

Chapter 4 - Introduction to Electronic Communication

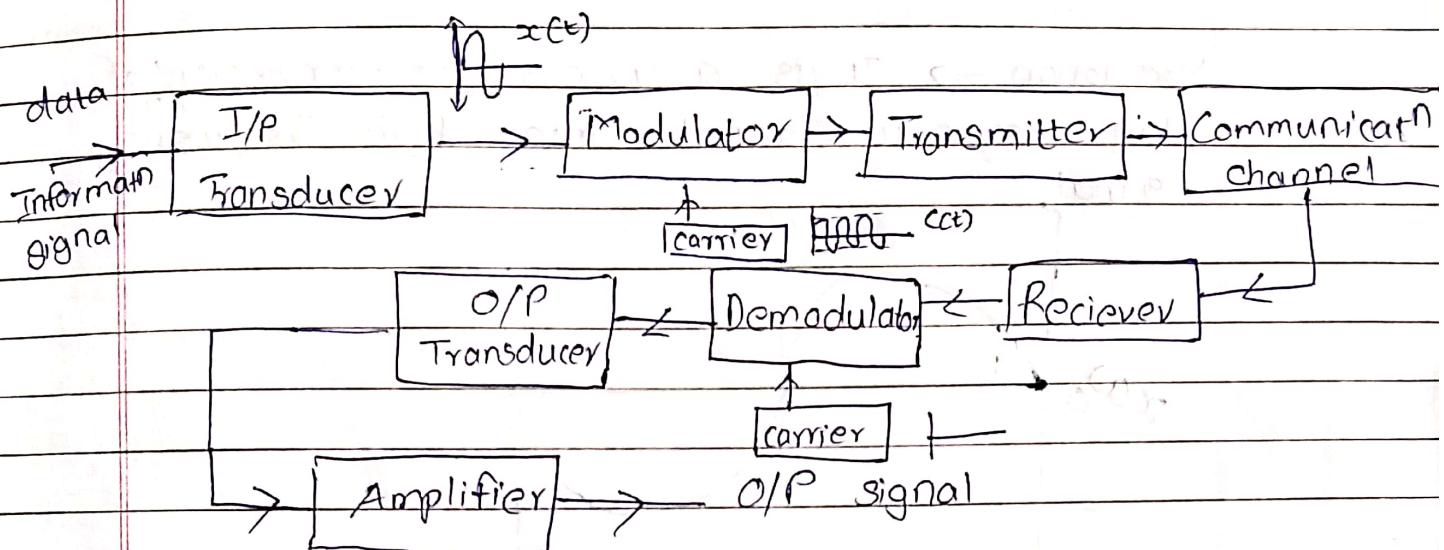
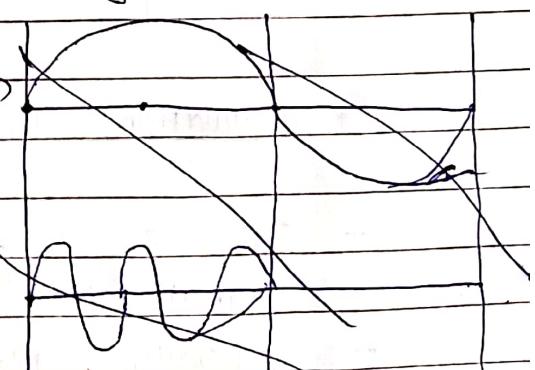
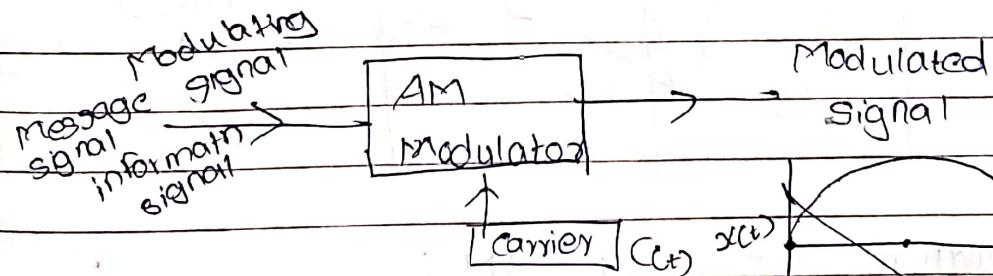


fig:- Basic Block Diagram of communication system.

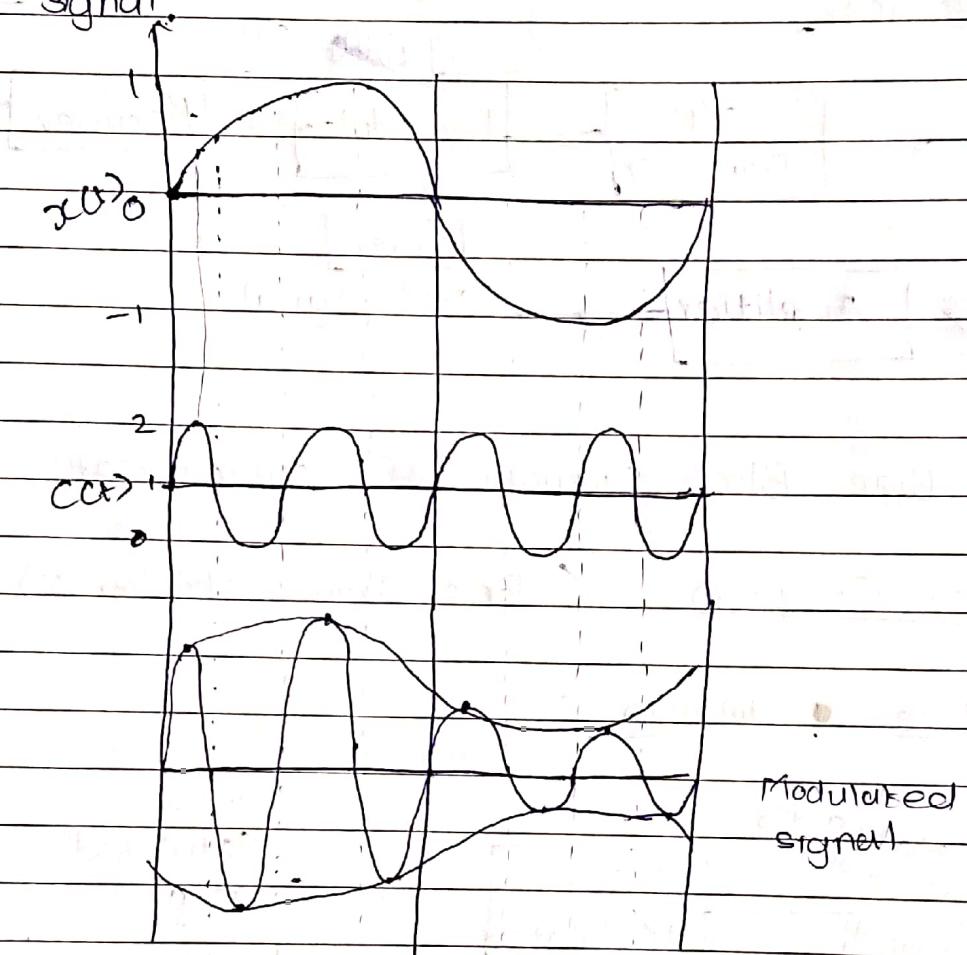
* Practical (10/07/23) (Topic from chapter 2)

Amplitude Modulation :-



- The change in amplitude of carrier signal ($C(t)$) with respect to instantaneous amplitude of message signal $x(t)$ by keeping frequency & phase of a carrier signal to be constant, this process is known as Amplitude Modulation.

Modulation → It is a process of conversion of low frequency signal into high frequency signal.



* Modulation Index of AM :

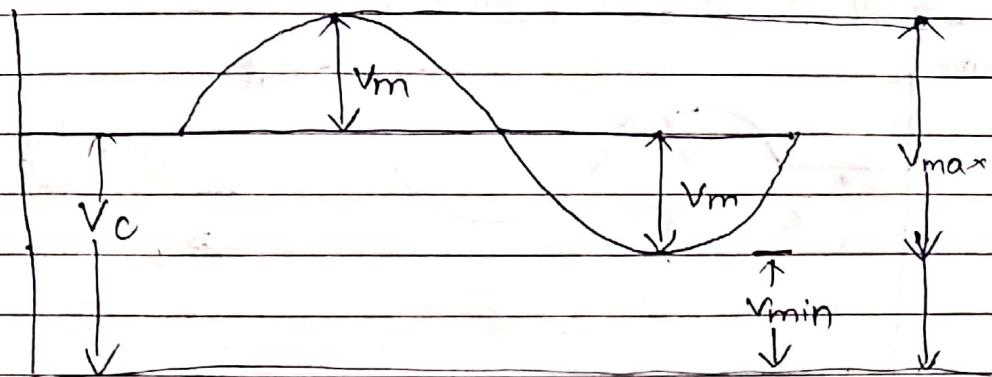
$$m = \frac{V_m}{V_c}$$

- It is a ratio of amplitude of message signal to the amplitude of a carrier signal.
- Generally, Modulation index of AM is less than $\neq 1$.

i) Under Modulation → $|m| << 1$ & $|V_m| \ll V_c$

ii) Perfect Modulation → $m = 1$ & $V_m = V_c$

iii) Over Modulation → $|m| > 1$ & $|V_m| \gg V_c$



Modulation Index :- (of air)

$$m = \frac{V_m}{V_c}$$

~~$V_c = V_m + V_{min}$~~

$$\therefore V_m = V_c - V_{min} = f \times$$

$$V_c = V_{max} - V_m$$

$$V_c = V_{max} - V_c + V_{min}$$

$$\boxed{V_c = \frac{V_{max} + V_{min}}{2}} \quad -(1)$$

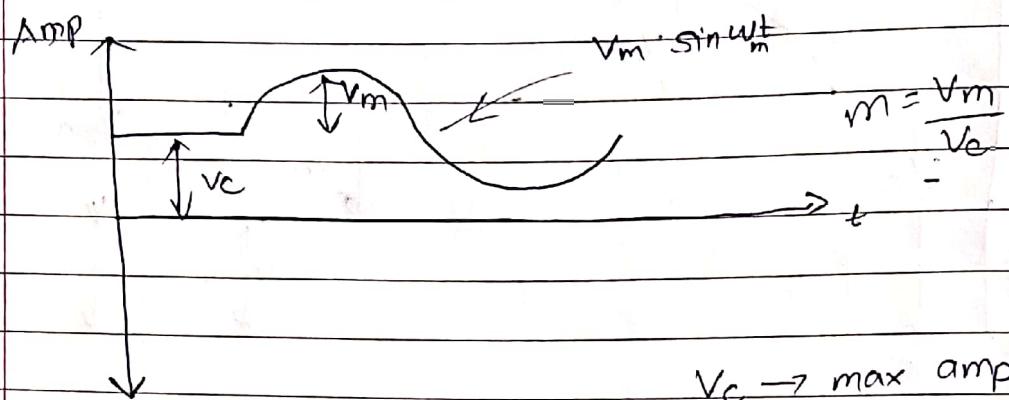
$$V_m = \frac{V_{max} + V_{min} - V_{min}}{2}$$

$$= \frac{V_{max} + V_{min} - 2V_{min}}{2}$$

$$\boxed{V_m = \frac{V_{max} - V_{min}}{2}} \quad -(2)$$

$$\boxed{m = \frac{V_{max} - V_{min}}{V_{max} + V_{min}}}$$

* Derivation of AM signal :-



$$V_c = V_c \sin \omega_c t$$

$$V_m = V_m \sin \omega_m t$$

$$A = V_c + V_m$$

$$= V_c + V_m \sin \omega_m t$$

$$A = V_c + m V_c \sin \omega_m t$$

$$A = V_c (1 + m \sin \omega_m t)$$

$$= V_c \sin \omega_c t (1 + m \sin \omega_m t)$$

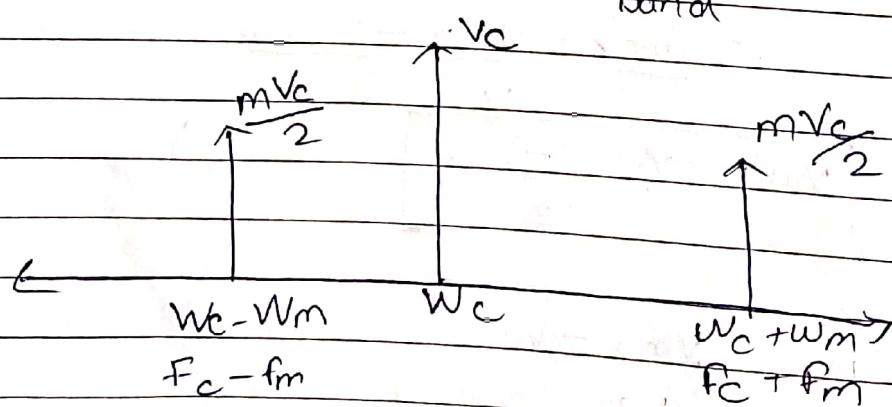
$$= V_c \sin \omega_c t + m V_c \sin \omega_m t \cdot \sin \omega_c t$$

$$= V_c \sin \omega_c t + \frac{m V_c}{2} [\cos(\omega_c - \omega_m)t + \cos(\omega_c + \omega_m)t]$$

