

INTERNAL MEMORY

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LEARNING OBJECTIVES

After studying this chapter, you should be able to:

- ◆ Present an overview of the principle types of semiconductor main memory.
- ◆ Understand the operation of a basic code that can detect and correct single-bit errors in 8-bit words.
- ◆ Summarize the properties of contemporary **DDR DRAM** organizations.
- ◆ Understand the difference between **NOR** and **NAND** flash memory.
- ◆ Present an overview of the newer nonvolatile solid-state memory technologies.

We begin this chapter with a survey of semiconductor main memory subsystems, including ROM, DRAM, and SRAM memories. Then we look at error control techniques used to enhance memory reliability. Following this, we look at more advanced DRAM architectures.

5.1 SEMICONDUCTOR MAIN MEMORY

In earlier computers, the most common form of random-access storage for computer main memory employed an array of doughnut-shaped ferromagnetic loops referred to as *cores*. Hence, main memory was often referred to as *core*, a term that persists to this day. The advent of, and advantages of, microelectronics has long since vanquished the magnetic core memory. Today, the use of semiconductor chips for main memory is almost universal. Key aspects of this technology are explored in this section.

Organization

The basic element of a **semiconductor memory** is the memory cell. Although a variety of electronic technologies are used, all semiconductor memory cells share certain properties:

- They exhibit two stable (or semistable) states, which can be used to represent binary 1 and 0.
- They are capable of being written into (at least once), to set the state.
- They are capable of being read to sense the state.

Figure 5.1 depicts the operation of a memory cell. Most commonly, the cell has three functional terminals capable of carrying an electrical signal. The select terminal, as the name suggests, selects a memory cell for a read or write operation. The control terminal indicates read or write. For writing, the other terminal provides an electrical signal that sets the state of the cell to 1 or 0. For reading, that terminal is used for output of the cell's state. The details of the internal organization, functioning, and timing of the memory cell depend on the specific integrated circuit technology used and are beyond the scope of this book, except for a brief summary. For our purposes, we will take it as given that individual cells can be selected for reading and writing operations.

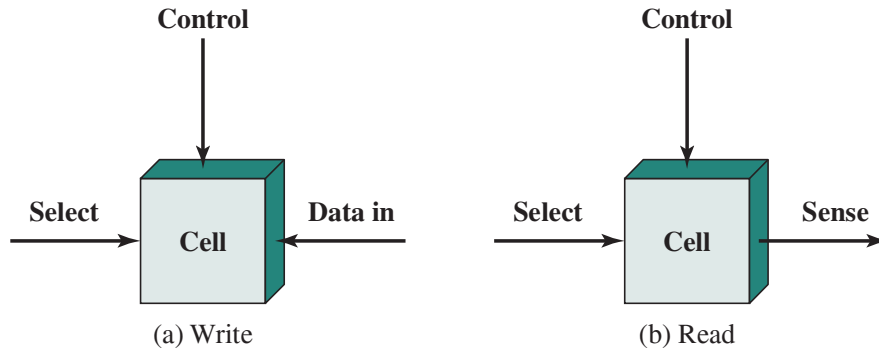


Figure 5.1 Memory Cell Operation

DRAM and SRAM

All of the memory types that we will explore in this chapter are random access. That is, individual words of memory are directly accessed through wired-in addressing logic.

Table 5.1 lists the major types of semiconductor memory. The most common is referred to as **random-access memory (RAM)**. This is, in fact, a misuse of the term, because all of the types listed in the table are random access. One distinguishing characteristic of memory that is designated as RAM is that it is possible both to read data from the memory and to write new data into the memory easily and rapidly. Both the reading and writing are accomplished through the use of electrical signals.

The other distinguishing characteristic of traditional RAM is that it is volatile. A RAM must be provided with a constant power supply. If the power is interrupted, then the data are lost. Thus, RAM can be used only as temporary storage. The two traditional forms of RAM used in computers are DRAM and SRAM. Newer forms of RAM, discussed in Section 5.5, are nonvolatile.

DYNAMIC RAM RAM technology is divided into two technologies: dynamic and static. A **dynamic RAM (DRAM)** is made with cells that store data as charge on capacitors. The presence or absence of charge in a capacitor is interpreted as a binary 1 or 0. Because capacitors have a natural tendency to discharge, dynamic RAMs require periodic charge refreshing to maintain data storage. The term

Table 5.1 Semiconductor Memory Types

Memory Type	Category	Erasure	Write Mechanism	Volatility
Random-access memory (RAM)	Read-write memory	Electrically, byte-level	Electrically	Volatile
Read-only memory (ROM)	Read-only memory	Not possible	Masks	Nonvolatile
Programmable ROM (PROM)			Electrically	
Erasable PROM (EPROM)	UV light, chip-level			
Electrically Erasable PROM (EEPROM)	Electrically, byte-level			
Flash memory	Electrically, block-level			

dynamic refers to this tendency of the stored charge to leak away, even with power continuously applied.

Figure 5.2a is a typical DRAM structure for an individual cell that stores one bit. The address line is activated when the bit value from this cell is to be read or written. The transistor acts as a switch that is closed (allowing current to flow) if a voltage is applied to the address line and open (no current flows) if no voltage is present on the address line.

For the write operation, a voltage signal is applied to the bit line; a high voltage represents 1, and a low voltage represents 0. A signal is then applied to the address line, allowing a charge to be transferred to the capacitor.

For the read operation, when the address line is selected, the transistor turns on and the charge stored on the capacitor is fed out onto a bit line and to a sense amplifier. The sense amplifier compares the capacitor voltage to a reference value and determines if the cell contains a logic 1 or a logic 0. The readout from the cell discharges the capacitor, which must be restored to complete the operation.

Although the DRAM cell is used to store a single bit (0 or 1), it is essentially an analog device. The capacitor can store any charge value within a range; a threshold value determines whether the charge is interpreted as 1 or 0.

STATIC RAM In contrast, a **static RAM (SRAM)** is a digital device that uses the same logic elements used in the processor. In a SRAM, binary values are stored using traditional flip-flop logic-gate configurations (see Chapter 11 for a description of flip-flops). A static RAM will hold its data as long as power is supplied to it.

Figure 5.2b is a typical SRAM structure for an individual cell. Four transistors (T_1 , T_2 , T_3 , T_4) are cross connected in an arrangement that produces a stable logic

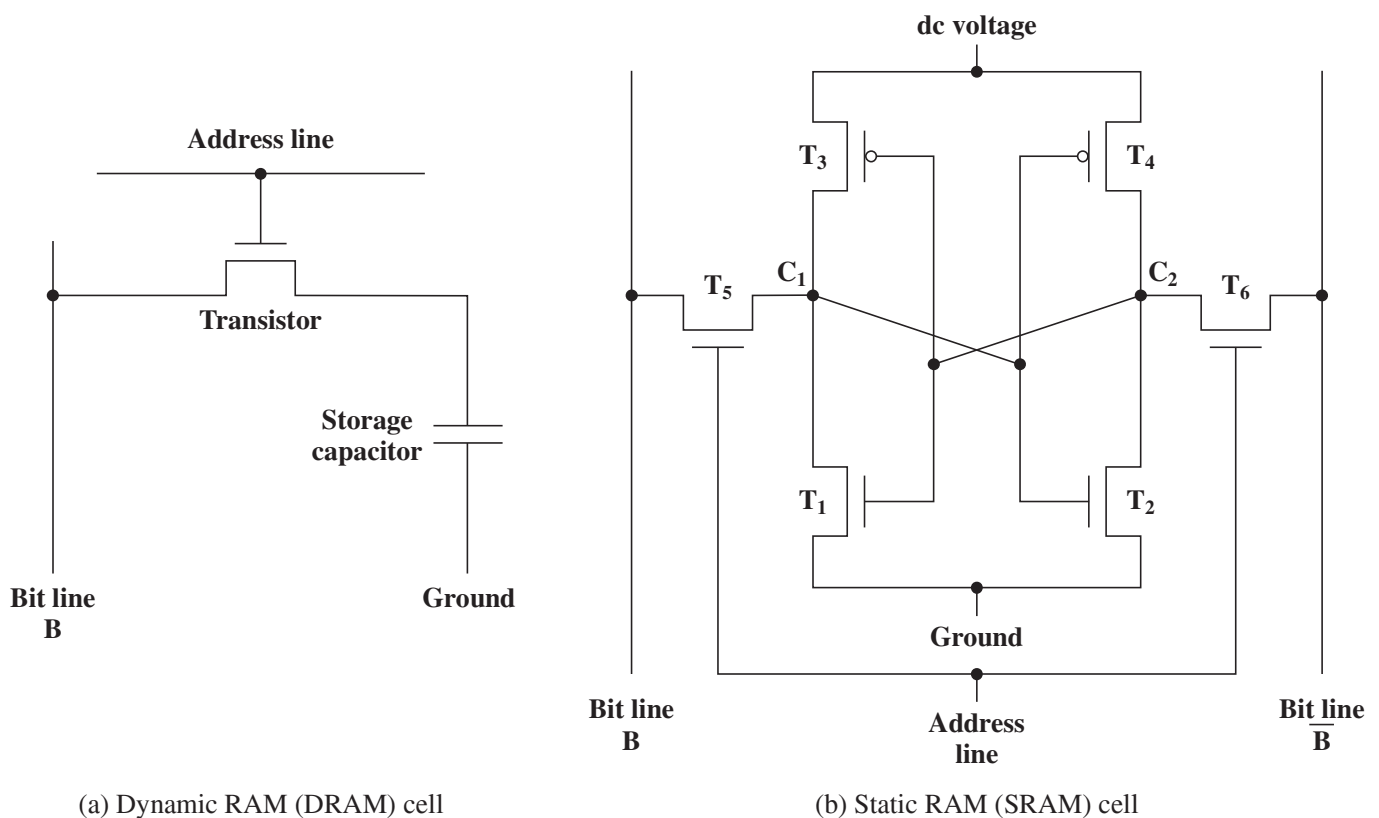


Figure 5.2 Typical Memory Cell Structures

state. In logic state 1, point C_1 is high and point C_2 is low; in this state, T_1 and T_4 are off and T_2 and T_3 are on.¹ In logic state 0, point C_1 is low and point C_2 is high; in this state, T_1 and T_4 are on and T_2 and T_3 are off. Both states are stable as long as the direct current (dc) voltage is applied. Unlike the DRAM, no refresh is needed to retain data.

As in the DRAM, the SRAM address line is used to open or close a switch. The address line controls two transistors (T_5 and T_6). When a signal is applied to this line, the two transistors are switched on, allowing a read or write operation. For a write operation, the desired bit value is applied to line B, while its complement is applied to line \bar{B} . This forces the four transistors (T_1, T_2, T_3, T_4) into the proper state. For a read operation, the bit value is read from line B.

