

An Efficient Channel Estimation Scheme for Short Frame OTFS Using Impulse-Train Pilots

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Outline



- **Motivations**
- DT-Based CE
- Simulations

Orthogonal Time Frequency Space (OTFS)



- Properties ^[1]

- Symbols are modulated in the delay-Doppler (DD) domain
- Exploits the time-frequency diversity to improve the error rate
- With lower peak-to-average power ratio (PAPR) than OFDM

- DD-based channel estimation (CE)

- To estimate the DD path parameters **as precisely as possible**
- **DD sparsity** assumption is usually introduced to exploit super-resolution algorithms
- Tough to be achieved when **the number of Doppler samples is relatively small**

Orthogonal Time Frequency Space (OTFS)



- DD-based channel estimation (CE)

- To estimate the DD path parameters **as precisely as possible**

No need in symbol demodulation since symbol **pulse is also bandwidth-limited**

- **DD sparsity** assumption is usually introduced to exploit super-resolution algorithms

DD channels are **not always sparse**, and the number of paths is also to be estimated

- Tough to be achieved when **the number of Doppler samples is relatively small (short frame OTFS)**

Due to the requirement of **low-latency** and **no range migration phenomena**

Proposed Delay-Time (DT) Based CE for OTFS



- Main ideas

- Using *impulse-train pilot (ITP)* pattern

↖
A **comb-type** pilot for time domain channel estimation

- Implemented in the DT domain, to achieve a **non-parametric** estimation

↖
DD Sparsity assumption and **super-resolution** processing are **not need**

- Simple *extra-impulse (EI)* method to improve the estimation performance

↖
To match the symbol duration and pilot duration

- Performs effectively in the **short frame** OTFS systems, **combined with the MRC-Rake detector** [1]

↖
DT-based OTFS signal processing

Outline



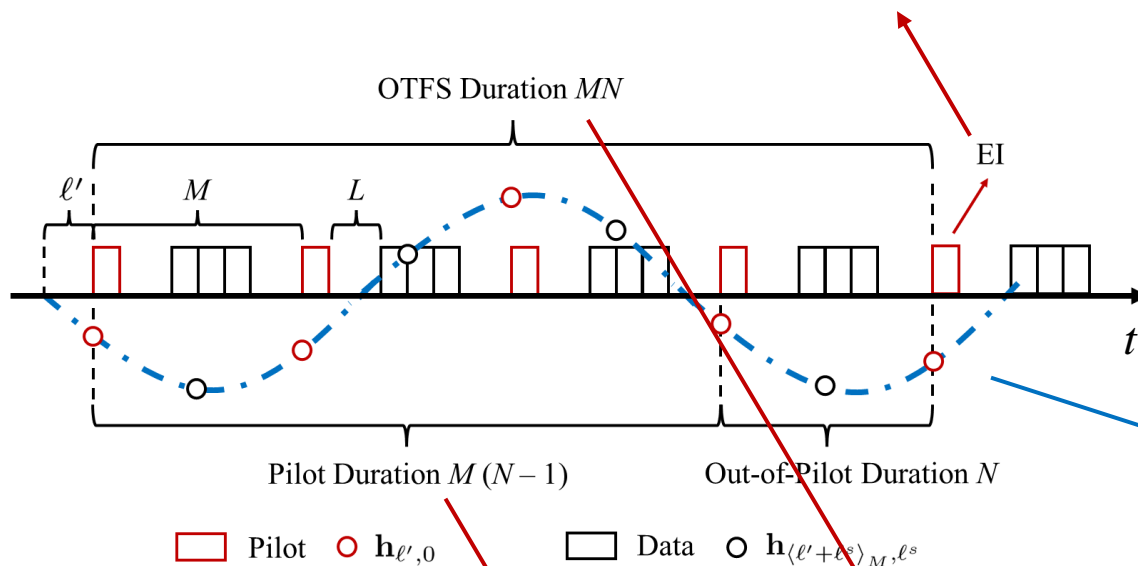
- Motivations
- **DT-Based CE**
- Simulations