Lower side band

Upper side band

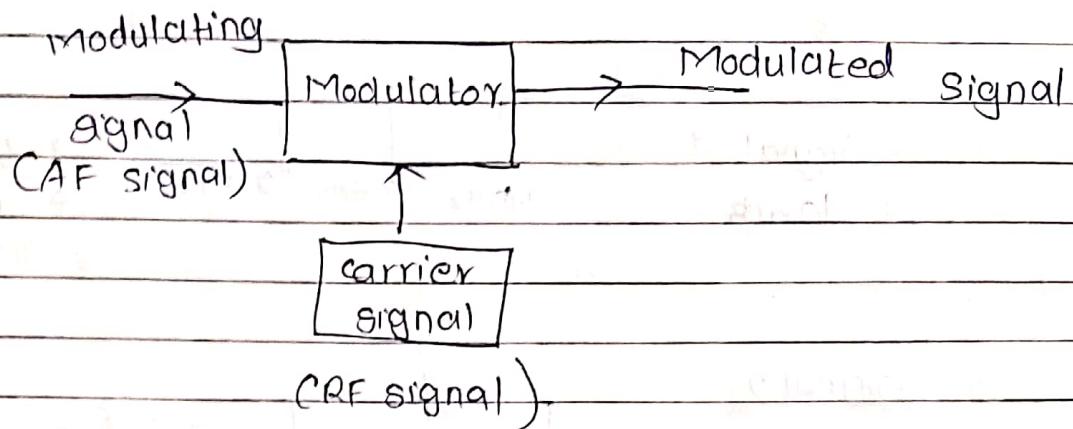
band



$$\text{BW} = f_c + f_m - f_c + f_m$$

$$\text{BW} = 2 P_c$$

* Modulation



Need of modulation :-

- To reduce the height of antenna

$$h = \frac{\lambda}{4}$$

$h \rightarrow$ height of antenna

$$\text{as, } \lambda = \frac{c}{f}$$

$\lambda \rightarrow$ Wavelength

$c \rightarrow$ speed of light

$$\therefore h = \frac{c}{4f}$$

$f \rightarrow$ Frequency,

$$h \propto \frac{1}{f}$$

case 1 :- If $F_1 = 10\text{ kHz}$

$$h_1 = \frac{3 \times 10^8}{4 \times 10^4} = 75\text{ km}$$

$$h_1 = 75\text{ km}$$

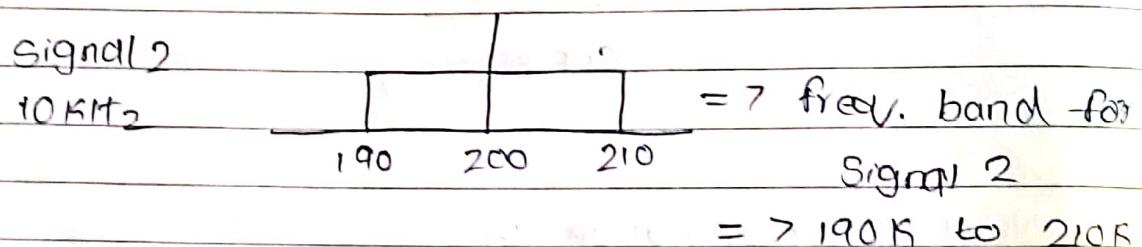
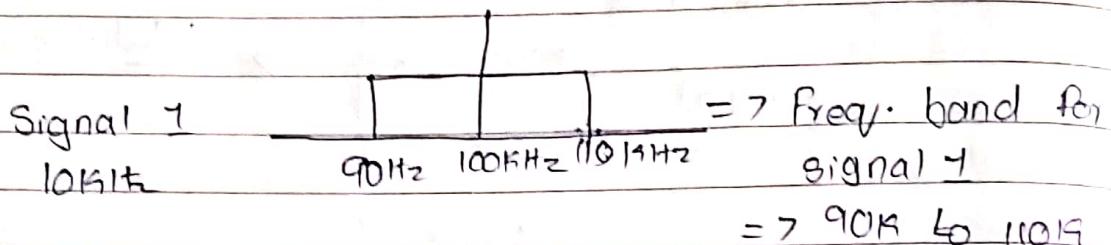
case 2 :- If $F_2 = 10\text{ MHz}$

$$h_2 = \frac{3 \times 10^8}{4 \times 10^7} = 75\text{ m}$$

As the frequency of signal increase then height of antenna decreases.

ii) To avoid mixing of the signal

100 KHz - 500 KHz



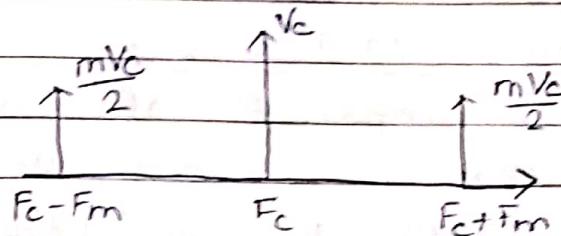
iii) Noise Minimize interfacing of noise :-

iv) Provide Multiplexing

v) Long distance communication

Page No.	/ /
Date	/ /

* Power calculation for AM wave :-



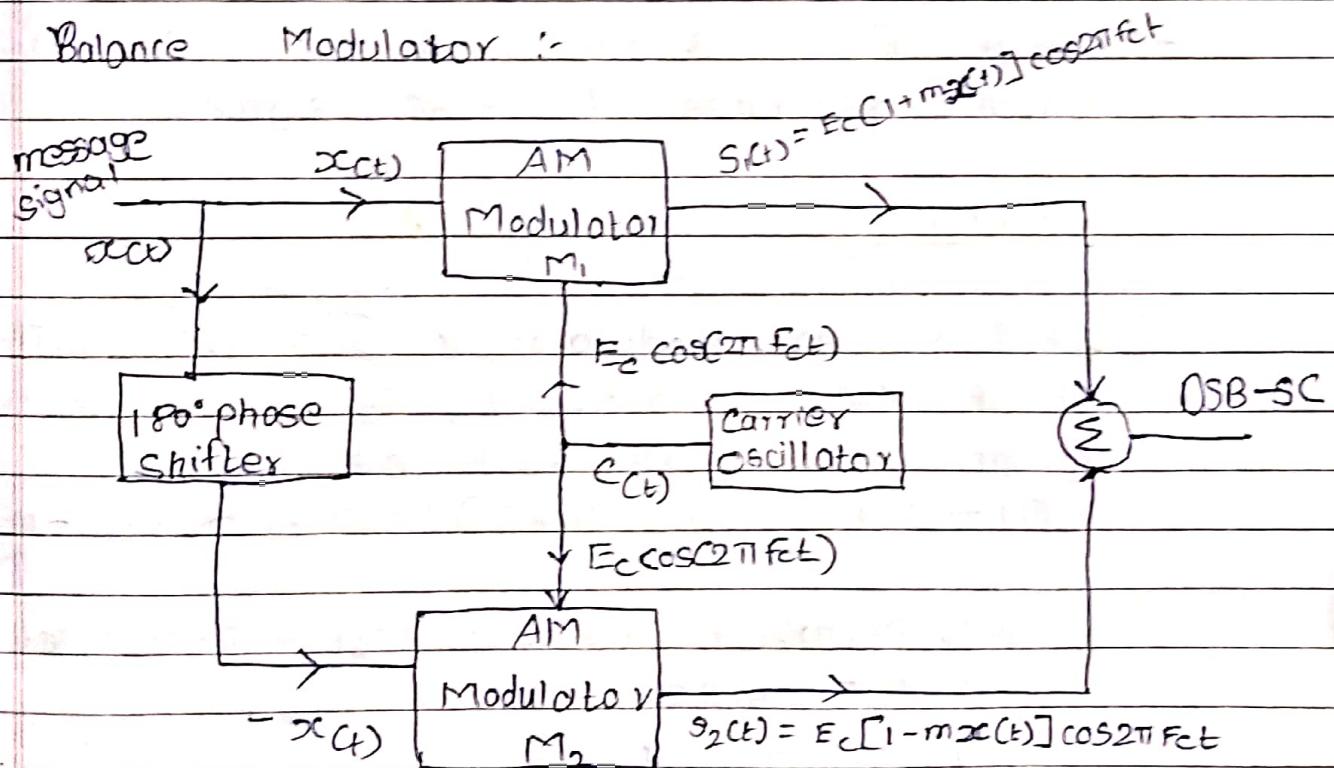
$$P_t = \frac{(mVc)^2}{4R} + \frac{Vc^2}{R} + \frac{(mVc)^2}{4R}$$

$$= \frac{Vc^2}{R} \left(1 + \frac{m^2}{4} + \frac{m^2}{4} \right)$$

$$\boxed{P_t = \frac{Vc^2}{R} \left(1 + \frac{m^2}{2} \right)}$$

* DSB-SC (Double side band suppress carrier)

Balance Modulator :-

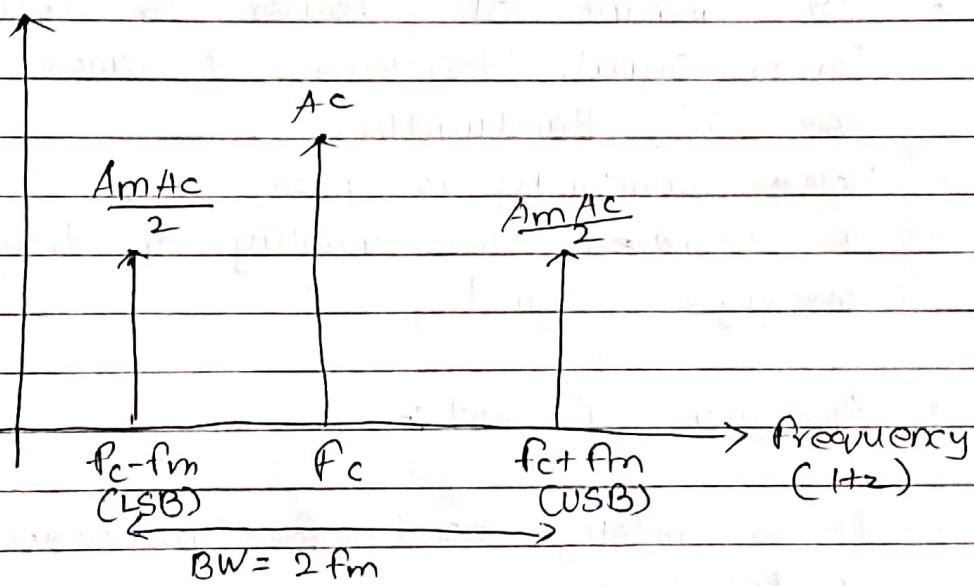
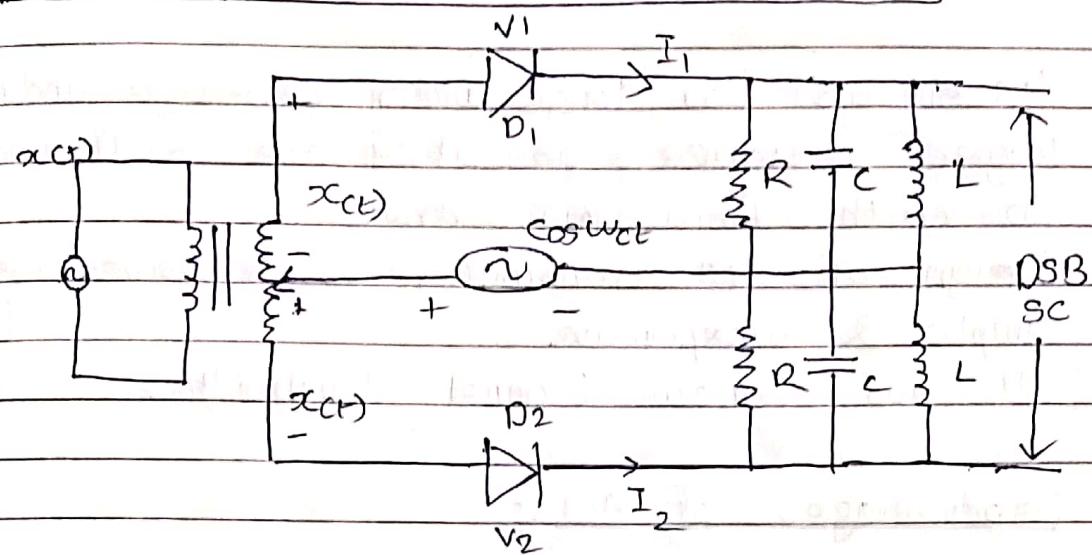


constructions →

- Principle of balance Modulator :-
- The principle of operation of balance modulator states that if two signals at diff. freq. are passed through a non-linear device at the o/p, we will get an AM signal with suppressed carrier.
- Construction :-
- Balance modulator consists of 2 AM modulators.
- For AM modulator 1, inputs are message signal $x_c(t)$ and carrier signal $c_c(t) = E_c \cos(2\pi f_c t)$.
- For AM modulator 2, inputs are 180° phase shifted message signal $(-x_c(t))$ and carrier signal $c_2(t) = E_c \cos(2\pi f_c t)$.
- Output of M₁ & M₂ given to adder blocks which generates DSB-SC signal.
- Working :-
- O/P of AM modulator 1, $s_1(t) = E_c [1 + m_x(t)] \cos 2\pi f_c t$
- O/P of AM modulator 2, $s_2(t) = E_c [1 - m_x(t)] \cos 2\pi f_c t$
- O/P of Adder block will be,
$$S(t) = s_1(t) - s_2(t) = E_c (1 + m_x(t)) \cos 2\pi f_c t - E_c (1 - m_x(t)) \cos 2\pi f_c t$$

$$= 2 E_c \cos 2\pi f_c t + m_x(t) E_c \cos 2\pi f_c t = E_c \cos 2\pi f_c t + 2 m_x(t) E_c \cos 2\pi f_c t$$

• Circuit diagram of balance Modulator is:



$$\omega_m = 2\pi f_m$$

$$m(t) = A_m \cos \omega_m t$$

$$c_{ct} = A_c \cos \omega_{ct}$$

- Advantages of AM :-

- AM signals has ~~large~~ wider coverage and travels longer distance as they are reflected back to earth from ionosphere.
- Design of AM transmitter & receiver is relatively simple & inexpensive.
- It has narrow channel bandwidth.

- Disadvantages of AM :-

- More amount of power is wasted in transmitting Carrier signal. Therefore, it causes inefficient use of Bandwidth.
- Noise immunity is poor.
- It deteriorates the quality of the original message signal.

