CH 4

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4.1 Fourier Series Representation of Continuoustime Periodic Signals

Trigonometric Fourier Series

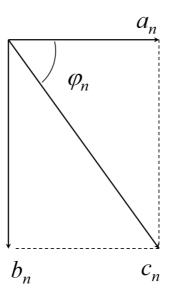
Periodic Signal x(t), with period T_1

$$x(t) = a_0 + \sum_{n=1}^{\infty} a_n \cos\left(n\omega_1 t
ight) + b_n \sin\left(n\omega_1 t
ight)$$

- Constant Component: $a_0=\frac{1}{T_1}\int_{t_0}^{t_0+T_1}x(t)\mathrm{d}t$ Cosine Coefficients: $a_n=\frac{2}{T_1}\int_{t_0}^{t_0+T_1}x(t)\cos{(n\omega_1 t)}\mathrm{d}t$ Sine Coefficients: $b_n=\frac{2}{T_1}\int_{t_0}^{t_0+T_1}x(t)\sin{(n\omega_1 t)}\mathrm{d}t$

An Alternative Form for the Fourier Series

$$x(t)=c_0+\sum_{n=0}^\infty c_n\cos\left(n\omega_1 t+\phi_n
ight)$$
 $c_0=a_0$ $c_n=\sqrt{a_n^2+b_n^2}$ $\phi_n=rctan(-rac{b_n}{a_n})$ $a_n=c_n\cos(\phi)$ $b_n=-c_n\sin(\phi_n)$



• x(t) is real and even

$$b_n=0$$
 $c_n=a_n$ $\phi_n=egin{cases} 0 & a_n>0 \ \pm\pi & a_n<0 \end{cases}$

• x(t) is real and odd

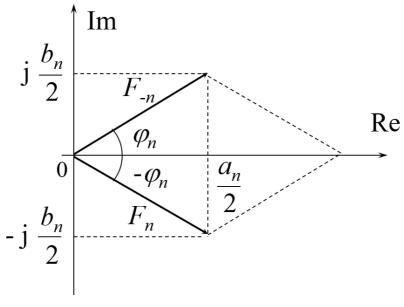
$$a_n=0 \qquad \phi_n=\pm rac{\pi}{2}$$

Complex Exponentially Fourier Series

$$egin{align} x(t) &= a_0 + \sum_{n=1}^\infty a_n \cos(n\omega_1 t) + b_n \sin(n\omega_1 t) \ &= a_0 + \sum_{n=1}^\infty \left[rac{a_n - jb_n}{2} e^{jn\omega_1 t} + rac{a_n + jb_n}{2} e^{-jn\omega_1 t}
ight] \end{aligned}$$

Let

$$F(n\omega_1) = rac{a_n - jb_n}{2}$$
 $F(-n\omega_1) = rac{a_n + jb_n}{2}$
 $\therefore \sum_{n=1}^{\infty} F(-n\omega_1)e^{-jn\omega_1 t} = \sum_{n=-1}^{-\infty} F(n\omega_1)e^{jn\omega_1 t}$
 $\therefore x(t) = \sum_{n=-\infty}^{\infty} F(n\omega_1)e^{jn\omega_1 t}$
 $F(n\omega_1) = rac{1}{T_1} \int_0^{T_1} x(t)e^{-jn\omega_1 t} \, \mathrm{d}t$



$$F(n\omega_1)=|F(n\omega_1)|e^{j\phi_n}$$

$$|F(n\omega_1)|=rac{1}{2}\sqrt{a_n^2+b_n^2}=rac{1}{2}c_n \qquad \phi_n=rctan(-rac{b_n}{a_n})$$

$$|F_n| = |F_{-n}| = rac{1}{2}c_n \quad |F_n| + |F_{-n}| = c_n \quad F_n + F_{-n} = a_n$$

since the complex exponential Fourier series is

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t} \qquad \omega_0 = rac{2\pi}{T}$$

- ullet $a_1 e^{\pm j \omega_0 t}$: fundamental components (first harmonic components)
- ullet $a_n e^{\pm jn\omega_0 t}$: Nth harmonic components

Names of the Equations and Coefficients

Given that $\omega_0=rac{2\pi}{T}$

• Synthesis Equation:

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}$$

• Analysis Equation:

$$a_k = rac{1}{T} \int_T x(t) e^{-jk\omega_0 t} \mathrm{d}t$$

- Fourier Series Coefficients (Spectral Coefficients): $\{a_k\}$
- Magnitude Spectrum: $|a_k|$
- Phase Spectrum: $\angle a_k$

the magnitude spectrum and phase spectrum all have **discrete property**, **harmonic property** and **convergence property**

• Constant Component:

$$a_0 = rac{1}{T} \int_T x(t) \mathrm{d}t$$

The Effect of Symmetry

Even Function

The Fourier Series only contains **constant coefficients** and **cosine coefficients**, and $F(n\omega)$ is the **real function**

$$b_n=0$$
 $a_n\neq 0$

$$F_n=rac{1}{2}a_n \qquad \phi_n=0$$

Odd Function

The Fourier Series function only contains **sine coefficients**, and $F(n\omega_1)$ is the **virtual function**

$$a_0 = 0$$
 $a_n = 0$

$$b_n
eq 0 \qquad F_n = -rac{1}{2}jb_n$$

Odd Harmonic Function

$$x(t) = -x(t \pm rac{T}{2})$$

$$\begin{cases} a_0 = 0 \\ a_n = b_n = 0 \end{cases} \quad n = 2k$$

$$\left\{egin{aligned} a_n = rac{4}{T}\int_0^{rac{T}{2}}x(t)\cos(n\omega_1t) & \ b_n = rac{4}{T}\int_0^{rac{T}{2}}x(t)\sin(n\omega_1t) & \end{aligned}
ight.$$

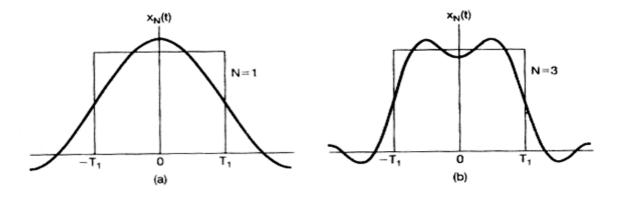
4.2 Convergence of the Fourier Series

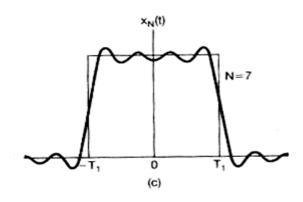
Dirichlet Conditions

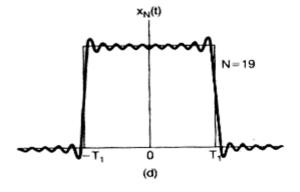
- must be integrable
- bounded variation
- finite number of discontinuities

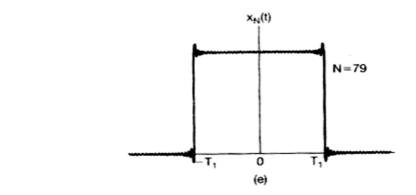
Gibbs Phenomenon

the truncated Fourier series approximation $x_N(t)$ of a discontinuous signal x(t) will in general exhibit high-frequency ripples and overshoot x(t) near the discontinuities









4.3 Parseval's Relation

the **total average power** in a periodic signal equals the sum of the average powers in **all of its harmonic components**