Proposed Delay-Time (DT) Based CE for OTFS



• Impulse-Train-Pilot-OTFS

A pilot impulse of the next OTFS frame,
To match the pilot duration and OTFS duration



Pilot power Pilot sequence

$$\mathbf{x}_{\ell,T} = \begin{cases} \sqrt{P_{\pi}} \mathbf{s}_N & , \ell = 0 \\ \mathbf{0}_N & , 1 \leq \ell \leq L \text{ or } \ell \geq M - L \\ \mathbf{F}_N^H \mathbf{x}_{\ell,D} & , \text{otherwise} \end{cases}$$

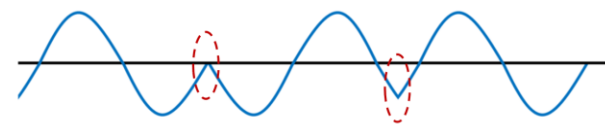
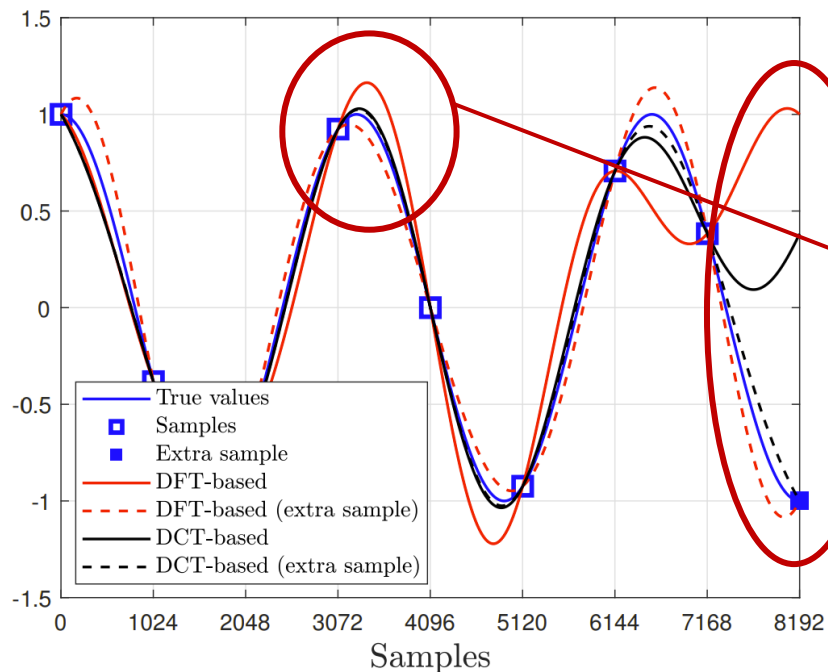
Channel gain $\mathbf{h}_{\langle \ell' + \ell^s \rangle_M, \ell^s}$ at data position can be estimated via sample **interpolation** using $\mathbf{h}_{\ell',0}$ at pilot positions

Mismatch between pilot duration and OTFS duration

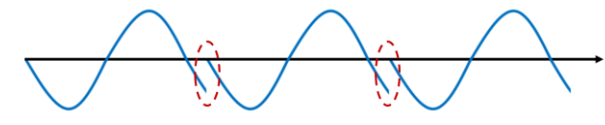
Proposed Delay-Time (DT) Based CE for OTFS



- Interpolation methods



DCT – mirror-extension



DFT – periodic extension

Less Gibbs' effect using **DCT-based interpolation** ^[1]

Improved using an **extra-impulse (EI)**

[1] Y.-H. Yeh, and S.-G. Chen, "Efficient channel estimation based on discrete cosine transform," *IEEE ICASSP*, Hong Kong, China, Apr. 2003

Proposed Delay-Time (DT) Based CE for OTFS



- Proposed DT-based CE method for OTFS

DT pilot samples with an extra-impulse

$$\mathbf{y}_{\ell',T}^{\pi} = \begin{cases} \sqrt{P_{\pi}} \mathbf{h}_{\ell',0} \odot \mathbf{s}_N + \mathbf{w}_{\ell',T} & , \ell' \in \mathcal{L} \\ \mathbf{w}_{\ell',T} & , \text{otherwise} \end{cases}$$