- Application of AM :-

- It is mostly used for broadcasting by radio & television.
- In airband radio, used for ground to air communication to
- It is also used in Police & navy dispatch radio system.

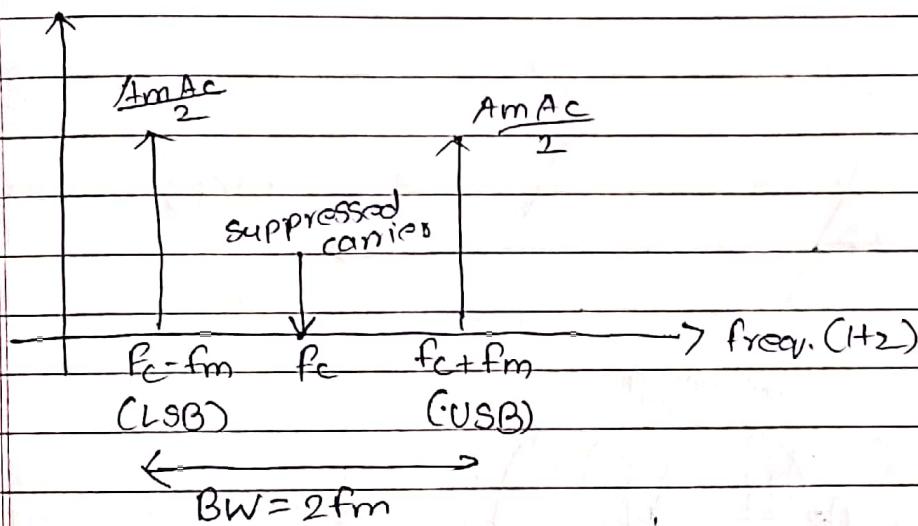
- Types of AM :-

- Depending on freq. components present in the spectrum, AM can be classified as :-

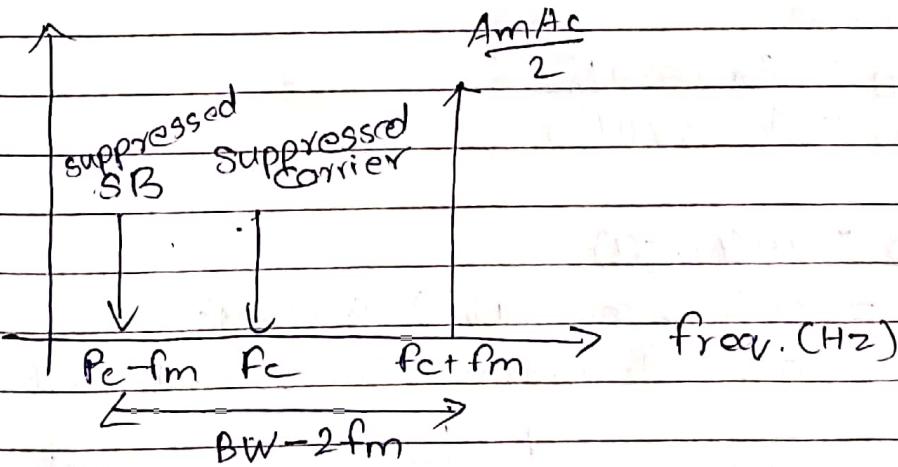
i) DSB - SC

ii) SSB - SC

i) DSB - SC :-

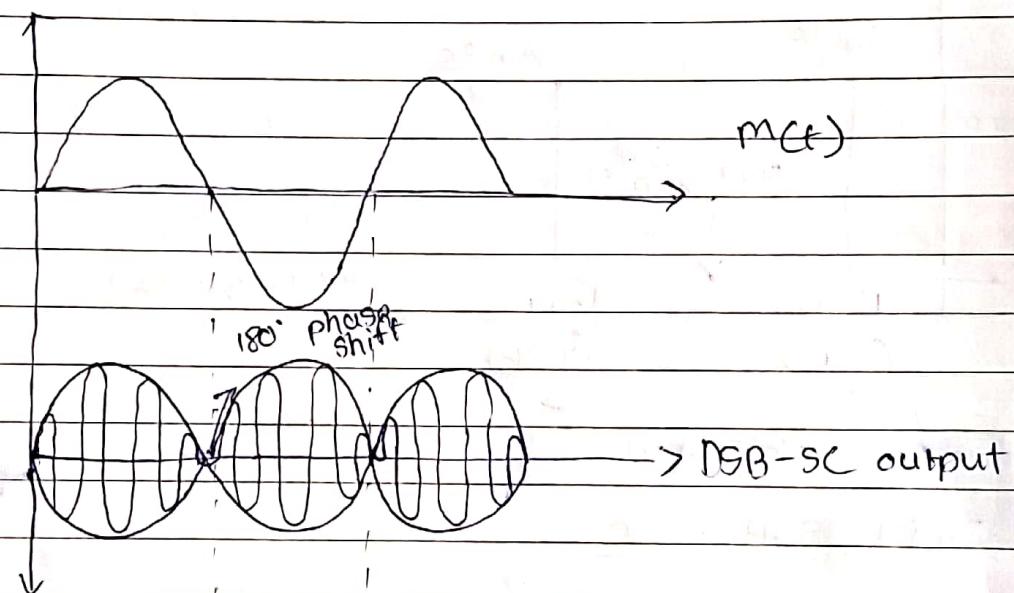


ii) SSB - SC



DSB-SC

- In DSB-SC only upper side band & lower side band are transmitted, while suppressing the carrier component.
- Since, carrier component does not transmit any message, it is suppressed and only LSB & USB are transmitted.
- Hence, total power is utilized only for transmission of side bands, resulting in modulation efficiency to be 100%.



$$m(t) = A_m \cos 2\pi f_m t$$

$$c(t) = A_c \cos 2\pi f_c t$$

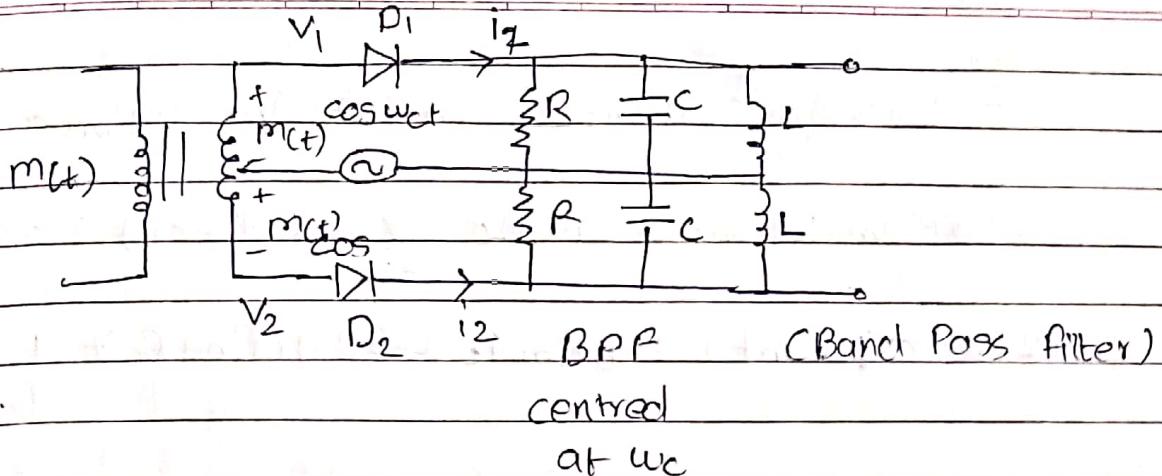
$$s(t) = m(t) \cdot c(t)$$

$$= A_c A_m \cdot \cos(2\pi f_m t) \cdot \cos(2\pi f_c t)$$

$$= \frac{A_c A_m}{2} \cos 2\pi (f_c + f_m) t + \frac{A_c A_m}{2} \cos 2\pi (f_c - f_m) t$$

USB

LSB



$$V_1 = m(t) + \cos w_c t$$

$$= A_m \cos w_m t + \cos w_c t$$

$$V_2 = m(t) - \cos w_c t$$

$$= A_m \cos w_m t - \cos w_c t$$

$$i_1 = a V_1 + b V_1^2 ; \quad i_2 = a V_2 + b V_2^2$$

$$\begin{aligned} i_1 &= a [A_c \cos w_c t + m(t)] + b [A_c \cos w_c t + m(t)]^2 \\ &= a A_c \cos w_c t + a m(t) + b A_c^2 \cos^2 w_c t + b m^2(t) + \\ &\quad 2 b A_c \cos w_c t m(t) \end{aligned}$$

$$\begin{aligned} i_2 &= a [A_c \cos w_c t - m(t)] + b [A_c \cos w_c t - m(t)]^2 \\ &= a A_c \cos w_c t - a m(t) + b A_c^2 \cos^2 w_c t + m^2(t) \\ &\quad - 2 b A_c \cos w_c t m(t) \end{aligned}$$

$$\begin{aligned} V_o &= i_1 R - i_2 R \\ &= R [i_1 - i_2] \\ &= R [2 a m(t) + b A_c \cos w_c t m(t)] \\ &\quad \text{message signal} \quad \text{DSB-SC signal} \\ &\quad [\text{modulated}] \end{aligned}$$

$$= 2aR_Am \cos(\omega_m t) + 4bR_Ac A_m \cos(\omega_m t + \omega_c t)$$

$$= 2aR_Am \cos(\omega_m t) + 2bR_Ac (\cos(\omega_m t + \omega_c t) + \cos(\omega_m t - \omega_c t))$$

$$= 2R_Am \cos(\omega_m t) + 2R_b A_m A_c \cos 2\pi(f_m + f_c)t + 2R_b A_m A_c \cos 2\pi(f_m - f_c)t$$

On Passing this (f_m) through LC Band Pass Filter,
the modulating ^{signal} term will be eliminated.

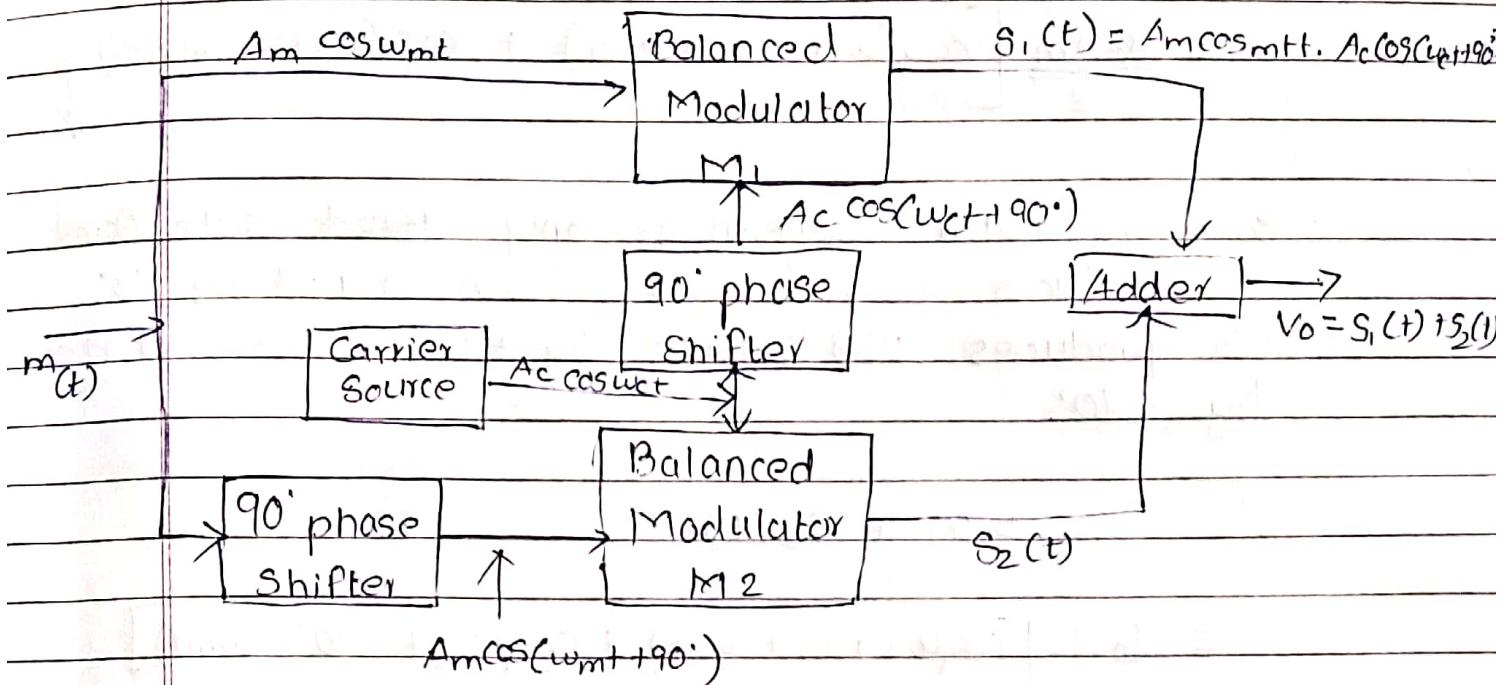
Advantages :-

- i) Transmission power is same as the carrier is supposed.
- ii) Generation & detection is relatively simple & inexpensive.