SRAM VERSUS DRAM Both static and dynamic RAMs are volatile; that is, power must be continuously supplied to the memory to preserve the bit values. A dynamic memory cell is simpler and smaller than a static memory cell. Thus, a DRAM is more dense (smaller cells = more cells per unit area) and less expensive than a corresponding SRAM. On the other hand, a DRAM requires the supporting refresh circuitry. For larger memories, the fixed cost of the refresh circuitry is more than compensated for by the smaller variable cost of DRAM cells. Thus, DRAMs tend to be favored for large memory requirements. A final point is that SRAMs are somewhat faster than DRAMs. Because of these relative characteristics, SRAM is used for cache memory (both on and off chip), and DRAM is used for main memory.

Types of ROM

As the name suggests, a **read-only memory (ROM)** contains a permanent pattern of data that cannot be changed. A ROM is nonvolatile; that is, no power source is required to maintain the bit values in memory. While it is possible to read a ROM, it is not possible to write new data into it. An important application of ROMs is micro-programming, discussed in Part Four. Other potential applications include

- Library subroutines for frequently wanted functions
- System programs
- Function tables

For a modest-sized requirement, the advantage of ROM is that the data or program is permanently in main memory and need never be loaded from a secondary storage device.

A ROM is created like any other integrated circuit chip, with the data actually wired into the chip as part of the fabrication process. This presents two problems:

- The data insertion step includes a relatively large fixed cost, whether one or thousands of copies of a particular ROM are fabricated.
- There is no room for error. If one bit is wrong, the whole batch of ROMs must be thrown out.

When only a small number of ROMs with a particular memory content is needed, a less expensive alternative is the **programmable ROM (PROM)**. Like the

¹The circles associated with T_3 and T_4 in Figure 5.2b indicate signal negation.

ROM, the PROM is **nonvolatile** and may be written into only once. For the PROM, the writing process is performed electrically and may be performed by a supplier or customer at a time later than the original chip fabrication. Special equipment is required for the writing or “programming” process. PROMs provide flexibility and convenience. The ROM remains attractive for high-volume production runs.

Another variation on read-only memory is the **read-mostly memory**, which is useful for applications in which read operations are far more frequent than write operations but for which nonvolatile storage is required. There are three common forms of read-mostly memory: EPROM, EEPROM, and flash memory.

The optically **erasable programmable read-only memory (EPROM)** is read and written electrically, as with PROM. However, before a write operation, all the storage cells must be erased to the same initial state by exposure of the packaged chip to ultraviolet radiation. Erasure is performed by shining an intense ultraviolet light through a window that is designed into the memory chip. This erasure process can be performed repeatedly; each erasure can take as much as 20 minutes to perform. Thus, the EPROM can be altered multiple times and, like the ROM and PROM, holds its data virtually indefinitely. For comparable amounts of storage, the EPROM is more expensive than PROM, but it has the advantage of the multiple update capability.

A more attractive form of read-mostly memory is **electrically erasable programmable read-only memory (EEPROM)**. This is a read-mostly memory that can be written into at any time without erasing prior contents; only the byte or bytes addressed are updated. The write operation takes considerably longer than the read operation, on the order of several hundred microseconds per byte. The EEPROM combines the advantage of nonvolatility with the flexibility of being updatable in place, using ordinary bus control, address, and data lines. EEPROM is more expensive than EPROM and also is less dense, supporting fewer bits per chip.

Another form of semiconductor memory is **flash memory** (so named because of the speed with which it can be reprogrammed). First introduced in the mid-1980s, flash memory is intermediate between EPROM and EEPROM in both cost and functionality. Like EEPROM, flash memory uses an electrical erasing technology. **An entire flash memory can be erased in one or a few seconds, which is much faster than EPROM.** In addition, it is possible to erase just blocks of memory rather than an entire chip. Flash memory gets its name because the microchip is organized so that a section of memory cells are erased in a single action or “flash.” However, flash memory does not provide byte-level erasure. Like EPROM, flash memory uses only one transistor per bit, and so achieves the high density (compared with EEPROM) of EPROM.

Chip Logic

As with other integrated circuit products, semiconductor memory comes in packaged chips (Figure 1.11). Each chip contains an array of memory cells.

In the memory hierarchy as a whole, we saw that there are trade-offs among speed, density, and cost. These trade-offs also exist when we consider the organization of memory cells and functional logic on a chip. For semiconductor memories, one of the key design issues is the number of bits of data that may be read/written at a time. At one extreme is an organization in which the physical arrangement of cells in the array is the same as the logical arrangement (as perceived by the processor) of words in memory. The array is organized into W words of B bits each.

For example, a 16-Mbit chip could be organized as 1M 16-bit words. At the other extreme is the so-called 1-bit-per-chip organization, in which data are read/written one bit at a time. We will illustrate memory chip organization with a DRAM; ROM organization is similar, though simpler.

Figure 5.3 shows a typical organization of a 16-Mbit DRAM. In this case, 4 bits are read or written at a time. Logically, the memory array is organized as four square arrays of 2048 by 2048 elements. Various physical arrangements are possible. In any case, the elements of the array are connected by both horizontal (row) and vertical (column) lines. Each horizontal line connects to the Select terminal of each cell in its row; each vertical line connects to the Data-In/Sense terminal of each cell in its column.

Address lines supply the address of the word to be selected. A total of $\log_2 W$ lines are needed. In our example, 11 address lines are needed to select one of 2048 rows. These 11 lines are fed into a row decoder, which has 11 lines of input and 2048 lines for output. The logic of the decoder activates a single one of the 2048 outputs depending on the bit pattern on the 11 input lines ($2^{11} = 2048$).

An additional 11 address lines select one of 2048 columns of 4 bits per column. Four data lines are used for the input and output of 4 bits to and from a data buffer. On input (write), the bit driver of each bit line is activated for a 1 or 0 according to the value of the corresponding data line. On output (read), the value of each bit line is passed through a sense amplifier and presented to the data lines. The row line selects which row of cells is used for reading or writing.

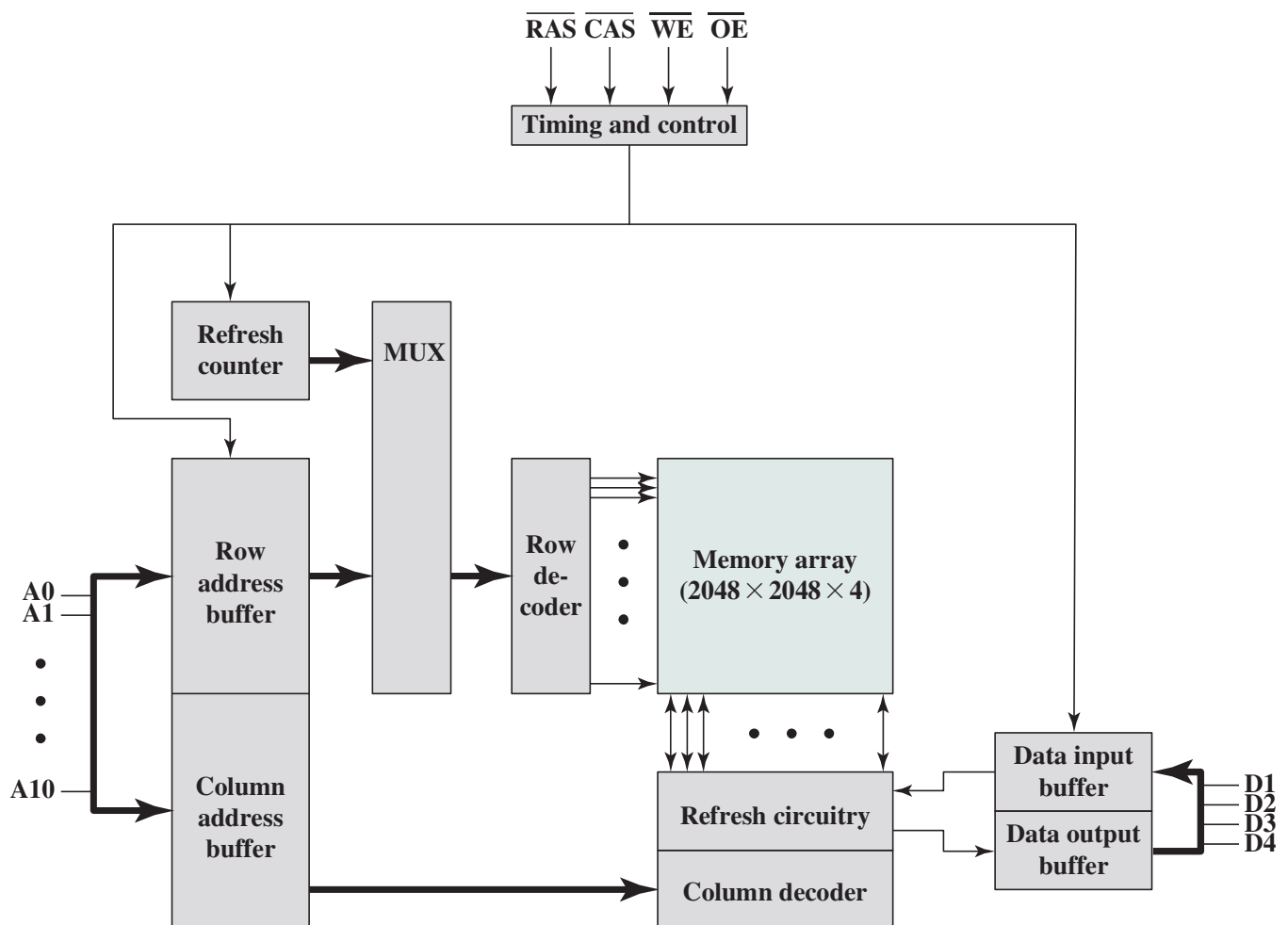


Figure 5.3 Typical 16-Mbit DRAM ($4\text{M} \times 4$)

Because only 4 bits are read/written to this DRAM, there must be multiple DRAMs connected to the memory controller to read/write a word of data to the bus.

Note that there are only 11 address lines (A0–A10), half the number you would expect for a 2048×2048 array. This is done to save on the number of pins. The 22 required address lines are passed through select logic external to the chip and multiplexed onto the 11 address lines. First, 11 address signals are passed to the chip to define the row address of the array, and then the other 11 address signals are presented for the column address. These signals are accompanied by row address select ($\overline{\text{RAS}}$) and column address select ($\overline{\text{CAS}}$) signals to provide timing to the chip.

The write enable ($\overline{\text{WE}}$) and output enable ($\overline{\text{OE}}$) pins determine whether a write or read operation is performed. Two other pins, not shown in Figure 5.3, are ground (V_{ss}) and a voltage source (V_{cc}).

As an aside, multiplexed addressing plus the use of square arrays result in a quadrupling of memory size with each new generation of memory chips. One more pin devoted to addressing doubles the number of rows and columns, and so the size of the chip memory grows by a factor of 4.

Figure 5.3 also indicates the inclusion of refresh circuitry. All DRAMs require a refresh operation. A simple technique for refreshing is, in effect, to disable the DRAM chip while all data cells are refreshed. The refresh counter steps through all of the row values. For each row, the output lines from the refresh counter are supplied to the row decoder and the RAS line is activated. The data are read out and written back into the same location. This causes each cell in the row to be refreshed.

