# **Digital Systems**

## 2nd May

#### What are the advantages of digital systems?

- Resilience to noise → Noise forms a Gaussian distribution around the actual value. Filtering analog signals is hard, especially with noise.
- Ease of storage → Flash/CD/DVD
  - Previously stored in analog form in magnetic tapes
- More secure (encrypted)
- Flexible signal processing (e.g. noise removal)

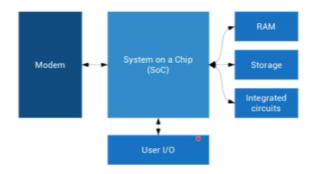
#### What are some examples and applications of digital systems?

- · Digital thermometers
- · automated driving
- · digital computer stuff
- · augmented reality, etc.
- The Al

#### What is the anatomy of a digital system?

- Top down methodology: we talk about big things and break it into smaller pieces
- Implement
  - · Sequential logic
  - Combinational logic

#### **Blocks in a Mobile Phone!**



#### What is a system on a chip?

Microprocesser	Microcontroller	
It's used in computer systems like	It's used in embedded systems like washing	
desktops, etc.	machine, etc.	

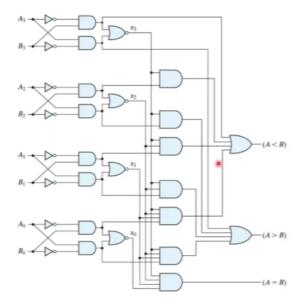
## Anatomy of a smartphone- What is the iPhone SOC?

Huge number of logic blocks

CLB→ configurable logic block

Logic blocks → consists of logic gates like NOR gates, AND gates, etc. and clocks

### **Combinational logic gate that does comparison**



Sequential logic: uses clock cycles and does stuff, and synchronizes stuff!

## **3rd May**

### Any base to decimal conversion → just multiply it out

$$7392 = 7 \times 10^3 + 3 \times 10^2 + 9 \times 10^1 + 2 \times 10^0$$

$$a_n a_{n-1} \ldots a_0 \ldots a_{-1} \ldots a_{-m} = \sum_{i=-m}^n a_i \mathrm{base}^i$$

#### **Decimal to binary:**

Eg. 41

41	Dividing by 2
20	1

41	Dividing by 2
10	0
5	0
2	1
1	0
0	1

Then write it from bottom up Answer is 100101.

#### **Decimal to Octal**

Eg. 153

153	Dividing by 8			
19	1			
2	3			
0	2			

Answer 231

0.6875 in binary:

 $0.6875 \times 2 = 1, 0.3750$ 

 $0.3750 \times 2 = 0, 0.7500$ 

 $0.7500 \times 2 = 1, 0.5$ 

 $0.5000\times 2=1,0$ 

The number is written top-down.

0.1011

## Binary, Hex, Octal conversion is lite, use 2, 4, and 3 digits respectively

8 bits makes up one byte (little children know this)

### **Summary**

Start	Binary	Decinal	Hexa
Binary	×	for non digit	4 bing -> Hern
Decimal	+ 2 (Ret) x2 (ngt)	×	+16 (Reft) ×16 (right)
Hesca	Each hera → 4 binary	x 16 <sup>n-1</sup> for n <sup>th</sup> digit	×

## 5th May

## **Complements**

Given two n-digit numbers in base 'r'

Nice video: https://www.youtube.com/watch?v=4qH4unVtJkE

**Negative numbers:** 

- Sign-magnitude
  - First bit for sign, remaining digits as usual binary
  - There's a problem: 5 + -5 in this representation is NOT zero.
- Ones' complement
  - Positive numbers as usual, negative numbers are just 1s complement (which is toggling every bit)
  - 5 → 0101
  - -3 → 1100
  - $5 3 \rightarrow 10001 \rightarrow 0001 \rightarrow 1$
  - So, it is offset by one and we have to do some juggling around while carrying out subtraction
- 2s complement
  - Take twos complement of every number to get its negative
  - you get one extra number
  - there are *no* signed zeroes
  - The numbers are the same as unsigned numbers mod (2^N)
  - First digit represent -2^N digit really
  - Number + its negative is 0

**Addition of numbers: lite** 

## Subtraction of numbers: M - N, where M and N are unsigned numbers

- M + (r^n N)
- if M >= N:

- Produces end carry, and leftmost carry digit corresponds to r^n, which
  needs to be discarded (n carry is formed when the first digit is +ve)
- M N
- if M < N:
  - Doesn't produce n carry, and result is r^n (N M) → We have to take r's complement and place negative sign at front

Eg. 6 - 13

- +6 → 00000110
- -13 →
  - Binary representation: (+13) → 00001101
  - Toggle: 1s complement: (-13) → 11110010
  - Add 1 to ones complement → Twos complement → 11110011
- Adding them both: 11111001 (first bit, i.e. sign bit is 1)
- end carry not there, so now do twos complement: 00000111
- This is **-7**!