Disadvantages :

- i) Wastage of Power in Transmitting to side Band which carrying same information.
- ii) Less noise immunity.
- iii) Require same bandwidth as AM i.e $BW = 2f_w$

* Generation of SSB-SC using phase shift method:-



- Figure shows, block diagram of phase shift method of to general USB of SSB-SC (LSB suppression).
- It consists of two balanced modulators M_1 & M_2 .
- For $M_1 \rightarrow$ Carrier signal shifted by 90° + Modulating signal is applied.
- For $M_2 \rightarrow$ Carrier signal + Modulating signal shifted by 90° .

$$2\cos A \cos B = \cos(A+B) + \cos(A-B)$$

$$S_1(t) = A_m \cos \underline{\omega_{mt}} \cdot A_c \cos \underline{\omega_{ct} + 90^\circ}$$

$$= \frac{A_m A_c}{2} [\cos(\omega_{ct} + 90^\circ + \omega_{mt}) + \cos(\omega_{ct} + 90^\circ - \omega_{mt})]$$

$$S_2(f) = A \cos(\omega_c t) A_m \cos(\omega_m t + 90^\circ)$$

$$= \frac{A \cdot A_m}{2} [\cos(\omega_c t + \omega_m t + 90^\circ) + \cos(\omega_c t - \omega_m t - 90^\circ)]$$

$$= \frac{A \cdot A_m}{2} [\cos(\omega_c t + \omega_m t + 90^\circ) + \cos(\omega_c t - \omega_m t - 90^\circ)]$$

- Both modulator generates only ~~band~~ side bands.
- M₁ produces USB & LSB, Both shifted by 90°.
- M₂ produces USB shifted by +90° & LSB shifted by -90°.

$$\therefore V_o = S_1(f) + S_2(f)$$

$$= \frac{A_m A_c}{2} [\cos(\omega_c t + 90^\circ + \omega_m t) + \cos(\omega_c t + 90^\circ - \omega_m t)]$$

$$+ \frac{A_m A_c}{2} [\cos(\omega_c t + \omega_m t - 90^\circ) + \cos(\omega_c t - \omega_m t - 90^\circ)]$$

$$\cos(90^\circ + \theta) = -\sin \theta$$

$$\cos(-\omega_c t + \omega_m t + 90^\circ)$$

$$\cos(\omega_c t + \omega_m t + 90^\circ)$$

$$= \frac{A_m A_c}{2} [-\sin(\omega_c t + \omega_m t) + \sin(\omega_c t - \omega_m t)]$$

$$+ \frac{A_m A_c}{2} [-\sin(\omega_c t + \omega_m t) - \sin(\omega_c t - \omega_m t)]$$

$$= \frac{A_m A_c}{2} [-\sin(\omega_c t + \omega_m t) - \sin(\omega_c t - \omega_m t)]$$

$$\sin(\omega_c t + \omega_m t)$$

$$+ \frac{A_m A_c}{2} [-\sin(\omega_c t + \omega_m t) + \sin(\omega_c t - \omega_m t)]$$

$$= -\frac{A_m A_c}{2} [2 \sin(\omega_c t + \omega_m t)]$$

$$= -A_m A_c \sin(\omega_c t + \omega_m t)$$

$$\therefore V_o = A_m A_c \cos(\omega_c t + \omega_m t + 90^\circ)$$

- Output of Balanced modulator are added by summing amplifier.
- two USBs are in phase so they are added, two LSBs are out of phase. (180° phase diff) so they cancel each other.

* For USB suppression :-

- The phase shifted version of both the carrier and modulating signal are fed to one of the modulator.
(Feed 90° shifted signals to M_2 & plain to M_1)

* Advantages of SSB

- Required bandwidth is smaller in comparison with AM & PSB-SC.
- As only one power is saved as only one band is to be transmitted.
- Will give more noise immunity / Amount of noise is less.
- Less loss of signal.

* Disadvantages of SSB-SC

- i) Detection & generation process of SSB is complex.
- ii) Signal quality will degrade.

* Applications of SSB-SC

- Air, land, marine communication.
- Radio Broadcasting.
- It is used TV & radar.
- It is used in Military communication.

$$c(t) = A \cos(\omega t + \phi)$$

$A \rightarrow$ amplitude \rightarrow AM
 $\omega \rightarrow$ angular freq. \rightarrow FM
 $\phi \rightarrow$ phase angle \rightarrow Angle modulation.

FM \rightarrow frequency of ^{msg} carrier signal changes w.r.t carrier signal by keeping amplitude and phase constant.

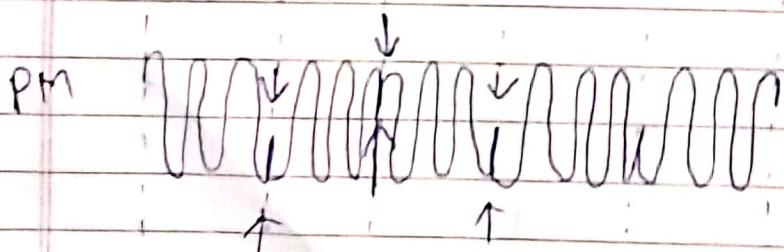
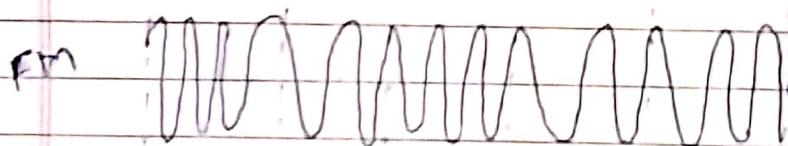
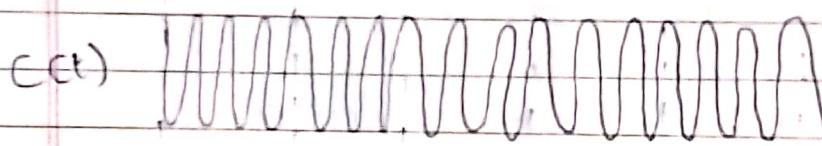
$$y(t) = A_c \cos(\omega_c t + \phi)$$

where $\omega_c = F[m(t)]$

PM \rightarrow

$$y(t) = A_c \cos(\omega_c t + \phi(t))$$

where $\phi(t) = F[m(t)]$



* Advantages of Angle Modulation:

- i) Noise Reduction
- ii) Better system fidelity
- iii) Efficient use of power
- iv) Operates in very high frequency range
(88 MHz - 108 MHz)

* Disadvantages

- Increased bandwidth.
- Used of complex circuits.

* Applications of Angle Modulation

- i) Radio Broadcasting.
- ii) TV sound transmission.
- iii) Cellular Radio.
- iv) Microwave Communication.
- v) Satellite communication.

* Frequency Modulation :-

$$\text{Unmodulated carrier } c_{(t)} = A \cos(\omega_c t + \theta) \quad (i)$$

$$c_{(t)} = A \cos \phi \quad (ii)$$

where $\phi = \omega_c t + \theta$

ϕ is the total phase angle of the unmodulated carrier.

FM wave :- $s_{(t)} = A \cos \phi;$
 $(\phi; \rightarrow \text{Instantaneous value of the})$
 phase angle

$$\Phi = \omega_0 t + \theta$$

$$\frac{d\Phi}{dt} = \omega_0 \rightarrow \Phi = \int \omega_0 dt - \textcircled{3}$$

Instantaneous frequency $\Rightarrow \omega_i$

$$\omega_i = \omega_0 + k_f m(t)$$

$$\Phi_i = \int \omega_i dt$$

$$= \int (\omega_0 + k_f m(t)) dt$$

$$\boxed{\Phi_i = \omega_0 t + k_f \int m(t) dt} - \textcircled{4}$$

$$s_a(t) = A \cos(\omega_0 t + k_f \int^t m(u) du) - \textcircled{5}$$

~~s_a(t) = A \cos(\omega_0 t + k_f \int^t m(u) du)~~
where ω_0

- Frequency sensitivity of modulator is expressed in Hz/bolt
- This is expression for frequency modulated signal in time domain
- Eqn. 5 shows $s_a(t)$ is a non linear function of $m(t)$ which makes F.M or non linear modulation technique

from 4,

$$\omega_i = \omega_0 + k_f m(t)$$

$$f_i = f_0 + k_f A_m \cos(2\pi f_m t)$$

Δf

$\Delta f \rightarrow$ frequency deviation which represents the difference between min & max amplitude of FM & signal & the carrier frequency.

$$\Delta f \propto A_m \quad \therefore \Delta f = K_f A_m$$

$$\therefore \Delta f \propto A_m$$

Δf is directly proportional A_m but ~~is not independent~~ of f_m .

From 3,

$$\phi = \int_0^t \omega(t) \cdot dt$$

$$\phi_i = 2\pi \int_0^t f_i(t) \cdot dt$$

$$= 2\pi \int_0^t (f_c + \Delta f \cos(2\pi f_m t)) dt$$

$$\phi_i = 2\pi f_c t + \frac{\Delta f \sin(2\pi f_m t)}{2\pi f_m}$$

$$\phi_i = 2\pi f_c t + \left(\frac{\Delta f}{f_m} \right) \sin(2\pi f_m t)$$

$$\phi_i = 2\pi f_c t + \beta \sin 2\pi f_m t$$

$$s(t) = A_c \cdot \cos(2\pi f_c t + \beta \sin 2\pi f_m t)$$

$\beta \rightarrow$ represents phase deviation of fm signal i.e. max dif. of fm of ϕ_i from the angle $2\pi f_c t$

Therefore, β is measured in radians.

$$\beta < 1 \rightarrow \text{NBFM}$$

$$\beta \geq 1 \rightarrow \text{WBFM}$$

Diff. betn Narrow band fm & wide band fm.

Parameters	Narrow band	Wide band
i) Modulation Index (β)	less than 1	Greater than 1
ii) Max. deviation	5 kHz	75 kHz
iii) Range of modulated freq.	30 Hz - 3 kHz	30 Hz - 15 kHz
iv) Band width	Small ($BW = 2f_m$) (approx. same as of A_m)	Large by 15 times greater than NBFM $[BW = 2(\Delta f + f_m)]_{\text{max}}$
v) Applications	Entertainment broadcasting / high quality music transmission.	FM mobile communication like police wireless.

* Spectral Analysis of FM :-

$$S(t) = A_c \cos(2\pi f_c t + \beta \sin 2\pi f_m t)$$

expanding $\cos(A+B)$

$$\text{using } \cos(A+B) = \cos A \cos B - \sin A \cdot \sin B$$

$$S(t) = A_c \left[\cos(2\pi f_c t) \cdot \cos(\beta \sin 2\pi f_m t) - \sin(2\pi f_c t) \cdot \sin(\beta \sin 2\pi f_m t) \right]$$

In the above eqn, we have terms having or consisting of $\cos(\sin)$ and $\sin(\sin)$, which on expansion gives complex terms.

The soln to such equation can be found using bessel's function.

$$S(t) = A_c \left\{ J_{-n}(\beta) \cos(w_r - n w_m)t \right\}$$

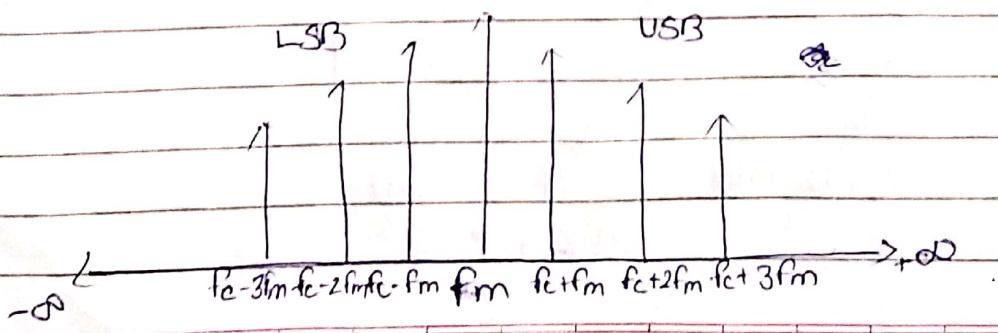
where $J_{-n}, \dots, J_2, J_1, J_0, J_1, J_2, \dots, J_n$ are Bessel's function.

$J_0 \rightarrow$ Amplitude of carrier signal

$J_1 \rightarrow$ 1st USB & LSB

$J_2 \rightarrow$ 2nd USB & LSB

$J_n \rightarrow$ Amplitude of n^{th} side band.



PAGE NO.	11
DATE	

- Amplitude of each side band depends on the Bessel's function and modulation index β .
- Amplitude of each side band is determined by using Bessel table.
- In FM, theoretically there are infinite no. of sidebands and therefore bandwidth in FM is also finite.
- FM bandwidth can be determined by using Bessel's Table or Carson's Rule.
- FM bandwidth by using Table: $BW = 2n f_{max}$ where n is no. of signif.

(Q) In a FM system, when the audio frequency is ~~is~~ 400Hz, the AF voltage is 4V, the deviation is 4.8KHz. Calculate the modulation Index and the bandwidth required. If the modulation frequency is ~~is~~ half, what is the bandwidth?

Given :- $f_m = 400\text{Hz}$, $A_m = 4\text{V}$, $\Delta f = 4.8\text{KHz}$

To find :- $\beta = ?$

Soln:-

$$\beta = \frac{\Delta f}{f_m}$$

$$= \frac{4.8 \times 10^3}{400} = \frac{12}{10} = 1.2$$

$$\boxed{\beta = 1.2}$$

$$\therefore BW = 2(\Delta f + f_m)$$

$$= 2(4.8 \times 10^3 + 400)$$

$$= 10400 = 10.4\text{KHz}$$

Q) A 100 MHz carrier signal is frequency modulated by an analog modulating signal. The Max frequency deviation is 100 kHz. Determine the Approx. transmission B.W of fm signal using Carson's B.W rule. If the freq. of modulating signal is i) 1 kHz ii) 500 kHz.

Given :- $f_c = 100 \text{ MHz}$

$$\Delta f = 100 \text{ kHz}$$

i) $f_m = 1 \text{ kHz}$

$$BW = 2(\Delta f + f_m)$$

$$= 2(100 \text{ K} + 1 \text{ K})$$

$$= 2(101 \text{ K})$$

$$= 202 \text{ kHz}$$

ii) $f_m = 500 \text{ kHz}$

$$BW = 2(\Delta f + f_m)$$

$$= 2(100 \text{ K} + 500 \text{ K})$$

$$= 1200 \text{ kHz}$$

Q) A modulating signal $15 \cos(2\pi 15 \times 10^3 t)$ angle modulates a carrier. Find modulation index and B.W of the system.

i) Determine the change in B.W & Modulation Index for FM, if modulating signal frequency is reduced to 5 kHz. Assume $K_f = 15 \text{ kHz/V}$

ii) Find B.W & modulation Index, IF Amplitude is reduced to half.