Chip Packaging

As was mentioned in Chapter 2, an integrated circuit is mounted on a package that contains pins for connection to the outside world.

Figure 5.4a shows an example EPROM package, which is an 8-Mbit chip organized as $1\text{M} \times 8$. In this case, the organization is treated as a one-word-per-chip package. The package includes 32 pins, which is one of the standard chip package sizes. The pins support the following signal lines:

- The address of the word being accessed. For 1M words, a total of 20 ($2^{20} = 1\text{M}$) pins are needed (A0–A19).
- The data to be read out, consisting of 8 lines (D0–D7).
- The power supply to the chip (V_{cc}).
- A ground pin (V_{ss}).
- A chip enable (CE) pin. Because there may be more than one memory chip, each of which is connected to the same address bus, the CE pin is used to indicate whether or not the address is valid for this chip. The CE pin is activated by logic connected to the higher-order bits of the address bus (i.e., address bits above A19). The use of this signal is illustrated presently.
- A program voltage (V_{pp}) that is supplied during programming (write operations).

A typical DRAM pin configuration is shown in Figure 5.4b, for a 16-Mbit chip organized as $4\text{M} \times 4$. There are several differences from a ROM chip. Because a RAM can be updated, the data pins are input/output. The write enable (WE) and output enable (OE) pins indicate whether this is a write or read operation.

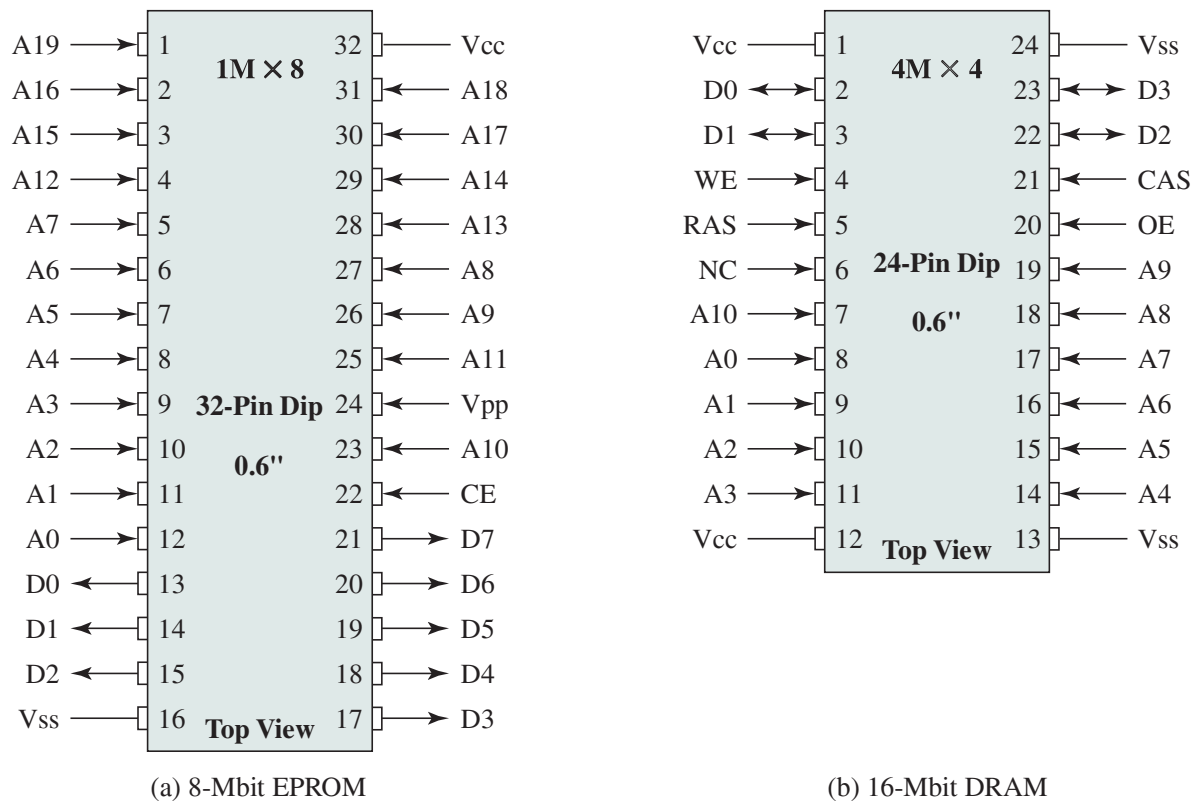


Figure 5.4 Typical Memory Package Pins and Signals

Because the DRAM is accessed by row and column, and the address is multiplexed, only 11 address pins are needed to specify the 4M row/column combinations ($2^{11} \times 2^{11} = 2^{22} = 4M$). The functions of the row address select (RAS) and column address select (CAS) pins were discussed previously. Finally, the no connect (NC) pin is provided so that there are an even number of pins.

Module Organization

If a RAM chip contains only one bit per word, then clearly we will need at least a number of chips equal to the number of bits per word. As an example, Figure 5.5 shows how a memory module consisting of 256K 8-bit words could be organized. For 256K words, an 18-bit address is needed and is supplied to the module from some external source (e.g., the address lines of a bus to which the module is attached). The address is presented to 8 $256K \times 1$ -bit chips, each of which provides the input/output of one bit.

This organization works as long as the size of memory equals the number of bits per chip. In the case in which larger memory is required, an array of chips is needed. Figure 5.6 shows the possible organization of a memory consisting of 1M word by 8 bits per word. In this case, we have four columns of chips, each column containing 256K words arranged as in Figure 5.5. For 1M word, 20 address lines are needed. The 18 least significant bits are routed to all 32 modules. The high-order 2 bits are input to a group select logic module that sends a chip enable signal to one of the four columns of modules.

Interleaved Memory

Main memory is composed of a collection of DRAM memory chips. A number of chips can be grouped together to form a *memory bank*. It is possible to organize the memory

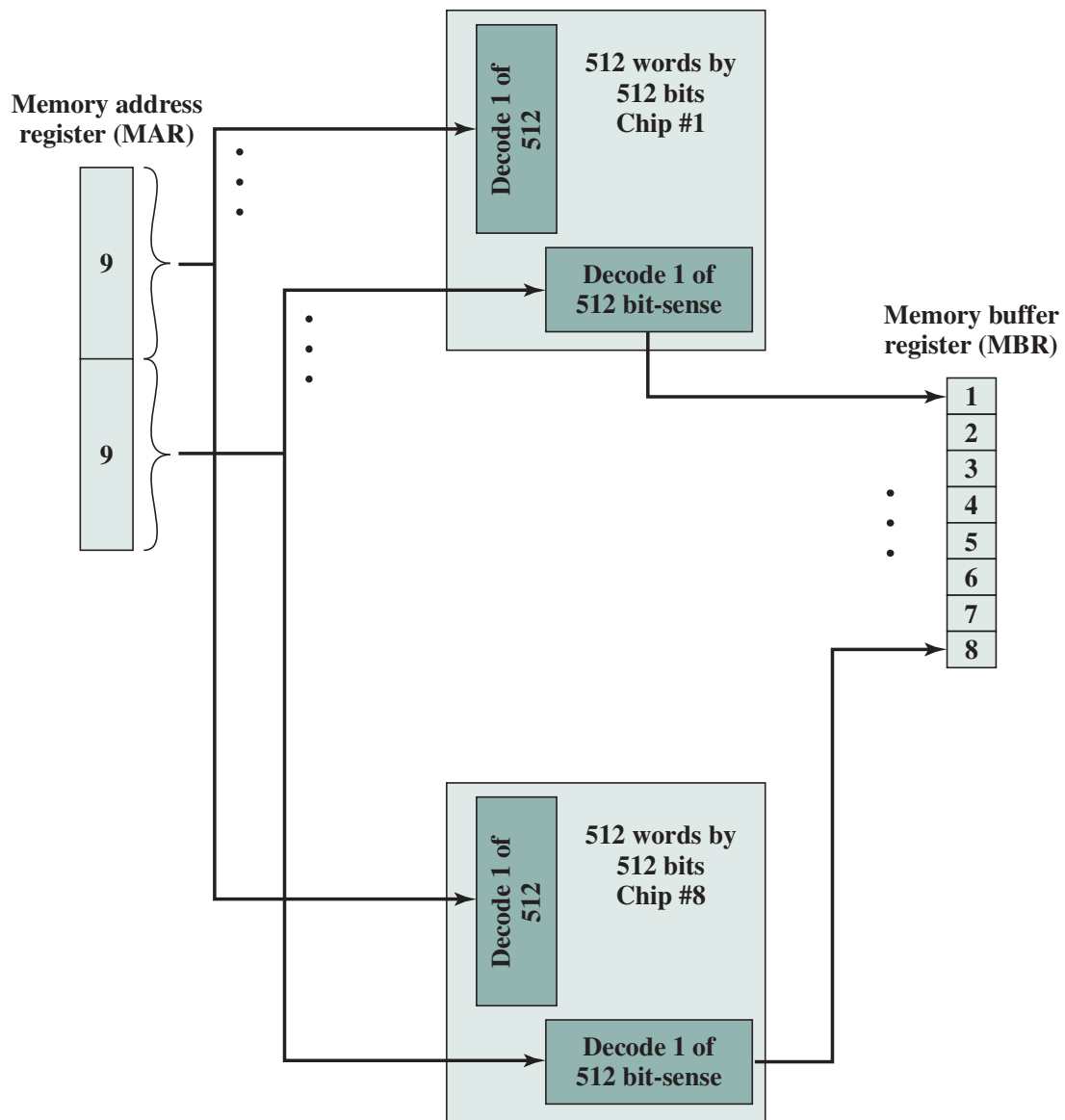


Figure 5.5 256-KByte Memory Organization

banks in a way known as interleaved memory. Each bank is independently able to service a memory read or write request, so that a system with K banks can service K requests simultaneously, increasing memory read or write rates by a factor of K . If consecutive words of memory are stored in different banks, then the transfer of a block of memory is speeded up. Appendix G explores the topic of interleaved memory.