Least-square-based estimation

$$\hat{\mathbf{h}}_{\ell',0} = \frac{1}{\sqrt{P_{\pi}}} \mathbf{y}_{\ell',T}^{\pi} \oslash \mathbf{s}_N \quad \hat{\mathbf{h}}_{\ell',0}^{\text{EI}(\kappa)} = \begin{bmatrix} \hat{\mathbf{h}}_{\ell',0}^{(\kappa)} \\ \left[\hat{\mathbf{h}}_{\ell',0}^{(\kappa+1)} \right]_0 \end{bmatrix}$$

Discrete cosine transforming

$$\left[\hat{\mathbf{c}}_{\ell',0}^{\text{EI}(\kappa)} \right]_k = \lambda_k \sum_{n=0}^N \left[\hat{\mathbf{h}}_{\ell',0}^{\text{EI}(\kappa)} \right]_n \cos \frac{\pi (2n+1)k}{2(N+1)}$$

Threshold **denoising** in DCT domain

$$\left[\tilde{\mathbf{c}}_{\ell',0}^{\text{EI}(\kappa)} \right]_k = \begin{cases} \left[\hat{\mathbf{c}}_{\ell',0}^{\text{EI}(\kappa)} \right]_k & , \left| \left[\hat{\mathbf{c}}_{\ell',0}^{\text{EI}(\kappa)} \right]_k \right| > \gamma \\ 0 & , \text{otherwise} \end{cases}$$

Interpolation using DCT [1]

$$\left[\tilde{\mathbf{h}}_{\langle \ell' + \ell^s \rangle_M, \ell^s}^{\text{EI}(\kappa)} \right]_n = \sum_{k=0}^N \lambda_k \left[\tilde{\mathbf{c}}_{\ell',0}^{\text{EI}(\kappa)} \right]_k \cos \frac{\pi (2(\ell^s + Mn) + M)k}{2M(N+1)}$$

Outline



- Motivations
- DT-Based CE
- **Simulations**

Numerical Results



- Simulation conditions

Num. of subcarriers	1024
Subcarrier spacing	30KHz
Carrier frequency	5GHz
Denoising threshold	$3\tilde{\sigma}$
Channel model	EVA
Doppler model	Jakes
Num. of Doppler samples	8/16/32
Num. of Doppler taps each delay tap	200

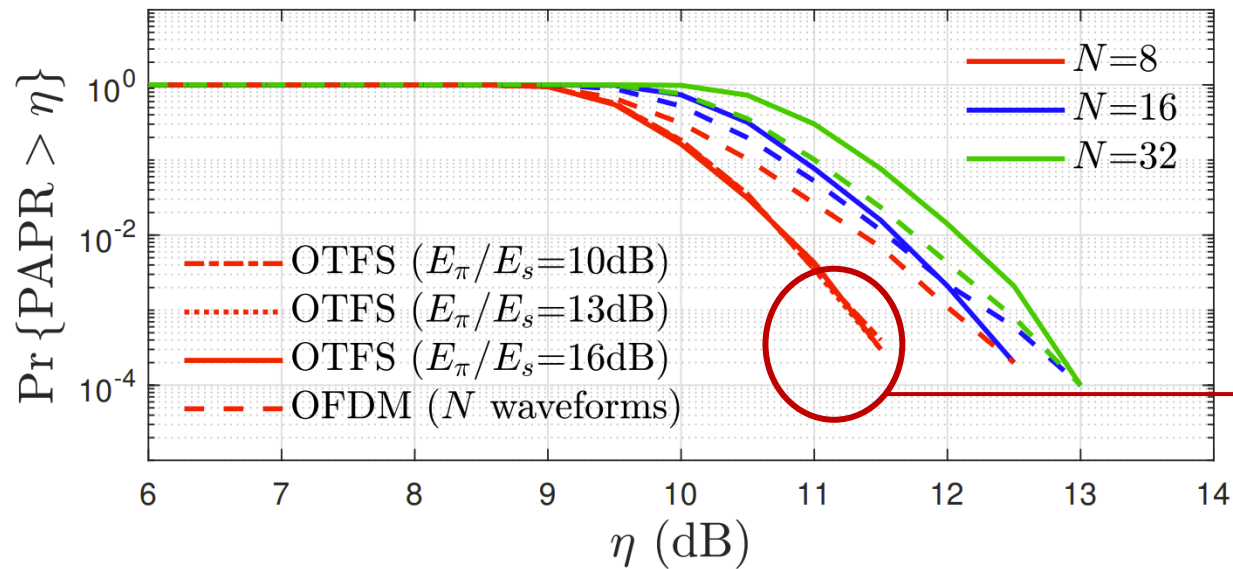
Non-sparse and fractional in Doppler domain

