

LABREPORT LINEAR SYSTEM THEORY

“BOAT LAB ASSIGNMENT”

Delivered to
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Technology



Written by
Group 50

Magnus Bolstad
Mads-Emil Kvammen
Per Arne Kjelsvik

Faculty of Information Technology, Mathematics and Electrical
Engineering

NTNU

Høyskoleringen 1, Trondheim - 7491

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0 System

In this report, the ship is modeled using the following equations.

$$\dot{\xi}_w = \psi_w \quad (1a)$$

$$\dot{\psi}_w = -\omega_0^2 \xi_w - 2\lambda\omega_0 \psi_w + K_w w_w \quad (1b)$$

$$\dot{\psi} = r \quad (1c)$$

$$\dot{r} = -\frac{1}{T}r + \frac{K}{T}(\delta - b) \quad (1d)$$

$$\dot{b} = w_b \quad (1e)$$

$$y = \psi + \psi_w + v \quad (1f)$$

Equation 1: Equations used in the model

In this system ψ is the average heading of the ship, ψ_w is a high-frequency component in the heading due to wave disturbance, r is the rotation velocity about the z-axis, b is bias to the rudder angle, δ is the rudder angle relative to the BODY frame, w_b is white noise disturbance from the current and w_w is a zero mean white noise process with unity variance from the waves.

Furtermore, the system can be written as:

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}u + \mathbf{E}\mathbf{w}, \quad y = \mathbf{C}\mathbf{x} + v \quad (2)$$

Equation 2: Equation for the system

where \mathbf{x} , u and \mathbf{w} is given by:

$$\mathbf{x} = \begin{bmatrix} \xi_w \\ \psi_w \\ \psi \\ r \\ b \end{bmatrix}, \quad u = [\delta] \quad \text{and} \quad \mathbf{w} = \begin{bmatrix} w_w \\ w_b \end{bmatrix} \quad (3)$$

1 Part I - Identification of the boat parameters

1.a Transfer function from δ to ψ

In this problem, the transfer function from δ to ψ is calculated. Using the ship model and assuming that there is no effect of the current on the rudder, reveals:

$$\dot{\psi} = r \quad (4)$$

$$\ddot{\psi} = \dot{r} = -\frac{1}{T}r + \frac{K}{T}(\delta - b), \quad b = 0 \quad (5)$$

$$\ddot{\psi} = -\frac{1}{T}\dot{\psi} + \frac{K}{T}\delta \quad (6)$$

$$\Downarrow \mathcal{L}\{\ddot{\psi}\}$$

$$\psi(s)s^2 = -\frac{1}{T}\psi(s)s + \frac{K}{T}\delta(s) \quad (7)$$

$$\psi(s)(s^2 + \frac{1}{T}s) = \frac{K}{T}\delta(s) \quad (8)$$

$$\mathbf{H}(s) = \frac{\psi(s)}{\delta(s)} = \frac{K}{s(Ts + 1)} \quad (9)$$

Equation 9 is the transfer function parametrized by T and K , and it represents how the course will respond to a change in the rudder angle set point. In equation 5, we can see that $\ddot{\psi}$ has unit $[\frac{rad}{s^2}]$, which means that T must have unit $[s]$ and K must have unit $[\frac{1}{s}]$. δ and b are in $[deg]$, which can be converted to rad with the scalar product $\frac{\pi}{180}$.

1.b Boat parameters

In this task all the disturbances in the model are turned off in order to identify the boat parameters T and K . A sine input with amplitude 1 and given frequency is implemented in the **SIMULINK** model. To find both T and K , two input frequencies are required. Given frequencies are $\omega_1 = 0.005$ and $\omega_2 = 0.05$. The amplitude of the output signal for both frequencies will then be equal to $|H_1(j\omega_1)|$ and $|H_2(j\omega_2)|$.

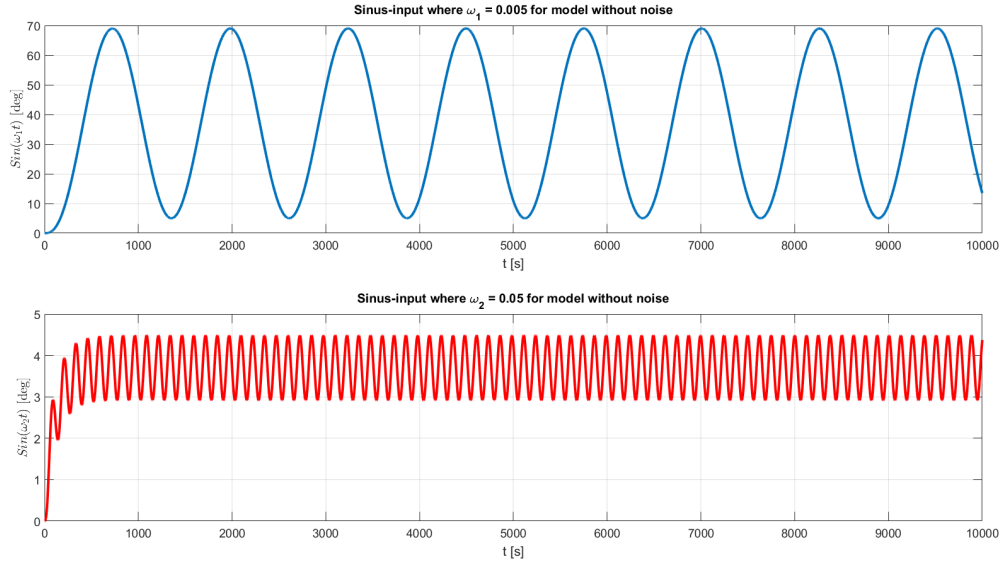


Figure 1: Simulation without disturbances, two different ω

Finding $|H_1(j\omega_1)|$ and $|H_2(j\omega_2)|$ using ω_1 and ω_2

Using the functions max and min in MATLAB for the output signal after steady-state is obtained reveals the maximum and minimum values for both frequencies. This is then further used in order to obtain the amplitude in both cases.

$$A_1 = |H_1(j\omega_1)| = \frac{68.96 - 5.022}{2} = 31.9787 \quad (10)$$

$$A_2 = |H_2(j\omega_2)| = \frac{4.485 - 2.916}{2} = 0.7847 \quad (11)$$

$$\begin{aligned}
 A_1 &= \left| \frac{K}{j\omega_1(Tj\omega_1 + 1)} \right| \\
 &= \left| \frac{K}{-T\omega_1^2 + j\omega_1} \right| \\
 &= \frac{K}{\sqrt{(-T\omega_1^2)^2 + \omega_1^2}} \\
 &= \frac{K}{\omega_1 \sqrt{T^2\omega_1^2 + 1}} \\
 &\Downarrow \\
 K &= A_1\omega_1 \sqrt{T^2\omega_1^2 + 1} \quad (12)
 \end{aligned}$$

$$\begin{aligned}
A_2 &\Rightarrow K = A_2 \omega_2 \sqrt{T^2 \omega_2^2 + 1} \\
T^2 \omega_2^2 + 1 &= \left(\frac{K}{A_2 \omega_2} \right)^2 \\
T &= \sqrt{\frac{\left(\frac{K}{A_2 \omega_2} \right)^2 - 1}{\omega_2^2}} \\
T &= \sqrt{\frac{K^2 - A_2^2 \omega_2^2}{\omega_2^2 A_2^2 \omega_2^2}} \\
T &= \frac{\sqrt{K^2 - A_2^2 \omega_2^2}}{A_2 \omega_2^2} \tag{13}
\end{aligned}$$

Finding values for K and T

Manipulating equation 12 and 13 and inserting the values found in 10 and 11 in order to find the values for T and K results in:

$$K = 0.1742 \text{ s}^{-1} \tag{14}$$

$$T = 86.5256 \text{ s} \tag{15}$$

1.c Waves and measurement noise

In this problem wave and measurement noise is turned on in the simulation. Using the same equations and approach as in section 1.b, new values for T and K are calculated.

New values for A_1 A_2 K T with disturbances

$$A_1 = 35.1054 \tag{16}$$

$$A_2 = 2.7057 \tag{17}$$

$$K = 0.1761 \text{ s}^{-1} \tag{18}$$

$$T = 16.6751 \text{ s} \tag{19}$$

It is difficult to find good new values for T and K using this approach due to the noise in the signal with lots of random spikes. As a result of this, it is not possible to get good estimates of the boat parameters, as seen in figure 2.

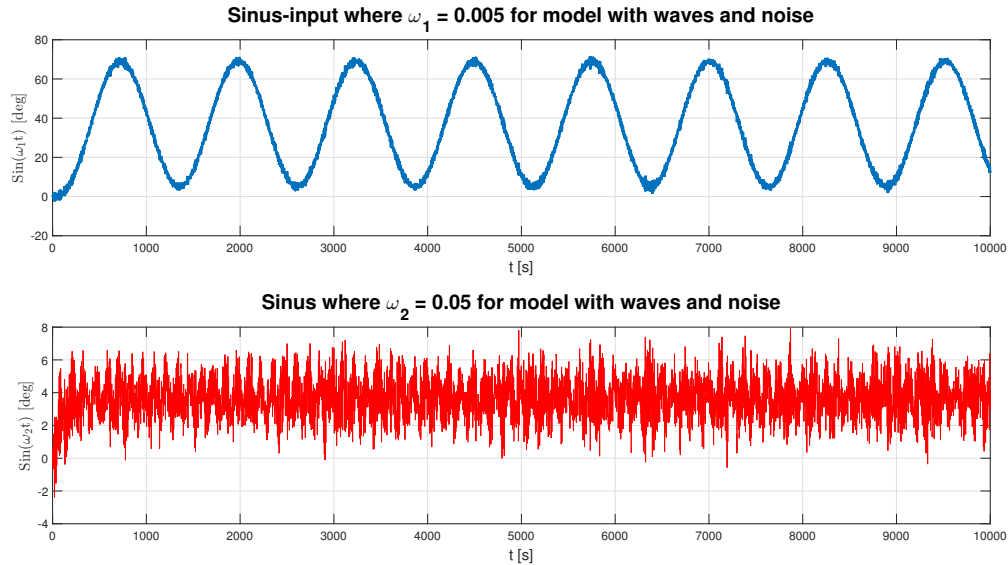


Figure 2: Simulation with disturbances, two different ω

1.d Step response

In this problem a step input of 1° to the rudder at $t = 0$ is applied to test the validity of the model. Comparing the step response of the model with the step response of the ship as shown in figure 3, reveals an offset in the response that grows with time. For small time periods, the model is good. The problem is when t gets larger, the model is more inaccurate. Anyway, this inaccuracy is not significant before after 2000-3000 seconds, which is an unrealistic long time to hold a constant rudder angle. Hence, the model is a good enough approximation for our purpose. The model could probably be improved further, by tuning the constants T and K , but by trial and error, no significant better results were obtained. Increasing K for instance, resulted in diverging path lines. Hence we decided to use the obtained values from the measurements in the rest of the assignment.

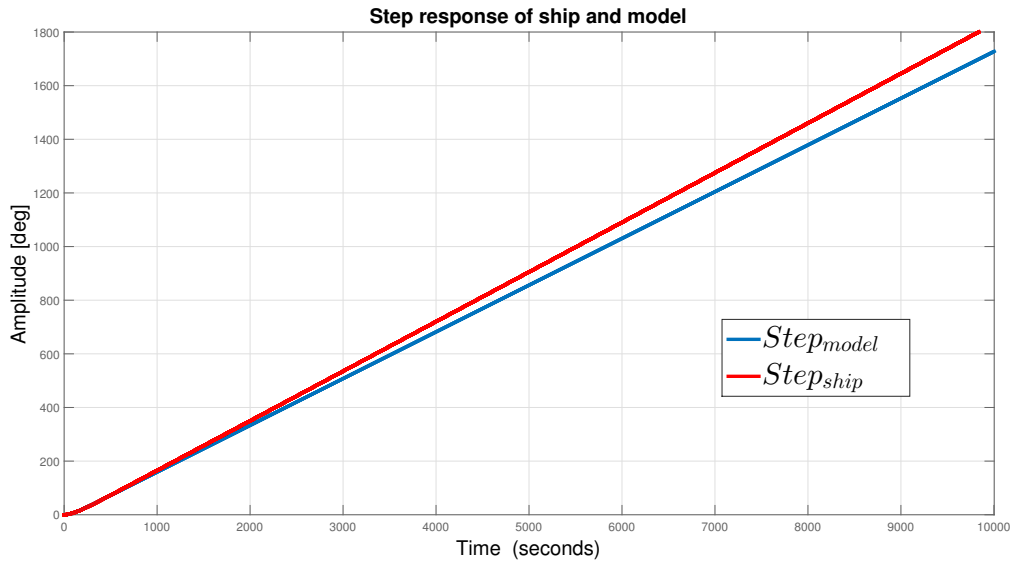


Figure 3: Step response of simulated boat and transfer function of boat model

2 Part II - Identification of wave spectrum model

2.a Power spectral density

In this problem, an estimate of the power spectral density function is to be found. In order to do so, the MATLAB function `pwelch` was used. This is a function that returns the PSD estimate of the input signal found using Welch's overlapped segment averaging estimator [3]. The PSD describes the distribution of power into frequency components. Figure 4 shows how the estimate of the PSD is in the model. While obtaining this plot and the estimate of the PSD, the input signal had to be modified so that the input was given in Hz and the output power was given in units power s/rad and the output frequency was given in rad/s. The scaling factors implemented to convert the outputs to the required units was $\frac{1}{2\pi}$ and 2π , respectively. This is the code for the estimated PSD:

Listing 1: MATLAB code for estimated psd

```

10 %% TASK 5.2.a — Estimate PSD
11 F_s = 10;
12 window = 4096;
13 noverlap = [];
14 nfft = [];
15
16 % \\ Find estimated PSD. Input converted from deg to rad
17 [pxx,f] = pwelch(psi_w(2,:).*(pi/180),window,noverlap,nfft,F_s);
18
19 % \\ f (Hz) to w (rad/s), and psd (power/pr. HZ) to psd (power s/rad)
20 omega = 2*pi.*f;
21 pxx = pxx./(2*pi);

```

2.b Transfer function from w_w to ψ_w

Finding the transfer function of the wave response (from w_w to ψ_w) using equation 20 and 21:

$$\dot{\xi}_w = \psi_w \quad (20)$$

$$\dot{\psi}_w = -\omega_0^2 \xi - 2\lambda\omega_0 \psi_w + K_w w_w \quad (21)$$

Taking the laplace transform on equation 20:

$$\begin{aligned} \xi_w(s) &= \psi_w(s) \\ \Downarrow \\ \xi_w(s) &= \frac{\psi_w(s)}{s} \end{aligned} \quad (22)$$

Taking the laplace transform on equation 21:

$$\begin{aligned} s\psi_w(s) &= -\omega_0^2 \xi_w(s) - 2\lambda\omega_0 \psi_w(s) + K_w w_w(s) \\ \Downarrow \text{Inserting equation 22 to find } \frac{\psi_w}{w_w} \\ s\psi_w(s) &= -\omega_0^2 \frac{\psi_w(s)}{s} - 2\lambda\omega_0 \psi_w(s) + K_w w_w(s) \\ \psi_w(s) \left(s + 2\lambda\omega_0 + \frac{\omega_0^2}{s} \right) &= K_w w_w(s) \\ H_{w_w, \psi_w}(s) = \frac{\psi_w(s)}{w_w(s)} &= \frac{K_w}{s + 2\lambda\omega_0 + \frac{\omega_0^2}{s}} \\ &= \frac{sK_w}{s^2 + 2\lambda\omega_0 s + \omega_0^2} \end{aligned} \quad (23)$$

Finding an analytical expression for the Power Spectral Density function of ψ_w using the transfer function of the wave response model:

$$P_{\psi_w}(j\omega) = P_{w_w}(j\omega) \cdot |H_{w_w, \psi_w}(j\omega)|^2 \quad (24)$$

It is stated that w_w is a zero mean white noise process with unity variance. Using this and the example given in [4, p. 117], the PSD for w_w can be found.

$$P_{w_w}(j\omega) = \mathcal{F}\{\mathbf{R}_{w_w}(\tau)\} \quad (25)$$

$$\begin{aligned} \mathbf{R}_{w_w} &= A\delta(\tau) \quad A = 1 \text{ at unity variance} \\ &= \delta(\tau) \quad \text{where } \delta(\tau) \text{ is the dirac delta function} \end{aligned} \quad (26)$$

$$P_{w_w}(j\omega) = \mathcal{F}\{\delta(\tau)\} = 1 \quad (27)$$

$$\begin{aligned} \Rightarrow P_{\psi_w}(j\omega) &= 1 \cdot |H_{w_w, \psi_w}(j\omega)|^2 \\ &= H(j\omega) \cdot H(-j\omega) \\ &= \frac{j\omega K_w}{(j\omega)^2 + 2\lambda\omega_0(j\omega) + \omega_0^2} \cdot \frac{-j\omega K_w}{(-j\omega)^2 + 2\lambda\omega_0(-j\omega) + \omega_0^2} \\ &= \frac{(\omega K_w)^2}{\left((\omega_0^2 - \omega^2) + j2\lambda\omega_0\omega\right)\left((\omega_0^2 - \omega^2) - j2\lambda\omega_0\omega\right)} \\ &= \frac{(\omega K_w)^2}{(\omega_0^2 - \omega^2)^2 + (2\lambda\omega_0\omega)^2} \\ &= \frac{(\omega K_w)^2}{\omega^4 + \omega_0^4 + 2\omega_0^2\omega^2(2\lambda^2 - 1)} \end{aligned} \quad (28)$$

2.c Resonance frequency, ω_0

By investigating the plot of the Power Spectral Density in 2.a, the resonance frequency, ω_0 , is found. The resonance frequency is the frequency where the Power Spectral Density has its maximum value. Hence, $\omega_0 = 0.7823$.