Interleaved Memory Simulator

5.2 ERROR CORRECTION

A semiconductor memory system is subject to errors. These can be categorized as hard failures and soft errors. A **hard failure** is a permanent physical defect so that the memory cell or cells affected cannot reliably store data but become stuck at 0 or 1 or

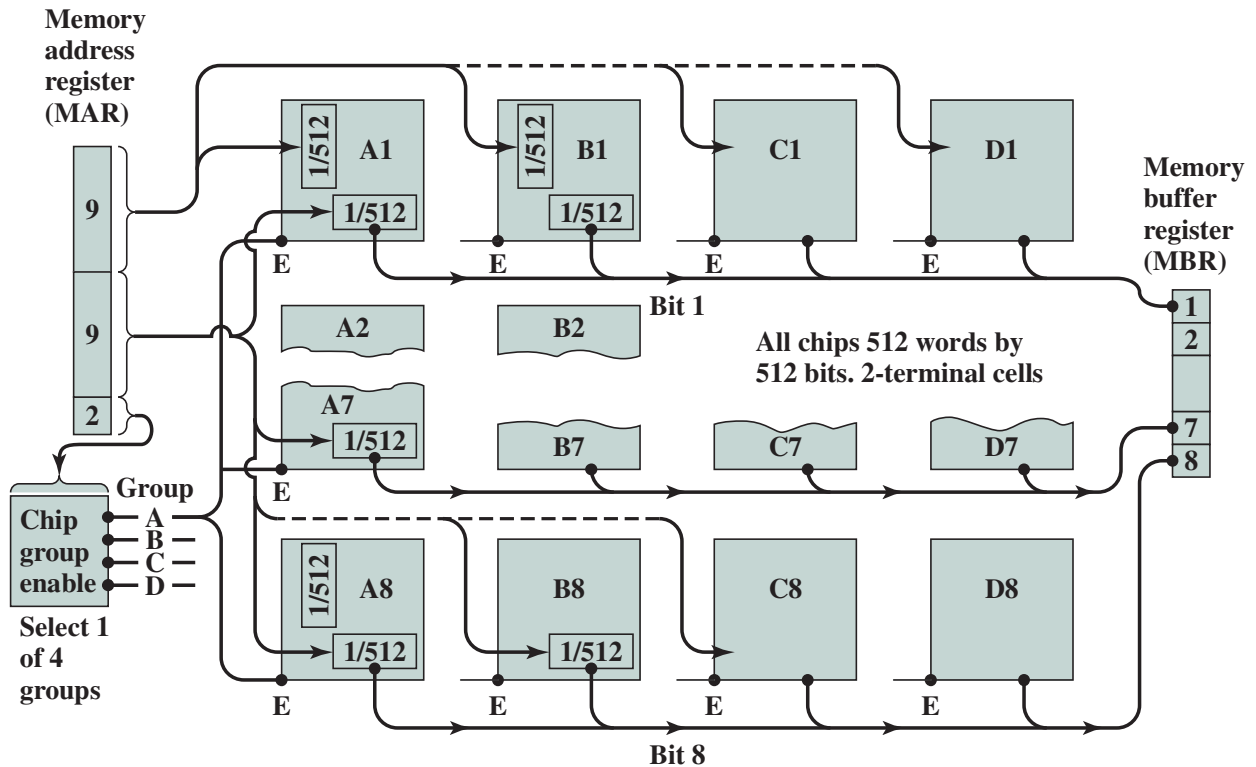


Figure 5.6 1-MB Memory Organization

switch erratically between 0 and 1. Hard errors can be caused by harsh environmental abuse, manufacturing defects, and wear. A **soft error** is a random, nondestructive event that alters the contents of one or more memory cells without damaging the memory. Soft errors can be caused by power supply problems or alpha particles. These particles result from radioactive decay and are distressingly common because radioactive nuclei are found in small quantities in nearly all materials. Both hard and soft errors are clearly undesirable, and most modern main memory systems include logic for both detecting and correcting errors.

Figure 5.7 illustrates in general terms how the process is carried out. When data are to be written into memory, a calculation, depicted as a function f , is performed on the data to produce a code. Both the code and the data are stored. Thus, if an M -bit word of data is to be stored and the code is of length K bits, then the actual size of the stored word is $M + K$ bits.

When the previously stored word is read out, the code is used to detect and possibly correct errors. A new set of K code bits is generated from the M data bits and compared with the fetched code bits. The comparison yields one of three results:

- No errors are detected. The fetched data bits are sent out.
- An error is detected, and it is possible to correct the error. The data bits plus **error correction** bits are fed into a corrector, which produces a corrected set of M bits to be sent out.
- An error is detected, but it is not possible to correct it. This condition is reported.

Codes that operate in this fashion are referred to as **error-correcting codes**. A code is characterized by the number of bit errors in a word that it can correct and detect.

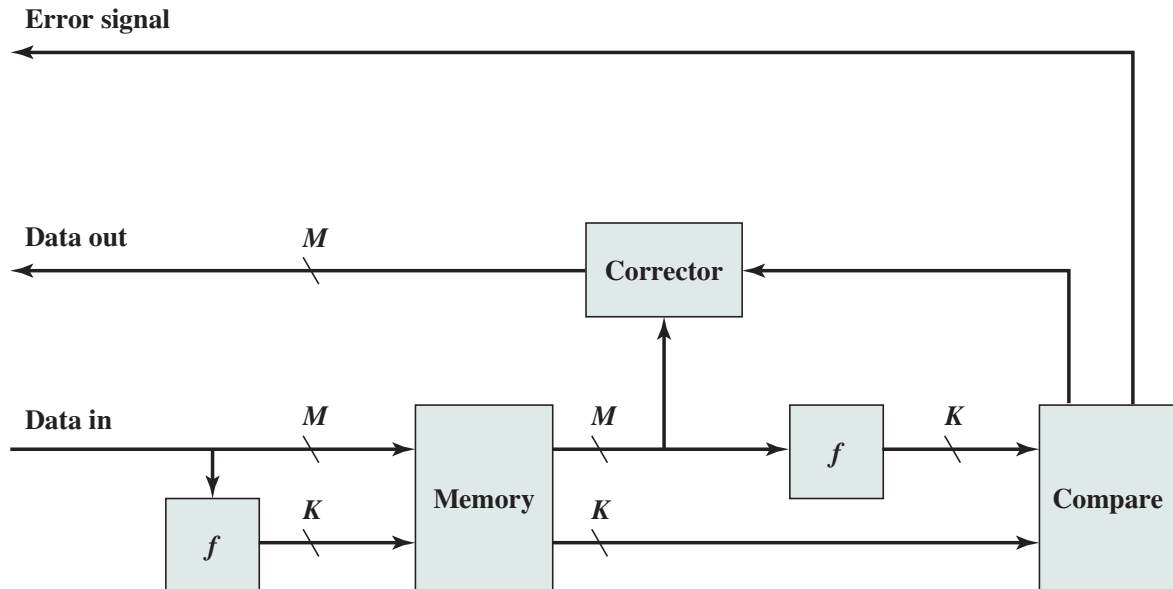


Figure 5.7 Error-Correcting Code Function

The simplest of the error-correcting codes is the **Hamming code** devised by Richard Hamming at Bell Laboratories. Figure 5.8 uses Venn diagrams to illustrate the use of this code on 4-bit words ($M = 4$). With three intersecting circles, there are seven compartments. We assign the 4 data bits to the inner compartments (Figure 5.8a). The remaining compartments are filled with what are called *parity bits*. Each parity bit is chosen so that the total number of 1s in its circle is even (Figure 5.8b). Thus, because circle A includes three data 1s, the parity bit in that circle is set to 1. Now, if an error changes one of the data bits (Figure 5.8c), it is easily found. By checking the parity bits, discrepancies are found in circle A and circle C but not in circle B. Only one of the seven compartments is in A and C but not B (Figure 5.8d). The error can therefore be corrected by changing that bit.

To clarify the concepts involved, we will develop a code that can detect and correct single-bit errors in 8-bit words.

To start, let us determine how long the code must be. Referring to Figure 5.7, the comparison logic receives as input two K -bit values. A bit-by-bit comparison is done by taking the exclusive-OR of the two inputs. The result is called the *syndrome word*. Thus, each bit of the **syndrome** is 0 or 1 according to if there is or is not a match in that bit position for the two inputs.

The syndrome word is therefore K bits wide and has a range between 0 and $2^K - 1$. The value 0 indicates that no error was detected, leaving $2^K - 1$ values to indicate, if there is an error, which bit was in error. Now, because an error could occur on any of the M data bits or K check bits, we must have

$$2^K - 1 \geq M + K$$

This inequality gives the number of bits needed to correct a single bit error in a word containing M data bits. For example, for a word of 8 data bits ($M = 8$), we have

- $K = 3: 2^3 - 1 < 8 + 3$
- $K = 4: 2^4 - 1 > 8 + 4$

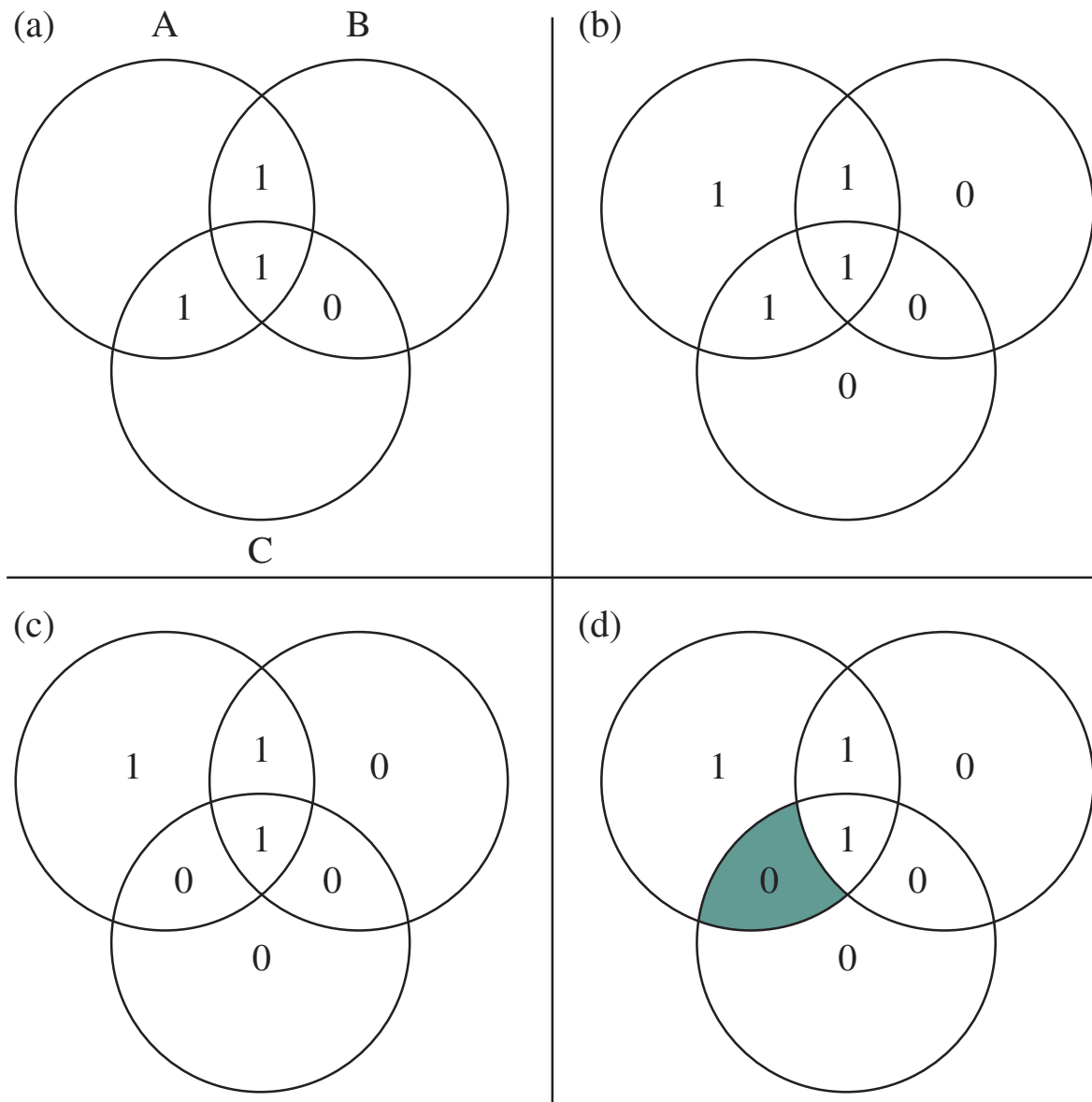


Figure 5.8 Hamming Error-Correcting Code

Thus, eight data bits require four check bits. The first three columns of Table 5.2 lists the number of check bits required for various data word lengths.