Short frame OTFS system

Numerical Results



- PAPRs of ITP-OTFS waveform under E_π/E_s or N



Number of Doppler samples

Single symbol energy

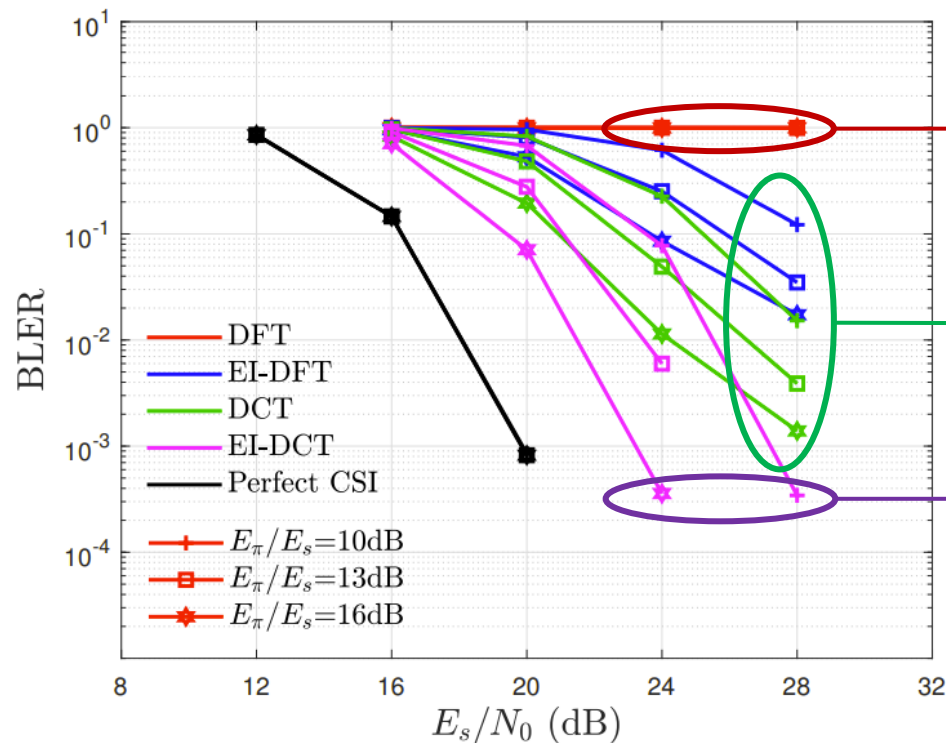
Total pilot energy

Lower PAPR than OFDM waveform when
short frame ITP-OTFS waveform is adopted

Numerical Results



- BLERs vs. E_s/N_0 under different E_π/E_s (16-QAM, $N=8$, $v_{\max}=500\text{kmh}$)