SOLN:-

$$\frac{15 \cos(2\pi 15 \times 10^3 t)}{A_m} \quad W_m \rightarrow \quad A_m = 15, \quad W_m = 2\pi 15 \times 10^3$$

$$\therefore f_m = 15 \times 10^3$$

$$\Delta f = K_f A_m$$

$$= 225 \text{ kHz}$$

$$\therefore \beta = \frac{\Delta f}{f_m} = \frac{225 \times 10^3}{15 \times 10^3} = 15$$

$$B.W = 2(\Delta f + f_m)$$

$$= 2(225 \text{ K} + 15 \text{ K})$$

$$= 480 \text{ K}$$

$$BW = 2(225 \times 10^3 + 15 \times 10^3)$$

i) $f_m' = 5 \text{ kHz}$

$$\beta' = \frac{\Delta f}{f_m} = \frac{225 \text{ K}}{5} = 45'$$

$$\begin{aligned} B.W' &= 2(\Delta f + f_m') \\ &= 2(225 \text{ K} + 5 \text{ K}) \\ &= 2(230 \text{ K}) \\ &= 460 \text{ kHz} \end{aligned}$$

$$\text{change in } \beta = 45 - 15 = 30$$

$$\text{change in } BW = 20 \text{ kHz}$$

ii) If amplitude is Half

$$\therefore A_m = 7.5$$

$$BW = 2(\Delta f + f_m)$$

$$= 2(112.5 \text{ K} + 15 \text{ K})$$

$$= 2(127.5 \text{ K})$$

$$BW = 255 \text{ kHz}$$

$$\Delta f = k_p \cdot A_m$$

$$= 15 \times 7.5$$

$$= 112.5 \text{ K}$$

$$\begin{aligned} \beta &= \frac{\Delta f}{f_m} = \frac{112.5 \times 10^3}{15 \times 10^3} \\ &= 7.5 \end{aligned}$$



Q) A single tone fm signal is given by

$$e_{FM}(t) = 20 \cos(16\pi 10^6 t + 25 \sin 2\pi 10^3 t)$$

Find Modulation Index, modulating freq, deviation carrier freq & power in fm signal.

Soln:-

$$e_{FM}(t) = 20 \cos(16\pi 10^6 t + 25 \sin 2\pi 10^3 t)$$

\downarrow
 A_m

\downarrow
 w_m

$$A_m = 20$$

compare with,

$$s(t) = A_c \cos(\omega_c t + \beta \sin w_m t)$$

$$A_c = 20; \omega_c = 16\pi 10^6; \beta = 25, w_m = 2\pi 10^3$$

$$\therefore f_c = \frac{\omega_c}{2\pi} = \frac{16\pi \times 10^6}{2\pi} = 8 \text{ MHz}$$

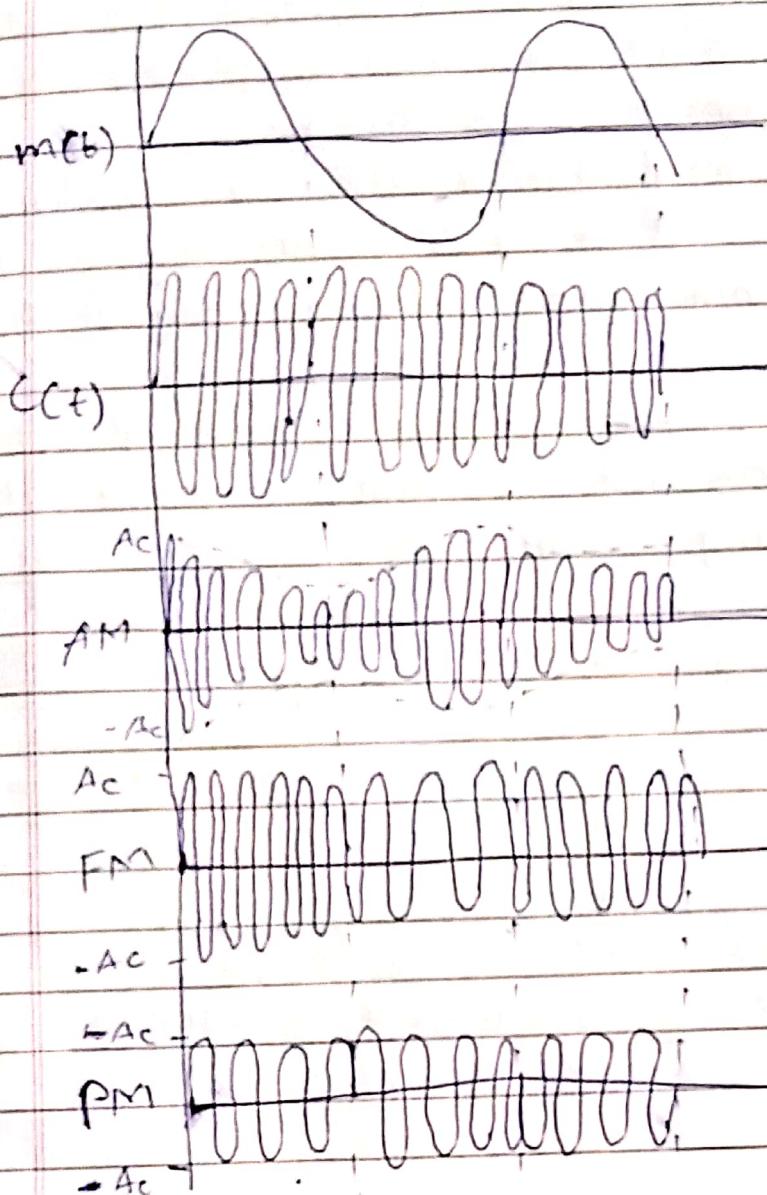
$$\therefore f_m = \frac{w_m}{2\pi} = \frac{2\pi 10^3}{2\pi} = 1 \text{ kHz}$$

$$\therefore \Delta f = \beta f_m \\ = 25 \times 1 \text{ kHz} = 25 \text{ kHz}$$

$$\therefore P_{FM} = \frac{A_c^2}{2} = \frac{400}{2} = 200$$

(Power)

Parameter	Am	Fm	Pm
Variable Parameter of carrier wave	Amplitude	Frequency	Phase (angle)
Mathematical eqn for S.C.I	$A_c \cos(\omega_c t + m_p \sin \omega_m t)$ $m A_c \cos(\omega_c - \omega_m)t$ $\frac{m^2 A_c}{2} \cos(\omega_c + \omega_m)t$	$A_c \cos(\omega_c t + m_p \sin \omega_m t)$	$A_c \cos(\omega_c t + m_p \sin \omega_m t)$
Noise immunity	Low / Poor	Better	Better
B.W	2 fm (constant)	$2(\Delta f + f_m)$ (Carson's Rule)	Moderate
no. of sidebands	2	infinite - 2 (Ideally)	More than 2 side bands
Application	Radio, TV broadcasting	Radio, TV & Police wireless.	Data communication
Signal to noise ratio	Poor	Best	Better



* Noise Triangle FM :-

- Noise Triangle is a study effect of noise on the carrier signal of the FM way.
- It is a triangular noise distribution for FM in which signal/noise ratio versus frequency is plotted.
- The carrier and noise voltages will mix and if the difference is significant (audible); it will naturally interfere with reception of message signal producing an unwanted deviation of carrier freq.

NAME	
DATE	/ /

- The magnitude of this unwanted frequency deviation depends on the amplitude of noise with respect to carrier.
- When this unwanted carrier deviation is demodulated it becomes noise if it has frequency components that falls within the message signal frequency spectrum.
- The effect of noise at the output of FM demodulator does not remain constant, it increases linearly with the increase in FM.

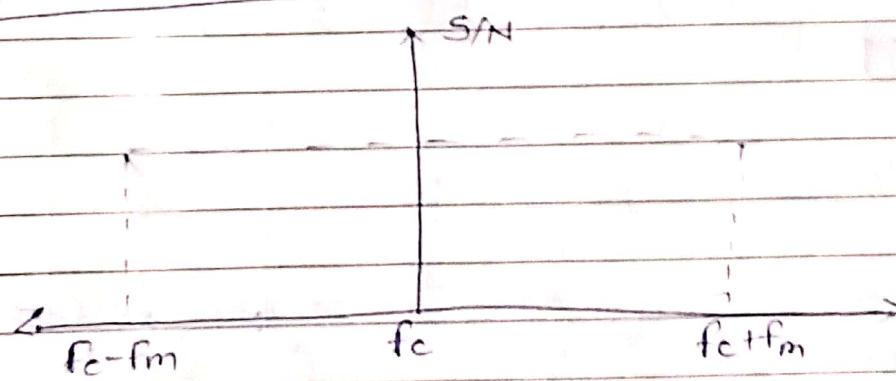
- The demodulated noise voltage is higher for higher modulating (msg) signal freq.

- as the modulating freq fm increases, $\beta \downarrow$ & effect of noise for lower $f_c - f_m$

modulation index is more. Therefore to reduce effect of noise on higher freq. of msg signal it β should be constant.

- which is possible by increasing freq. deviatn. Noise Spectrum for in Fm

$$\beta = \frac{\Delta f}{f_m}$$



Noise distribution for AM & PPM

S/N remains constant for all modulating freq.

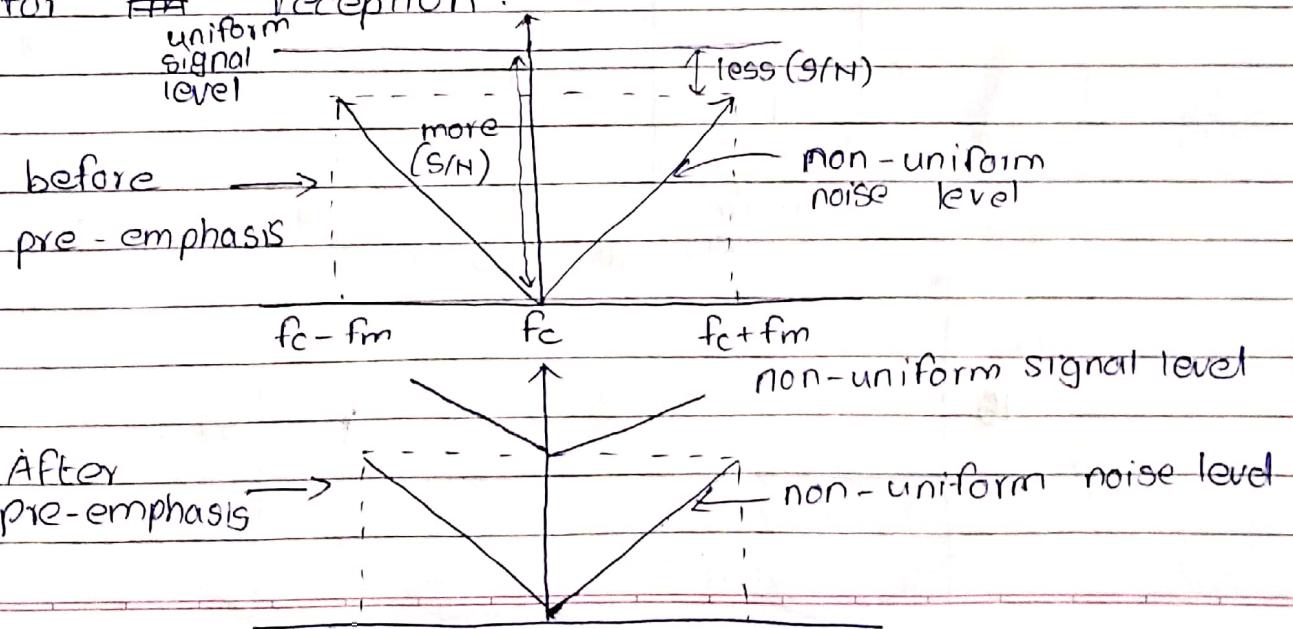
* Pre-emphasis & De-emphasis:

- An alternative way to increase the signal to noise ratio (SNR)

$$\begin{array}{c} \text{(Noise Voltage)} \\ \text{high freq.} \end{array} > \begin{array}{c} \text{(Noise Voltage)} \\ \text{low freq.} \end{array}$$

$$\begin{array}{c} \text{(S/N)} \\ \text{high freq.} \end{array} < \begin{array}{c} \text{(S/N)} \\ \text{low freq.} \end{array}$$

- To compensate this high modulated freq. (FM) (2KHz - 15Khz) are boosted (emphasised) in amplitude in the transmitter before performing modulation.
- This artificial boosting is called pre-emphasis.
- To neutralize (compensate) for this boost the high frequency signal are attenuated (de-emphasized) at the receiver after demodulation. This is called de-emphasis.
- Objective of these two technique is to ~~increas~~ improve signal to noise ratio SNR for FM^{FM} reception.



- Q) An AM modulator modulates modulating signal of 25 kHz, 10 Vp with carrier signal of 800 kHz, 40 Vp. Calculate i) USB & LSB frequency.
ii) Modulation coefficient &
iii) Draw O/P freq. spectrum.
iv) Draw the sig. envelope & label it.

