-introduction.ipynb`

## “Our” Lotka-Volterra

$$\begin{cases} \dot{y}_{\text{prey}} &= 10y_{\text{prey}} - 1y_{\text{prey}}y_{\text{pred}} - u_{\text{prey}} \\ \dot{y}_{\text{pred}} &= -0.1y_{\text{pred}} + y_{\text{prey}}y_{\text{pred}} - u_{\text{pred}} \end{cases}$$

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- 5 understand the controllability and observability properties of the system (*not in this course*)
- 6 design minimal human interventions that make the system behave as desired (*not in this course*)

?

## Complex numbers - introduction

# Roadmap

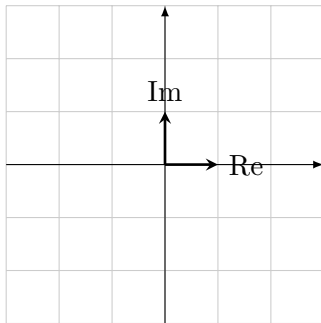
- definition
- sum, subtraction, multiplication, division
- why is this important?

What is a complex number, and why did we introduce them?

# What is a complex number, and why did we introduce them?

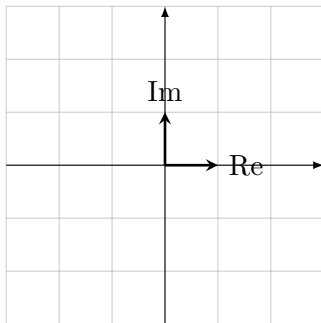
In essence:

- 1 a point in the Cartesian plane
- 2 to be sure to find all the roots of polynomials (*i.e., be able to write polynomials in convenient forms*)



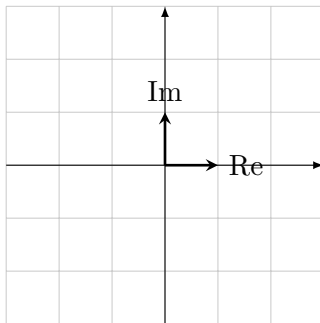


## The “imaginary unit”



$$i \quad : \quad i^2 = -1$$

## Simple operations with complex numbers: sums

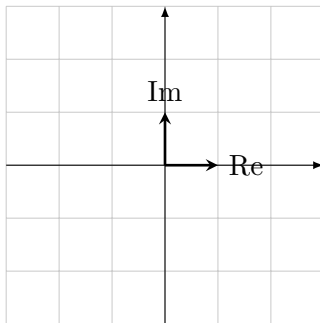


$$z_1 = a_1 + ib_1 \quad z_2 = a_2 + ib_2$$

implies

$$z_1 + z_2 = (a_1 + ib_1) + (a_2 + ib_2) = (a_1 + a_2) + i(b_1 + b_2)$$

## Simple operations with complex numbers: subtractions

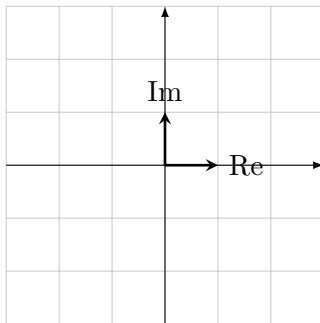


$$z_1 = a_1 + ib_1 \quad z_2 = a_2 + ib_2$$

implies

$$z_1 - z_2 = (a_1 + ib_1) - (a_2 + ib_2) = (a_1 - a_2) + i(b_1 - b_2)$$

## Simple operations with complex numbers: multiplication

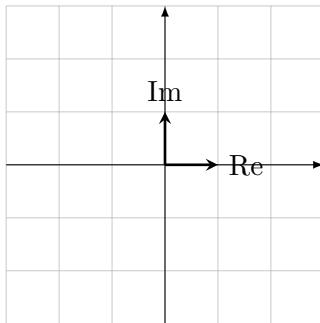


$$z_1 = a_1 + ib_1 \quad z_2 = a_2 + ib_2$$

implies

$$z_1 z_2 = (a_1 + ib_1)(a_2 + ib_2) = (a_1 a_2 - b_1 b_2) + i(a_1 b_2 + a_2 b_1)$$