Eg. 15 - 7:

- 15 → 00001111
- -7 → 11111001
- Add up:
   100001000 → 8!

## May 6

## **Boolean Algebra**

### **Postulates of Boolean Algebra:**

For a set B (not necessarily {0,1} btw):

- 1. Closure
- 2. Identity

$$1. x + 0 = x$$

$$2. x.1 = x$$

- 3. Commutative
- 4. Distributive

1. 
$$x + yz = (x + y)(x + z)$$

$$2. x (y + z) = xy + xz$$

5. Complement

$$5. x + x' = 1$$

$$6. x x' = 0$$

6. There are at least two elements x and y such that  $x \neq y$ 

To solve question, we need identity, distributive, complement

х	y	x · y
0	0	0
0	1	0
1	0	0
1	1	1

x	y	x + y
0	0	0
0	1	1
1	0	1
1	1	1

X	<b>x</b> '
0 1	1 0

### **Theorems of Boolean Algebra**

1.  $x + x = x, x \cdot x = x$ 

=  $x (1 + 1) \rightarrow Distributive Law$ 

 $= x . 1 \rightarrow OR$  definition

 $= x \rightarrow Identity element property$ 

2.  $x + 1 = 1, x \cdot 0 = 0$ 

= x + 1

 $= x + x + x' \rightarrow Complement$ 

 $= x + x' \rightarrow Theorem 1$ 

= 1 → Complement definition

3. Involution: (x')' = x

(x')' = (x')' + 0

=  $(x')' + x x' \rightarrow Complement definition$ 

=  $(x + (x')')(x' + (x')') \rightarrow Distributive Law$ 

 $= x + (x')' \rightarrow Complement$ 

=  $(x + (x')')(x + x') \rightarrow Complement$ 

=  $x + (x') x \rightarrow Property again!$ 

=  $x (1 + x') \rightarrow Distributive$ 

= X

## 4. Associativity

xy = yx (Commutativity)

$$x = (a + b) + c$$

$$y = a + (b + c)$$

$$xy = ((a + b) + c)(a + (b + c)) = y$$

By symmetry, yx = x

So, 
$$x = y$$

## 5. De morgan!

Lemma:

#### Given:

- a + b = a + a' = 1
- ab = aa' = 0 To prove:
- a = b'

Multiply first one by b

$$ab + b = ab + a'b$$

$$0 + b = a'a + a'b$$

$$b = a'(a + b) = a' 1 = a'$$

#### Now

$$x'y' + (x + y) = 1$$
 (Using lemma)

$$(x + y)' + (x + y) = 1$$

Therefore:

• 
$$(x + y)' = x'y'$$

#### 6. Absorption

$$X + XY$$

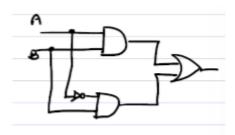
= 
$$x (1 + y) \rightarrow Distributive$$

$$= x$$

## Usefulness of Boolean algebra: simplifying Boolean expression

The electrical idea is to keep number of gates minimum

a) 
$$AB + A'B = (A + A')B = B$$



### **Canonical Form of Boolean Expressions**

Boolean expression consists of variables in either normal form or its complement

n variables with AND (lowercase)

- We can have 2<sup>n</sup> "minterms" (minterm is just every possibility basically)
- minterms are ordered as if they were binary numbers (the image below makes it clear)
- $m_0m_1m_2...m_{2^n-1}$

- We consider "1" here, as the output 1 is unique n variables with OR (uppercase):
- We have 2'n "maxterms"
- $\bullet \quad M_0M_1M_2\dots M_{2^n-1}$
- We consider "0" here, as the output 0 is unique

عد	J	Z	Minterms	Multerns
0	0	0	2' y' z' (Mo)	sc+ += (M0)
0	0	١	x' y' z (M,)	sc+y+z' (M,)
0	ι	0	x' y z' (M2)	x + y'+z (M2)
0	ı	١	x' y ~ (M)	$x+y'+z'$ $(M_2)$
-1	0	0	x y' z' (M4)	x' +y+ 2 (M4)
- 1	0	(	× y 2 (M5)	x'+ y+z' (Ms)
ı	ı	0	x y z' (M,)	x'+y'+2 (M4)
1	ı	ı	z y z (m,)	$x'+y'+z'$ $(M_7)$

## (i) Boolean

Any boolean function can be expressed as

- sum of minterms
- products of maxterms!