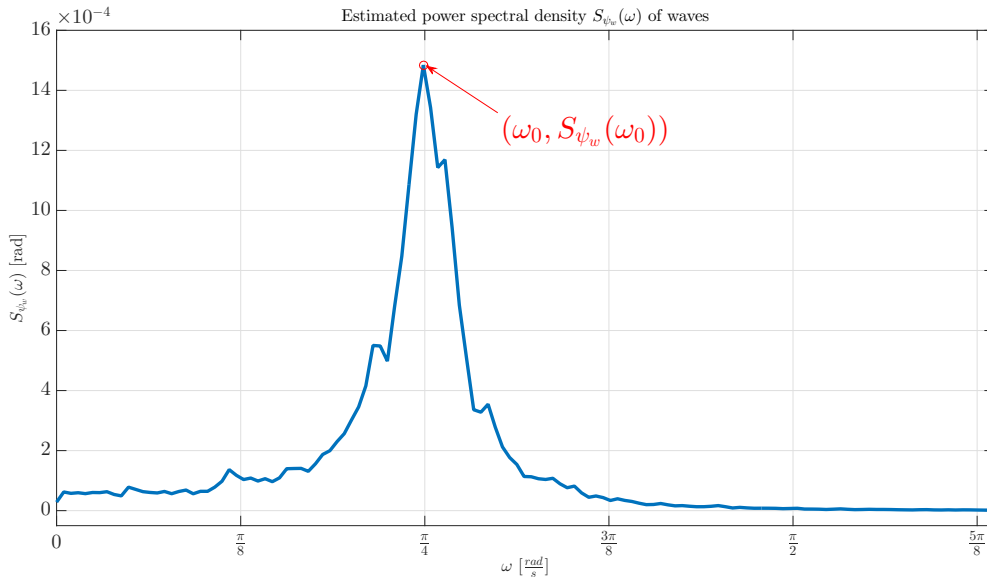


Figure 4: ω_0 was read from the plot and also computed in MATLAB

This value was also confirmed by finding it in MATLAB like this (see appendix A.2 for full code):

Listing 2: MATLAB code for ω_0

```

46 % Find resonance frequency from estimated PSD
47 [maxPSD, indexAtMaxPSD] = max(pxx);
48 omega_0 = omega(indexAtMaxPSD);

```

2.d Damping factor λ

To find a complete model for the wave response, the damping factor λ needs to be identified. To do so, K_w was defined as $2\lambda\omega_0\sigma$, where σ^2 is the peak value of $P_{\psi_w}(w)$. By choosing σ^2 as the peak of the estimated PSD, $S_{\psi_w}(w)$, the value could be set. This resulted in only λ having to be chosen in order to do the curve-fitting of $P_{\psi_w}(w)$ and $S_{\psi_w}(w)$. By trial and error in MATLAB, the value for λ was found to be 0.08 when the curves was the closest, as shown in figure 5. In the trial and error testing, closer values to 0.08 than shown in figure 5 was tested, although they are not displayed in the figure since that would make it hard to look at the difference between the high and low values of λ . See appendix A.2 for complete code for 5 and the more detailed plot that is not included.

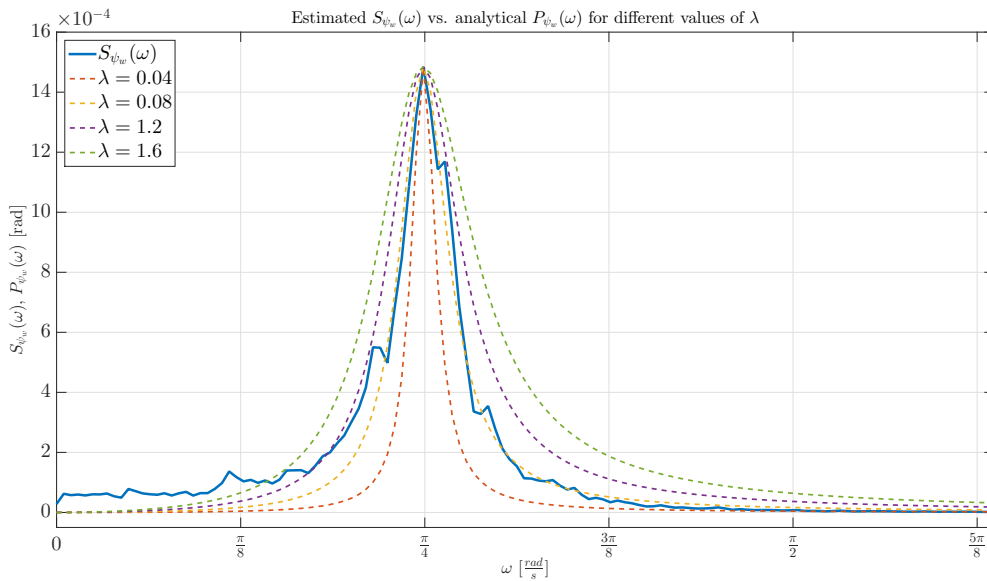


Figure 5: Plot of analytical PSD with different λ . $\lambda = 0.08$ was chosen

It was observed that when λ was set to a higher value, such as 0.8, the wave filtering performed later in the assignment,5.e, became more oscillating. As discussed in 5.e, this is not desirable.

3 Part III - Control system design

3.a PD controller

A PD-controller, $H_{pd} = K_{pd} \frac{1+T_d s}{1+T_f s}$, is to be designed from the transfer function 9. The derivative time constant, T_d , is chosen to be equal to the transfer function time constant, T in 15, such that they cancel. This gives the following transfer function for the open loop system:

$$h_0 = H_{pd}(s)H_{ship}(s) = \frac{KK_{pd}}{(1+T_f s)s} \quad (29)$$

To obtain the desired phase margin and cross frequency $\varphi = 50$ and $\omega_c = 0.1$ and, the following equations were solved by looking at the angle of the transfer function:

$$\varphi = \angle h_0(j\omega_c) - (-180^\circ) \quad (30)$$

$$50^\circ - 180^\circ = \angle h_0(j\omega_c) \quad (31)$$

$$\angle \frac{KK_{pd}}{(1+T_f s)s} = -130^\circ \quad (32)$$

$$0 - \angle -\omega_c^2 T_f + j\omega_c = -130^\circ \quad (33)$$

$$\tan^{-1}\left(\frac{\omega_c}{-\omega_c^2 T_f}\right) = 130^\circ \quad (34)$$

$$T_f = -\frac{1}{\omega_c \tan(130^\circ)} \quad (35)$$

$$T_f = 8.39s \quad (36)$$

K_{pd} can then be found by

$$|h_0(j\omega_c)| = 1 \quad (37)$$

$$\sqrt{\frac{K^2 K_{pd}^2}{(1+T_f s)^2 s^2}} = 1 \quad (38)$$

Inserting $K = 0.1742$ into 38 gives:

$$K_{pd} = 0.7494 \quad (39)$$

This gives the following Bode plot for the open loop system:

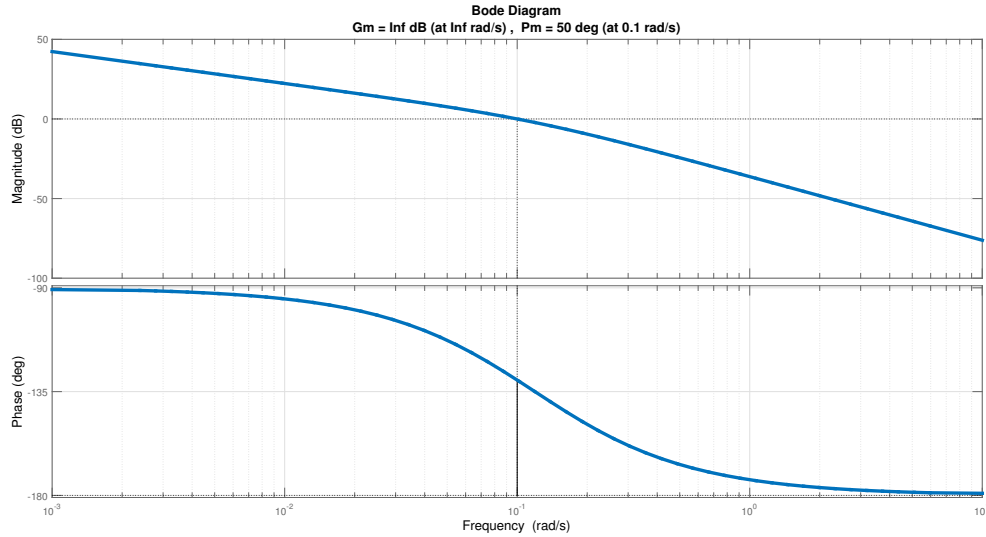


Figure 6: Bode plot of transfer function where $\omega_c = 0.1$ and $\varphi = 50^\circ$ as desired

The T_f constant is what limits the derivative-effect. A higher T_f would for example limit the derivative-effect more than it currently does, which means that the controller would low-pass less frequencies, or rather, allow more frequencies to pass. This would in turn introduce oscillations in the response, and perhaps overshooting, while lowering T_f would make the response slower. Lowering T_d would perhaps make the response faster, since T_d is derivative-effect. However, T_f and T_d remains as defined throughout the rest of the assignment to meet requirements to phase margin φ and cross frequency ω_c .

3.b Simulation without disturbance

The implemented PD-controller gives the following compass course with a course reference of 30° , when no disturbances are present, only measurement noise. To make sure that the rudder set-point stayed within its limits of $\pm 35^\circ$, a saturation block was added to the SIMULINK model for the rudder input δ . After approximately 320 seconds, the ship has obtained the reference course angle, which is an acceptable result for a large ship. This means that the autopilot works for smooth weather conditions. Figure 7 shows the autopilot with only measurement noise and no other disturbances. For the rest of the assignment, the symbol δ will be used for the rudder input angle, and not the actual rudder angle, which δ was interpreted as before.

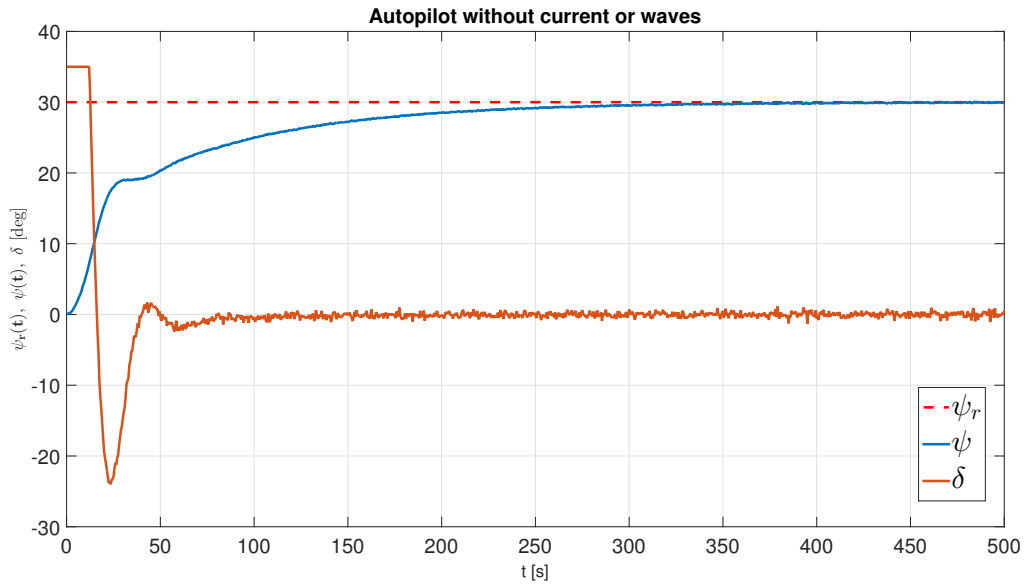


Figure 7: Autopilot without any disturbances

It is interesting to see how the rudder changes direction even before the heading has reached the reference. This is done by the derivative effect because the course is changing too fast, and otherwise it would result in an overshoot off the planned change in the course. This has to be done because of the momentum and the inertia of the ship.

The effect of removing the saturation block entirely was investigated. The rudder control started at an angle of 220° , which is unphysical, as a real rudder can only rotate $\pm 90^\circ$.

3.c Simulation with current disturbance

The autopilot is now tested with current disturbance present. It is assumed that the only effect of the current is rudder angle bias. The course angle has the same shape as without the current disturbance, but will now have a steady state error as seen in figure 8. The rudder set-point will get a non-zero value, because of the error from the reference course angle, but the current bias will act in the opposite direction, such that the actual rudder angle will be zero, and the ship will head forward with a stationary error in the compass course. The rudder doesn't reach a high enough value to overcome the current, and turn the ship to its desired course. The ship will head approximately 3° off the desired 30° . This is not satisfactory, and further adjustment have to be made. This steady state error could have been removed by the integral effect in an PI or PID- controller. The PI or PID controller would have used some time to accumulate the error, and correct it, but on the other hand, a Kalman filter will spend some time finding the estimations by iteration. In the end of this assignment, Kalman filtering together with the PD controller is used instead, with sufficiently fast sampling time. Estimate of the current bias is used to cancel the current bias disturbance. The assumption that the current only will affect the rudder angle is a simplification, and in real, the hull of the ship would also be influenced.