## Simple operations with complex numbers: inversion

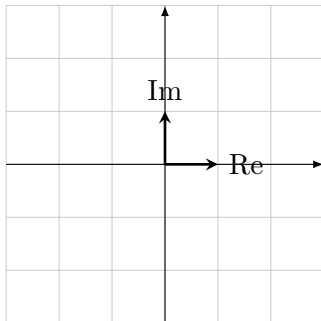


$$z_1 = a_1 + ib_1$$

implies

$$z_1^{-1} = \frac{a_1}{a_1^2 + b_1^2} - i \frac{b_1}{a_1^2 + b_1^2}$$

## Simple operations with complex numbers: division

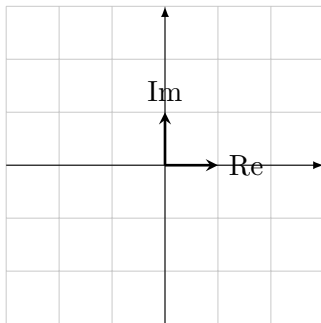


$$z_1 = a_1 + ib_1 \quad z_2 = a_2 + ib_2$$

implies

$$\frac{z_1}{z_2} = \frac{a_1 + ib_1}{a_2 + ib_2} = \frac{(a_1 + ib_1)(a_2 - ib_2)}{(a_2 + ib_2)(a_2 - ib_2)} = \frac{a_1a_2 + b_1b_2}{a_2^2 + b_2^2} + i \frac{b_1a_2 - a_1b_2}{a_2^2 + b_2^2}$$

## Simple operations with complex numbers: conjugation



$$z_1 = a_1 + ib_1$$

implies

$$\overline{z_1} = a_1 - ib_1$$

Conjugacy: a good way of simplifying the previous operations



## Conjugacy: a good way of simplifying the previous operations

- addition:  $z + \bar{z} = a + ib + a - ib = 2a$ , thus  $\operatorname{Re}(z) = \frac{1}{2}(z + \bar{z})$

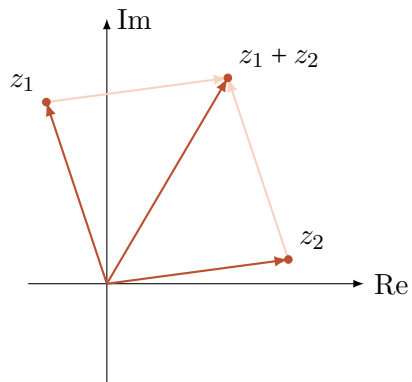
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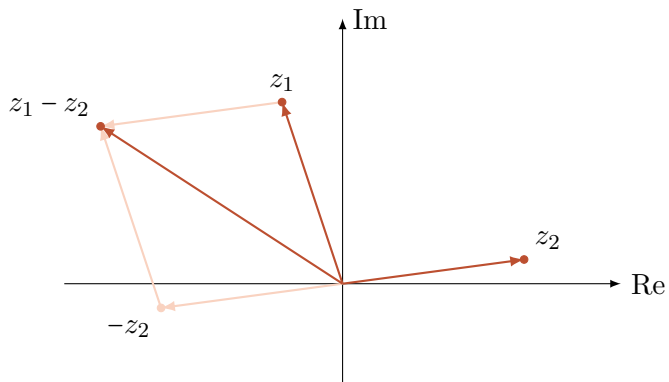
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- subtraction:  $z - \bar{z} = a + ib - a + ib = 2ib$ , thus  $\operatorname{Im}(z) = \frac{1}{2i}(z - \bar{z})$
- multiplication:  $z\bar{z} = (a + ib)(a - ib) = a^2 + b^2$ , thus  $|z|^2 = z\bar{z}$