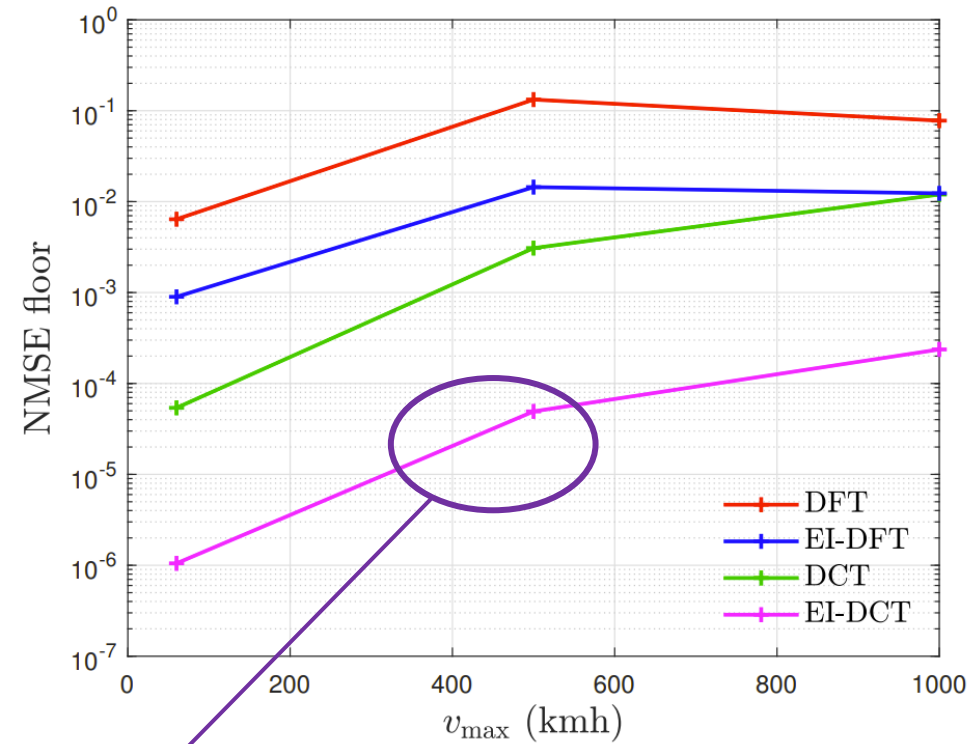
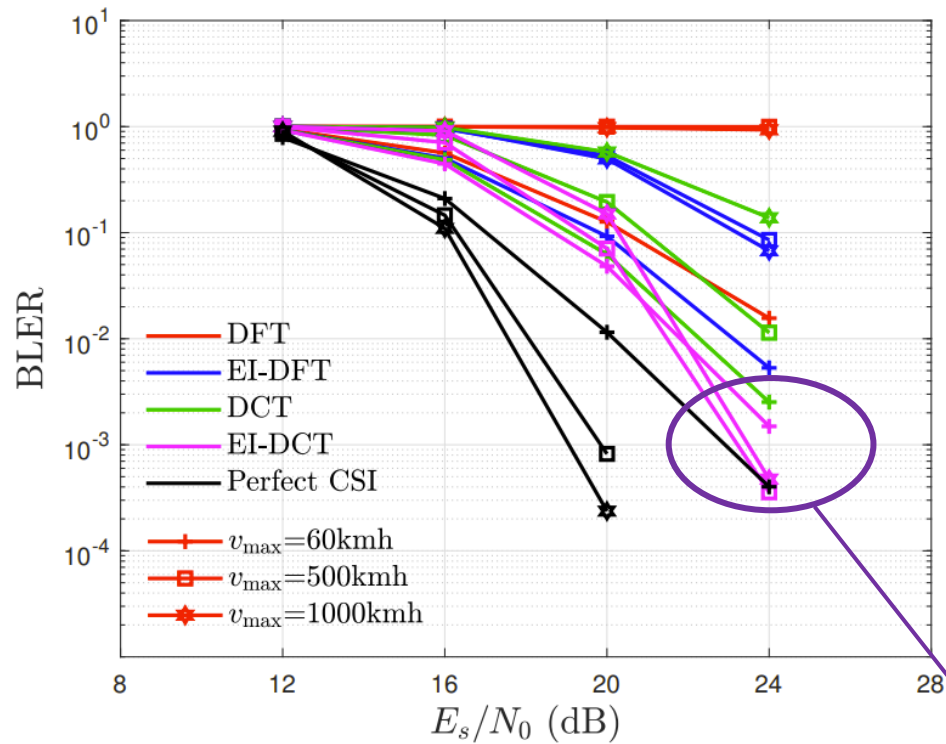