Given : $A_m = 10 \text{ V}$

$$f_m = 25 \text{ kHz}$$

$$f_c = 800 \text{ kHz}$$

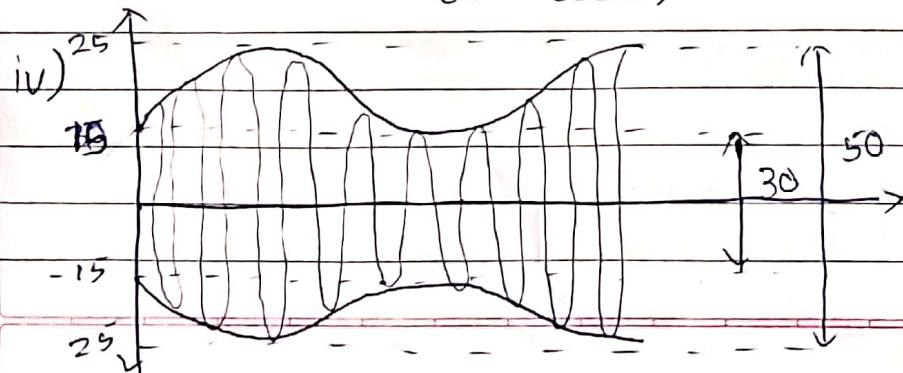
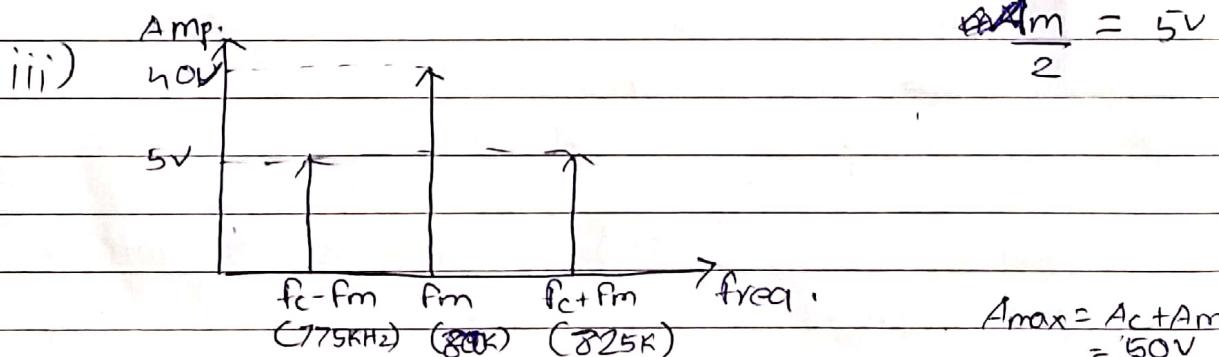
$$A_c = 40 \text{ V}$$

$$\text{i) } f_{USB} = f_c - \frac{A_c^2}{2R} \rightarrow f_{LSB} = f_c - f_m \\ = 800 \text{ kHz} - 25 \text{ kHz} \\ = 775 \text{ kHz}$$

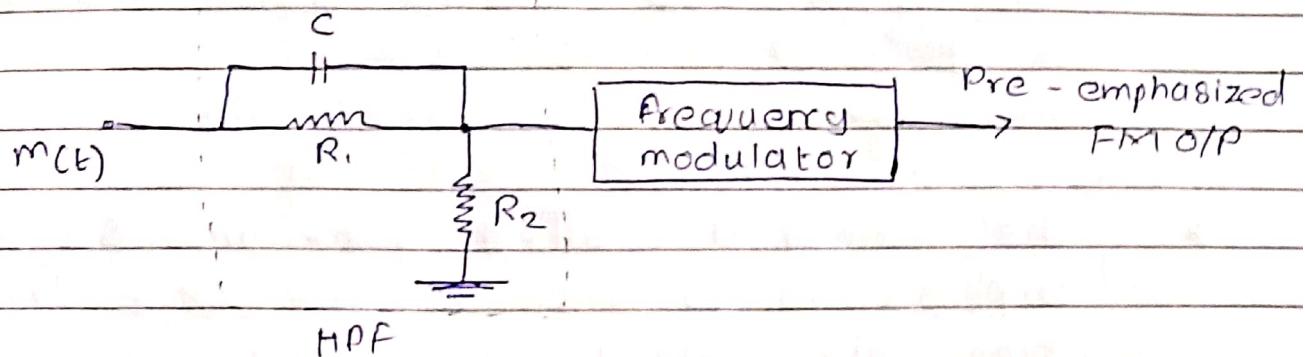
$$f_{USB} = f_c + f_m = 800 \text{ kHz} + 25 \text{ kHz} = 825 \text{ kHz}$$

$$\text{ii) } m = \frac{A_m}{A_c} = \frac{10}{40} = 0.25$$

$$\therefore m = 25\%$$



* Pre-emphasis circuit :-

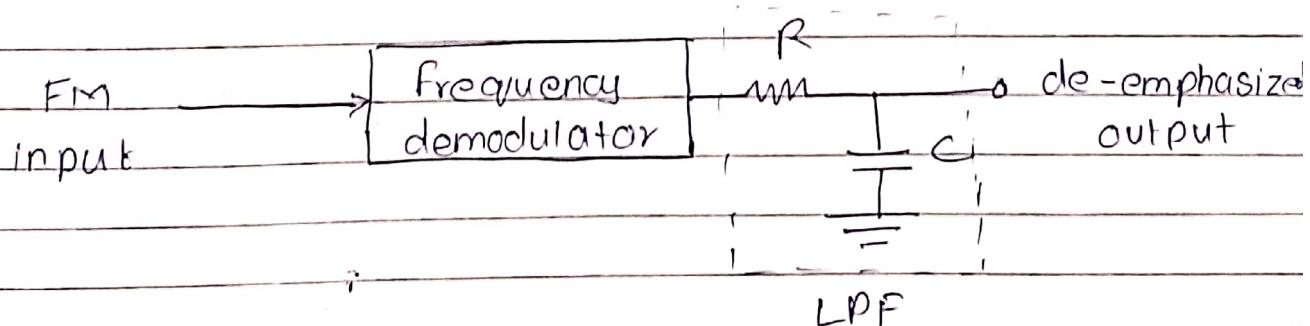


- Let the transmitter $m(t)$ is passed through a high pass filter which amplifies high frequency components more than low freq. components.

- + Specifications :- time constant, $T = 75\text{ }\mu\text{s}$
 $f = \frac{1}{T} = 13.3\text{ kHz}$

- Any combination of RNC giving this time constant can be used.
- Such a circuit has a cut off frequency means frequencies higher than f_{cutoff} will be linearly enhanced or boosted.

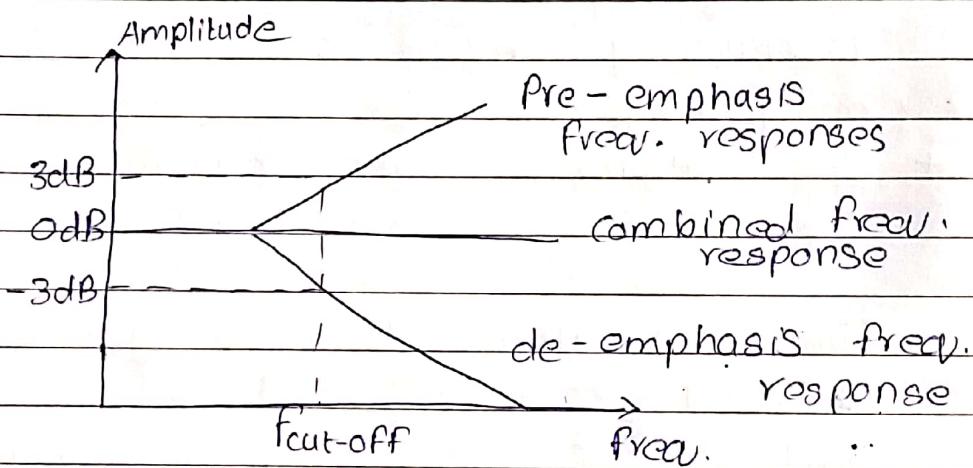
* De-emphasis Circuit:-



- At the receiver, the demodulated o/p is passed through a low pass filter with Time

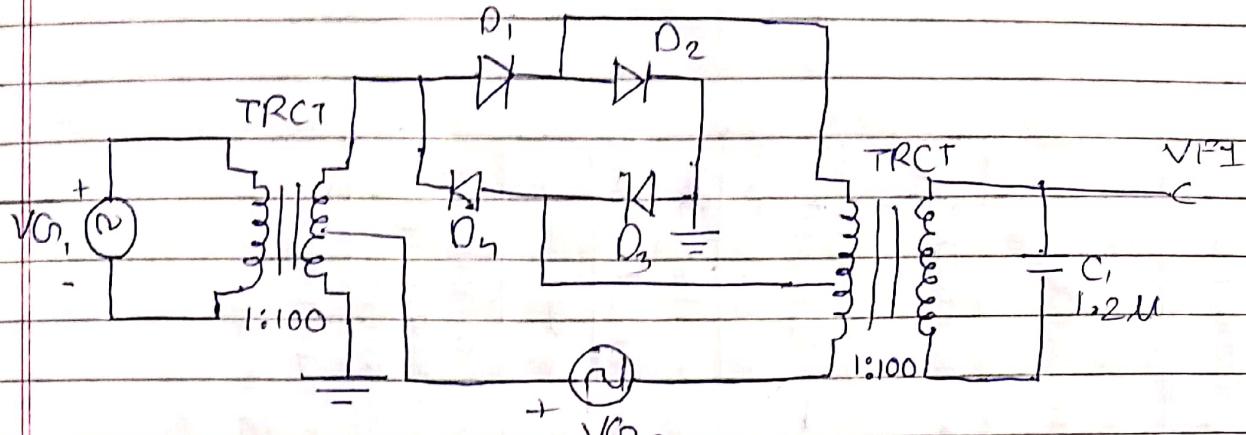
constant of 75.15.

- Such a circuit features a cut-off frequency that causes ~~singt~~ signals above this frequency to ~~be~~ attenuated.
 - ~~The rem~~
- * The combined effect pre-emphasis & de-emphasis is to increase the high frequency components during transmission so that they will be stronger and ~~not~~ masked by the noise.



Exp - P3

IN 4007



/ / open loop sys

V_G1 : 5, 10K

V_G2 : 1, 115

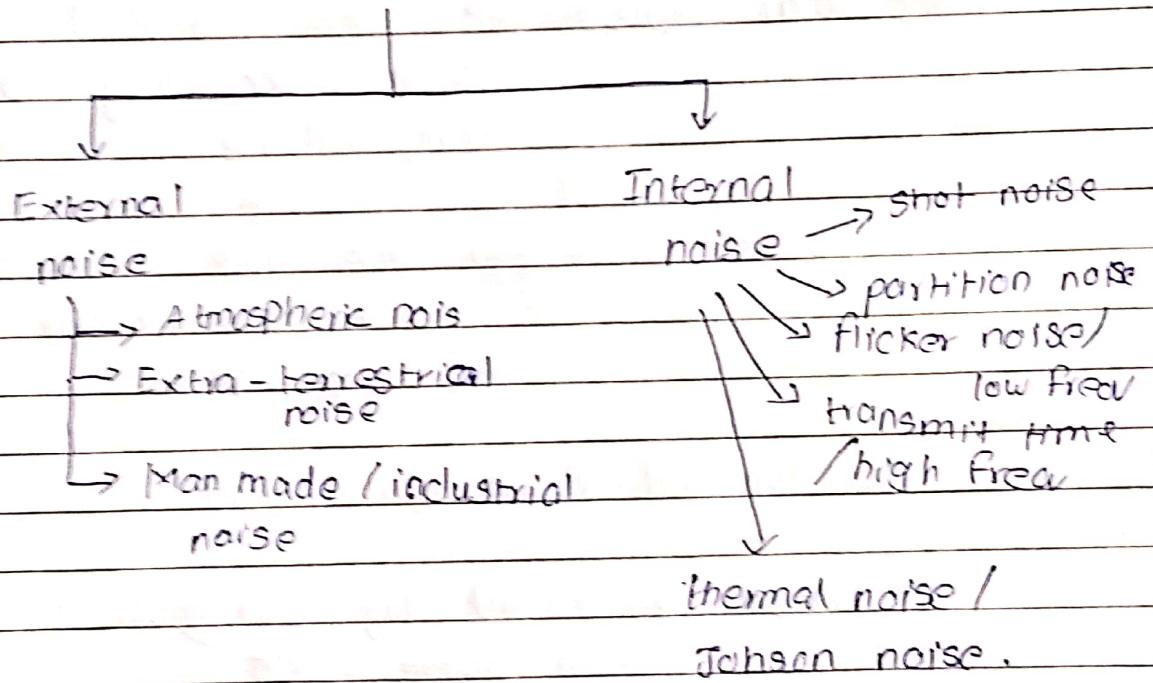
- Range of wave of energy that are oscillating electric & magnetic fields which ~~towards~~ travels and spreads out as it goes.

READ MY INSTRUCTIONS VISIBLE UNDER X-RAY GLASSES.
 Radio wave region Micro wave region infra ray Visible rays UV X-RAY Gamma

Region	Frequency	Wavelength	Source	Applications
R	$< 3 \times 10^{11}$	$> 10\text{cm}$	Oscillating currents in vacuum tubes	TV, Radio communication
I	$3 \times 10^{11} - 5 \times 10^{14}$	$10 - 0.1\text{cm}$	Oscillation of electrons in gases	TV, micro oven
II	$5 \times 10^{14} - 7.5 \times 10^{15}$	$1000 - 0.74\text{nm}$	Excitation of atoms & molecules	Gives information about the structure of molecules & atoms
U	$7.5 \times 10^{15} - 3 \times 10^{16}$	$740 - 100\text{nm}$	The rotation of atoms, spins & orientation of a atom in vacuum	The rotation of atoms, spins & orientation of a atom in vacuum
X	$3 \times 10^{16} - 3 \times 10^{19}$	$10 - 0.01\text{nm}$	Sudden release of high energy of electrons	X-ray theory, industrial radiography, cancer treatment
Y	$> 3 \times 10^{19}$	$< 0.1\text{nm}$	Decay of radioactive nuclei	Gives information about the structure of nuclei

> Noise :-

- It is an unwanted form of energy or electrical disturbance that can cause audible or visual disturbance in communication system and errors in digital communication system.
- It limits the performance of the communication system.
- Noise gets superimposed on the message signal & makes it impossible to separate the signal from the noise.