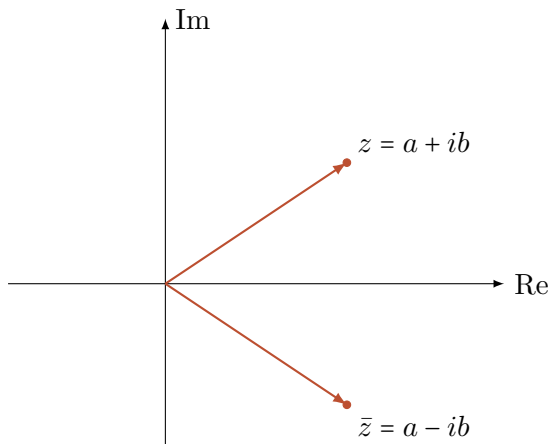
Once again, graphically: addition



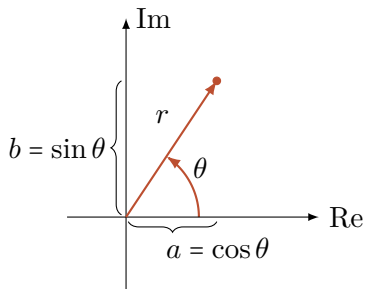
Once again, graphically: subtraction



## Once again, graphically: conjugation



## Polar coordinates



$$z = a + ib$$

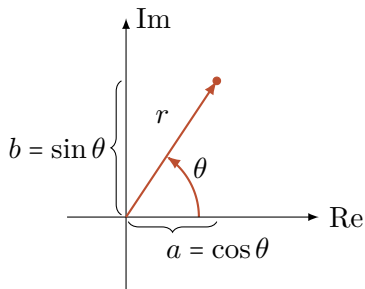
can be rewritten through  $r$  and  $\theta$  so that

$$a = r \cos \theta \text{ and } b = r \sin \theta$$

so that

$$z = r (\cos \theta + i \sin \theta)$$

# Polar coordinates



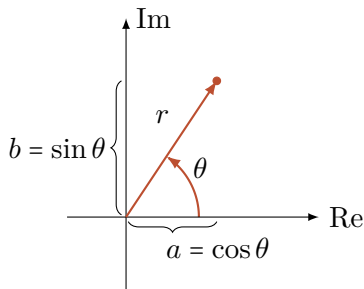
Equations:

$$r = |z| = \sqrt{a^2 + b^2} = \sqrt{z\bar{z}}$$

$$\theta = \arg z = \text{atan}(b, a) = \tan^{-1}\left(\frac{b}{a}\right)$$



# Polar coordinates



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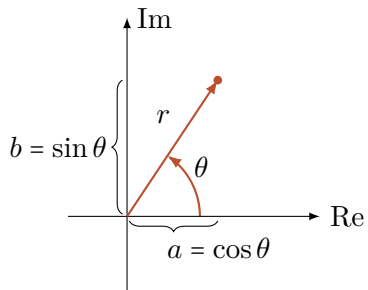
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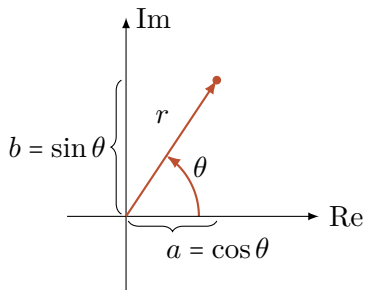
Notation:

- $r$  = absolute value or modulus of  $z$
- $\theta$  = argument, angle, or phase of  $z$

Problem: different  $\theta$ 's lead to the same  $z$

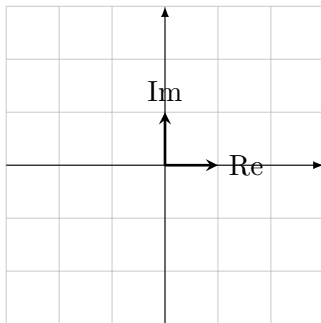


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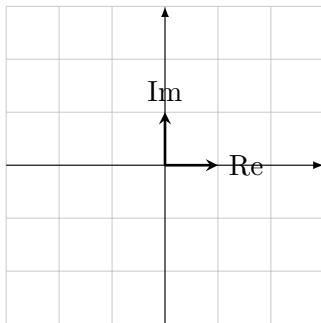
*Definition:* principal value of  $z =$  that value of  $\theta$  that is in  $[-\pi, \pi]$

## Usefulness of polar forms: the multiplication is immediate



$$z_1 z_2 = r_1 r_2 [\cos (\theta_1 + \theta_2) + i \sin (\theta_1 + \theta_2)]$$

Implication: the division is immediate



$$\frac{z_1}{z_2} = \frac{r_1}{r_2} [\cos (\theta_1 - \theta_2) + i \sin (\theta_1 - \theta_2)]$$

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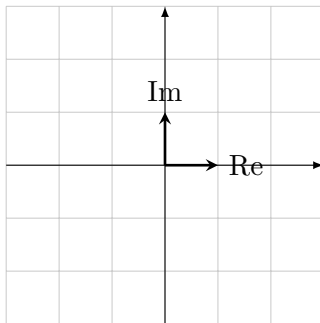
E.g., thus

$$e^z = 1 + z + \frac{1}{2!}z^2 + \frac{1}{3!}z^3 + \dots$$



?