For convenience, we would like to generate a 4-bit syndrome for an 8-bit data word with the following characteristics:

- If the syndrome contains all 0s, no error has been detected.
- If the syndrome contains one and only one bit set to 1, then an error has occurred in one of the 4 check bits. No correction is needed.
- If the syndrome contains more than one bit set to 1, then the numerical value of the syndrome indicates the position of the data bit in error. This data bit is inverted for correction.

To achieve these characteristics, the data and check bits are arranged into a 12-bit word as depicted in Figure 5.9. The bit positions are numbered from 1 to 12. Those bit positions whose position numbers are powers of 2 are designated as check

Table 5.2 Increase in Word Length with Error Correction

Data Bits	Single-Error Correction		Single-Error Correction/ Double-Error Detection	
	Check Bits	% Increase	Check Bits	% Increase
8	4	50.0	5	62.5
16	5	31.25	6	37.5
32	6	18.75	7	21.875
64	7	10.94	8	12.5
128	8	6.25	9	7.03
256	9	3.52	10	3.91

bits. The check bits are calculated as follows, where the symbol \oplus designates the exclusive-OR operation:

$$\begin{aligned}
 C1 &= D1 \oplus D2 \oplus D4 \oplus D5 \oplus D7 \\
 C2 &= D1 \oplus D3 \oplus D4 \oplus D6 \oplus D7 \\
 C4 &= D2 \oplus D3 \oplus D4 \oplus D8 \\
 C8 &= D5 \oplus D6 \oplus D7 \oplus D8
 \end{aligned}$$

Each check bit operates on every data bit whose position number contains a 1 in the same bit position as the position number of that check bit. Thus, data bit positions 3, 5, 7, 9, and 11 ($D1, D2, D4, D5, D7$) all contain a 1 in the least significant bit of their position number as does $C1$; bit positions 3, 6, 7, 10, and 11 all contain a 1 in the second bit position, as does $C2$; and so on. Looked at another way, bit position n is checked by those bits C_i such that $\sum_i i = n$. For example, position 7 is checked by bits in position 4, 2, and 1; and $7 = 4 + 2 + 1$.

Let us verify that this scheme works with an example. Assume that the 8-bit input word is 00111001, with data bit $D1$ in the rightmost position. The calculations are as follows:

$$\begin{aligned}
 C1 &= 1 \oplus 0 \oplus 1 \oplus 1 \oplus 0 = 1 \\
 C2 &= 1 \oplus 0 \oplus 1 \oplus 1 \oplus 0 = 1 \\
 C4 &= 0 \oplus 0 \oplus 1 \oplus 0 = 1 \\
 C8 &= 1 \oplus 1 \oplus 0 \oplus 0 = 0
 \end{aligned}$$

Bit position	12	11	10	9	8	7	6	5	4	3	2	1
Position number	1100	1011	1010	1001	1000	0111	0110	0101	0100	0011	0010	0001
Data bit	D8	D7	D6	D5		D4	D3	D2		D1		
Check bit					C8				C4		C2	C1

Figure 5.9 Layout of Data Bits and Check Bits

Suppose now that data bit 3 sustains an error and is changed from 0 to 1. When the check bits are recalculated, we have

$$C1 = 1 \oplus 0 \oplus 1 \oplus 1 \oplus 0 = 1$$

$$C2 = 1 \oplus 1 \oplus 1 \oplus 1 \oplus 0 = 0$$

$$C4 = 0 \oplus 1 \oplus 1 \oplus 0 = 0$$

$$C8 = 1 \oplus 1 \oplus 0 \oplus 0 = 0$$

When the new check bits are compared with the old check bits, the syndrome word is formed:

	C8	C4	C2	C1
	0	1	1	1
\oplus	0	0	0	1
	0	1	1	0

The result is 0110, indicating that bit position 6, which contains data bit 3, is in error.

Figure 5.10 illustrates the preceding calculation. The data and check bits are positioned properly in the 12-bit word. Four of the data bits have a value 1 (shaded in the table), and their bit position values are XORed to produce the Hamming code 0111, which forms the four check digits. The entire block that is stored is 001101001111. Suppose now that data bit 3, in bit position 6, sustains an error and is changed from 0 to 1. The resulting block is 001101101111, with a Hamming code of 0001. An XOR of the Hamming code and all of the bit position values for nonzero data bits results in 0110. The nonzero result detects an error and indicates that the error is in bit position 6.

The code just described is known as a **single-error-correcting (SEC) code**. More commonly, semiconductor memory is equipped with a **single-error-correcting, double-error-detecting (SEC-DED) code**. As Table 5.2 shows, such codes require one additional bit compared with SEC codes.

Figure 5.11 illustrates how such a code works, again with a 4-bit data word. The sequence shows that if two errors occur (Figure 5.11c), the checking procedure goes astray (d) and worsens the problem by creating a third error (e). To overcome

Bit position	12	11	10	9	8	7	6	5	4	3	2	1
Position number	1100	1011	1010	1001	1000	0111	0110	0101	0100	0011	0010	0001
Data bit	D8	D7	D6	D5		D4	D3	D2		D1		
Check bit					C8				C4		C2	C1
Word stored as	0	0	1	1	0	1	0	0	1	1	1	1
Word fetched as	0	0	1	1	0	1	1	0	1	1	1	1
Position number	1100	1011	1010	1001	1000	0111	0110	0101	0100	0011	0010	0001
Check bit					0				0		0	1

Figure 5.10 Check Bit Calculation

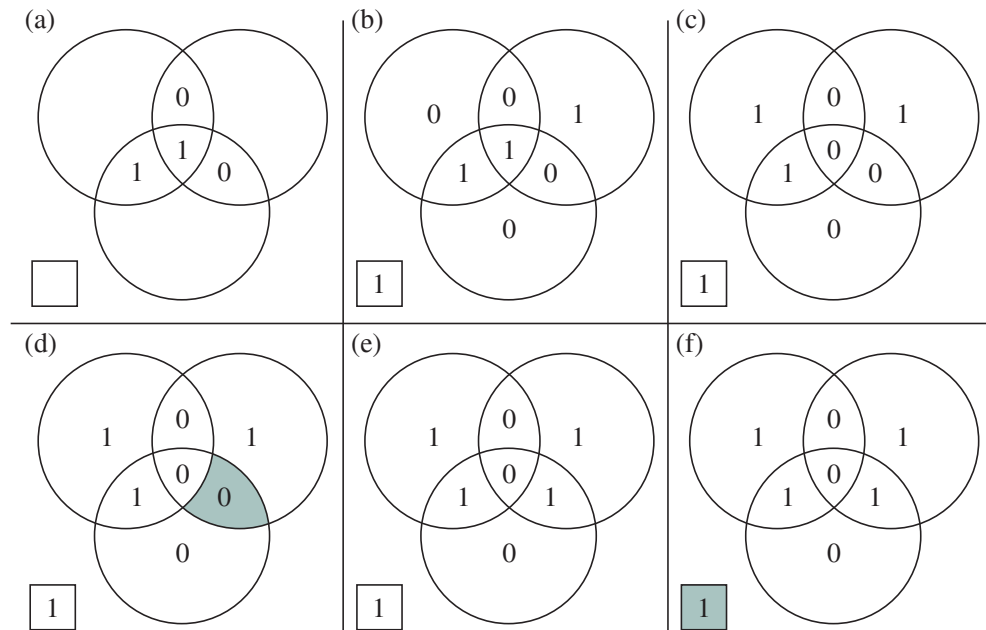


Figure 5.11 Hamming SEC-DEC Code

the problem, an eighth bit is added that is set so that the total number of 1s in the diagram is even. The extra parity bit catches the error (f).

An error-correcting code enhances the reliability of the memory at the cost of added complexity. With a 1-bit-per-chip organization, an SEC-DED code is generally considered adequate. For example, the IBM 30xx implementations used an 8-bit SEC-DED code for each 64 bits of data in main memory. Thus, the size of main memory is actually about 12% larger than is apparent to the user. The VAX computers used a 7-bit SEC-DED for each 32 bits of memory, for a 22% overhead. Contemporary DRAM systems may have anywhere from 7% to 20% overhead [SHAR03].

5.3 DDR DRAM

As discussed in Chapter 1, one of the most critical system bottlenecks when using high-performance processors is the interface to internal main memory. This interface is the most important pathway in the entire computer system. The basic building block of main memory remains the DRAM chip, as it has for decades; until recently, there had been no significant changes in DRAM architecture since the early 1970s. The traditional DRAM chip is constrained both by its internal architecture and by its interface to the processor's memory bus.

We have seen that one attack on the performance problem of DRAM main memory has been to insert one or more levels of high-speed SRAM cache between the DRAM main memory and the processor. But SRAM is much costlier than DRAM, and expanding cache size beyond a certain point yields diminishing returns.

In recent years, a number of enhancements to the basic DRAM architecture have been explored. The schemes that currently dominate the market are SDRAM and DDR-DRAM. We examine each of these in turn.

Synchronous DRAM

One of the most widely used forms of DRAM is the **synchronous DRAM (SDRAM)**. Unlike the traditional DRAM, which is asynchronous, the SDRAM exchanges data with the processor synchronized to an external clock signal and running at the full speed of the processor/memory bus without imposing wait states.

In a typical DRAM, the processor presents addresses and control levels to the memory, indicating that a set of data at a particular location in memory should be either read from or written into the DRAM. After a delay, the access time, the DRAM either writes or reads the data. During the access-time delay, the DRAM performs various internal functions, such as activating the high capacitance of the row and column lines, sensing the data, and routing the data out through the output buffers. The processor must simply wait through this delay, slowing system performance.

With synchronous access, the DRAM moves data in and out under control of the system clock. The processor or other master issues the instruction and address information, which is latched by the DRAM. The DRAM then responds after a set number of clock cycles. Meanwhile, the master can safely do other tasks while the SDRAM is processing the request.