**Note: EI-DCT in these slides refers to the DCT approach with the extra-impulse, rather than the extendible inverse DCT.*

Numerical Results



- BLERs vs. E_s/N_0 under different v_{\max} (16-QAM, $E_{\pi}/E_s=16\text{dB}$, $N=8$)

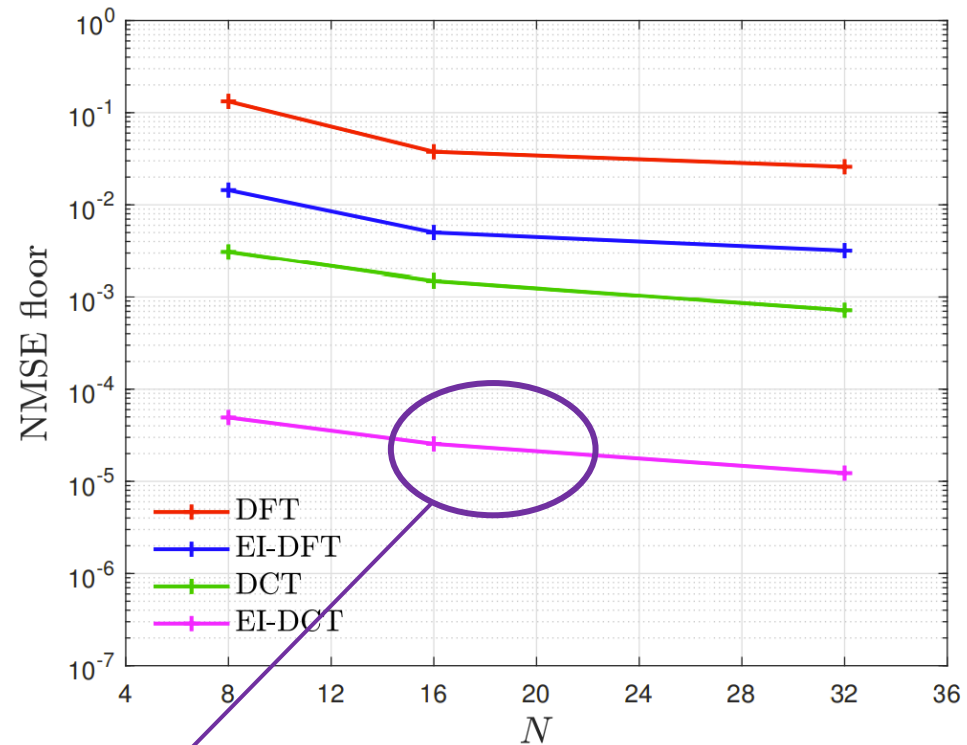
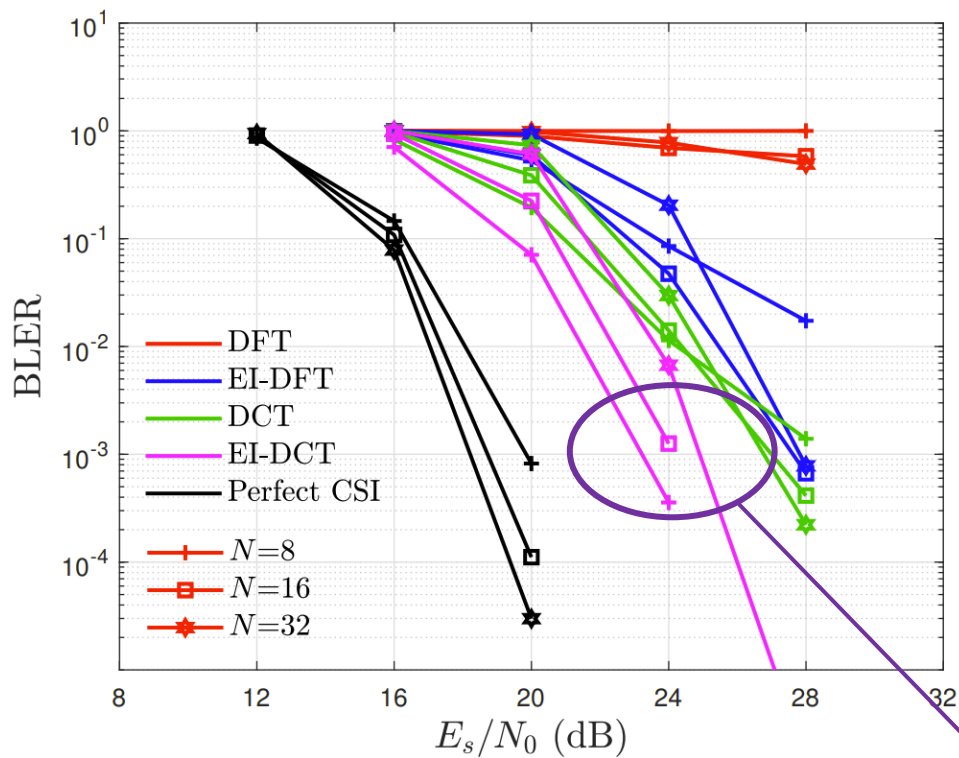