## The absolute value of a complex number



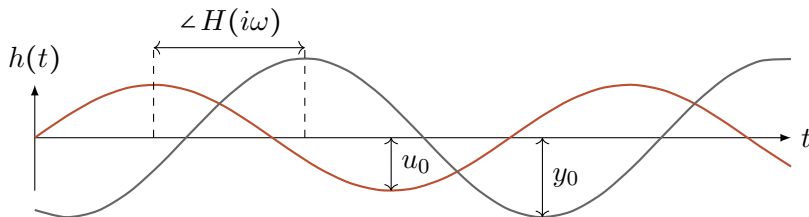
Meaning: Euclidean length of the vector. Very important for control, since very often we compute the absolute value of a transfer function at a specific  $s = i\omega$  (*and very very often the transfer function is rational*)

Thus, chain of implications highlighting the importance of complex numbers for control perspectives

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1: LTI systems have sinusoidal fidelity

$$u(t) = u_0 \sin(\omega t) \implies y(t) = u_0 |H(i\omega)| \sin(\omega t + \angle H(i\omega))$$



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3: we very often consider rational  $H$ 's

this means that to compute  $|H(i\omega)|$  we need to do multiplications and divisions among the complex numbers  $s - \star_i$

Example:  $H(s) = \frac{1 + 2s}{1 + 2s + s^2}$ . What is  $|H(i\omega)|$ ?

$$\begin{aligned}|H(i\omega)| &= \sqrt{H(i\omega)\overline{H(i\omega)}} \\&= \sqrt{\frac{1 + 2i\omega}{1 + 2i\omega - \omega^2} \cdot \frac{1 - 2i\omega}{1 - 2i\omega - \omega^2}} \\&= \sqrt{\frac{1 + 4\omega^2}{(1 - \omega^2)^2 + 4\omega^2}} \\&= \sqrt{\frac{1 + 4\omega^2}{\omega^4 + 2\omega^2 + 1}}\end{aligned}$$

?



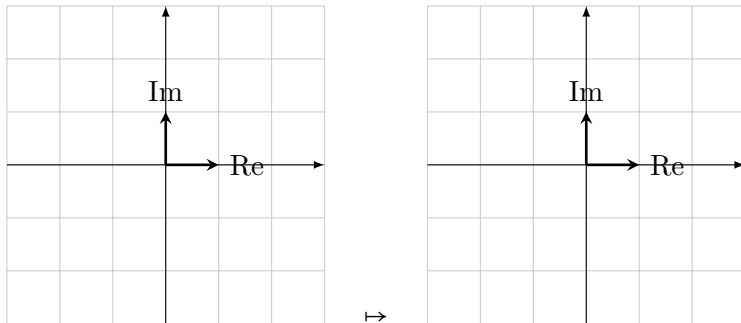
# Complex functions

# Roadmap

- definition
- why are they important?

## Complex function: definition

$$f : \mathbb{C} \mapsto \mathbb{C}$$



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thus, also in polar representations,  $(r, \theta) \mapsto (r', \theta')$  with in general both  $r'$  and  $\theta'$  functions of both  $r$  and  $\theta$

Example: if  $f(z) = z^2 + 3z$  then what is  $f(1 + 3j)$ ?

$$\begin{aligned} f(z) &= (x + iy)(x + iy) + 3x + 3iy \\ &= x^2 + 2ixy - y^2 + 3x + 3iy \\ &= x^2 - y^2 + 3x + i(2xy + 3y) \end{aligned}$$

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thus

$$\begin{aligned}f(1 + 3j) &= u(1, 3) + iv(1, 3) \\&= 1^3 - 3^2 + 3 + i(2 \cdot 1 \cdot 3 + 3 \cdot 3) \\&= -5 + 15i\end{aligned}$$



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Recall: the forced evolution is given by

$$Y(s) = H(s)U(s)$$

with  $H(s)$  very often rational, i.e., ratio of polynomials.