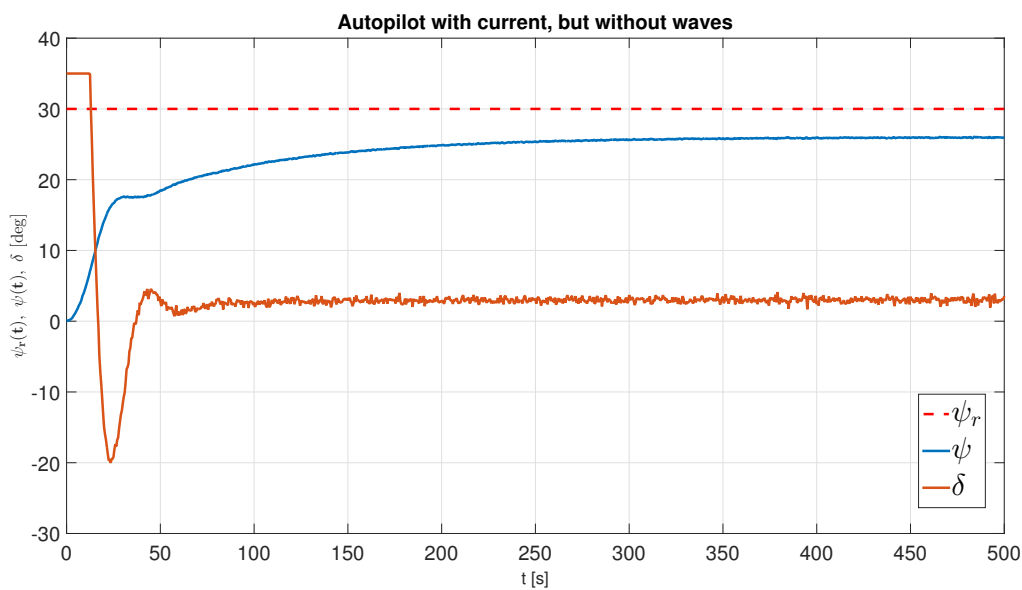


Figure 8: Autopilot with current disturbance

3.d Simulation with wave disturbance

The current disturbance is now turned off, and wave disturbance is included to the SIMULINK model. Figure 9 shows the ship in this setting. It can be seen from the figure that the ship reaches its heading. The problem here is that the rudder tries to counter the effect each wave has on the ship. The combination of the rudder changing rapidly and the rocking waves results in lots of fluctuations in the heading of the ship around its desired value.

The rapid changing of the rudder as a result of the waves is important to deal with. In the worst cases the rudder set-point changes by more than 40 degrees in about 4 seconds. This is an unrealistic high value for a large rudder, and by doing this over a period of time, the rudder will be destroyed. It's obvious that this is a matter of importance and needs to be fixed to get a desired behavior from the ship. To solve these problems, a Kalman Filter which uses a feed-forward loop to cancel out the offset from the current disturbance shown in section 3.c and a filtered feedback that smooths the rudder input, δ , so that the rudder doesn't fluctuate as much as shown in figure 9 is implemented, see section 5.

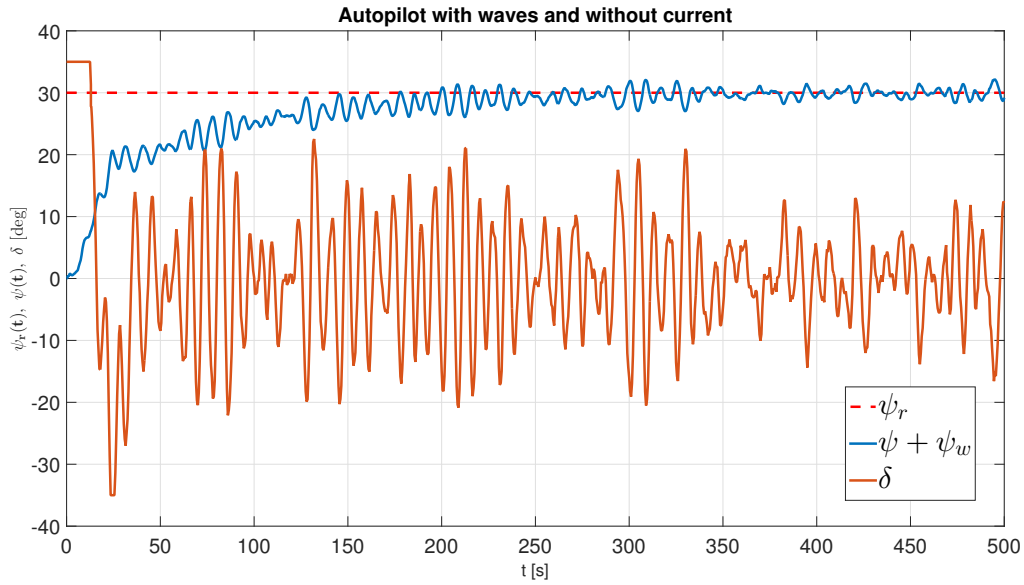


Figure 9: Autopilot with both current and wave disturbance

The rudder input under the different conditions previously discussed are summarized in fig. 10. Current causes the rudder input to have a stationary error, or offset while the waves causes the rudder input to change constantly to counteract the wave disturbance. In the situation with measurement noise only, the rudder input will converge to zero when the reference course is obtained. The effect of the measurement noise, which is present during all of the three conditions, can be seen as small, high frequent oscillations. This is easiest seen on the plot without external disturbance, and the one with current disturbance, after they have reached their steady state error.

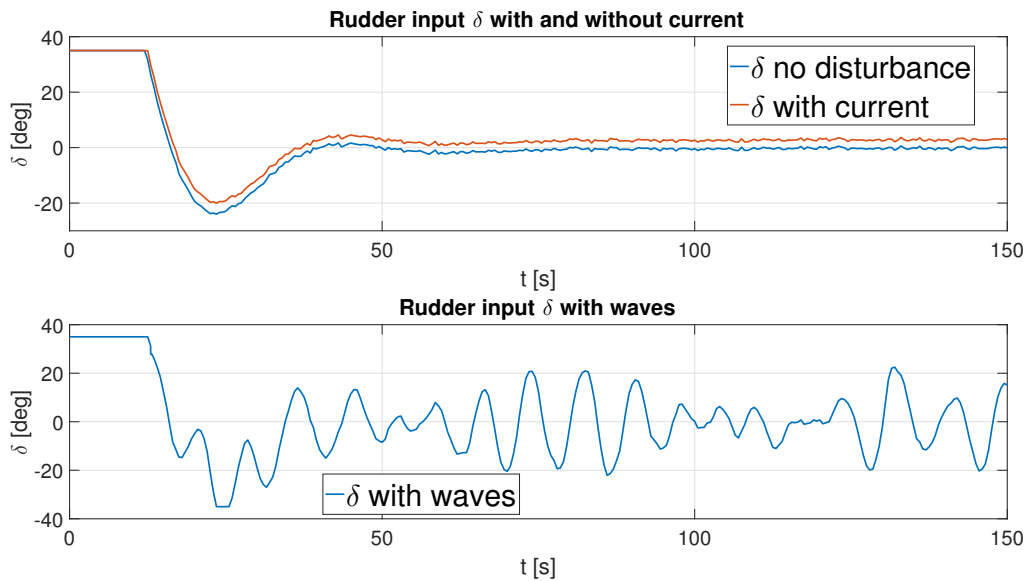


Figure 10: Rudder input in the three cases: No disturbances (except measurement noise), current disturbance and wave disturbance

4 Part IV - Observability

4.a State space model

Using the equations stated in 1 along with equation 2 and 3 in order to find the matrices A, B, C and E:

$$\dot{\mathbf{x}} = \underbrace{\begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ -\omega_0^2 & -2\lambda\omega_0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & -\frac{1}{T} & -\frac{K}{T} \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix}}_{=\mathbf{A}} \mathbf{x} + \underbrace{\begin{bmatrix} 0 \\ 0 \\ 0 \\ \frac{K}{T} \\ 0 \end{bmatrix}}_{=\mathbf{B}} u + \underbrace{\begin{bmatrix} 0 & 0 \\ K_w & 0 \\ 0 & 0 \\ 0 & 0 \\ 0 & 1 \end{bmatrix}}_{=\mathbf{E}} \mathbf{w} \quad (40)$$

$$\mathbf{y} = \underbrace{\begin{bmatrix} 0 & 1 & 1 & 0 & 0 \end{bmatrix}}_{=\mathbf{C}} \mathbf{x} + v \quad (41)$$

4.b Without disturbance

When there are no disturbances in the system, $b = 0$ and all the wave disturbances are neglected. The state vector is then reduced to:

$$\mathbf{x} = \begin{bmatrix} \psi \\ r \end{bmatrix} \quad (42)$$

Without the disturbances, \mathbf{w} is also removed from the system while u is the same. This gives the following system equation:

$$\dot{\mathbf{x}} = \underbrace{\begin{bmatrix} 0 & 1 \\ 0 & -\frac{1}{T} \end{bmatrix}}_{=\mathbf{A}} \mathbf{x} + \underbrace{\begin{bmatrix} 0 \\ \frac{K}{T} \end{bmatrix}}_{=\mathbf{B}} u \quad (43)$$

$$\mathbf{y} = \underbrace{\begin{bmatrix} 1 & 0 \end{bmatrix}}_{=\mathbf{C}} \mathbf{x} + v \quad (44)$$

To find out whether the system is observable or not, the observability matrix is computed:

$$\mathcal{O} = \begin{bmatrix} \mathbf{C} \\ \mathbf{CA} \\ \vdots \\ \mathbf{CA}^{n-1} \end{bmatrix} \quad (45)$$

[1, p. 197] The code for section 4 is shown in appendix A.4

This can easily be computed by using the MATLAB function `obsv(A,C)`, where the A matrix and C matrix are input. This gives the following observability matrix:

$$\mathcal{O} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \quad (46)$$

This matrix has rank equal two, which is full rank. Thus, the system is observable without disturbances.

4.c Current disturbance

In this problem only the current disturbance is included in the system. The wave disturbances are still neglected. This results in the following state vectors:

$$\mathbf{x} = \begin{bmatrix} \psi \\ r \\ b \end{bmatrix}, \quad u = \delta, \quad w = w_b \quad (47)$$

With the new state vectors the system can be reduced to:

$$\dot{\mathbf{x}} = \underbrace{\begin{bmatrix} 0 & 1 & 0 \\ 0 & -\frac{1}{T} & -\frac{K}{T} \\ 0 & 0 & 0 \end{bmatrix}}_{=\mathbf{A}} \mathbf{x} + \underbrace{\begin{bmatrix} 0 \\ \frac{K}{T} \\ 0 \end{bmatrix}}_{=\mathbf{B}} u + \underbrace{\begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}}_{=\mathbf{E}} w \quad (48)$$

$$\mathbf{y} = \underbrace{\begin{bmatrix} 1 & 0 & 0 \end{bmatrix}}_{=\mathbf{C}} \mathbf{x} + v \quad (49)$$

The observability matrix with current disturbance becomes:

$$\mathcal{O} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & -0.0116 & -0.0020 \end{bmatrix} \quad (50)$$

This observability matrix has rank equal to three, which is also full rank. Thus the system is observable also with the current disturbance.

4.d Wave disturbance

In this problem the wave disturbance is included in the system. The current disturbances is then neglected. This results in the following state vectors:

$$\mathbf{x} = \begin{bmatrix} \xi_w \\ \psi_w \\ \psi \\ r \end{bmatrix}, \quad u = \delta, \quad w = w_w \quad (51)$$

With the new state vectors the system can be reduced to:

$$\dot{\mathbf{x}} = \underbrace{\begin{bmatrix} 0 & 1 & 0 & 0 \\ -\omega_0^2 & -2\lambda\omega_0 & 0 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & -\frac{1}{T} \end{bmatrix}}_{=\mathbf{A}} \mathbf{x} + \underbrace{\begin{bmatrix} 0 \\ 0 \\ 0 \\ \frac{K}{T} \end{bmatrix}}_{=\mathbf{B}} u + \underbrace{\begin{bmatrix} 0 \\ K_w \\ 0 \\ 0 \end{bmatrix}}_{=\mathbf{E}} w \quad (52)$$

$$\mathbf{y} = \underbrace{\begin{bmatrix} 0 & 1 & 1 & 0 \end{bmatrix}}_{=\mathbf{C}} \mathbf{x} + v \quad (53)$$

The observability matrix with wave disturbance becomes:

$$\mathcal{O} = \begin{bmatrix} 0 & 1 & 1 & 0 \\ -0.6120 & -1.2517 & 0 & 1 \\ 0.7661 & 0.9548 & 0 & -0.0116 \\ -0.5844 & -0.4290 & 0 & 0.0001 \end{bmatrix} \quad (54)$$

This observability matrix has rank equal to four, which is also full rank. Thus the system is observable also when there is wave disturbances.

4.e Both current and wave disturbance

For this part, the observability for the complete system was to be investigated. This was investigated based on the system shown in equation 40 and 41 as well as the equation for computing the observability matrix shown in equation 45

$$\mathcal{O} = \begin{bmatrix} 0 & 1 & 1 & 0 & 0 \\ -61.20 & -1.2517 & 0 & 1 & 0 \\ 0.7661 & 0.9548 & 0 & -0.0116 & -0.0020 \\ -0.5844 & -0.4290 & 0 & 0.0001 & 0 \\ 0.2626 & -0.0473 & 0 & 0 & 0 \end{bmatrix} \quad (55)$$

This observability matrix has rank equal to five, which is also full rank. Thus the system is observable with all disturbances.

In part 4 it has been observed that the system is observable in all cases. This means that for a measured output it is possible to determine the initial states as well as the behavior of the system. This is useful for the ship model, as it makes it possible to find where the ship has been and how it has behaved just from the measured outputs and the input during the time. The observability also make it possible to make use of estimators, which will be used in the next part of the assignment, to improve the PD controller made in section 3a.

5 Part V - Discrete Kalman filter

5.a Discretization

To improve the behavior of the ship, a Kalman filter is included to the PD controller. By including this, it's possible to make better estimates of the state variables that are affected by the noise in the system. In this part, exact discretization is used in the Kalman filter based on the model from 4.a. By discretizing the model, the continuous model is transferred into a discrete counterpart. This is done by using the MATLAB function `c2d`, which converts a model from continuous to discrete time. The sampling frequency was given as 10 Hz which gives a sampling time of 0.1 s.

The MATLAB function `c2d` only allows two inputs from the state space equations. It is given that $\mathbf{C}_d = \mathbf{C}$ and $\mathbf{D}_d = \mathbf{D}$ [1, p. 110], which means that only \mathbf{A} , \mathbf{B} and \mathbf{E} has to be discretized using the `c2d` function. This should be okay, since it is a linear time-invariant system, which means that the superposition property holds. Superposition says that the net response of two or more inputs at a give time and place is the same as the sum of the response caused by each input by itself. The result of this is that the `c2d` can be simply be run twice. The discretized matrices was found to be:

$$\mathbf{A}_d = \begin{bmatrix} 0.9970 & 0.0993 & 0 & 0 & 0 \\ -0.0608 & 0.9845 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0.0999 & -0.0000 \\ 0 & 0 & 0 & 0.9988 & -0.0002 \\ 0 & 0 & 0 & 0 & 1 \end{bmatrix}, \mathbf{B}_d = \begin{bmatrix} 0 \\ 0 \\ 0.0000 \\ 0.0002 \\ 0 \end{bmatrix},$$

$$\mathbf{C}_d = [0 \ 1 \ 1 \ 0 \ 0], \mathbf{D}_d = 0, \mathbf{E}_d = \begin{bmatrix} 0 & 0 \\ 0.0007 & 0 \\ 0 & -0.0000 \\ 0 & -0.0000 \\ 0 & 0.1 \end{bmatrix} \quad (56)$$

The MATLAB code for this is as follows:

Listing 3: Using `c2d`-function to compute discretized matrices

```

44 % \\\ Exact discretization
45 [Ad, Bd] = c2d(A,B,T_s);    Cd = C;
46 [Ad, Ed] = c2d(A,E,T_s);    Dd = D;

```

5.b Variance of measurement noise

To find an estimate of the variance of the measurement noise, the MATLAB-function `var` was used. This function returns the variance of a vector. In SIMULINK, the input δ is set to zero while the compass value is measured. As a result of this, only the measurement

noise on the compass is measured. This measurement is imported to MATLAB as a vector, and the `var`-function is run based on this. The resulting variance of the measurement noise, R , was:

$$\sigma^2 = 6.1614 \cdot 10^{-7} \quad (57)$$

5.c Implementation of discrete Kalman filter

The discrete Kalman filter is implemented by using MATLAB.