• External Noise :-

- "Generated external to a communication system"

1) Atmospheric noise

source → Lighting discharge

Field strength → inverse approx. inverse with the frequency.

- It becomes less severe & significant at freq. above 30MHz.

ii) Extra-Terrestrial Noise

- "Originates from sources outside the earth atmosphere." hence called ~~deep~~ space noise.

Types :-

- (i) Solar Noise → Noise radiated by sun varies according to its surface temperature
- (ii) Cosmic Noise → The noise radiated by distant stars also known as black body noise.

iii) Man made / Industrial Noise

- Noise produced by automobile, electrical motor, switch ~~gates~~, gears, etc.

iv) Internal Noise / Fundamental Noise

- Noise produced internally or within an electronic equipment. - can be reduced by proper system design.

i) Shot Noise :-

- Arising in active devices due to random behaviour of charge carriers.
- Can be generated due to random emission of electrons from cathodes or due to random diffusion of minority carriers or simply random generation and recombination of electron hole pairs.
- For amplifying devices shot noise is inversely proportional to the trans conductance of the device & directly proportional to the output current.

ii) Partition Noise :-

- Generated in the circuit where a current has to divide between two or more paths.
- Its freq. spectrum is a flat spectrum.
- Partition noise in a transistor will be higher than in a diode.
- Mosfet has low partition noise.

iii) Low Frequency / Flicker Noise

- In semiconductor devices, flicker noise is generated due to fluctuations in charge carrier density
- Power spectral density of this noise increases as freq. decreases.
- Flicker Noise appears at freq. below a few kHz and hence is sometimes called "1/f" noise.

iv) High frequency / transient time noise :-

- Time taken by a current carrier (electrons/holes) to travel from input to output of a device.
- High freq. is generally observed in semiconductor devices when transient time of charge carriers crossing a junction is comparable with time period of signal which is been amplified.
- Once this noise appears it goes on increasing with frequency at a rate of $6P$ dB per octave

v) Thermal Noise

- It is called as white noise / Johnson noise
- Produced due to random thermal agitation of electrons in a conductor which increases with freq. and temperature
- Avg thermal noise power $P_n = kTB$

$K \Rightarrow$ Boltzmann constant

$T \Rightarrow$ Temp. of conductor in Kelvin

$B \Rightarrow$ Bandwidth of noise spectrum.



* Signal to Noise Ratio (SNR)

- It is defined as the Ratio of signal power to the noise power.

$$\text{SNR} =$$

$$\frac{S}{N} = \frac{P_s}{P_n} = \frac{V_s^2/R}{V_n^2/R} = \frac{V_s^2}{V_n^2}$$

- It's expressed in dB.
- The higher the value of $\frac{S}{N}$ ratio better is the system performance in presence of noise.

$$\begin{aligned} \left(\frac{S}{N}\right)_{\text{dB}} &= 10 \log \left(\frac{P_s}{P_n}\right) \\ &= 10 \log \left(\frac{V_s^2}{V_n^2}\right) \\ &= 10 \log \left(\frac{V_s}{V_n}\right) \end{aligned}$$

- SNR is a key parameter for any radio receiver as it determines the noise performance of the receiver which implies that the receiver is able to reproduce and adequate / sufficient quality signal even in the presence of noise.

* Noise figure :-

- It is denoted by F.

$$F = \frac{\text{S/N ratio at input}}{\text{S/N ratio at output}}$$

$$= \frac{P_{Si}}{P_{Ni}} \div \frac{P_{No}}{P_{No}}$$

$$= \frac{P_{Si}}{P_{Ni}} \times \frac{P_{No}}{P_{No}}$$

$$\frac{P_{Si}}{P_{No}} = a \quad (\text{Power gain})$$

- Noise factor is a means to measure the amount of noise added by the device and ^{will} always be greater than 1.

$$= \frac{P_{No}}{P_{Ni} G}$$

as, we know

$$P_{Ni} = kTB \quad - (1)$$

$$\therefore F = \frac{P_{No}}{P_{Ni} G}$$

$$P_{No} = P_{Ni} FG$$

From (1)

$$P_{No} = kTBFG$$

or

$$\frac{P_{No}}{G} = kTBF = P_{Ni}(\text{Total})$$

$$P_{Ni}(\text{Total}) = P_{Ni} + P_{Na}$$

(input source) (device)

$$P_{Na} = P_{Ni}(\text{Total}) - P_{Ni}(\text{input source})$$

$$= kTBFG - kTB$$

$$P_{Na} = (F-1) kTB \quad - (2)$$

* Noise Temperature :-

- It is the temp. which generates noise power in the system.
- Noise power by the device :

$$P_{na} = K T_0 B (F-1) \quad - (2)$$

(device)

where T_0 is the environment temperature

- Input Noise power in terms of noise temperature .

$$P = K T_{eq} B$$

$$K T_{eq} B = K T_0 B (F-1)$$

$$\boxed{T_{eq} = T_0 (F-1)} \quad - (3)$$

Module 4 - Pulse Modulation

* Pulse Modulation:

- Sampling - To transmit Analog signal that is used to in real world such as voice & video signal by digital mean. This has to be converted to a digital signal.
- While an analog signal is continuous in both time and magnitude, a digital signal is discrete in both time & amplitude.
- To make a signal discrete, value of a signal is measured in certain interval of time.
- Each measurement is called as sample.
- In order to preserve the information contained in the signal it is necessary to get samples at a particular time interval or specific frequency. → Nyquist rate

* Sampling Theorem

- Let $m(t)$ be a band limited signal with maximum frequency f_m .
- The continuous time signal can be represented as $m(nT_s)$ using sampling with the help of periodic train of impulses.
- * With period T_s seconds
- From discrete time signal $m(nT_s)$, the original signal $m(t)$ can be regenerated without any distortion using low pass filter, in (reconstruction Filter).

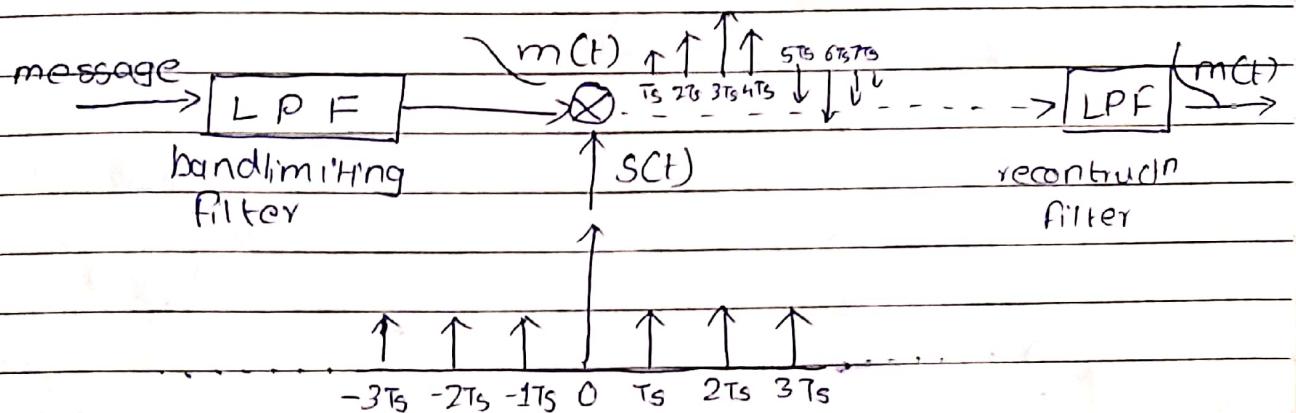
- With cutoff freq. ~~CUT~~ fm Hz, If:-

$$f_s \geq 2 f_m$$

OR

$$T_s \geq \frac{1}{2 f_m}$$

- where, $f_s \Rightarrow$ Sampling OR Nyquist Rate
 $T_s \Rightarrow$ Sampling interval



$$f_s \leq 2 f_m$$

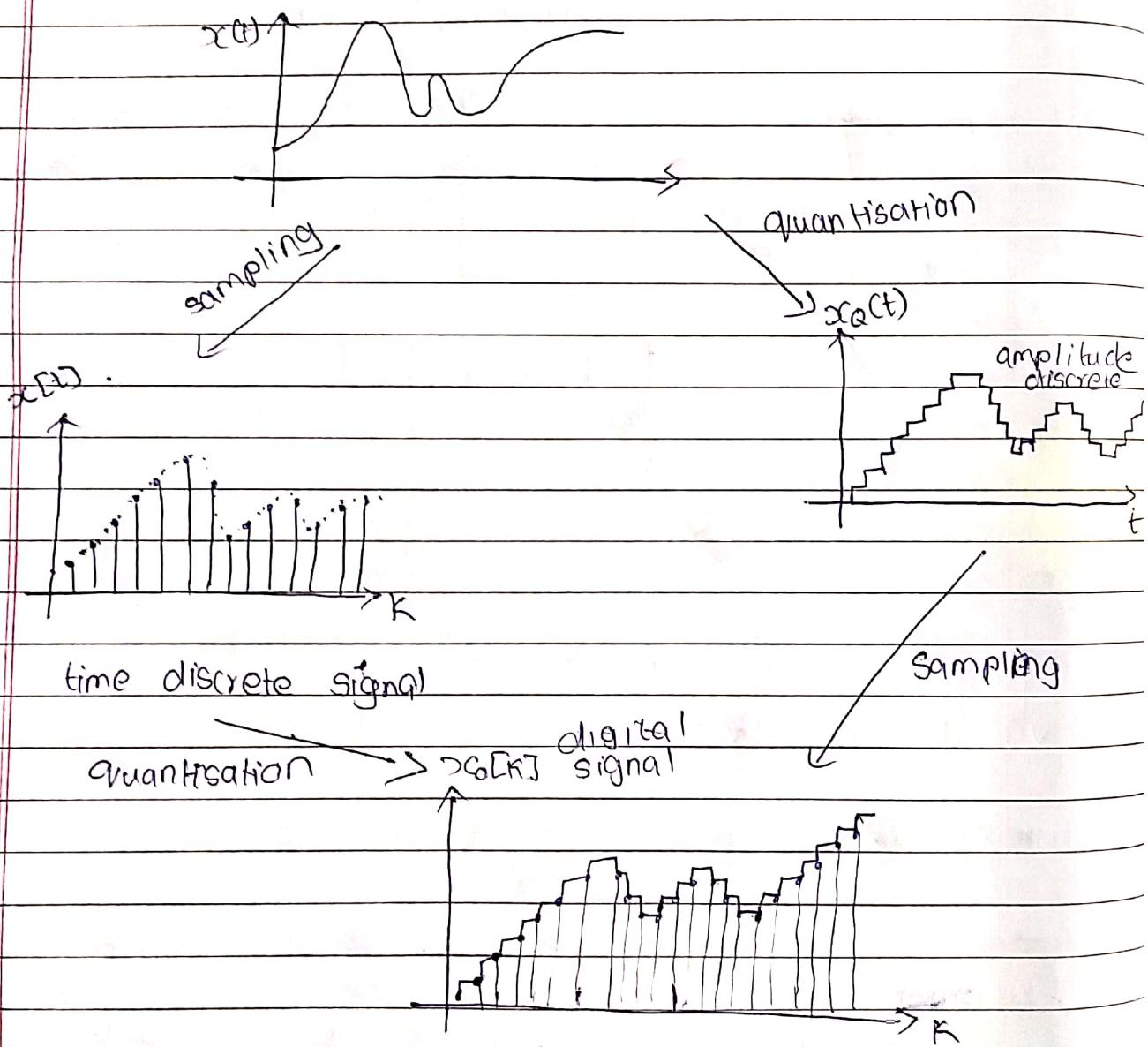
$f_s > 2 f_m \rightarrow$ under sampling

$$f_s = 2 f_m$$

* Quantisation :-

- The sample values measured during sampling must be quantised to produce a digital representation of the analog signal i.e each value is approximated to its nearest quantisation level.
- The process of digitizing the domain (time) is called sampling and the process of digitizing the amplitude range is called the Quantisation.

- A ~~Q~~-level quantiser compares the discrete type input $m(nT_s)$ with its fixed digital levels.
- It assigns anyone of the digital levels to $m(nT_s)$ which results in minimum distortion or error.



- Q) The AM transmitter develops an unmodulated power output of 400W. Across a 50Ω resistive load, the carrier is modulated by a sinusoidal signal with a modulation index of 0.8. Assuming, $F_M = 5\text{kHz}$, $F_c = 1\text{MHz}$.
- Obtain the value of carrier amplitude & hence write the expression for AM signal.
 - Find the total sideband power.
 - Draw the AM wave for the given modulatn index.