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Recall: the forced evolution is given by

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with  $H(s)$  very often rational, i.e., ratio of polynomials. *Essential tool for automatic control people: roots of complex polynomials*

# Roots of complex functions

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## Primary definition: root of a complex number

if  $z \in \mathbb{C}$  and  $n \in \mathbb{N}$ , then the  $n$  complex roots of  $z$  are the  $n$  complex numbers  $z_0, \dots, z_{n-1}$  for which  $z_k^n = z$ , i.e.,

$$\sqrt[n]{z} = \sqrt[n]{r} \left( \cos \frac{\theta + 2k\pi}{n} + i \sin \frac{\theta + 2k\pi}{n} \right) \quad \text{for } k = 0, 1, \dots, n-1$$

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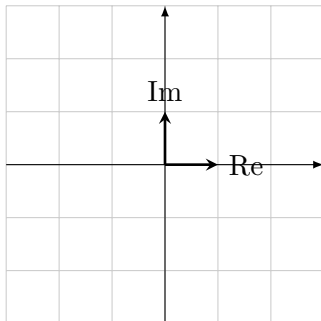
$$\sqrt[n]{z} = \sqrt[n]{r} \left( \cos \frac{\theta + 2k\pi}{n} + i \sin \frac{\theta + 2k\pi}{n} \right) \quad \text{for } k = 0, 1, \dots, n-1$$

The intuition on how to get them follows from:

$$z_1 z_2 = r_1 r_2 [\cos(\theta_1 + \theta_2) + i \sin(\theta_1 + \theta_2)]$$

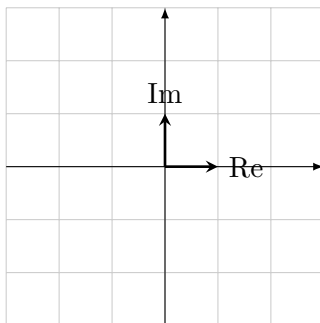
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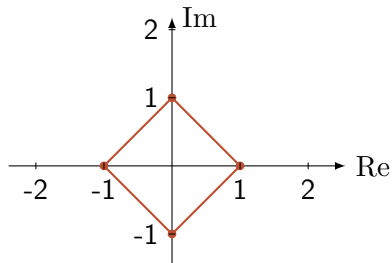
these  $n$  roots always exist



## Roots of complex functions, example: quartic roots of 1

$$\sqrt[4]{1} = \{1, i, -1, -i\}$$

*(note that only two of them are in  $\mathbb{R}$ )*



## IMPORTANT: ONE SHOULD CONSIDER THE PRINCIPAL VALUE

... otherwise one may artificially add  $2\pi k$  to the phase of  $w = \sqrt[n]{z}$  and have an infinite number of roots ...

Why are we using so much time on this?

## Why are we using so much time on this?

Because we often have to do with objects of the type  $z^n + a_{n-1}z^{n-1} + \dots + a_1z + a_0 = 0$ , thus we need to know what we are dealing with!

## Why are we using so much time on this?

Because we often have to do with objects of the type  $z^n + a_{n-1}z^{n-1} + \dots + a_1z + a_0 = 0$ , thus we need to know what we are dealing with! *Essential results:*

- $n$ -order polynomials have always from 0 to  $n$  real roots
- $n$ -order polynomials have always  $n$  complex roots

## Example

$$z^4 - 6iz^2 + 16 = 0$$

implies

$$z_1 = 2 + 2i \quad z_2 = -2 - 2i \quad z_3 = -1 + i \quad z_4 = 1 - i$$

*(to get the solution let  $y = z^2$ , and then do a bit of massaging)*

?

# Complex exponentials



# Roadmap

- intuitions
- definition
- Euler's identities
- complex logarithms

## In the previous episodes ...

- complex sums and multiplications
- complex roots
- complex polynomials

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- complex sums and multiplications
- complex roots
- complex polynomials

→ generalizing everything, even the functions

## Discussion

why are exponentials important in control?