Notation:

A priori is denoted as $\hat{\mathbf{x}}^-[k]$

A posteriori is denoted as $\hat{\mathbf{x}}[k]$

The filter is initialized at:

$$\hat{\mathbf{x}}^-(0) = E[\mathbf{x}(0)] = \mathbf{m}_{\mathbf{x}_0} \quad (58)$$

$$\begin{aligned} \mathbf{P}^-(0) &= E[\mathbf{E}^-(0)\mathbf{E}^-(0)^T] \\ &= E[(\mathbf{x}(0) - \mathbf{m}_{\mathbf{x}_0})(\mathbf{x}(0) - \mathbf{m}_{\mathbf{x}_0})^T] \\ &= \mathbf{C}_{\mathbf{x}_0} \end{aligned} \quad (59)$$

The stepwise solution implemented in MATLAB was:

$$\mathbf{L}[k] = \mathbf{P}^-[k]\mathbf{C}^T(\mathbf{C}\mathbf{P}^-[k]\mathbf{C}^T + \bar{\mathbf{R}}_{\mathbf{v}})^{-1} \quad (60)$$

\Downarrow

$$\hat{\mathbf{x}}[k] = \hat{\mathbf{x}}^-[k] + \mathbf{L}[k](\mathbf{y}[k] - \mathbf{C}\hat{\mathbf{x}}^-[k]) \quad (61)$$

\Downarrow

$$\mathbf{P}[k] = (\mathbf{I} - \mathbf{L}[k]\mathbf{C})\mathbf{P}^-[k](\mathbf{I} - \mathbf{L}[k]\mathbf{C})^T + \mathbf{L}[k]\bar{\mathbf{R}}_{\mathbf{v}}[k]\mathbf{L}[k]^T \quad (62)$$

\Downarrow

$$\hat{\mathbf{x}}^-[k+1] = \bar{\mathbf{A}}\hat{\mathbf{x}}[k] + \bar{\mathbf{B}}\mathbf{u}[k] \quad (63)$$

$$\mathbf{P}^-[k+1] = \bar{\mathbf{A}}\mathbf{P}[k]\bar{\mathbf{A}}^T + \bar{\mathbf{Q}}_{\mathbf{w}} \quad (64)$$

The initial a priori estimate error covariance and the initial a priori state estimate is given in the assignment [2]. This is implemented in a MATLAB function block in SIMULINK, with compass course and rudder angle as input, as shown in fig. 26 in appendix 5.d. The output is the estimated values, which are used as feedback and feedforward. The code for the MATLAB function block is as follows:

Listing 4: MATLAB function block Kalman algorithm

```

1 function [b,psi] = Kalman_matlab_fnc(u, y, data)
2 %#codegen
3 persistent init_flag A B C E Q R P_ x_ I
4
5 if (isempty(init_flag))
6     init_flag = 1;
7
8     % Initialization for system
9     [A,B,C,E,Q,R,P_,x_, I] = deal(data.Ad,data.Bd,data.Cd,data.Ed, ...
10         data.Q, data.R, data.P_0, data.X_0, data.I);
11 end
12
13 % 1 – Compute the Kalman Gain
14 L = (P_*C')/((C*P_*C'+R));
15 % 2 – Update estimate with measurment
16 x = x_ + L*(y-C*x_);
17 % 3 – Update error covariance matrix
18 P = (I - L*C)*P_*(I-L*C)' + L*R*L';
19 % 4 – Project ahead
20 x_ = A*x + B*u;
21 P_ = A*P*A' + E*Q*E';
22
23 psi = x(3); b = x(5);
24
25 end

```

The MATLAB function block was also edited to take in the variable `data` from the current workspace. In the MATLAB code for part 5 (found in appendix A.5) the variable `data` is defined like this:

Listing 5: Defining struct data for use in Kalman

```

66 % \\ Q is the Process noise covariance
67 Q = [30 0; 0 10^(-6)];
68 P_0 = [1 0 0 0 0; 0 0.013 0 0 0; 0 0 pi^2 0 0; 0 0 0 1 0; ...
69     0 0 0 0 2.5*10^-4];
70 X_0 = [0; 0; 0; 0; 0];
71 R = R/T_s;
72 I = diag([1 1 1 1 1]);
73
74 % \\ Put data in a struct for use in the Kalman filter
75 data = struct('Ad',Ad,'Bd',Bd,'Cd',Cd,'Ed', Ed, 'Q',Q,'R', R,'P_0',P_0, ...
76     'X_0',X_0, 'I', I);

```

The struct is then dealt to the persistent global variables of the MATLAB function block during first initialization where `init_flag` is an empty variable. The output of the function block is also changed to be of size 1 in both output ports. Also, a Memory block is applied to the output of the ship and the output of the boat, as in all other SIMULINK implementations, to avoid problems with algebraic loops. Lastly there is an zero order hold on the input of the Kalman filter as rudder angle and compass measurements are

continuous. The sample time on these blocks are also 0.1s. The noise covariance is obtained through an averaging convention.

Estimates for values in the ship model, that are not available for measuring (such as the rudder bias from the current) are now available through the estimators. This is utilized in section 5.d-5.e of the assignment. The effect of changing the sampling frequency was also investigated, and it was concluded that a higher sampling frequency did not improve the results significantly, while a lower sampling frequency resulted in inaccurate measurements.

5.d Feed forward from estimated bias

The a posteriori estimated bias, found by the algorithm in 5.c, is now used in a feed forward to the rudder, such that the bias is cancelled. The reference course, ψ_r , is 30° . With a sample frequency of 10 Hz, the following behavior is obtained with current disturbance:

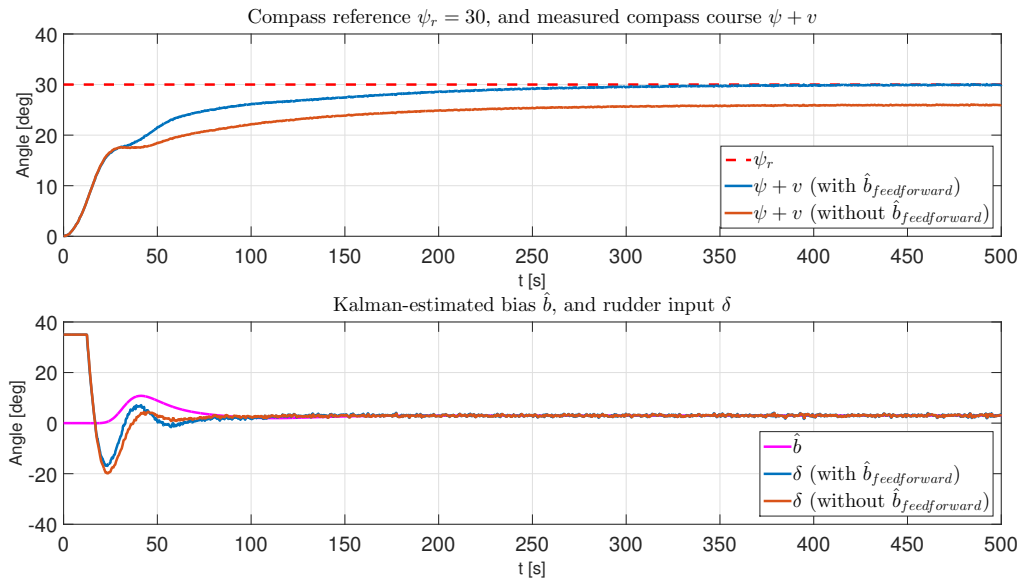


Figure 11: Comparing compass and rudder input angle with and without estimated feed-forward current bias

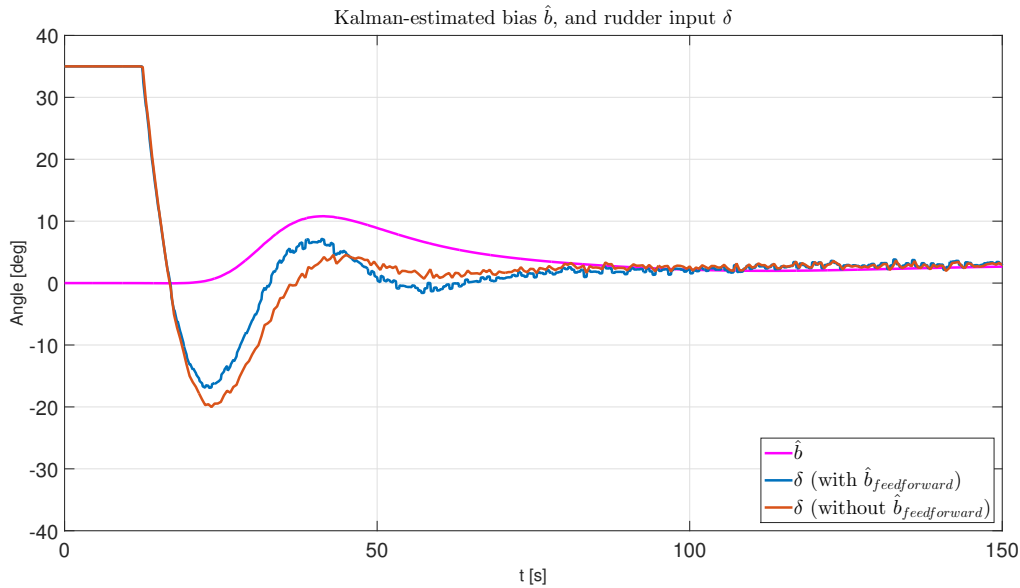


Figure 12: More detailed view of figure 11

The measured compass course is now equal to the reference angle after approximately 320 seconds, and the steady state error of approximately 3 degrees from problem 5.3, part c) is now removed. This is because the estimated bias is equal to the steady state error, and the estimated bias is added to the signal from the PD controller. Hence the current disturbance is cancelled. This performance is satisfactory, and much better than the one in problem 5.3, part c).

It is a difference in the rudder angle in the beginning of figure 11, where the plot picturing the rudder set-point angle with the estimated feed-forward current bias is slightly higher than the one without the feed-forward. This difference is enough to lift the compass angle to its desired value of 30 degrees, and the steady state error is avoided. The reason for the angle to be more accurate now, is that the estimation for the current bias cancels the effect of the current. After a while, both the rudder angles, with and without the feed forward, will be equal. In the case without the feed forward, there will be a steady state error, while in the case with a feed forward, the course angle will follow the reference perfectly.

It is seen from fig. 11 and fig. 12 how the rudder input angles become different from each other when the estimated bias becomes non-zero. This value is used as feed forward, and added to the rudder set-point angle in the case with a Kalman filter. After a while, this course angle is equal to the reference, and the error in the feedback loop is zero. The feed-forward will then control the rudder alone, which is confirmed by the fact that rudder and the estimated bias are equal after approximately 100 s. In the case without Kalman filter, the feedback error to the PD controller will control the rudder. Summarized; the feed forward gives the rudder set-point a large enough value to overcome the bias from the current.

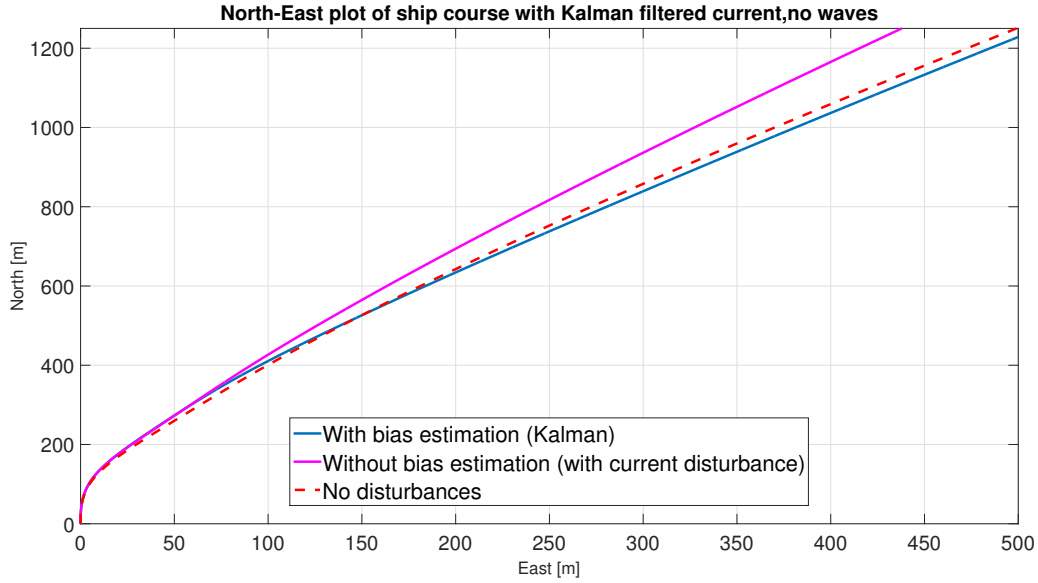


Figure 13: Ship path, both with and without the bias estimation

From fig. 13 it is obvious that the offset in course angle will result in a larger and larger offset from the actual destination, as the ship sails forward, while the controller with feed forward will follow approximately the same path as in the situation without disturbance.

5.e Wave filtering

The waves may change fast and frequently, and hence the measured compass course will also fluctuate a lot. Measurement noise will also contribute to this. Anyway, the average of the high frequency component due to wave disturbance and noise may be close to zero, and hence the course remains quite constant. With the controller using this measured angle as feedback, the rudder may unnecessarily change continuously because of the high frequency waves. To avoid this wear and tear on the rudder, the high frequency component due to wave disturbance, ψ_w , is substituted with the estimated $\hat{\psi}$, in the feedback. This is referred to as wave filtering, and only the dynamics of the course angle will influence the rudder. When both current and waves are present, the feed forward from the estimated bias will cancel the current bias, and the situation is approximately equal to the one in 3.d, where only waves are present.