#### **Consider the function**

f1(x, y, z) = xyz' + xy'z + xy'z'

This function is 1 when:

x	у	z
1	1	0
1	0	1
1	0	0

And 0 for anything else

f1(x, y, z) 
$$\rightarrow \Sigma 4, 5, 6$$

$$f1 \rightarrow \Pi 0, 1, 2, 3, 7$$

$$f2(x, y) \rightarrow xy + xy' + x'y'$$

$$\rightarrow \Sigma 0, 2, 3$$

$$\rightarrow \Pi 1 = (x' + y)$$

This is intuitive enough!

#### $f = \Pi$ Some maxterms = $\Sigma$ of other minterms

- It's easy enough to see that  $m_i' = M_i$ 
  - Say  $m_2 = x + y' + z$
  - $m_2' = x'yz' = M_2$
- $(\sum_{ ext{i runs over whatever}} m_i)' = \sum_{ ext{i runs over everything else}} m_i$ 
  - Proof:
    - $f = \sum m_i$
    - f is 1 for certain values of i
    - f' is 0 for those values of i
    - f' is 1 for all other values of i
    - $ullet f' = \sum_{ ext{j runs over everything else}} m_j$
- $ullet m_i = \Pi_{
  m j\,=\,everything\,\,else} M_{
  m j}$
- $f = \sum_{i \text{ runs over whatever}} m_i$
- $f' = \sum_{i \text{ runs over everything else}} m_i$
- $f'' = f = \prod_{\text{i runs over everything else}} m'_i = \prod_{\text{i runs over everything else}} M_i$

#### Dual of a function:

- Replacing all AND by OR and vice versa
- · Replacing all 1s by 0s

Taking complement of a function is the same as taking dual of the function and replacing each literal by it's complement (this follows from de Morgan's theorem).

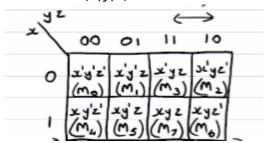
To go from positive logic to negative logic, we just have to take the dual!

## **May 9**

## **Gate level minimization: K-maps**

- The objective is find the optimal gate level implementation of the Boolean function
  - Complexity of digital logic circuit is related to the complexity of the algebraic expression
  - Truth tables are unique

- Pictorial representation is Karnaugh map
- Consider f(x,y,z)



Only one bit changes between adjacent cells (in both axes)

Adjacencies	Literals
1	3
2	2
4	1
8	0

### f2 = Sum(3,5,6,7)

x\yz	00	01	11	10
0	0	0	1	0
1	0	1	1	1

Adjacent cells in K-map represent whatever didn't change!

f2 doesn't change once in x direction, once in y direction, once in z direction So, f2=xy+yz+zx

## f1 = Sum(1,4,7)

x\yz	00	01	11	10
0	0	1	0	0
1	1	0	1	0

No adjacency, so it's simplified already

$$f3 = sum(3,4,6,7)$$

x\yz	00	01	11	10
0	0	0	1	0
1	1	0	1	1

$$f3 = yz + xy + xz' = yz + xz'$$

• Reason: xy terms is redundant!

So the aim is to cover all minterms.

$$\mathbf{f4} = \text{sum} (0,2,3,4,5)$$

x\yz	00	01	11	10
0	1	0	0	1
0	1	1	0	1

• Four cell adjacency!

$$f4 = z' + x y'$$

$$f5 = Sum(1,2,5,6,7) = Prod(0, 3, 4)$$

x\yz	00	01	11	10
0	0	1	0	1
1	0	1	1	1

$$f5 = y'z + yz' + xy$$

$$f5 = (y + z) \cdot (x + y' + z')$$

A\BC	00	01	11	10
0	0	1	1	1
1	0	1	1	0

$$f6 = C + A'B$$

### Four variable K-maps

wx\yz	00	01	11	10
00	0	0	0	0
01	0	0	0	0
11	0	0	0	0
10	0	0	0	0

Adjacencies	Literals
1	4
2	3
4	2
8	1
16	0

**f1** = Sum(0,1,2,4,5,6,8,9,12,13,14)

wx\yz	00	01	11	10
00	1	1	0	1
01	1	1	0	1
11	1	1	0	1
10	1	1	0	0

$$f1 = y' + w'z' + x z' = y' + z' (x + w')$$

$$f1' = yz + w x' y$$
  
 $f1 = (y' + z') (w' + x + y') = w'y' + w'z' + x y' + x z' + y'z' + y'$   
 $= y' + z' (x + w')$ 