Figure 5.12 shows the internal logic of a typical 256-Mb SDRAM typical of SDRAM organization, and Table 5.3 defines the various pin assignments. The

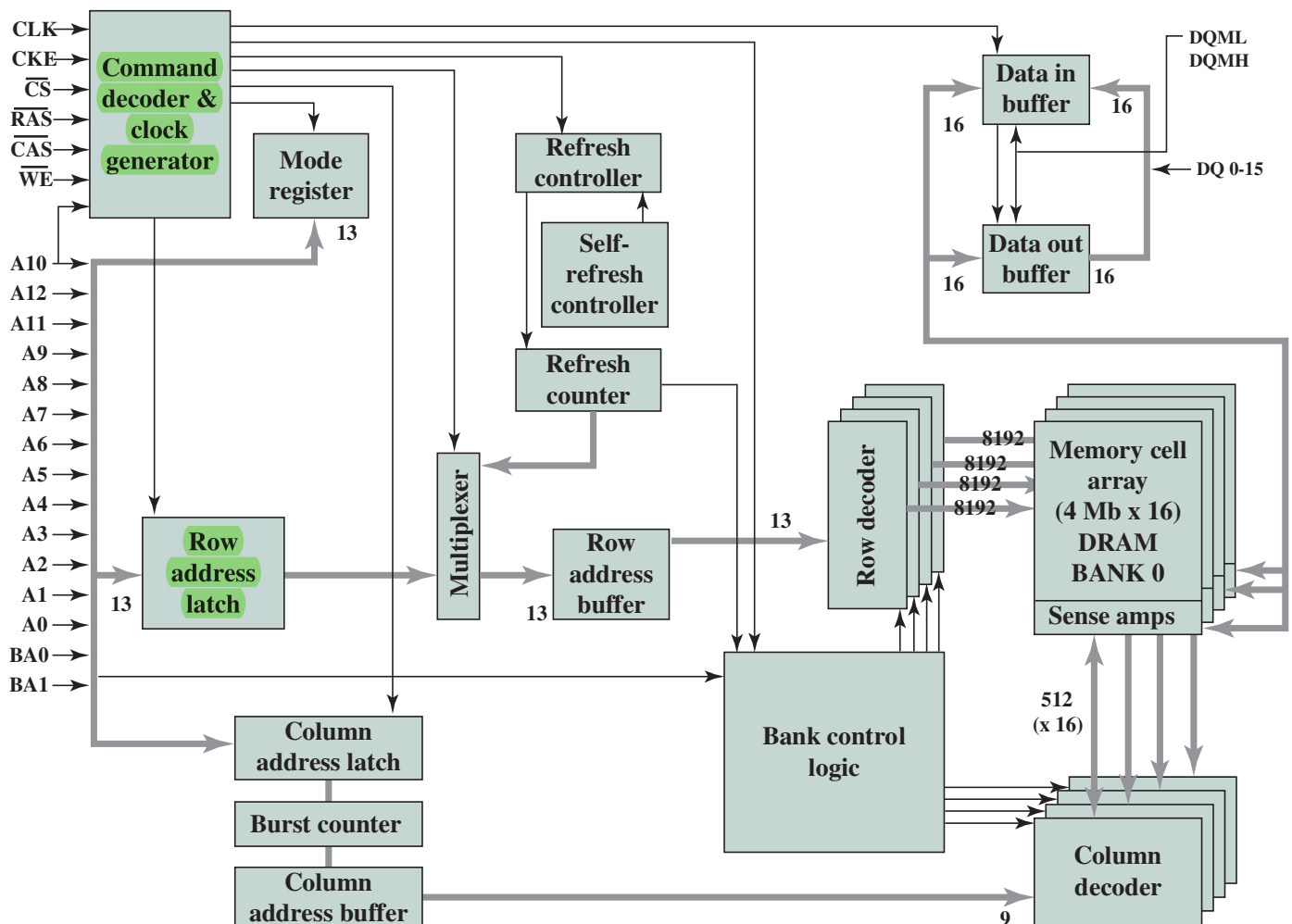


Figure 5.12 256-Mb Synchronous Dynamic RAM (SDRAM)

Table 5.3 SDRAM Pin Assignments

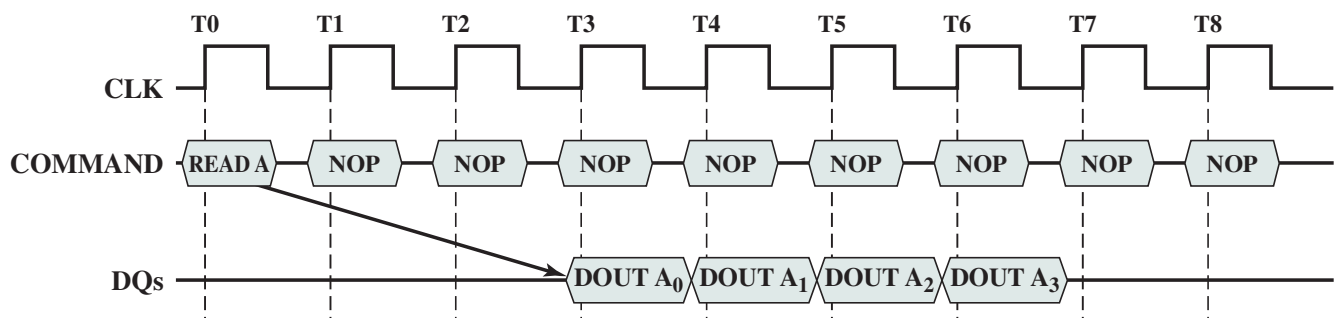
A0 to A13	Address inputs
BA0, BA1	Bank address lines
CLK	Clock input
CKE	Clock enable
$\overline{\text{CS}}$	Chip select
$\overline{\text{RAS}}$	Row address strobe
$\overline{\text{CAS}}$	Column address strobe
$\overline{\text{WE}}$	Write enable
DQ0 to DQ7	Data input/output
DQM	Data mask

SDRAM employs a burst mode to eliminate the address setup time and row and column line precharge time after the first access. In burst mode, a series of data bits can be clocked out rapidly after the first bit has been accessed. This mode is useful when all the bits to be accessed are in sequence and in the same row of the array as the initial access. In addition, the SDRAM has a multiple-bank internal architecture that improves opportunities for on-chip parallelism.

The mode register and associated control logic is another key feature differentiating SDRAMs from conventional DRAMs. It provides a mechanism to customize the SDRAM to suit specific system needs. The mode register specifies the burst length, which is the number of separate units of data synchronously fed onto the bus. The register also allows the programmer to adjust the latency between receipt of a read request and the beginning of data transfer.

The SDRAM performs best when it is transferring large blocks of data sequentially, such as for applications like word processing, spreadsheets, and multimedia.

Figure 5.13 shows an example of SDRAM operation. In this case, the burst length is 4 and the latency is 2. The burst read command is initiated by having $\overline{\text{CS}}$ and $\overline{\text{CAS}}$ low while holding $\overline{\text{RAS}}$ and $\overline{\text{WE}}$ high at the rising edge of the clock. The address inputs determine the starting column address for the burst, and the mode register sets the type of burst (sequential or interleave) and the burst length (1, 2, 4, 8, full page). The delay from the start of the command to when the data from the first cell appears on the outputs is equal to the value of the $\overline{\text{CAS}}$ latency that is set in the mode register.

**Figure 5.13** SDRAM Read Timing (burst length = 4, $\overline{\text{CAS}}$ latency = 2)

DDR SDRAM

Although SDRAM is a significant improvement on asynchronous RAM, it still has shortcomings that unnecessarily limit that I/O data rate that can be achieved. To address these shortcomings a newer version of SDRAM, referred to as double-data-rate DRAM (DDR DRAM) provides several features that dramatically increase the data rate. DDR DRAM was developed by the JEDEC Solid State Technology Association, the Electronic Industries Alliance's semiconductor-engineering-standardization body. Numerous companies make DDR chips, which are widely used in desktop computers and servers.

DDR achieves higher data rates in three ways. First, the data transfer is synchronized to both the rising and falling edge of the clock, rather than just the rising edge. This doubles the data rate; hence the term *double data rate*. Second, DDR uses higher clock rate on the bus to increase the transfer rate. Third, a buffering scheme is used, as explained subsequently.

JEDEC has thus far defined four generations of the DDR technology (Table 5.4). The initial DDR version makes use of a 2-bit prefetch buffer. The prefetch buffer is a memory cache located on the SDRAM chip. It enables the SDRAM chip to preposition bits to be placed on the data bus as rapidly as possible. The DDR I/O bus uses the same clock rate as the memory chip, but because it can handle two bits per cycle, it achieves a data rate that is double the clock rate. The 2-bit prefetch buffer enables the SDRAM chip to keep up with the I/O bus.

To understand the operation of the prefetch buffer, we need to look at it from the point of view of a word transfer. The prefetch buffer size determines how many words of data are fetched (across multiple SDRAM chips) every time a column command is performed with DDR memories. Because the core of the DRAM is much slower than the interface, the difference is bridged by accessing information in parallel and then serializing it out the interface through a multiplexor (MUX). Thus, DDR prefetches two words, which means that every time a read or a write operation is performed, it is performed on two words of data, and bursts out of, or into, the SDRAM over one clock cycle on both clock edges for a total of two consecutive operations. As a result, the DDR I/O interface is twice as fast as the SDRAM core.

Although each new generation of SDRAM results in much greater capacity, the core speed of the SDRAM has not changed significantly from generation to generation. To achieve greater data rates than those afforded by the rather modest increases in SDRAM clock rate, JEDEC increased the buffer size. For DDR2, a 4-bit buffer is used, allowing for words to be transferred in parallel, increasing the effective data rate by a factor of 4. For DDR3, an 8-bit buffer is used and a factor of 8 speedup is achieved (Figure 5.14).