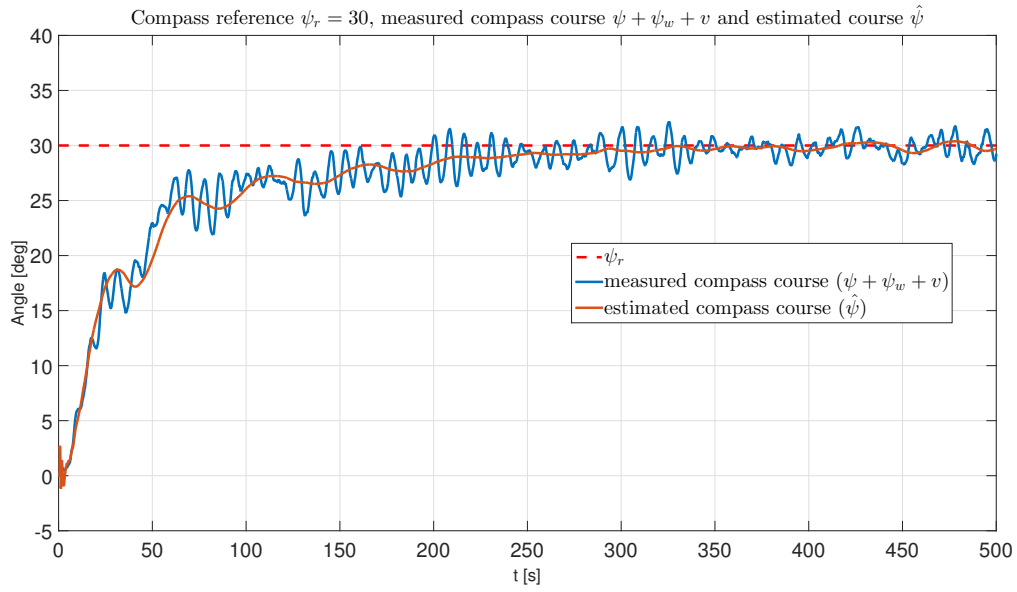


Figure 14: Measured and estimated courses

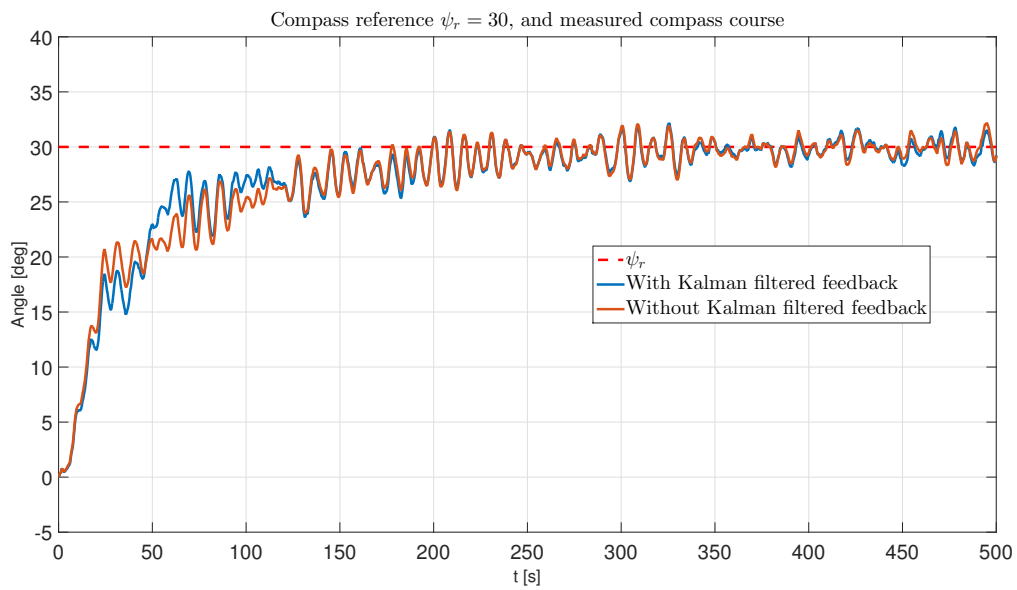


Figure 15: Ship course, with and without filtered feedback

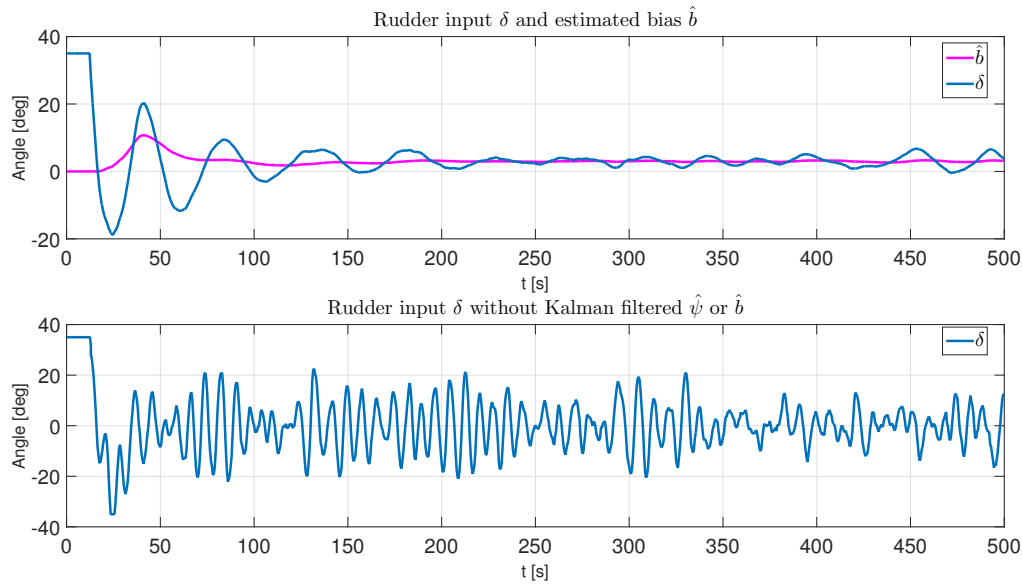


Figure 16: Estimated bias and rudder input angle in the situation with filtered (upper) and non-filtered (lower) feedback

It is obvious from fig. 14 how the estimated, wave filtered $\hat{\psi}$ follows the dynamics of the measured course, but avoids excessive oscillations. The high frequency component of the wave disturbance, as well as the measurement noise is filtered out. The remaining slow oscillations in the compass course, due to the waves rocking the ship, can not be filtered out.

Compared to the controller in section 3.d, the feed forward estimated bias will remove the steady state error, while the filtered feedback avoids unnecessary change in the rudder input.

As seen from fig. 15, the ship will sail with a constant compass course equal to the reference course after approximately four minutes, with small fluctuations because of the waves. It is observed how the ship will have approximately the same course during wave disturbance, independent on whether the actual measured, or the estimated compass course is used as feedback, but the latter alternative clearly avoids unnecessary wear and tear on the rudder. This is because a less fluctuating signal will be sent back to the rudder input, and hence cause a less fluctuating error, as seen from the estimated course in fig. 14. The impact of the estimated feedback can be seen by the rudder input angles in fig. 16. It is clear how the situation with filtered feedback will cause less wear and tear on the rudder due to the filtering of high-frequent noise, which was present in fig. 10, in part 3. The estimated bias has the same shape as in section 5.d.

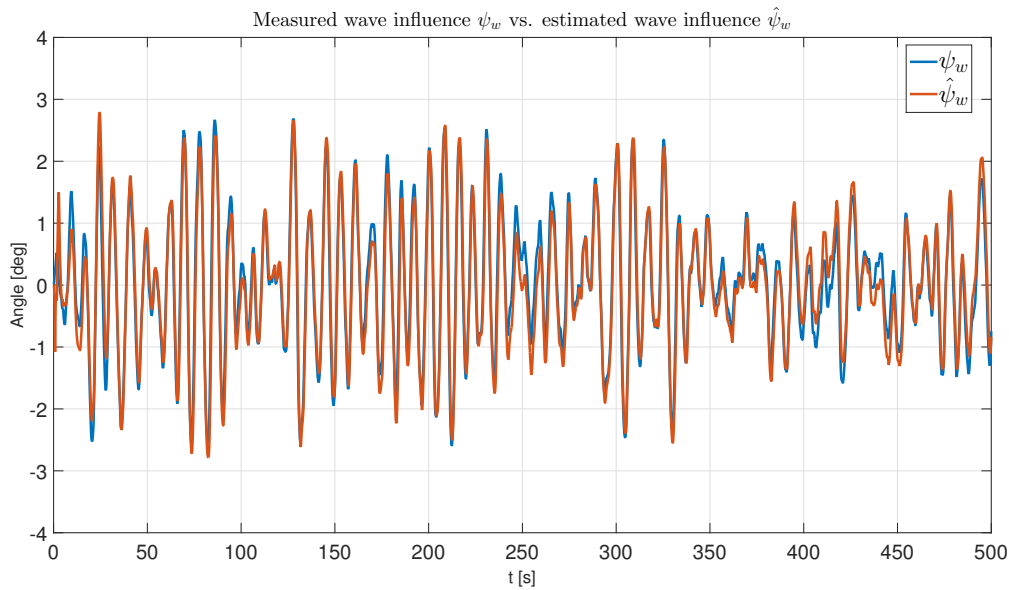


Figure 17: Actual wave influence and estimated and estimated wave influence. Rudder is set to zero to investigate the wave influence on the system only.

To measure the actual wave influence, the measurement noise and current disturbance must be turned off. The rudder input δ to the ship and Kalman filter must also be equal to zero, as seen in the SIMULINK implementation in figure 28, appendix B. The implemented Kalman MATLAB function block outputs the estimated compass heading, seen in appendix A.6. To output the estimated wave influence $\hat{\psi}_w$ instead, the last line of code must be changed like this

```
23 psi = x(3); b = x(5); % x(3) is average compass heading
```

```
23 psi_w = x(2); b = x(5); % Output name also changed to psi_w,
```

Now the ship outputs the wave influence ψ_w and the Kalman filter outputs the estimated wave influence $\hat{\psi}_w$. From 17 it is observed that the estimated high frequency wave influence is quite similar to the actual measured wave influence. This indicates that the estimator is quite good.

From fig. 18 it is seen that the filtered feedback does not improve the course significantly, as discussed above. It is of course important to note that the ship with Kalman filtering has to deal with both current and waves, but it is clear that this is no challenge for the filter. But the reason for why the situation from 3.d is slightly better than the one with Kalman filter may be due to the system with a pure PD-controller is faster. Tuning the PD-controller (as previously discussed), could improve the response time, which would result in the ship turning into correct course a little earlier. This is a minor trade-off compared with no current compensation or wave-filtering. We believe that the result is sufficient as is, and that Kalman-filtering has been a great success for improving the ship's performance.

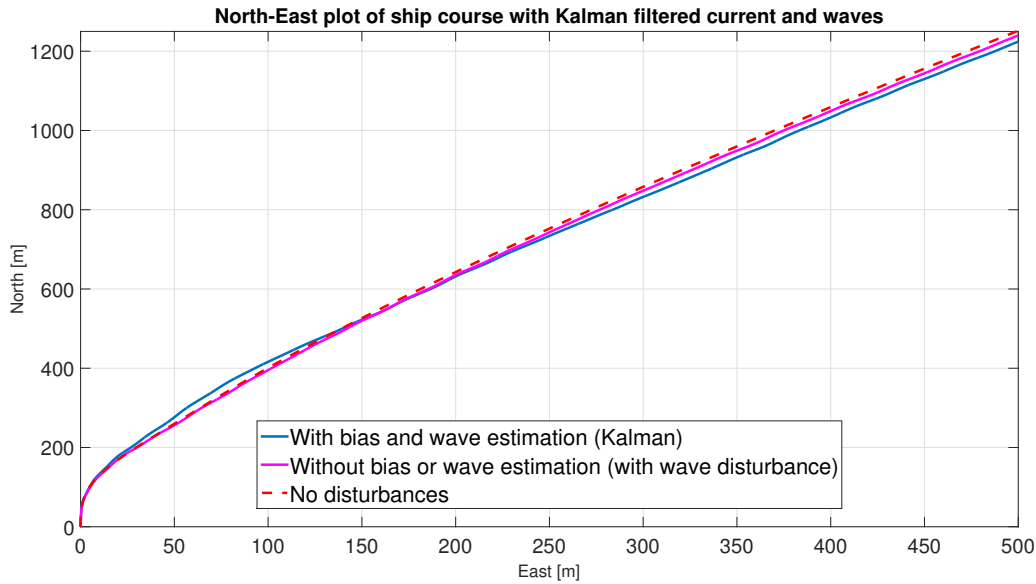


Figure 18: Ship path, both with and without the bias and wave estimation

6 Conclusion

In this assignment it was observed how waves and currents influence the behavior of a ship. It was observed how a simple PD controller with feedback is able to control the ship without disturbances, but it is not good enough to deal with waves and currents.

It was observed how the current created a steady state error in the compass course, compared to the reference. The disturbances from the waves resulted in rapid change in the rudder input angle, which again leads to faster wear and tear.

After confirming that the system was observable in all the considered situations, these problems were solved by using a Kalman filter. A feed forward loop canceled the current offset, and a filtered feedback smoothed the rudder operation. The Kalman filter filtered out the high frequency component of the wave disturbance, as well as measurement noise, which resulted in smoother operation of the rudder.

References

- [1] Chi-Tsong Chen. *Linear System Theory and Design*. Oxford University Press, international fourth edition, 2013.
- [2] Kristoffer Gryte. *TTK4115 - Discrete Kalman Filter Applied to a Ship Autopilot*, volume 1.8. Department of Engineering Cybernetics NTNU, October 2016.
- [3] MathWorks. pwelch. <https://se.mathworks.com/help/signal/ref/pwelch.html>, Accessed November 9, 2016.
- [4] Patrick Y.C. Hwang Robert Grover Brown. *Introduction Random Signals and Applied Kalman Filtering*. John Wiley & Sons, Inc., fourth edition edition, 2012.

A Appendix A - MATLAB codes

A.1 MATLAB - Part I

```

1  %% Init
2  close all
3  clc
4  clear variables
5
6  figNum = 1;           % Figure number-counter
7  w_1 = 0.005;         % For use in 5.1.b and 5.1.c
8  w_2 = 0.05;          % For use in 5.1.b and 5.1.c
9
10 % \\ Load previously simulated data from .mat-files
11 addpath('Data-files')
12 load('omega_1.mat')   % 5.1.b - omega_1 is system output
13 load('omega_2.mat')   % 5.1.b - omega_2 is system output
14 load('omega_1_E.mat') % 5.1.c - omega_1_E is system output
15 load('omega_2_E.mat') % 5.1.c - omega_2_E is system output
16 %
17
18 %% TASK 5.1.b — Without disturbance
19 H_1 = omega_1;
20 H_2 = omega_2;
21 t = length(H_1);
22
23 % \\ Plot system output without waves or noise
24 figure(figNum); figNum = figNum + 1;
25 subplot(2,1,1)
26 plot(H_1(:,1),H_1(:,2), 'LineWidth', 3)
27 xlabel('t [s]', 'FontSize', 18); ylabel('$\sin(\omega_1 t)$ [deg]', ...
28     'FontSize', 18, 'Interpreter', 'latex')
29 title('Sinus-input where \omega_1 = 0.005 for model without noise', ...
30     'FontSize', 24);
31 set(gca,'FontSize',14); grid on;
32
33 subplot(2,1,2)
34 plot(H_2(:,1),H_2(:,2),'r', 'LineWidth', 3)
35 xlabel('t [s]', 'FontSize', 18); ylabel('$\sin(\omega_2 t)$ [deg]', ...
36     'Interpreter', 'latex', 'FontSize', 18);
37 title('Sinus-input where \omega_2 = 0.05 for model without noise', ...
38     'FontSize', 24)
39 set(gca,'FontSize',14); grid on;
40
41 % \\ Find amplitude peaks
42 A_11 = max(H_1(3500:t,2));
43 A_12 = min(H_1(3500:t,2));
44 A_21 = max(H_2(3500:t,2));
45 A_22 = min(H_2(3500:t,2));
46