Given :- $P_c = 400\text{W}$

$R_L = 50\Omega$

$m = 0.8$

$f_m = 5\text{kHz}$

$f_c = 1\text{MHz}$

To find:- i) $\cancel{A_c^2} \text{ ?}_{\text{even}}$

ii) $P_{SB} = ?$

$$\text{i) } P_c = \frac{A_c^2}{2R}$$

$$\omega_m = 2\pi f_m$$

$$A_c^2 = 2P_c R$$

$$= 2 \times 3.14 \times 5\text{kHz}$$

$$A_c = \sqrt{2 \times 400 \times 50}$$

$$= 31.4 \times 10^4$$

$$A_c = 200\text{W}$$

$$\omega_c = 2\pi f_c$$

$$= 2 \times 3.14 \times 10^6$$

$$s(t) = A_c [1 + m \cos \omega_m t] \cos \omega_c t$$

$$= 2\pi 10^6$$

$$= 200 [1 + (0.8)(\pi 10^6)] \cos(2\pi 10^6 t)$$

$$\text{ii) } P_{SB} = \frac{m^2 A_c^2}{8R} = \frac{m^2 P_c}{4}$$

$$= \frac{(0.8)^2 \times 400}{4}$$

$$\boxed{P_{SB} = 64\text{W}}$$

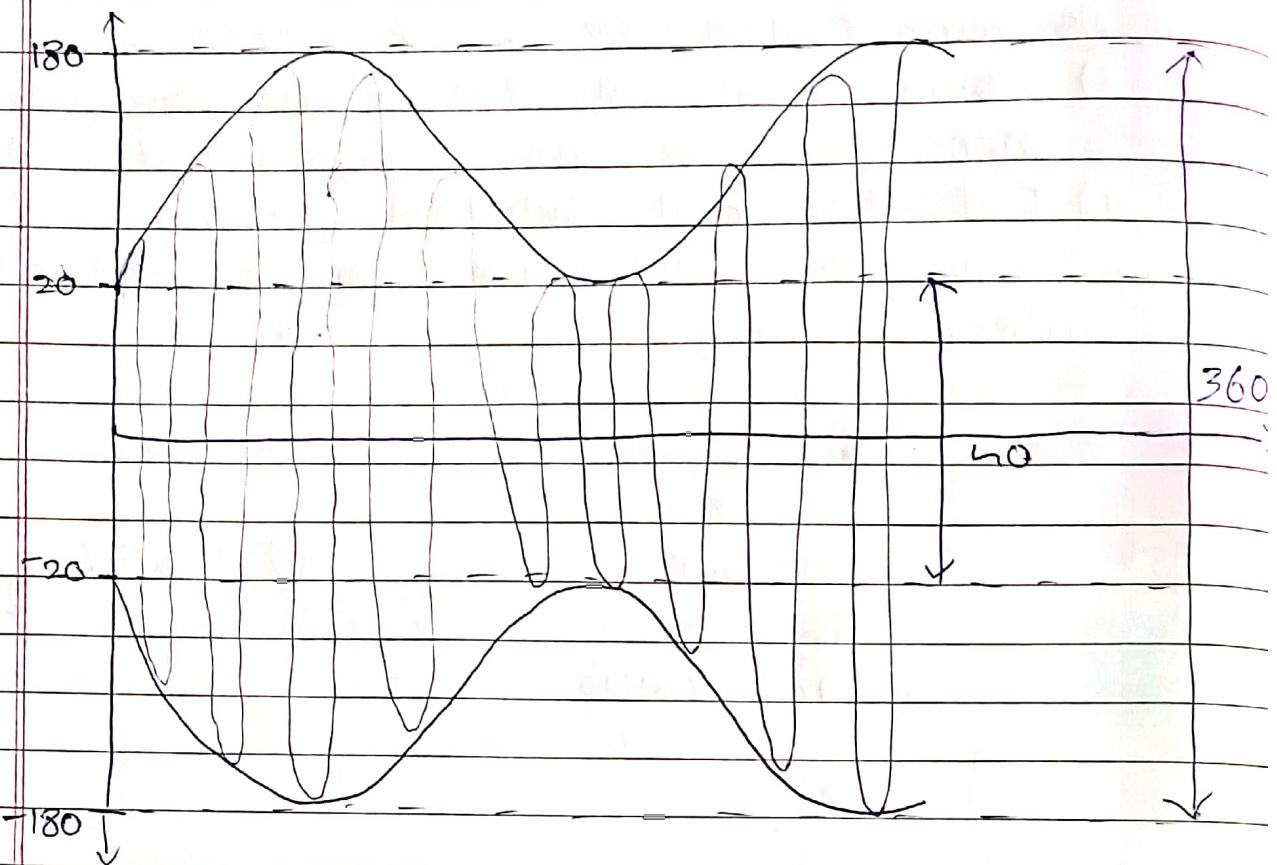
$$\text{iii) } A_{\max} = A_c + A_m = 360 \text{ V}$$

$$m = \frac{A_m}{A_c}$$

$$A_{\min} = A_c - A_m = 40 \text{ V}$$

$$A_m = 0.8 \times 200$$

$$A_m = 160 \text{ V}$$



Q) An AM signal is produced by modulating a carrier signal with peak voltage of 10V & frequency of 100 kHz by a sinusoidal signal of Amplitude 4V & freq. 4 kHz. Calculate
i) modulation index & write its mathematical expression.

ii) Bandwidth of Am & sketch its two sided spectrum.

iii) Total power of the modulated wave developed across load resistance, $R_L = 50$ & power content in each sideband and carrier.

Given:- $A_c = 10V$

$$f_c = 100\text{kHz}$$

$$A_m = 10\text{Vpp}$$

$$A_m = 4V$$

$$f_m = 4\text{kHz}$$

$$\omega_m = 2\pi f_m$$

$$= 8\pi \times 10^3$$

$$i) m = \frac{A_m}{A_c} = \frac{4}{10} = 0.4$$

$$\omega_c = 2\pi f_c$$

$$= 2\pi \times 10^5$$

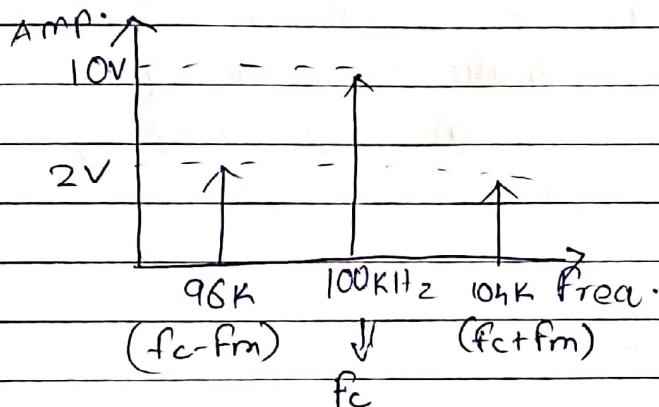
$$s(t) = [1 + m \cos \omega_m t] \cos \omega_c t$$

$$= [1 + (0.4) \cos (8\pi \times 10^3 t)] \cos (2\pi \times 10^5 t)$$

$$ii) \text{ BW} = 2f_m$$

$$= 2 \times 4$$

$$= 8\text{kHz}$$



$$f_c - f_m = 96\text{kHz}$$

$$f_c + f_m = 104\text{kHz}$$

$$\frac{A_m}{2} = \frac{4}{2} = 2$$

$$\text{Amp. of SB} = \frac{A_m}{2}$$

$$iii) \text{ Total Power} = P_{LSB} + P_c + P_{USB}$$

$$P_c = \frac{A_c^2}{2R} = \frac{100}{2 \times 50} = 1\text{W}$$

$$P_{LSB} = \frac{m^2 P_c}{4} = \frac{0.4 \times 0.4 \times 1}{4} = 0.04\text{W}$$

$$\therefore P_{USB} = 0.04\text{W}$$

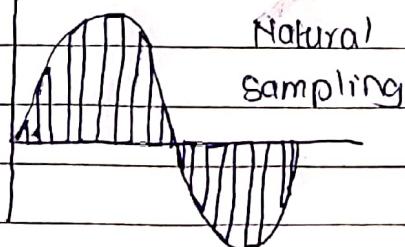
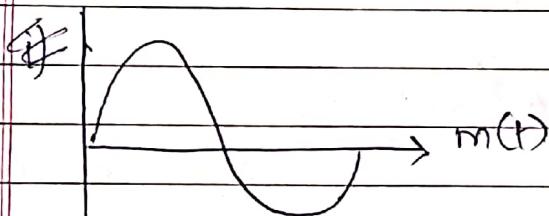
$$\therefore \text{Total Power} = 0.04 + 1 + 0.04 \\ = 1.08\text{W}$$

* Pulse Amplitude Modulation :-

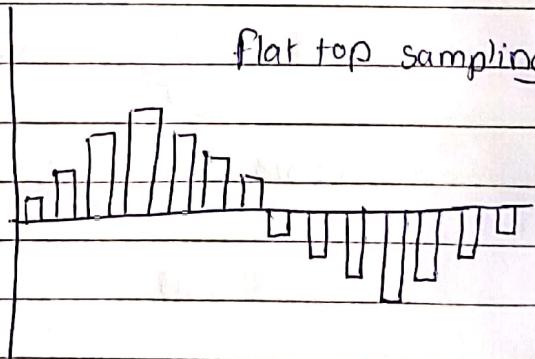
- It is a modulating technique in which the amplitude of each pulse is controlled by the amplitude of the modulating signal.
- Signal is sampled at regular interval & each sample is made proportional to the amplitude of the signal, at the instant of sampling.

* diff. betn continuous signal & PM is that continuous signal has sinusoidal wave & PM has square si wave.

- The sampling of the signal is basically done in two ways - i)natural sampling
ii) flat top sampling



flat top sampling

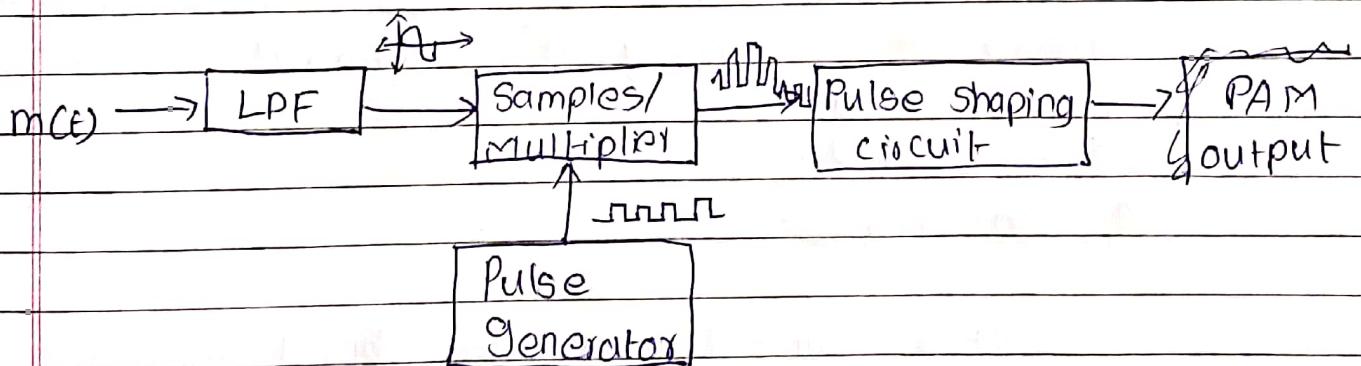


i) Natural Sampling

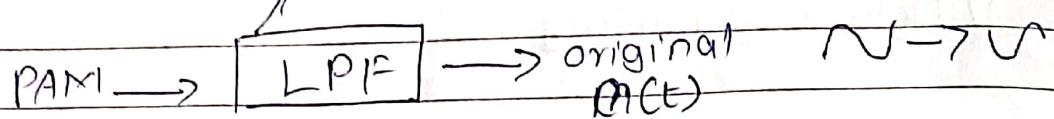
→ Here, the variation in the amplitude of each pulse is directly proportional to the modulating signal amplitude at the time of pulse occurrence.

ii) Flat top Sampling

- Here, the top of each sample/pulse is flat.
- flat top sampling is widely used because at the time of signal transmission the channel noise introduces some form of distortion which can be easily eliminated in case of flat tops.
- As against natural sampling where the top of pulses varies according to the modulating signal.
- The elimination of noise component from the sampled signal becomes difficult comparatively during detectn. (Demodulation)
- Hence, flat top sampling is preferred & also called as practical sampling.



Detection :- eliminates high frequency noise components



- The low passed filter used for detection eliminates high freq. noise components.
- This $m(t)$ is then passed to an inverting amplifier to amplify its signal level to have the demodulated output almost same as $m(t)$.

Advantages of PAM :-

- High efficiency
- Construction of transmitter & receiver in PAM is not complex.
- As, it doesn't require synchronization of pulses PAM has low noise.
- Quick transmission & reception

Disadvantages of PAM :-

- PAM is quantized in both amplitude & time which increases noise interference - Effects the quality of information being transmitted.
- Large B.W.

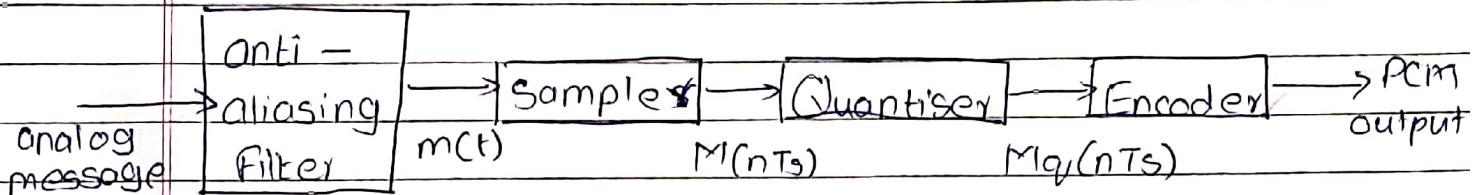
Applications

- Ethernet connectivity broadband interface communication.
- To control signal in micro controllers.
- For energy efficient lighting in LED drivers.

- For better signal clarity & clearer picture in digital television.

* Pulse Code Modulation :- (PCM)

- PCM or digital technique that involves sampling & analog signals at regular interval & coding the measured amplitude into a series of binary values (0s & 1s)
- Generation of PCM:



Anti-Aliasing Filter

- It is a low pass filter used to band limit the input message signal with maximum freq. fm Hz.

Sampler

- It is used to convert continuous time $m(t)$ to discrete time $M(nTs)$. This process is called as sampling. According Sampling theorem the sampling Frequency $f_s \geq 2f_m$ to regenerate the original signal $m(t)$ from $M(nTs)$.

Quantiser :

→ It is used to convert continuous amplitude $M(nT_s)$ to discrete amplitude $M_q(nT_s)$. This process is called Quantisation, which is an amplitude approximation process.