Prime Implicant: Total number of adjacencies in the K-map (including redundancies) Essential Prime Implicant: Prime implicants that cover a minterm no other implicant covers

Redundant Implicant: Implicant for which each minterm is covered by some other implicant

AB\CD	00	01	11	10
00	1	0	1	1
01	0	1	1	0
11	0	1	1	0
10	1	1	1	1

Eg.

- m5 is covered only by one prime implicant, BD is essential prime implicant
- m0 is covered only by B' D'

### **Simplification of Kmap**

K. Map Simplification Process:

Step 1: Represent all the minterna/macterna

Step 2: Identify the extential prime implicants

Step 3: Cover all other minterna as/ max adjacency (prime implicant)

Step 4: Ensure no reduced terms

#### F

1.

AB\CD	00	01	11	10
00	1	1	0	1
01	0	1	0	0
11	0	0	0	0
10	1	1	0	1

- 2. Essential implicants:
- B'D' (for 0010 and 1010)
- B'C'
- A'C D
- 3. All minterms are covered!
- 4. F = B'D' + B'C' + A' C' D

# **May 12**

#### Don't care conditions:

Eg. Binary coded decimal (BCD)

Representation of decimals in binary numbers

396 → Each value is represented by corresponding value

396 → 0011 1001 0110

BCD doesn't care about remaining inputs like 1111 (corresponding to 15). So, we don't care about that part.

Don't cares are represented by X in K-map

$$\begin{aligned} & \text{F} = \Sigma(1, 3, 7, 11, 15) \\ & \text{d} = \Sigma(0, 2, 5) \end{aligned}$$

wx\yz	00	01	11	10
00	Χ	1	1	Χ
01	0	Χ	1	0
11	0	0	1	0
10	0	0	1	0

SOP

F = yz + w'x' or yz + w'z

POS

F = z (w' + y)

## **May 13**

#### NAND and NOR gates are easier to realize

AND, NOT, OR gates → NAND and NOR gates

NAND and NOR gate → universal gate → we can create any other gates using these gates

- NOT gate → NAND(x, x)
- AND gate → NOT(NAND(x, x))
- OR gate  $\rightarrow$  NAND(NOT(x), NOT(y))

## **(i)** Boolean function implementation

- 1. Represent function in K-map
- 2. Obtain simplified function in terms of Boolean operators
- 3. Convert function to NAND/NOR logic
- 4. Implement using NAND/NOR gates

Eg.

F = AB + CD

NAND logic

• (A' + B')' + (C' + D')' = NOR(NOR(NOR(NOR(A), NOR (B)), NOR(NOR(C), NOR(D))))

Eg. F = 
$$\Sigma(1, 2, 3, 4, 5, 7)$$

A\BC	00	01	11	10
0	0	1	1	1
1	1	1	1	0

$$F = A B' + A' B + C$$

2. Done

3.

NAND((NAND(A, NAND(B)), (NAND(NAND(A), B)), NAND(C)) NOR(NOR(C, NOR(NOR(A), B), NOR(A, NOR(B))))

Eg. 
$$F = ((A + B)(C + D)E)'$$
  
NOR(NOR(NOR(A, B)), NOR(C, D), NOR(E)))

In general, look at complement:

- AND-NOR ↔ NAND-AND → SOP form combining 1s
- OR-NAND ↔ NOR-OR → POS form combining 0s

x\yz	00	01	11	10
0	1	0	0	0
1	0	0	0	1

#### **SOP form**

From zeroes, we get F'

$$F' = z + xy' + x'y$$

$$F = (z + xy' + x'y)' \rightarrow AND-NOR realization$$

$$F = ((x'y)' (x'y)' z') \rightarrow NAND-AND realization$$

#### **POS form**

$$F' = (x + y + z)(x' + y + z')$$

$$F = ((x + y + z) (x' + y + z'))' \rightarrow OR-NAND realization$$

$$F = (x + y + z)' + (x' + y + z')' \rightarrow NOR-OR$$
 realization

## **XOR** gate

- Parity checking
- Binary adders
- Error detection and correction

$$x \oplus y = xy' + x'y = (x+y')(x'+y)$$

#### **Identities:**

• 
$$x \oplus 0 = x$$

- $x \oplus 1 = x'$
- $x \oplus x = 0$
- $x \oplus x' = 1$
- $x \oplus y' = x' \oplus y = (x \oplus y)'$
- Using NAND gate: NAND(NAND(x, y'), NAND(x', y))