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Because they are the essence of the modes of LTI systems with rational transfer functions, and LTI systems are often good approximations of nonlinear systems around their equilibria

## First usefulness of complex exponentials: simplify notation even further

Question: can we write  $z = r (\cos \theta + i \sin \theta)$  in a more complex way?

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with sin and cos?

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with sin and cos? Of course, Euler's formula!



# Why does Euler's formula work? (so that one may remember it more...)

Starting point:

$$e^z = e^{x+iy} = e^x e^{iy}$$

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$$e^{iy} = 1 + iy + \frac{1}{2!}(iy)^2 + \frac{1}{3!}(iy)^3 + \dots + \frac{1}{k!}(iy)^k + \dots$$

## Why does Euler's formula work? (so that one may remember it more...)

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## Why does Euler's formula work? (so that one may remember it more...)

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thus

$$e^z = e^x (\cos y + i \sin y)$$

## The new representation given by Euler's formula and polar representations

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$$r = \sqrt{x^2 + y^2} \quad \theta = \operatorname{atan}\frac{y}{x}$$

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that, by the way, implies  $(x + iy)i = -y + ix$ , i.e., a 90-degrees rotation

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thus

$$\sin y = \frac{1}{2i} (e^{iy} - e^{-iy}) \quad \cos y = \frac{1}{2} (e^{iy} + e^{-iy})$$

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- exponentials are never equal to 0, i.e.,  $e^z \neq 0$  independently of  $z$
- exponentials are periodic, i.e.,  $e^{z+2\pi i} = e^z$

Notation: “fundamental region of the exponential”

$$-\pi < \operatorname{Im}(z) \leq \pi$$

## Multiplications and divisions through the complex functions

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- $\frac{z_1}{z_2} = \frac{r_1}{r_2} e^{i(\theta_1 - \theta_2)}$

## Roots through the complex functions

$w = z^n$  is s.t.  $w = re^{i\theta+2\pi k}$  and is equal to

$$z_k = r^{1/n} e^{i\left(\frac{\theta}{n} + \frac{2k\pi}{n}\right)}$$

*(note that besides  $k = 0, 1, \dots, n-1$ , for other  $k$ 's we get the same roots as before)*

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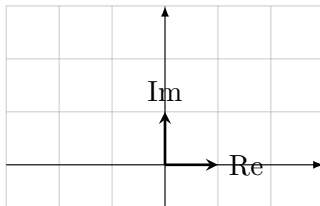
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Graphically:



## Very important difference

as soon as  $z = re^{i\theta}$  is s.t.  $r > 0$  then

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exists, thus the complex logarithm is defined for every  $z \neq 0$ !

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### Example

$$\ln(-10) = 2.30259 + i\pi.$$

?

## LTI filters - motivations

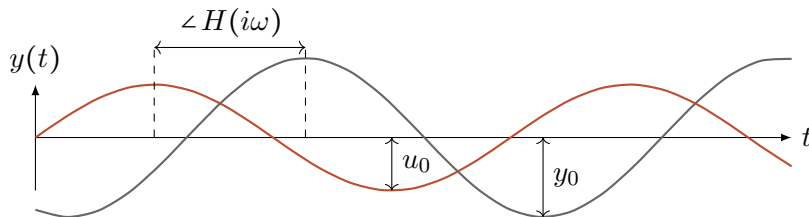
# Roadmap

- how to use Fourier to analyse LTI systems
- Bode plots

## Recall: why are Fourier transforms important for control people?

Among others:  $H(i\omega)$  says how to apply the sinusoidal fidelity property, i.e., the fact that

$$u(t) = u_0 \sin(\omega t) \implies y(t) = u_0 |H(i\omega)| \sin(\omega t + \angle H(i\omega))$$





## A motivating example: sinusoidal fidelity

$$\dot{y} = ay + u \quad u(t) = \sin \omega t \quad \Longrightarrow \quad y(t) = y_0 e^{at} + e^{at} \int_0^t e^{-a\tau} \sin(\omega\tau) d\tau$$