EI-DCT based scheme has the **lowest error rate** and **lowest NMSE floor**

Numerical Results



- BLERs vs. E_s/N_0 under different N (16-QAM, $E_\pi/E_s=16\text{dB}$, $v_{\max}=500\text{kmh}$)

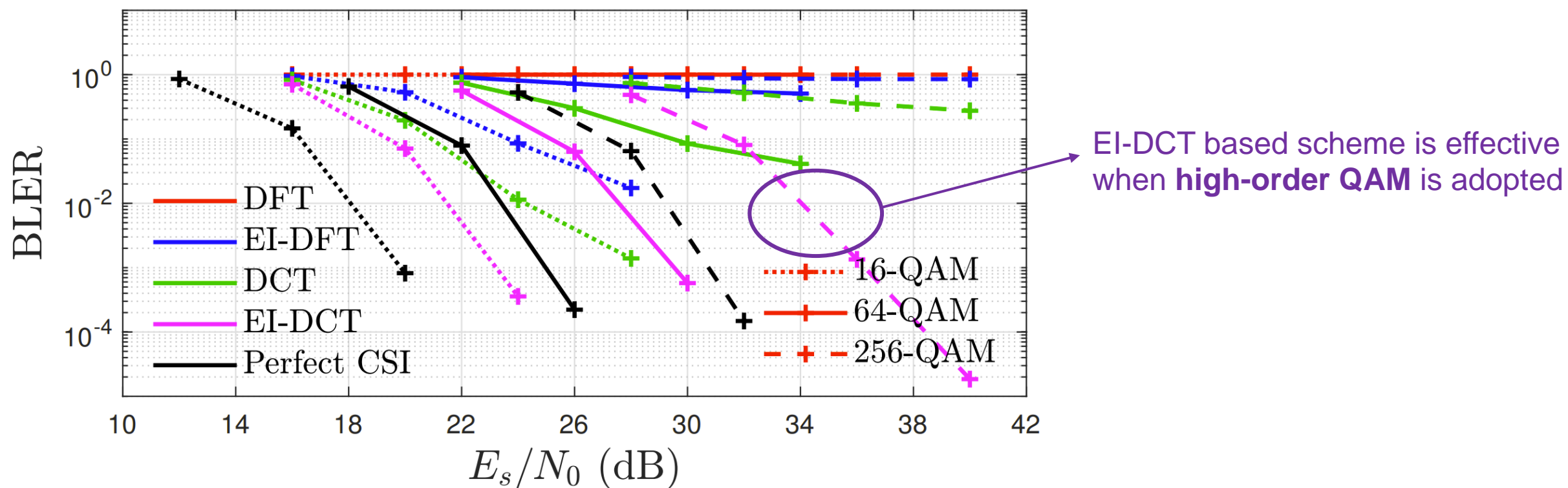


EI-DCT based scheme has the **lowest error rate** and **lowest NMSE floor**

Numerical Results



- BLERs vs. E_s/N_0 under different QAMs ($E_\pi/E_s=16\text{dB}$, $N=8$, $v_{\max}=500\text{kmh}$)



Thanks for your attention !

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