```

```

47 % \\ Average amplitude value of output
48 A_1 = (A_11-A_12)/2;
49 A_2 = (A_21-A_22)/2;
50
51 % \\ Finding K and T from system output
52 K = sqrt((A_1^2*w_1^2- ((A_1^2*A_2^2*w_1^4*w_2^2)/(w_2^4*A_2^2)))/...
53     /(1-(A_1^2*w_1^4)/(w_2^4*A_2^2)));
54 T = (sqrt(K^2 - A_2^2*w_2^2))/(w_2^2*A_2);
55 %
56
57 %% TASK 5.1.c — with disturbance
58 H_1_E = omega_1_E;
59 H_2_E = omega_2_E;
60 t_E = length(H_1_E);
61
62 % \\ Plot system output with waves and noise
63 figure(figNum)
64 figNum = figNum + 1;
65 subplot(2,1,1)
66 plot(H_1_E(:,1),H_1_E(:,2), 'LineWidth', 3)
67 xlabel('t [s]', 'FontSize', 18); ylabel('Sin($\omega_{1}$t) [deg]', ...
68     'Interpreter', 'latex', 'FontSize', 18)
69 title('Sin-input where \omega_1 = 0.005 for model with waves and noise',...
70     'FontSize', 24)
71 set(gca,'FontSize',14); grid on;
72
73 subplot(2,1,2)
74 plot(H_2_E(:,1),H_2_E(:,2),'r', 'LineWidth', 1)
75 xlabel('t [s]', 'FontSize', 18); ylabel('Sin($\omega_{2}$t) [deg]', ...
76     'Interpreter', 'latex', 'FontSize', 18)
77 title('Sinus where \omega_2 = 0.05 for model with waves and noise', ...
78     'FontSize', 24)
79 set(gca,'FontSize',14), grid on;
80
81 % \\ Find amplitude peaks
82 A_11_E = max(H_1_E(3500:t_E,2));
83 A_12_E = min(H_1_E(3500:t_E,2));
84 A_21_E = max(H_2_E(3500:t_E,2));
85 A_22_E = min(H_2_E(3500:t_E,2));
86
87 % \\ Average amplitude for w_1
88 A_1_E = (A_11_E - A_12_E)/2;
89
90 % \\ Attempt to find average value of amplitude for w_2
91 s_w_2 = 0;
92 for i=4000:8000
93     A_max = max(H_2_E(i-100:i+100,2));
94     A_min = min(H_2_E(i-100:i+100,2));
95     s_w_2 = s_w_2 + (A_max-A_min)/2;
96 end
97

```

```

98 A_2_E = s_w_2/4000;
99
100 % \\ Finding K and T from system output
101 K_E = sqrt((A_1_E^2*w_1^2- ((A_1_E^2*A_2_E^2*w_1^4*w_2^2)/...
102     (w_2^4*A_2_E^2)))/(1-(A_1_E^2*w_1^4)/(w_2^4*A_2_E^2)));
103 T_E = (sqrt(K_E^2 - A_2_E^2*w_2^2))/(w_2^2*A_2_E);
104 %
105
106 %% TASK 5.1.d — Step response
107 load('step_simulink.mat')
108
109 % \\ Define transfer function
110 H_tf = tf(K, [T 1 0]);
111
112 % \\ Plot step response of ship and model
113 figure(figNum);
114 figNum = figNum + 1;
115 H_tf_sim = step_simulink;
116 step(H_tf,length(H_tf_sim)/2)
117 hold on;
118 plot(H_tf_sim(:,1),H_tf_sim(:,2),'r', 'LineWidth', 4)
119 title('Step response of ship and model', 'FontSize', 24)
120 legend({'$$Step_{model}$$', '$$Step_{ship}$$'}, 'FontSize', 36, ...
121     'Interpreter', 'latex');
122 grid on; hold off;
123 xlabel('Time ', 'FontSize', 24); ylabel('Amplitude [deg]', 'FontSize', 24);
124 set(gca, 'FontSize', 14)
125 %

```

A.2 MATLAB - Part II

```

1 %% Init
2 close all
3 clc
4 clear variables
5
6 figNum = 1;           % Figure number-counter
7 load('wave.mat');     % Load wave disturbance
8 %
9
10 %% TASK 5.2.a — Estimate PSD
11 F_s = 10;
12 window = 4096;
13 noverlap = [];
14 nfft = [];
15
16 % \\ Find estimated PSD. Input converted from deg to rad
17 [pxx,f] = pwelch(psi_w(2,:).*(pi/180),window,noverlap,nfft,F_s);
18

```

```

19 % \\ f (Hz) to w (rad/s), and psd (power/pr. HZ) to psd (power s/rad)
20 omega = 2*pi.*f;
21 pxx = pxx./(2*pi);
22 %
23
24 %% TASK 5.2.c — Find omega_0
25 % Plot estimated PSD
26 figure(figNum)
27 figNum = figNum+1;
28 plot(omega,pxx, 'LineWidth', 4)
29 axis([0 2 -0.00005 16*10^(-4)])
30 hold on
31 xlabel('$\omega$ [$\frac{\text{rad}}{\text{s}}$]', 'FontSize', 20, ...
32     'Interpreter', 'latex')
33 ylabel('$S_{\psi_w}(\omega)$ [rad]', 'FontSize', 20, ...
34     'Interpreter', 'latex')
35 title(['Estimated power spectral density $S_{\psi_w}(\omega)$ of '...
36     'wave disturbance'], 'FontSize', 20, 'Interpreter', 'latex')
37 grid on;
38
39 ax = gca; ax.XTick = [0:pi/8:2];
40 ax.XTickLabel = {'$0$', '$\frac{\pi}{8}$', '$\frac{\pi}{4}$', ...
41     '$\frac{3\pi}{8}$', '$\frac{\pi}{2}$', '$\frac{5\pi}{8}$', ...
42     '$\frac{3\pi}{4}$'};
43 ax.TickLabelInterpreter = 'latex';
44 ax.FontSize = 24;
45
46 % Find resonance frequency from estimated PSD
47 [maxPSD, indexAtMaxPSD] = max(pxx);
48 omega_0 = omega(indexAtMaxPSD);
49
50 % Plot arrow to resonance frequency
51 plot(omega_0, maxPSD, 'ro', 'MarkerSize', 10);
52 a = annotation('textarrow', 2.15*[0.23 0.203], [0.79 0.866], 'String', ...
53     '($\omega_0, S_{\psi_w}(\omega_0)$)', 'Interpreter', 'latex');
54 a.Color = 'red';
55 a.FontSize = 36;
56 hold off
57 %
58
59 %% TASK 5.2.d — Finding lambda
60 % Plot estimated PSD
61 figure(figNum)
62 figNum = figNum+1;
63 plot(omega,pxx, 'LineWidth', 3)
64 axis([0 2 -0.00005 16*10^(-4)])
65 hold on
66 xlabel('$\omega$ [$\frac{\text{rad}}{\text{s}}$]', 'FontSize', 20, ...
67     'Interpreter', 'latex')
68 ylabel('$S_{\psi_w}(\omega)$, $P_{\psi_w}(\omega)$ [rad]', ...
69     'FontSize', 20, 'Interpreter', 'latex')

```

```

70 title(['Estimated  $S_{\psi(w)}(\omega)$  vs. analytical' ...
71       '$P_{\psi(w)}(\omega)$ for different values of  $\lambda$ '], ...
72       'FontSize', 20, 'Interpreter', 'latex')
73 grid on
74
75 ax = gca; ax.XTick = 0:pi/8:2;
76 ax.XTickLabel = {'$0$', '$\frac{\pi}{8}$', '$\frac{\pi}{4}$', ...
77                 '$\frac{3\pi}{8}$', '$\frac{\pi}{2}$', '$\frac{5\pi}{8}$', ...
78                 '$\frac{3\pi}{4}$'};
79 ax.TickLabelInterpreter = 'latex';
80 ax.FontSize = 24;
81
82 % Find analytical PSD for several lambdas
83 sigma = sqrt(maxPSD);
84 for lambda=0.04:0.04:0.16
85
86     K_w = 2*lambda*omega_0*sigma;
87     pxx_a = (omega.*K_w).^2./(omega.^4 + omega_0^4 + ...
88                2*omega_0^2*omega.^2*(2*lambda^2-1));
89     plot(omega, pxx_a, '—', 'LineWidth', 2);
90 end
91 legend({'$S_{\psi(w)}(\omega)$', '$\lambda = 0.04$', ...
92        '$\lambda = 0.08$', '$\lambda = 1.2$', '$\lambda = 1.6$'}, ...
93        'Interpreter', 'latex', 'FontSize', 24, 'Location', 'northwest');
94 hold off
95
96 % More accurate lambda plotting (not included in report, same result)
97 figure(figNum)
98 figNum = figNum+1;
99 plot(omega, pxx, 'LineWidth', 3)
100 axis([0 2 -0.00005 16*10^(-4)])
101 hold on
102 xlabel('$\omega$ [$\frac{\text{rad}}{\text{s}}$]', 'FontSize', 20, ...
103        'Interpreter', 'latex')
104 ylabel('$S_{\psi(w)}(\omega)$, $P_{\psi(w)}(\omega)$ [rad]', ...
105        'FontSize', 20, 'Interpreter', 'latex')
106 title(['Estimated  $S_{\psi(w)}(\omega)$  vs. analytical '...
107       '$P_{\psi(w)}(\omega)$'], 'FontSize', 20, 'Interpreter', 'latex')
108 grid on
109
110 ax = gca; ax.XTick = [0:pi/8:2];
111 ax.XTickLabel = {'$0$', '$\frac{\pi}{8}$', '$\frac{\pi}{4}$', ...
112                 '$\frac{3\pi}{8}$', '$\frac{\pi}{2}$', '$\frac{5\pi}{8}$', ...
113                 '$\frac{3\pi}{4}$'};
114 ax.TickLabelInterpreter = 'latex';
115 ax.FontSize = 24;
116
117 for lambda=0.07:0.01:0.10
118     K_w = 2*lambda*omega_0*sigma;
119     pxx_a = (omega.*K_w).^2./(omega.^4 + omega_0^4 + ...
120                2*omega_0^2*omega.^2*(2*lambda^2-1));

```

```

121 plot(omega, pxx_a, '—', 'LineWidth', 2);
122 end
123 legend({'$S_{\psi_w}(\omega)$', '$\lambda = 0.07$', ...
124         '$\lambda = 0.08$', '$\lambda = 0.9$', '$\lambda = 1.0$'}, ...
125         'Interpreter', 'latex', 'FontSize', 24, 'Location', 'northwest');
126 hold off
127
128 %choose lambda to be 0.08
129 lambda = 0.08;
130 %

```

A.3 MATLAB - Part III

```

1 %% Init
2 close all
3 clc
4 clear variables
5
6 figNum = 1;    % Figure number-counter
7 PSI_r = 30;    % Reference angle for simulation
8 sim_t = 500;    % Simulation time
9
10 % \\ Simulink models
11 addpath('Simulink models tasks')
12 %
13
14 %% TASK 5.3.a — Designing a PD-controller
15 % \\ Setting up transfer variable s and constants
16 s = tf('s');
17 K = 0.1742;
18 T = 86.5256;
19 T_d = T;
20 w_c = 0.1;
21
22 % \\ Defining the transfer function
23 T_f = -1/(tan(130*pi/180)*w_c);
24 K_pd = sqrt(w_c^2+T_f^2*w_c^4)/K;
25 H_0 = (K*K_pd)/(s*(1+T_f*s));
26
27 % \\ Bode margin plot of transfer function
28 figure(figNum)
29 figNum = figNum + 1;
30 margin(H_0); grid on;
31 %
32
33 %% TASK 5.3.b — Simulating without disturbances
34 load_system('task5_3_b.slx')
35 sim('task5_3_b.slx')
36

```



```

37 % \\ Plot of autopilot wihtout current and waves
38 figure(figNum)
39 figNum = figNum+1;
40 plot(t,sim_PSI_r, 'r—',t,sim_compass, t, delta, 'LineWidth',3);
41 title('Autopilot without current or waves', 'FontSize', 24);
42 xlabel('t [s]', 'FontSize', 20); grid on;
43 ylabel('$\mathbf{\psi_r}(t), \ \psi(t), \ \delta$ [deg]', ...
44     'FontSize', 20, 'Interpreter', 'latex');
45 legend({'$\psi_r$','$\psi$', '$\delta$'}, 'Location', ...
46     'best', 'FontSize', 36, 'Interpreter', 'latex')
47 ax = gca; ax.FontSize = 24;
48
49 % \\ Save delta and north-east data for later plotting
50 delta1 = delta; t1 = t; psi1 = sim_compass;
51 x1 = north_east(:,1); y1 = north_east(:,2);
52 %
53
54 %% TASK 5.3.c — Simulating with current, without waves
55 load_system('task5_3_c.slx')
56 sim('task5_3_c.slx')
57
58 % \\ Plot of autopilot with current, but without waves
59 figure(figNum)
60 figNum = figNum + 1;
61 plot(t,sim_PSI_r, 'r—',t,sim_compass, t, delta, 'LineWidth',3);
62 xlabel('t [s]', 'FontSize', 20); grid on;
63 ylabel('$\mathbf{\psi_r}(t), \ \psi(t), \ \delta$ [deg]', ...
64     'FontSize', 20, 'Interpreter', 'latex');
65 title('Autopilot with current, but without waves', 'FontSize', 24);
66 legend({'$\psi_r$','$\psi$', '$\delta$'}, 'Location', ...
67     'best', 'FontSize', 36, 'Interpreter', 'latex')
68 ax = gca; ax.FontSize = 24;
69
70 % \\ Save delta and north-east data for later plotting
71 delta2 = delta; t2 = t; psi2 = sim_compass;
72 x2 = north_east(:,1); y2 = north_east(:,2);
73 %
74
75 %% TASK 5.3.d — Simulating with waves, without current
76 load_system('task5_3_d.slx')
77 sim('task5_3_d.slx')
78
79 % \\ Plot of autopilot with waves and without current
80 figure(figNum)
81 figNum = figNum + 1;
82 plot(t,sim_PSI_r,'r—',t,sim_compass, t, delta, 'LineWidth',3);
83 xlabel('t [s]', 'FontSize', 20); grid on; ax = gca; ax.FontSize = 24;
84 ylabel('$\mathbf{\psi_r}(t), \ \psi(t), \ \delta$ [deg]', ...
85     'FontSize', 20, 'Interpreter', 'latex');
86 title('Autopilot with waves and without current', 'FontSize', 24);
87 legend({'$\psi_r$','$\psi + \psi_w$', '$\delta$'}, 'Location', ...