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Using Rottmann, equation 132 on page 144:

$$\int e^{-a\tau} \sin(\omega\tau) d\tau = \frac{e^{-a\tau}}{a^2 + \omega^2} (-a \sin \omega\tau - \omega \cos \omega\tau) + C$$

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## Sinusoidal fidelity of LTIs, in details

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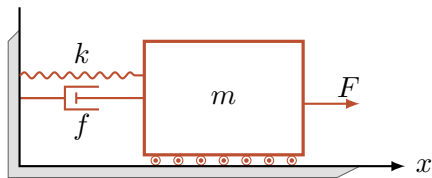
*Notation:*

- first term = free evolution
- second term = transient response (*part of the forced response*)
- third term = stationary response (*part of the forced response*)

# Why do we have free evolution, plus transient & stationary response?

Example: spring-mass system:

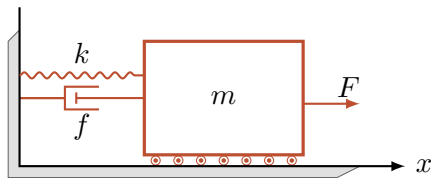
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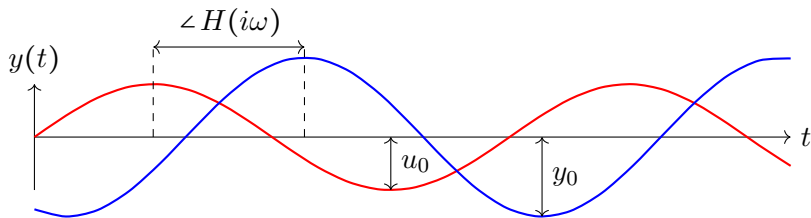
Answer: even if we have  $y_0 = 0$ , the machine needs to “warm up”

LTI means sinusoidal fidelity, and sinusoidal fidelity for every sine means LTI



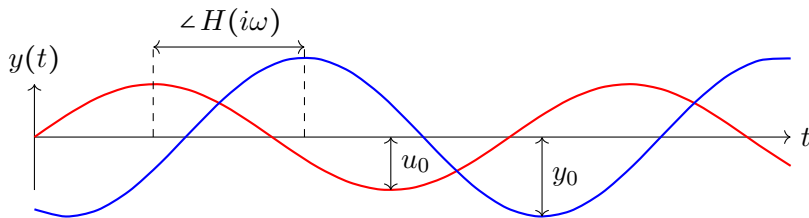
Idea: we can estimate  $H$  by repeating experiments with different sinusoidal  $u$ 's!

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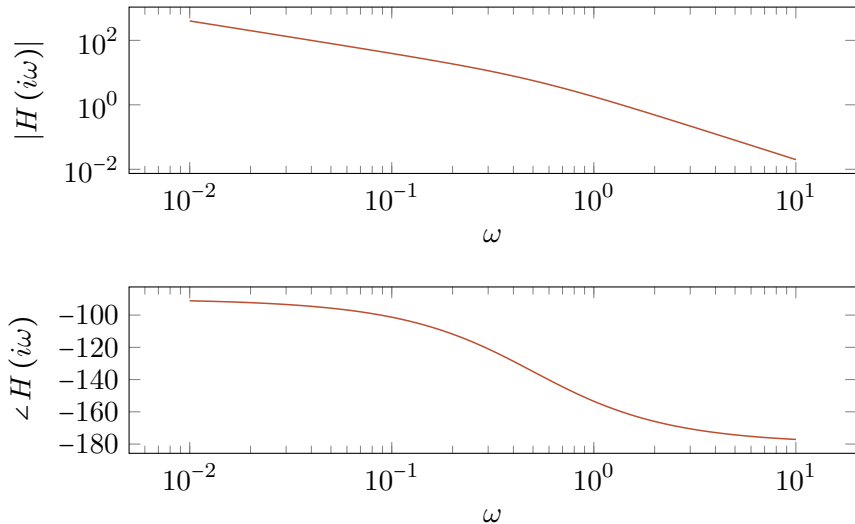


Good idea: check the “ETFE” system identification approach if you want to know more

Idea: we can estimate  $h$  by first estimating  $H$ , and then inverse-Laplacing!

... actually not. There are better ways of doing this → will be seen in courses that deal with system identification

Our goal: arrive at Bode plots



?