Table 5.4 DDR Characteristics

	DDR1	DDR2	DDR3	DDR4
Prefetch buffer (bits)	2	4	8	8
Voltage level (V)	2.5	1.8	1.5	1.2
Front side bus data rates (Mbps)	200–400	400–1066	800–2133	2133–4266

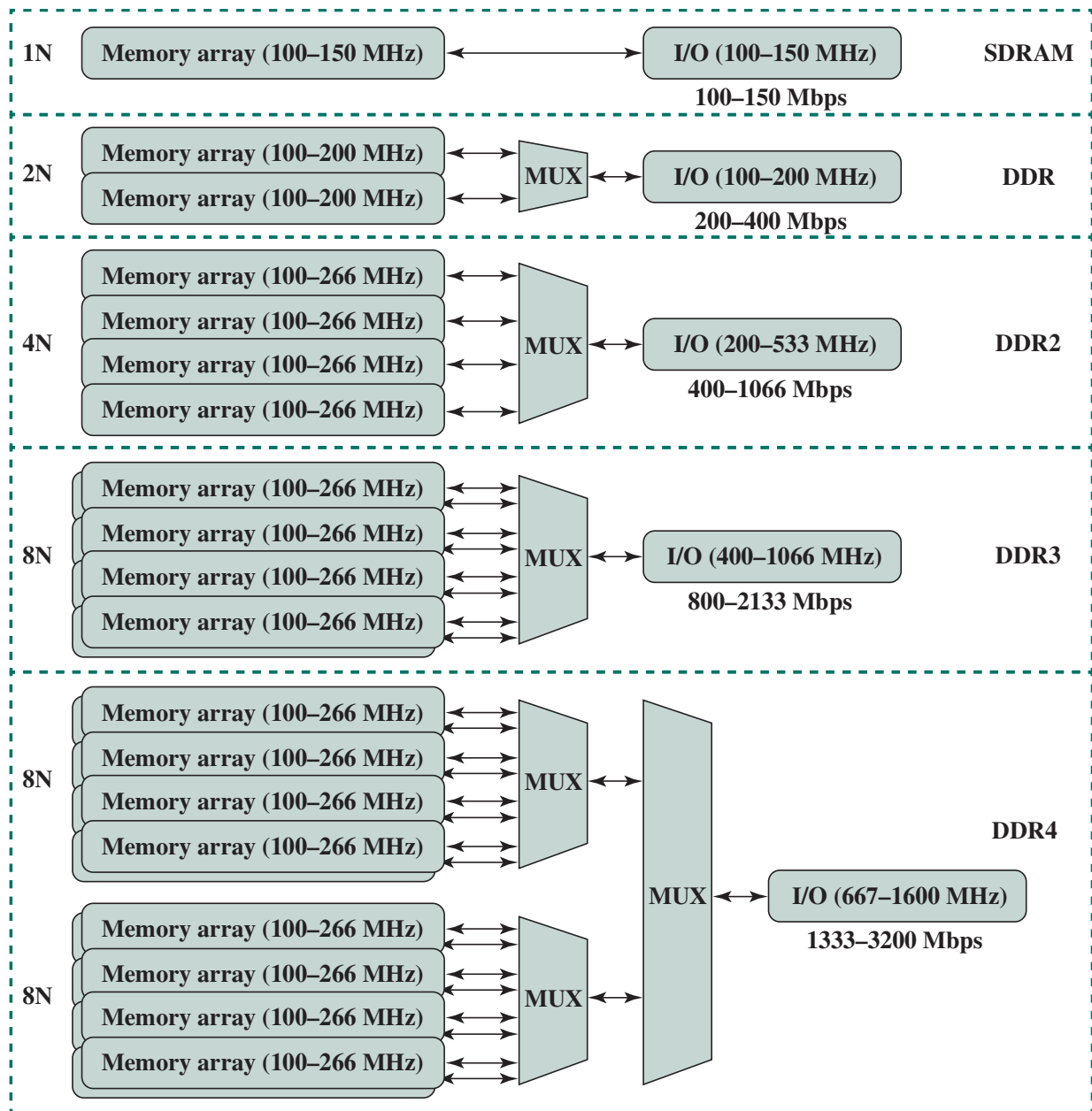


Figure 5.14 DDR Generations

The downside to the prefetch is that it effectively determines the minimum burst length for the SDRAMs. For example, it is very difficult to have an efficient burst length of four words with DDR3's prefetch of eight. Accordingly, the JEDEC designers chose not to increase the buffer size to 16 bits for DDR4, but rather to introduce the concept of a **bank group** [ALLA13]. Bank groups are separate entities such that they allow a column cycle to complete within a bank group, but that column cycle does not impact what is happening in another bank group. Thus, two prefetches of eight can be operating in parallel in the two bank groups. This arrangement keeps the prefetch buffer size the same as for DDR3, while increasing performance as if the prefetch is larger.

Figure 5.14 shows a configuration with two bank groups. With DDR4, up to 4 bank groups can be used.

5.4 FLASH MEMORY

Another form of semiconductor memory is flash memory. Flash memory is used both for internal memory and external memory applications. Here, we provide a technical overview and look at its use for internal memory.

First introduced in the mid-1980s, flash memory is intermediate between EPROM and EEPROM in both cost and functionality. Like EEPROM, flash memory uses an electrical erasing technology. An entire flash memory can be erased in one or a few seconds, which is much faster than EPROM. In addition, it is possible to erase just blocks of memory rather than an entire chip. Flash memory gets its name because the microchip is organized so that a section of memory cells are erased in a single action or “flash.” However, flash memory does not provide byte-level erasure. Like EPROM, flash memory uses only one transistor per bit, and so achieves the high density (compared with EEPROM) of EPROM.

Operation

Figure 5.15 illustrates the basic operation of a flash memory. For comparison, Figure 5.15a depicts the operation of a transistor. Transistors exploit the properties of semiconductors so that a small voltage applied to the gate can be used to control the flow of a large current between the source and the drain.

In a flash memory cell, a second gate—called a floating gate, because it is insulated by a thin oxide layer—is added to the transistor. Initially, the floating gate does not interfere with the operation of the transistor (Figure 5.15b). In this state, the cell is deemed to represent binary 1. Applying a large voltage across the oxide layer causes electrons to tunnel through it and become trapped on the floating gate, where they remain even if the power is disconnected (Figure 5.15c). In this state, the cell is deemed to represent binary 0. The state of the cell can be read by using external circuitry to test whether the transistor is working or not. Applying a large voltage in the opposite direction removes the electrons from the floating gate, returning to a state of binary 1.

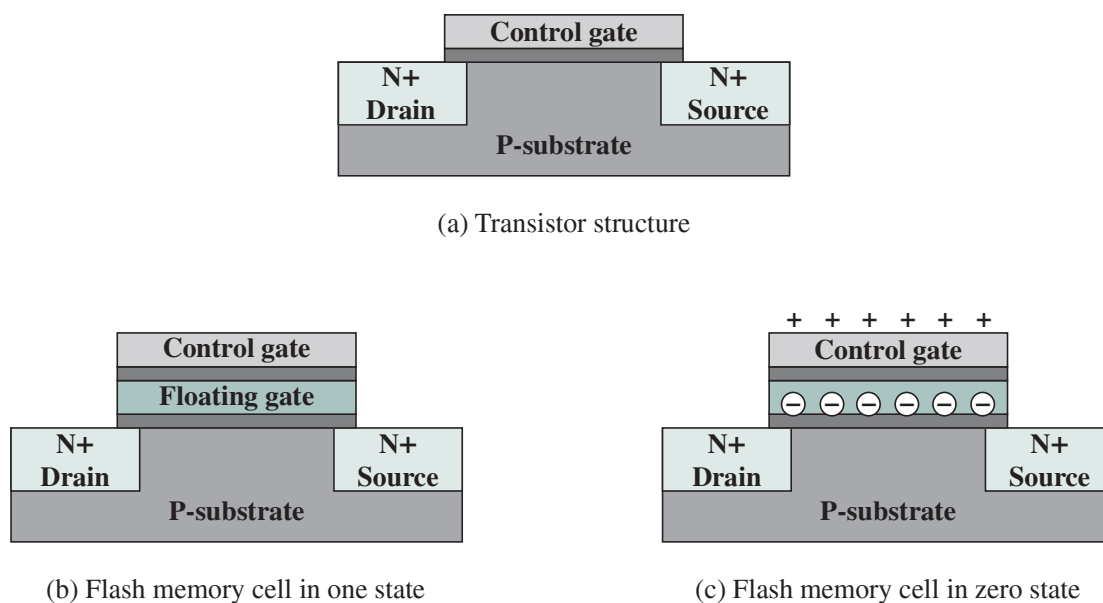


Figure 5.15 Flash Memory Operation

An important characteristic of flash memory is that it is persistent memory, which means that it retains data when there is no power applied to the memory. Thus, it is useful for secondary (external) storage, and as an alternative to random access memory in computers.

NOR and NAND Flash Memory

There are two distinctive types of flash memory, designated as NOR and NAND (Figure 5.16). In **NOR flash memory**, the basic unit of access is a bit, referred to as a *memory cell*. Cells in NOR flash are connected in parallel to the bit lines so that each cell can be read/write/erased individually. If any memory cell of the device is turned on by the corresponding word line, the bit line goes low. This is similar in function to a NOR logic gate.²

NAND flash memory is organized in transistor arrays with 16 or 32 transistors in series. The bit line goes low only if all the transistors in the corresponding word lines are turned on. This is similar in function to a NAND logic gate.

Although the specific quantitative values of various characteristics of NOR and NAND are changing year by year, the relative differences between the two types has remained stable. These differences are usefully illustrated by the Kiviat graphs³ shown in Figure 5.17.

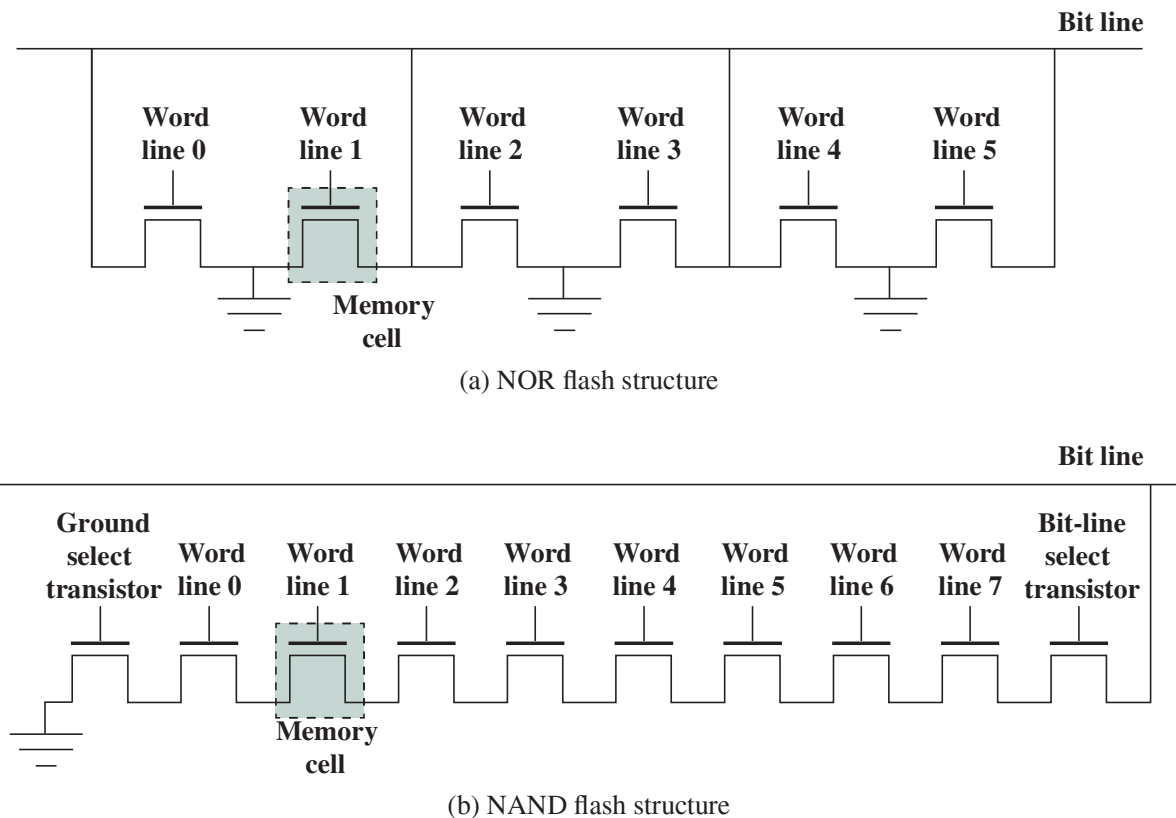


Figure 5.16 Flash Memory Structures

²The circles associated with and in Figure 5.2b indicate signal negation.