```

```

88     'best', 'FontSize', 36, 'Interpreter', 'latex')
89
90 % \\ Save delta and north-east data for later plotting
91 delta3 = delta; t3 = t; psi3 = sim_compass;
92 x3 = north_east(:,1); y3 = north_east(:,2);
93 %
94
95 %% Plotting the different rudder inputs against eachother
96 figure(figNum)
97 figNum = figNum+1;
98
99 % \\ Subplot for delta with no disturbance and current disturbance
100 subplot(2,1,1)
101 plot(t1,delta1,t2,delta2,'LineWidth', 2);
102 title('Rudder input \delta with and without current', 'FontSize', 24);
103 xlabel('t [s]', 'fontSize', 20); grid on;
104 ylabel('\delta [deg]', 'FontSize', 20);
105 legend({'\delta no disturbance', '\delta with current'}, 'Location', ...
106        'best', 'FontSize', 36)
107 axis([0 150 -30 40]); ax = gca; ax.FontSize = 24;
108
109 % \\ Subplot for delta with wave disturbance
110 subplot(2,1,2);
111 plot(t3,delta3,'LineWidth',2); xlabel('t [s]', 'FontSize', 20);
112 title('Rudder input \delta with waves', 'FontSize', 24);
113 ylabel('\delta [deg]', 'FontSize', 20); grid on;
114 legend({'\delta with waves'}, 'Location', 'best', 'FontSize', 36)
115 axis([0 150 -40 40]); ax = gca; ax.FontSize = 24;
116 %
117
118 %% Plotting North-East plot
119 figure(figNum)
120 figNum = figNum+1;
121
122 % \\ Plot North-East plot for previous cases
123 plot(y1,x1,'r—',y2,x2,y3,x3,'m', 'LineWidth', 3);
124 title('North-East plot of ship course', 'FontSize', 24);
125 ylabel('North [m]', 'FontSize', 20); grid on;
126 xlabel('East [m]', 'FontSize', 20);
127 ax = gca; ax.FontSize = 24; axis([0 150 0 600]);
128 legend({'Without disturbances', 'With current', ...
129        'With waves'}, 'Location', 'best', 'FontSize', 36)
130 %

```

A.4 MATLAB - Part IV

```

1 %% Init
2 close all
3 clc

```

```

4 clear variables
5
6 addpath('Data-files')
7 % \\ Constants from previous tasks (without disturbances)
8 load('constants_5.2.mat')
9 load('constants_5.3.mat')
10 %
11
12 %% TASK 5.4.a — Finding A, B, C and E
13 A_bw = [0 1 0 0 0; -omega_0^2 -2*lambda*omega_0 0 0 0; 0 0 0 1 0; ...
14         0 0 0 -1/T -K/T; 0 0 0 0 0];
15 B_bw = [0; 0; 0; K/T; 0];
16 C_bw = [0 1 1 0 0];
17 E_bw = [0 0; K_w 0; 0 0; 0 0; 0 1];
18 %
19
20 %% TASK 5.4.b — Observability without disturbances
21 A = [0 1; 0 -1/T];
22 B=[0; K/T];
23 C= [1 0];
24 O = obsv(A,C);
25
26 % \\ If rank(O) = 2, we have full rank <=> Observability
27 rank_O = rank(O);
28 %
29
30 %% TASK 5.4.c — Observability with current
31 A_b = [0 1 0; 0 -1/T -K/T; 0 0 0];
32 C_b = [1 0 0];
33 O_b = obsv(A_b,C_b);
34
35 % \\ If rank(O_b) = 3, we have full rank <=> Observability
36 rank_O_b = rank(O_b);
37 %
38
39 %% TASK 5.4.d — Observability with wave
40 A_w = [0 1 0 0; -omega_0^2 -2*lambda*omega_0 0 0; 0 0 0 1; 0 0 0 -1/T];
41 C_w = [0 1 1 0];
42 O_w = obsv(A_w,C_w);
43
44 % \\ If rank(O_w) = 4, we have full rank <=> Observability
45 rank_O_w = rank(O_w);
46 %
47
48 %% TASK 5.4.e — Observability with current and wave
49 O_bw = obsv(A_bw,C_bw);
50
51 % \\ If rank(O_bw) = 5, we have full rank <=> Observability
52 rank_O_bw = rank(O_bw);
53 %
54

```



```

33 A = [0 1 0 0 0; -omega_0^2 -2*lambda*omega_0 0 0 0; 0 0 0 1 0; ...
34       0 0 0 -1/T -K/T; 0 0 0 0 0];
35 B = [0; 0; 0; K/T; 0];
36 C = [0 1 1 0 0];
37 D = 0;
38 E = [0 0; K_w 0; 0 0; 0 0; 0 1];
39
40 % \\ Sample frequency
41 F_s = 10;
42 T_s = 1/F_s;
43
44 % \\ Exact discretization
45 [Ad, Bd] = c2d(A,B,T_s); Cd = C;
46 [Ad, Ed] = c2d(A,E,T_s); Dd = D;
47 %
48
49 %% Task 5.5.b — Estimate of var(v)
50 load_system('task5_5_b.slx')
51 sim('task5_5_b.slx')
52
53 % \\ R is the Measurment noise variance
54 R = var(sim_compass*pi/180);
55 %
56
57 %% Task 5.5.c — Discrete Kalman Filter
58 % \\ Q is the Process noise covariance
59 Q = [30 0; 0 10^(-6)];
60 P_0 = [1 0 0 0 0; 0 0.013 0 0 0; 0 0 pi^2 0 0; 0 0 0 1 0; ...
61        0 0 0 0 2.5*10^-4];
62 X_0 = [0; 0; 0; 0; 0];
63 R = R/T_s;
64 I = diag([1 1 1 1 1]);
65
66 % \\ Put data in a struct for use in the Kalman filter
67 data = struct('Ad',Ad,'Bd',Bd,'Cd',Cd,'Ed', Ed, 'Q',Q,'R', R,'P_0',P_0, ...
68              'X_0',X_0, 'I', I);
69 %
70
71 %% Task 5.5.d — Feed forward estimated bias
72 load_system('task5_5_d.slx')
73 sim('task5_5_d.slx')
74
75 % \\ Plot measured compass with and without estimated bias
76 figure(figNum)
77 figNum = figNum+1;
78 subplot(2,1,1)
79 plot(t,sim_PSI_r, 'r—', t, sim_compass, t2, psi2, 'LineWidth', 3)
80 title(['Compass reference $\psi_{r} = 30$, and measured compass course' ...
81        '$\psi + v$'], ...
82        'FontSize', 24, 'Interpreter', 'latex');
83 xlabel('t [s]', 'FontSize', 20); ylabel('Angle [deg]', 'FontSize', 20);

```

```

84 legend({'$\psi_r$', '$\psi + v$ (with $\hat{b}_{\text{feedforward}}$)', ...
85         '$\psi + v$ (without $\hat{b}_{\text{feedforward}}$)', 'FontSize', 24, ...
86         'Interpreter', 'latex', 'Location', 'SouthEast')
87 ax = gca; ax.FontSize = 24; grid on ;
88
89 subplot(2,1,2)
90 plot(t, b_filtered, 'm', t, delta, t2, delta2, 'LineWidth', 3)
91 title('Kalman-estimated bias $\hat{b}$, and rudder input $\delta$', ...
92       'FontSize', 24, 'Interpreter', 'latex');
93 xlabel('t [s]', 'FontSize', 20); ylabel('Angle [deg]', 'FontSize', 20);
94 legend({'$\hat{b}$', '$\delta$ (with $\hat{b}_{\text{feedforward}}$)', ...
95         '$\delta$ (without $\hat{b}_{\text{feedforward}}$)', 'FontSize', 24, ...
96         'Interpreter', 'latex', 'Location', 'SouthEast')
97 ax = gca; ax.FontSize = 24; grid on;
98
99 % \\ Plotting North-East plot for this task and part 5.3
100 figure(figNum)
101 figNum = figNum+1;
102 plot(north_east(:,2), north_east(:,1), y2, x2, 'm', y1, x1, 'r—', ...
103       'LineWidth', 3);
104 title(['North-East plot of ship course with Kalman filtered current, ...
105       'no waves'], 'FontSize', 24);
106 ylabel('North [m]', 'FontSize', 20); grid on;
107 xlabel('East [m]', 'FontSize', 20);
108 legend({'With bias estimation (Kalman)', ...
109         'Without bias estimation (with current disturbance)', ...
110         'No disturbances'}, 'FontSize', 26, 'Location', 'best');
111 ax = gca; ax.FontSize = 24; axis([0 500 0 1250]);
112 %
113
114 %% Task 5.5.e — Feed forward estimated bias and wave filtered psi
115 load_system('task5_5_e.slx')
116 sim('task5_5_e.slx')
117
118 % \\ Plot measured compass and estimated compass
119 figure(figNum)
120 figNum = figNum+1;
121 plot(t, sim_PSI_r, 'r—', t, sim_compass, t, psi_filtered, 'LineWidth', 3)
122 title(['Compass reference $\psi_r = 30^\circ$, measured compass course ...
123       '$\psi + \psi_w + v$ and estimated course $\hat{\psi}$'], ...
124       'FontSize', 24, 'Interpreter', 'latex');
125 xlabel('t [s]', 'FontSize', 20); ylabel('Angle [deg]', 'FontSize', 20);
126 legend({'$\psi_r$', 'measured compass course ($\psi + \psi_w + v$)' ...
127         ', 'estimated compass course ($\hat{\psi}$)', 'FontSize', 24, ...
128         'Interpreter', 'latex', 'Location', 'best')
129 ax = gca; ax.FontSize = 24; grid on; axis([0 500 -5 40])
130
131 % \\ Plot measured psi with and without Kalman filtered bias and waves
132 figure(figNum)
133 figNum = figNum+1;
134 plot(t, sim_PSI_r, 'r—', t, sim_compass, t3, psi3, 'LineWidth', 3)

```

```

135 title('Compass reference  $\psi_r = 30^\circ$ , and measured compass course', ...
136       'FontSize', 24, 'Interpreter', 'latex');
137 xlabel('t [s]', 'FontSize', 20); ylabel('Angle [deg]', 'FontSize', 20);
138 legend({' $\psi_r$ ', 'With Kalman filtered feedback', ...
139       'Without Kalman filtered feedback'}, 'FontSize', 24, ...
140       'Interpreter', 'latex', 'Location', 'best')
141 ax = gca; ax.FontSize = 24; grid on; axis([0 500 -5 40])
142
143 % \\ Plotting rudder input delta
144 figure(figNum)
145 figNum = figNum+1;
146 subplot(2,1,1)
147 plot(t, b_filtered, 'm', t, delta, 'LineWidth', 3)
148 title('Rudder input  $\delta$  and estimated bias  $\hat{b}$ ', ...
149       'FontSize', 24, 'Interpreter', 'latex');
150 xlabel('t [s]', 'FontSize', 20); ylabel('Angle [deg]', 'FontSize', 20);
151 legend({' $\hat{b}$ ', ' $\delta$ '}, 'FontSize', 24, ...
152       'Interpreter', 'latex', 'Location', 'best')
153 ax = gca; ax.FontSize = 24; grid on;
154
155 subplot(2,1,2)
156 plot(t3, delta3, 'LineWidth', 3)
157 title(['Rudder input  $\delta$  without Kalman filtered  $\psi$  '...
158       'or  $\hat{b}$ '], 'FontSize', 24, 'Interpreter', 'latex');
159 xlabel('t [s]', 'FontSize', 20); ylabel('Angle [deg]', 'FontSize', 20);
160 legend({' $\delta$ '}, ...
161       'FontSize', 24, 'Interpreter', 'latex', 'Location', 'best')
162 ax = gca; ax.FontSize = 24; grid on;
163
164 % \\ Plotting North-East plot compared to 5.3.d
165 figure(figNum)
166 figNum = figNum+1;
167 plot(north_east(:,2), north_east(:,1), y3, x3, 'm', y1, x1, 'r—', ...
168       'LineWidth', 3);
169 title(['North-East plot of ship course with Kalman filtered current '...
170       'and waves'], 'FontSize', 24);
171 ylabel('North [m]', 'FontSize', 20); grid on;
172 xlabel('East [m]', 'FontSize', 20);
173 legend({'With bias and wave estimation (Kalman)', ...
174       'Without bias or wave estimation (with wave disturbance)', ...
175       'No disturbances'}, 'FontSize', 26, 'Location', 'best');
176 ax = gca; ax.FontSize = 24; axis([0 500 0 1250]);
177
178
179 % \\ Wave influence (current turned off, delta = 0)
180 load_system('task5_5_e_2.slx')
181 sim('task5_5_e_2.slx')
182
183 % \\ Plotting wave influence on system
184 figure(figNum)
185 figNum = figNum+1;

```

```

186 plot(t, sim_compass, t, psi_filtered, 'LineWidth', 3)
187 title(['Measured wave influence $\psi_{w}$ vs. estimated wave influence'...
188       '$\hat{\psi}_{w}$'], 'FontSize', 24, 'Interpreter', 'latex');
189 xlabel('t [s]', 'FontSize', 20); ylabel('Angle [deg]', 'FontSize', 20);
190 legend({'$\psi_{w}$', '$\hat{\psi}_{w}$'}, 'FontSize', 24, ...
191        'Interpreter', 'latex', 'Location', 'best')
192 ax = gca; ax.FontSize = 24; grid on;
193 axis([0 500 -4 4]);
194 %

```

A.6 Kalman MATLAB function

```

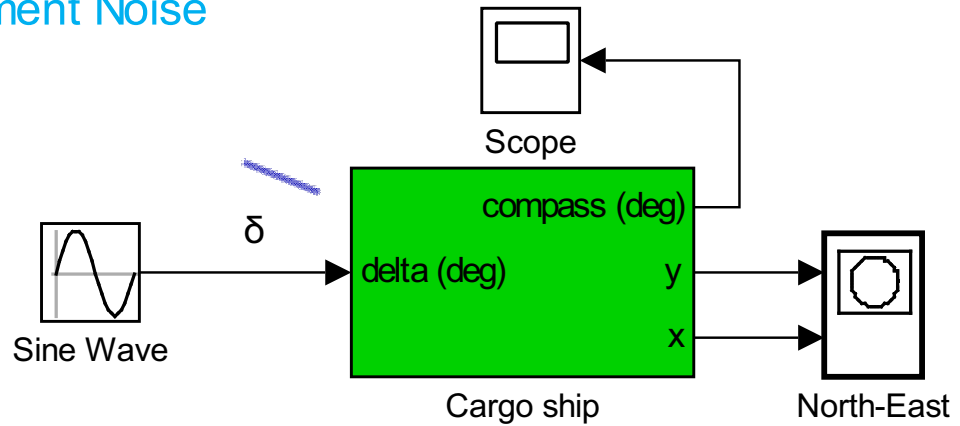
1 function [b,psi] = Kalman_matlab_fnc(u, y, data)
2 %#codegen
3 persistent init_flag A B C E Q R P_ x_ I
4
5 if (isempty(init_flag))
6     init_flag = 1;
7
8     % Initialization for system
9     [A,B,C,E,Q,R,P_,x_, I] = deal(data.Ad,data.Bd,data.Cd,data.Ed, ...
10      data.Q, data.R, data.P_0, data.X_0, data.I);
11 end
12
13 % 1 — Compute the Kalman Gain
14 L = (P_*C')/((C*P_*C'+R));
15 % 2 — Update estimate with measurment
16 x = x_ + L*(y-C*x_);
17 % 3 — Update error covariance matrix
18 P = (I - L*C)*P_*(I-L*C)' + L*R*L';
19 % 4 — Projeet ahead
20 x_ = A*x + B*u;
21 P_ = A*P*A' + E*Q*E';
22
23 psi = x(3); b = x(5);
24
25 end