Eg.  $A \oplus B$ 

AB\CD	00	01	11	10
00	0	0	0	0
01	1	1	1	1
11	1	1	1	1
10	0	0	0	0

 $A \oplus B \oplus C$ 

AB\CD	00	01	11	10
00	0	0	1	1
01	1	1	0	0
11	0	0	1	1
10	1	1	0	0

 $A\oplus B\oplus C\oplus D$ 

AB\CD	00	01	11	10
00	0	1	0	1
01	1	0	1	0
11	0	1	0	1
10	1	0	1	0

# 17th May

## **Combinatorial Logic Circuits**

Interconnection of logic gates to accomplish a logic operation

- Binary addition
- Binary subtraction

- Binary multiplication
- Comparison between binary numbers
- Encoding, decoding, etc.

### **Design Procedure**

- 1. Determine number of inputs and outputs
- 2. Derive truth table
- 3. Obtain simplified Boolean expressions
- 4. Implement logic circuit

## **One-bit adders**

### Half-adder

Carries out binary addition of two binary inputs

x	у	С	s
0	0	0	0
0	1	0	1
1	0	0	1
1	1	1	0

$$s = x \oplus y$$

$$c = xy$$

## **Full Adder**

X	у	Z	С	s
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

• K-map for s

•	x/yz	00	01	11	10

x/yz	00	01	11	10
0	0	1	0	1
1	1	0	1	0

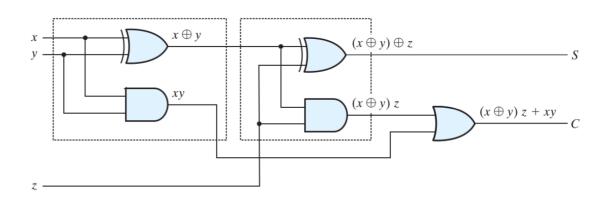
$$s=xy'z'+x'yz'+x'y'z+xyz=z'(x\oplus y)+z((x\oplus y)')=(x\oplus y)\oplus z$$

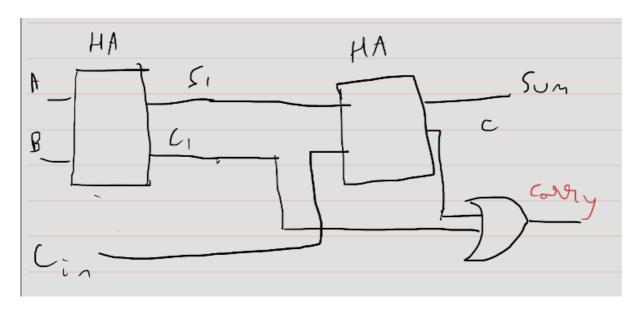
## • K-map for c

•	x/yz	00	01	11	10
	0	0	0	1	0
	1	0	1	1	1

$$c = xy + x'yz + xy'z = xy + z(x \oplus y)$$

We can just use the XOR used in s for c





Full Adder → 2 Half Adder + OR gate

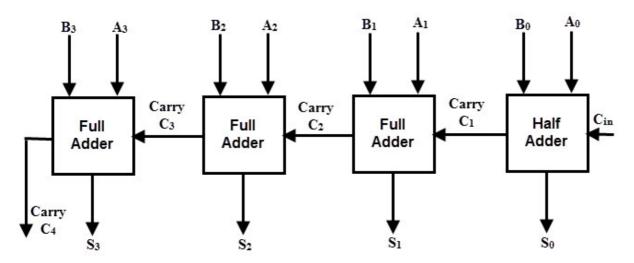
## **Many bit adders**

## 4-bit adder

#### **Ripple Carry adder**

Nice video <a href="https://www.youtube.com/watch?v=wvJc9CZcvBc&ab\_channel=BenEater">https://www.youtube.com/watch?v=wvJc9CZcvBc&ab\_channel=BenEater</a>
Augend + Addend

Output carry of ones place becomes input to next digit Output carry of twos place becomes input to fours digit, etc.