```


B Appendix B - SIMULINK models

B.1 SIMULINK Task5.1.b

☐ Measurment Noise
☐ Current
☐ Wave

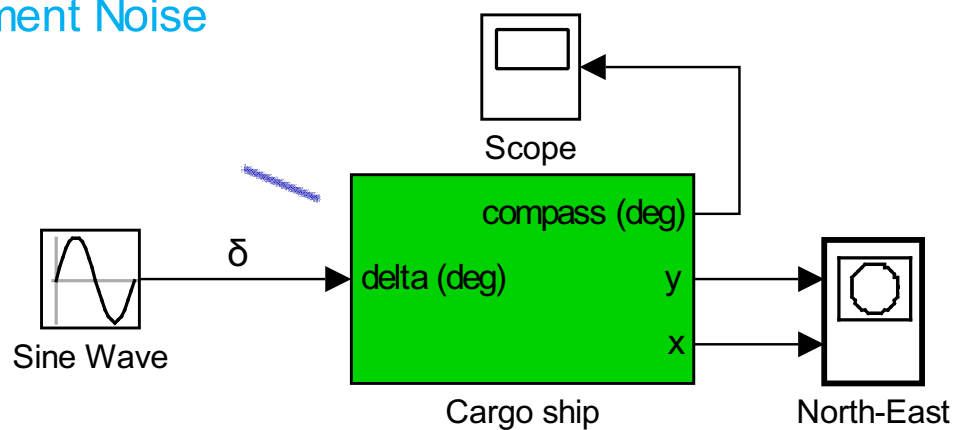


TASK 5.1.b

Figure 19: SIMULINK for section 1.b

B.2 SIMULINK Task5.1.c

☒ Measurment Noise
☐ Current
☒ Wave

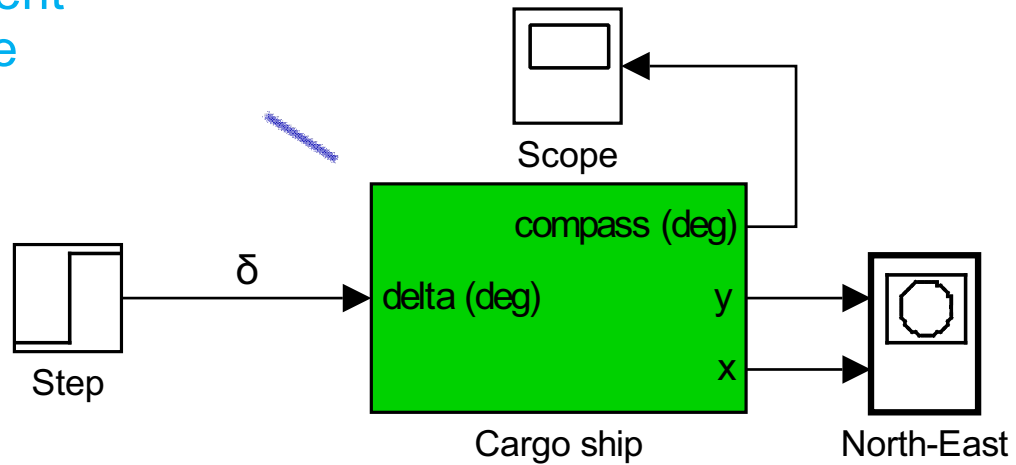


TASK 5.1.c

Figure 20: SIMULINK for section 1.c

B.3 SIMULINK Task5.1.d

[X] Measurment Noise
 [] Current
 [X] Wave



TASK 5.1.d

Figure 21: SIMULINK for section 1.d

B.4 SIMULINK Task5.3.b

TASK 5.3.b

[X] Measurment Noise
 [] Current
 [] Wave

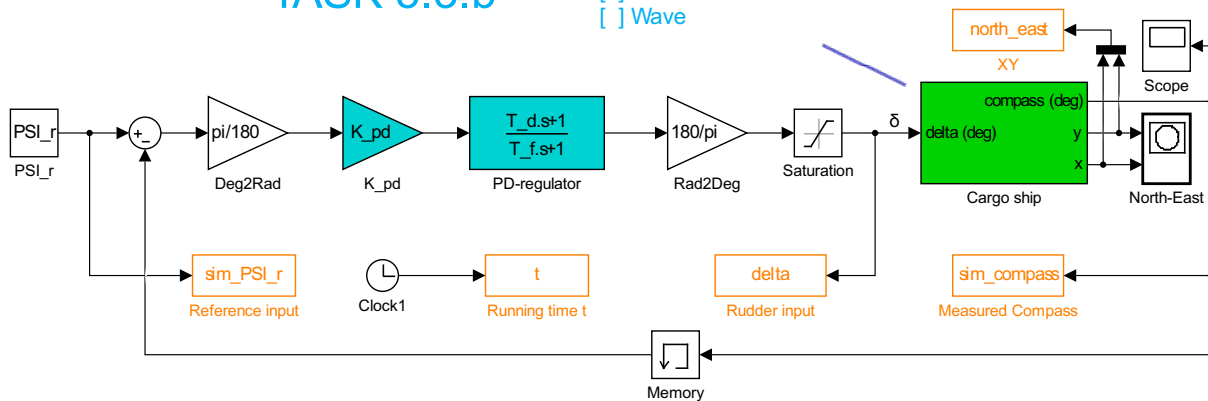


Figure 22: SIMULINK for section 3.b

B.5 SIMULINK Task5.3.c

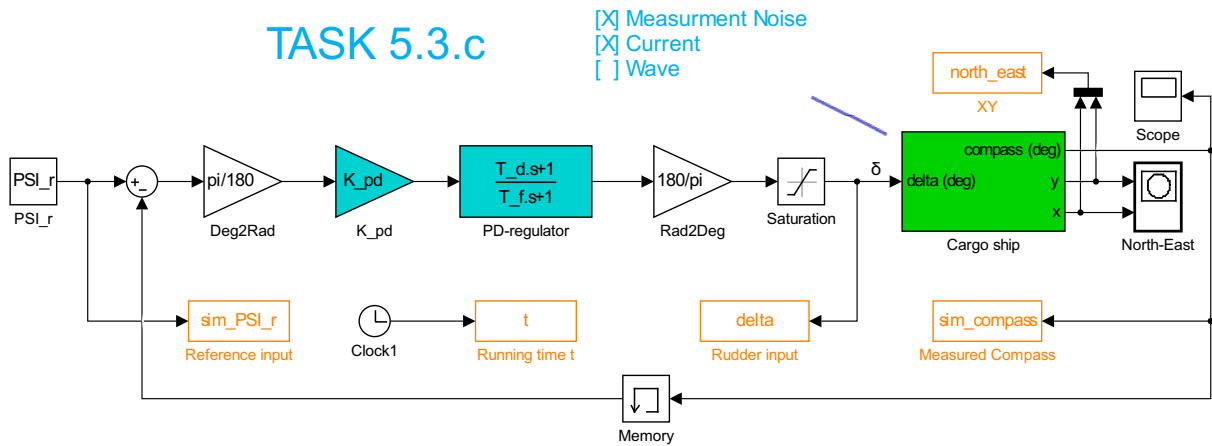


Figure 23: SIMULINK for section 3.c

B.6 SIMULINK Task5.3.d

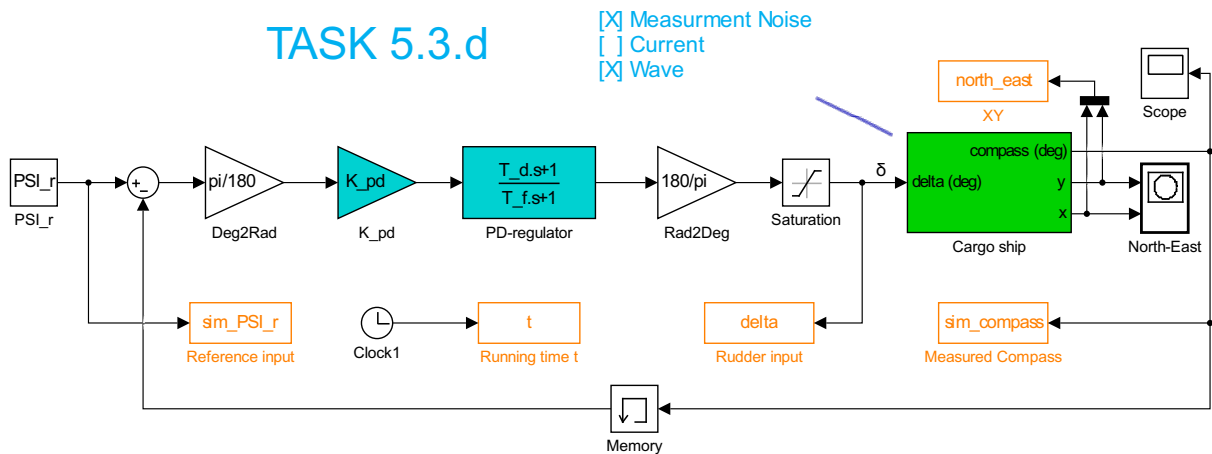


Figure 24: SIMULINK for section 3.d

B.7 SIMULINK Task5.5.b

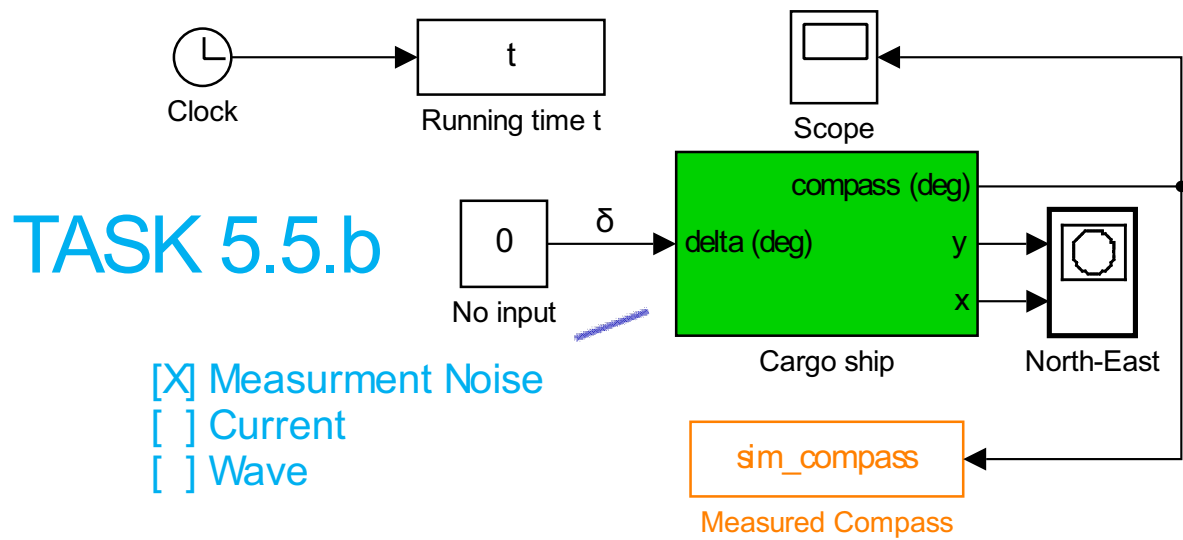


Figure 25: SIMULINK for section 5.b

B.8 SIMULINK Task5.5.d

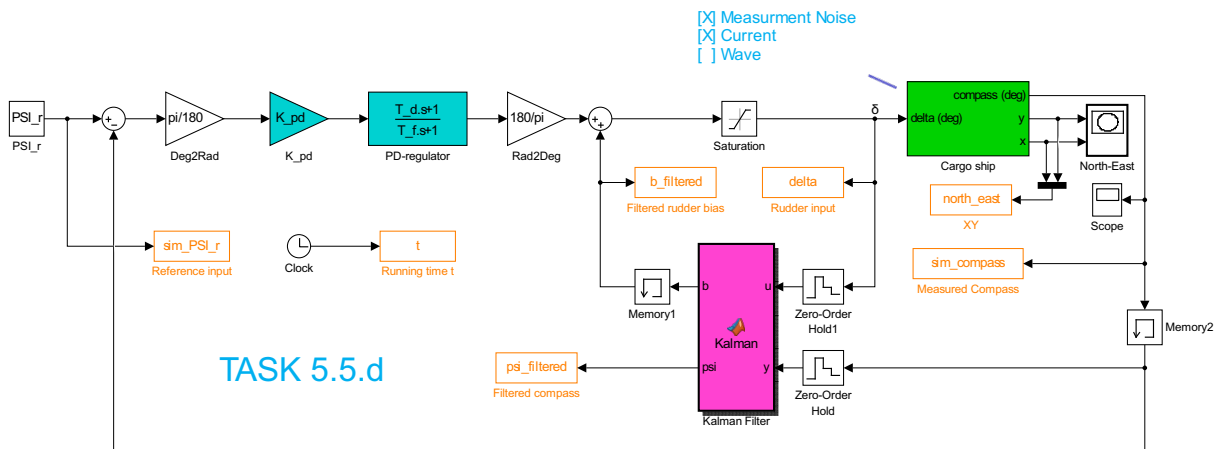


Figure 26: SIMULINK for section 5.d

B.9 SIMULINK Task5.5.e

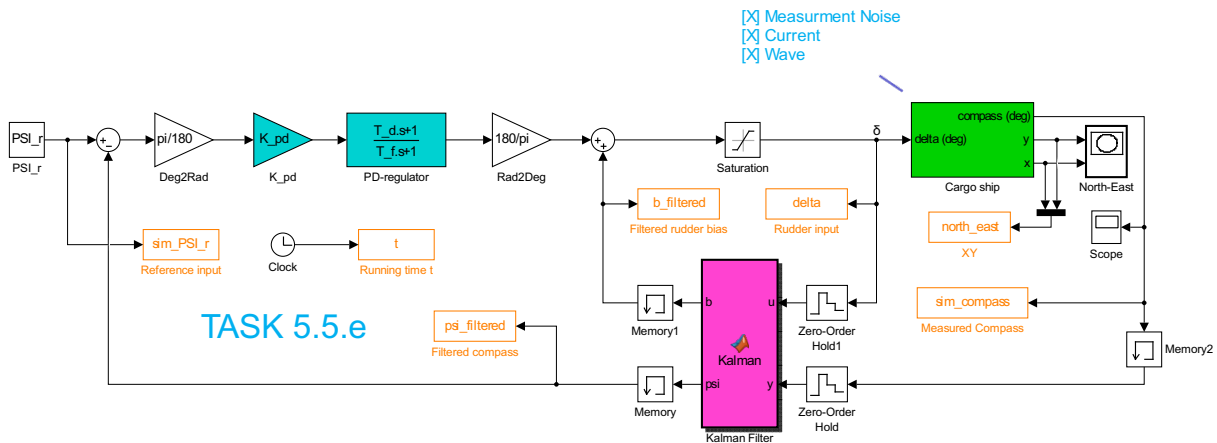


Figure 27: SIMULINK for section 5.e

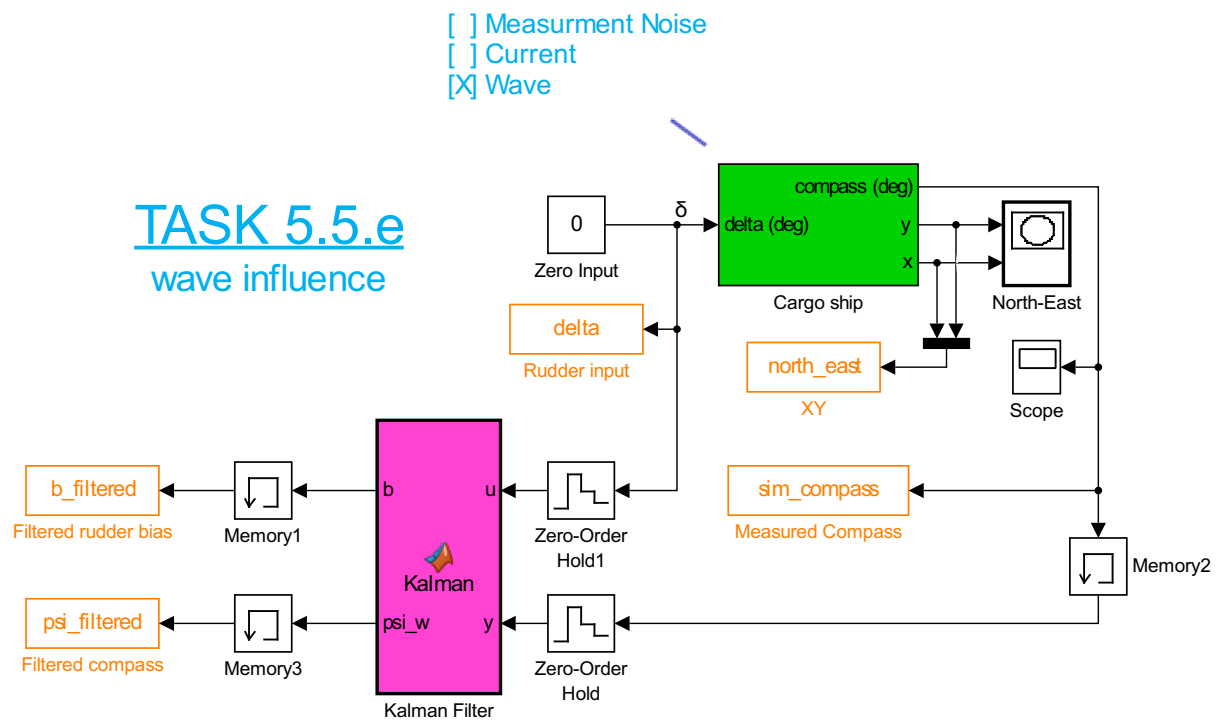


Figure 28: SIMULINK for section 5.e to find the wave influence