#### × Problems with this method

We have to do this sequentially, one by one, so it takes time. We have to wait for the carry

### Carry look ahead logic

$$C_{i+1} = A_i B_i + C_i (A_i \oplus B_i)$$

Carry Generate  $\rightarrow G_i = A_i B_i$ 

Carry Propagate  $\rightarrow P_i = (A_i \oplus B_i)$ 

$$S_i = P_i \oplus C_i$$

$$C_{i+1} = G_i + C_i P_i$$

 $C_0 \rightarrow \text{Input carry}$ 

$$C_1 \rightarrow G_0 + P_0 C_0$$

$$C_2 \rightarrow G1 + P1G_0 + P_1P_0C_0$$

$$C_3 \Rightarrow G_2 + P_2G_1 + P_2P_1G_0 + P_2P_1P_0C_0$$

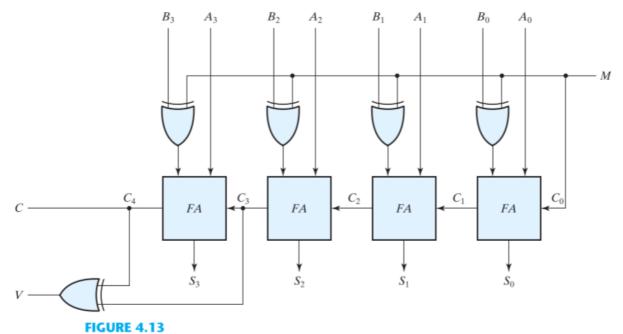
## **Binary Subtraction**

+49 →00110001

-49 →1s: 11001110 → Accomplished by XOR gate

2s: 11001111  $\rightarrow$  Can be done using initial carry  $C_0$ 

Extra sign bit  $\mathbf{M} \rightarrow \mathbf{M} = 0$ , add;  $\mathbf{M} = 1$ , subtract



Four-bit adder-subtractor (with overflow detection)

- Thinking about the addition part:
  - If M is 0,  $0 \oplus A = A$ . So, we just get out regular 4-bit ripple carry adder
- Think about the subtraction part:
  - If M is 1, 1  $\oplus$  A = A'. So, each bit first gets toggled (which gives 1s complement). Also, M acts as the input carry  $C_0$ , so we account for the 1 added in 2s complement by setting  $C_0 = 1$

#### The overflow bit V

If  $V = C_4 \oplus C_3 = 0 \rightarrow \mathsf{Correct}$ 

Else → Wrong

If we're adding a positive and negative number, our answer is always correct

- A3 is 0, B3 is 1
- If C3 is 0, C4 is 0, so V is 0 and we're good
- If C3 is 1, C4 is also 1, so V is 0 and we're good

If we're adding two positive numbers, overflow might occur.

- C4 is 0 if we're adding two positive numbers
- 0011, 0010
  - C3 is 0
  - C4 is 0
  - So, the output is correct
- 0111, 0100
  - C3 is 1
  - C4 is 0
  - So, the output is wrong and overflow has occurred

If we're adding two positive numbers, overflow might occur.

- C4 is 1 if we're adding two positive numbers
- 1011, 1010
  - C3 is 0 (since we're toggling bits here, C3 being 0 means A3 and B2 are like 0, which means they're "large" in magnitude)
  - C4 is 1
  - So, the output is wrong and overflow has occurred
- 1111, 1100
  - C3 is 1
  - C4 is 1
  - So, the output is correct

Eg.

70 → 01000110

80 → 01010000

**10010110** = -106

-70 → 10111010

-80 → 10110000

01101010 (C8 = 1) → 106

## **May 26**

#### **Half Subtractor**

A - B

 $B_0$  is a boolean that keeps track of if we have to borrow from the next digit

A	В	D	$B_0$
0	0	0	0
0	1	1	1
1	0	1	0
1	0	0	0

$$D = A \oplus B$$

$$C = A'B$$

#### **Full Subtractor**

A - B - C

A	В	С	D	$B_0$
0	0	0	0	0

A	В	С	D	$B_0$
0	0	1	1	1
0	1	0	1	1
0	1	1	0	1
1	0	0	1	0
1	0	1	0	0
1	1	0	0	0
1	1	1	1	1

$$\mathsf{D} = A \oplus B \oplus C$$

## **Binary Multiplication**

Multiplicand  $\rightarrow$  B1 B0 Multiplier  $\rightarrow$  A1 A0

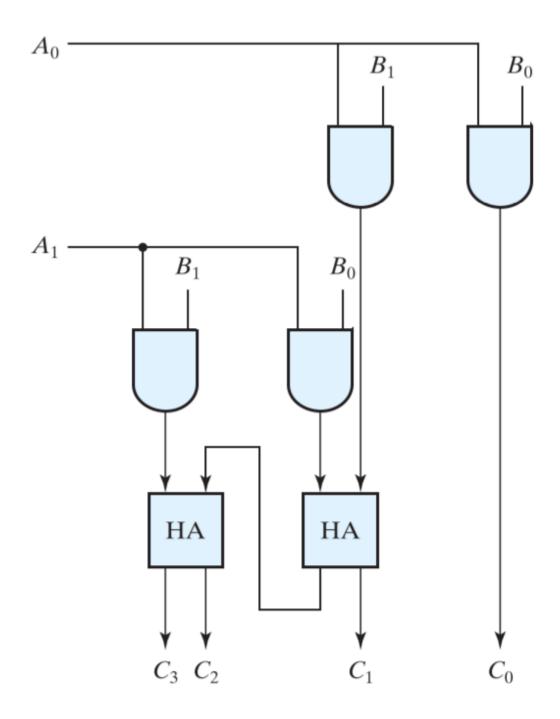
B1 B0 A1 A0

P0 = A0B0

P1 = A0B1 + A1B0

P2 = Carry + A1B1

P3 = Carry



2×2 multiplication →4 AND gates and 2 HA

# **May 20**

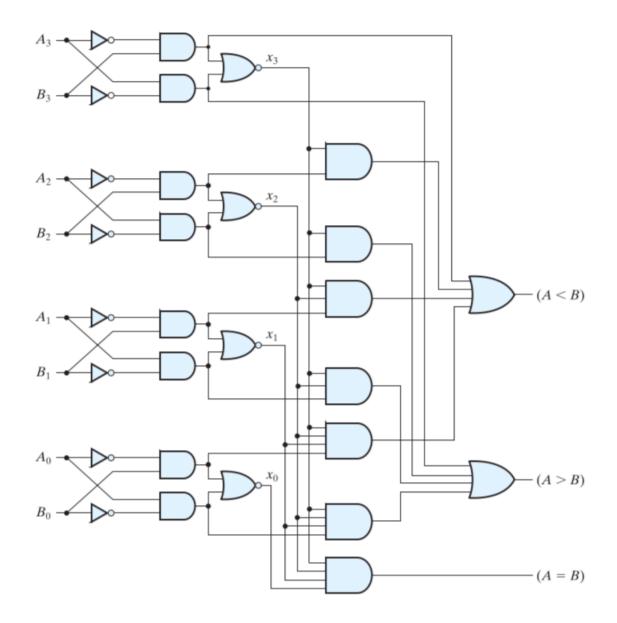
## **Magnitude Comparator**

A = A3A2A1A0

B = B3B2B1B0

 $A_i \odot B_i = x_i$ , where  $\odot$  is XNOR gate

- ullet A = B if  $A_i=B_i orall i$ 
  - A = B if  $\Pi A_i \odot B_i = \Pi x_i = 1$
- $A > B \rightarrow A_3B_3' + x_3A_2B_2' + x_3x_2A_1B_1' + x_3x_2x_1A_0B_0'$
- A < B  $\Rightarrow$   $A_3'B_3 + x_3A_2'B_2 + x_3x_2A_1'B_1 + x_3x_2x_1A_0'B_0$



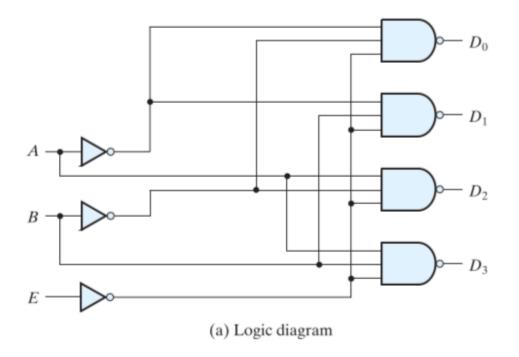
## **Decoder / Demultiplexer**

## 2 × 4 Decoder with Enable Pin

Enable  $1 \rightarrow$  Chip not ready

Enable  $0 \rightarrow$  Chip ready to work

Enable	A	В	DO	D1	D2	D3
1	Х	Χ	1	1	1	1
0	0	0	0	1	1	1
0	0	1	1	0	1	1
0	1	0	1	1	0	1
0	1	1	1	1	1	0



### Also acts as demultiplexer

- if you want your data to go to say D3, we can give A = 1, B = 1. The output is exactly the same in the required pin, and remains always 1 in the other pins
- after two seconds say we want data to go to D1, we can give A = 0, B = 1

### Info

It's possible to use the enable pin as in input, in that case we get  $2^{(n+1)}$  outputs

## **Priority Encoders**

 $2^n$  bits to n bits

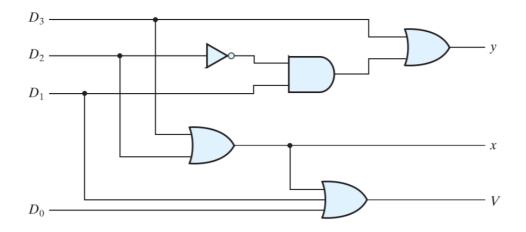
#### **V** → Validity indicator

- 0 if all 0s
- 1 otherwise

Least Priority → D0	D1	D2	Highest Priority → D3	x	у	v
0	0	0	0	X	Х	0
1	0	0	0	0	0	1
X	1	0	0	0	1	1
X	Χ	1	0	1	0	1
Х	Χ	Χ	1	1	1	1

• 
$$V = D0 + D1 + D2 + D3$$

• 
$$x = D2 + D3$$



## Multiplexer

S → Select pin

 $x \rightarrow Input$ 

y → Output

<b>S</b> 1	S0	у
0	0	х0
0	1	x1
1	0	x2
1	1	х3

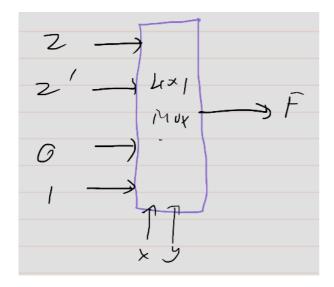
## **Boolean Function implementation**

## **Using Multiplexer**

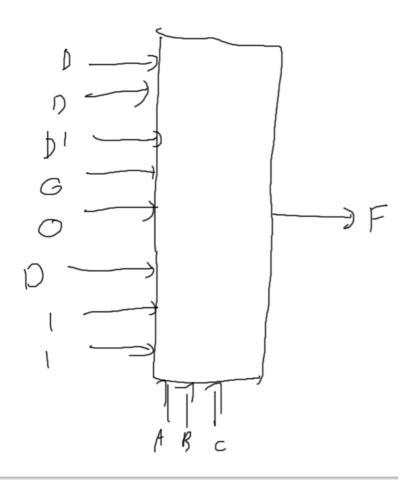
Implement  $f(x,y,z) = \sum (1,2,6,7)$  using 4×1 mux

х	у	Z	F	
0	0	0 1	0 1	F = z
0	1 1	0 1	1 0	F = z'
1 1	0	0 1	0	F = 0
1 1	1 1	0 1	1 1	F = 1

# (a) Truth table



 $F(A,B,C,D) = \sum (1,3,4,11,12,13,14,15)$ 



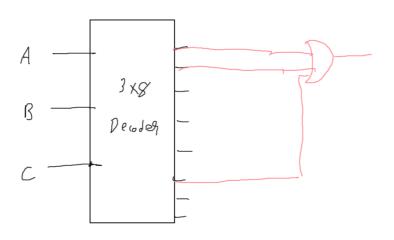
### General:

n variables

- n 1 select pins
- 2^(n 1) inputs

## **Using Decoder**

$$f(A, B, C) = A'BC + A B C' + ABC$$



## **Binary Coded decimal:**

Integers from 0 - 9  $\rightarrow$  as such 10  $\rightarrow$  15 are don't cares for any operation

Gray's Code: adjacent numbers vary only by one bit

	Gray code				
	4	3	2	1	
0	0	0	0	0	
1	0	0	0	1	
2	0	0	1	1	
3	0	0	1	0	
4	0	1	1	0	
5	0	1	1	1	
6	0	1	0	1	
7	0	1	0	0	
8	1	1	0	0	
9	1	1	0	1	
10	1	1	1	1	
11	1	1	1	0	
12	1	0	1	0	
13	1	0	1	1	
14	1	0	0	1	
15	1	0	0	0	

**3's excess:** 3's excess number = Binary number + 0011 (i.e. "3"), from 0 - 9

• Taking 9's complement is just toggling all bits!