³A Kiviat graph provides a pictorial means of comparing systems along multiple variables [MORR74]. The variables are laid out at as lines of equal angular intervals within a circle, each line going from the center of the circle to the circumference. A given system is defined by one point on each line; the closer to the circumference, the better the value. The points are connected to yield a shape that is characteristic of that system. The more area enclosed in the shape, the “better” is the system.

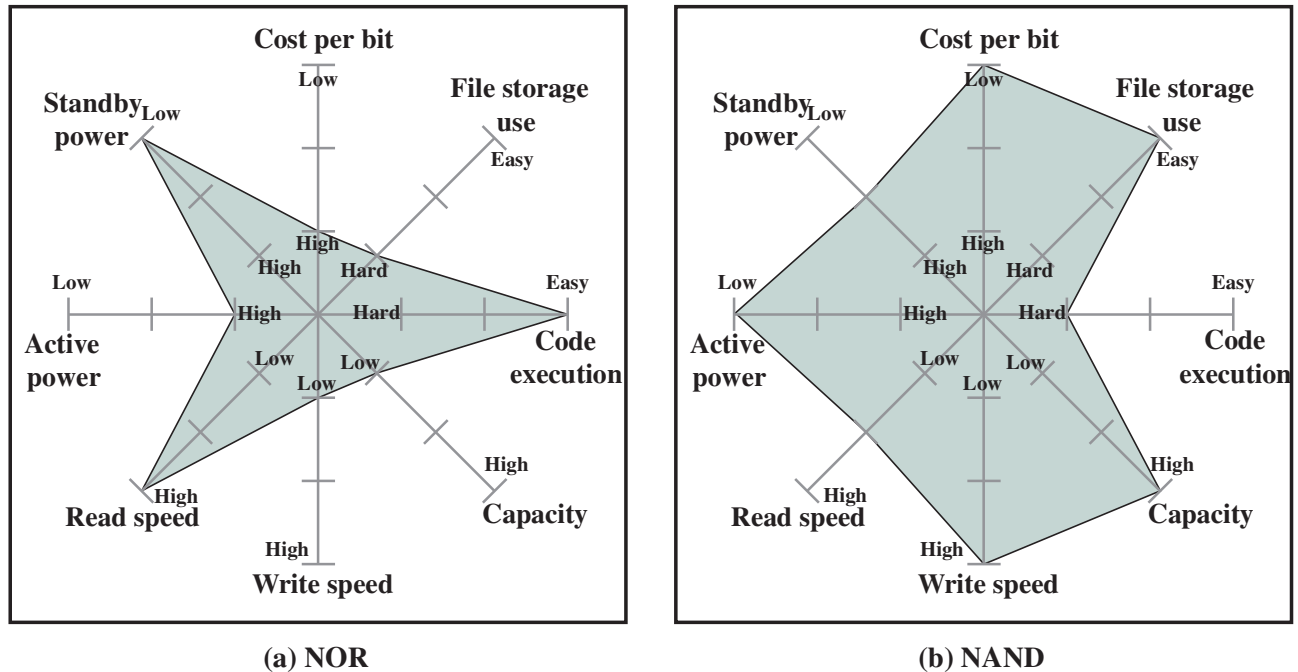


Figure 5.17 Kiviat Graphs for Flash Memory

NOR flash memory provides high-speed random access. It can read and write data to specific locations, and can reference and retrieve a single byte. NAND reads and writes in small blocks. NAND provides higher bit density than NOR and greater write speed. NAND flash does not provide a random-access external address bus so the data must be read on a blockwise basis (also known as page access), where each block holds hundreds to thousands of bits.

For internal memory in embedded systems, NOR flash memory has traditionally been preferred. NAND memory has made some inroads, but NOR remains the dominant technology for internal memory. It is ideally suited for microcontrollers where the amount of program code is relatively small and a certain amount of application data does not vary. For example, the flash memory in Figure 1.16 is NOR memory.

NAND memory is better suited for external memory, such as USB flash drives, memory cards (in digital cameras, MP3 players, etc.), and in what are known as solid-state disks (SSDs). We discuss SSDs in Chapter 6.

5.5 NEWER NONVOLATILE SOLID-STATE MEMORY TECHNOLOGIES

The traditional memory hierarchy has consisted of three levels (Figure 5.18):

- **Static RAM (SRAM):** SRAM provides rapid access time, but is the most expensive and the least dense (bit density). SRAM is suitable for cache memory.
- **Dynamic RAM (DRAM):** Cheaper, denser, and slower than SRAM, DRAM has traditionally been the choice off-chip main memory.
- **Hard disk:** A magnetic disk provides very high bit density and very low cost per bit, with relatively slow access times. It is the traditional choice for external storage as part of the memory hierarchy.

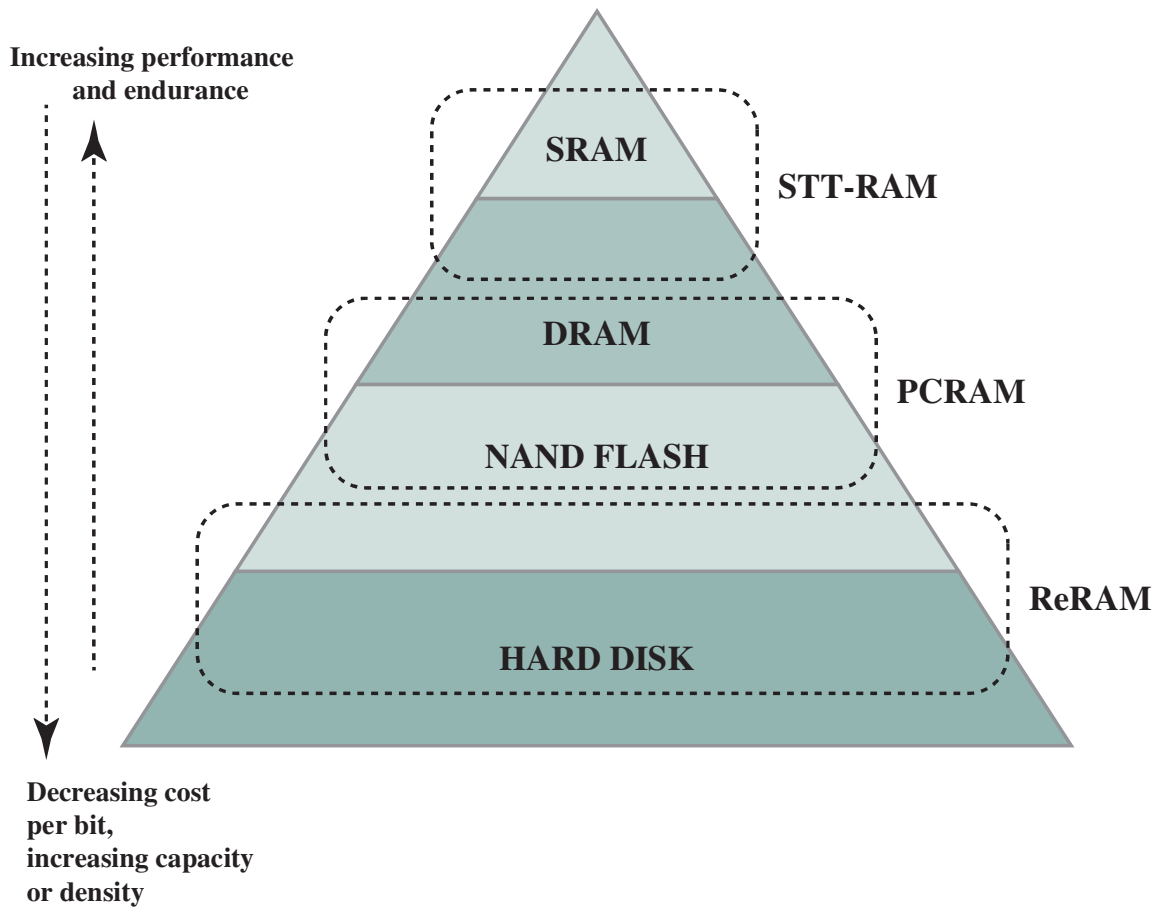


Figure 5.18 Nonvolatile RAM within the Memory Hierarchy

Into this mix, as we have seen, as been added flash memory. Flash memory has the advantage over traditional memory that it is nonvolatile. NOR flash is best suited to storing programs and static application data in embedded systems, while NAND flash has characteristics intermediate between DRAM and hard disks.

Over time, each of these technologies has seen improvements in scaling: higher bit density, higher speed, lower power consumption, and lower cost. However, for semiconductor memory, it is becoming increasingly difficult to continue the pace of improvement [ITRS14].

Recently, there have been breakthroughs in developing new forms of non-volatile semiconductor memory that continue scaling beyond flash memory. The most promising technologies are spin-transfer torque RAM (STT-RAM), phase-change RAM (PCRAM), and resistive RAM (ReRAM) ([ITRS14], [GOER12]). All of these are in volume production. However, because NAND Flash and to some extent NOR Flash are still dominating the applications, these emerging memories have been used in specialty applications and have not yet fulfilled their original promise to become dominating mainstream high-density nonvolatile memory. This is likely to change in the next few years.

Figure 5.18 shows how these three technologies are likely to fit into the memory hierarchy.

STT-RAM

STT-RAM is a new type of **magnetic RAM (MRAM)**, which features non-volatility, fast writing/reading speed (< 10 ns), and high programming endurance ($> 10^{15}$ cycles) and zero standby power [KULT13]. The storage capability or programmability of MRAM arises from magnetic tunneling junction (MTJ), in which a thin tunneling dielectric is sandwiched between two ferromagnetic layers. One ferromagnetic layer (pinned or reference layer) is designed to have its magnetization pinned, while the magnetization of the other layer (free layer) can be flipped by a write event. An MTJ has a low (high) resistance if the magnetizations of the free layer and the pinned layer are parallel (anti-parallel). In first-generation MRAM design, the magnetization of the free layer is changed by the current-induced magnetic field. In STT-RAM, a new write mechanism, called *polarization-current-induced magnetization switching*, is introduced. For STT-RAM, the magnetization of the free layer is flipped by the electrical current directly. Because the current required to switch an MTJ resistance state is proportional to the MTJ cell area, STT-RAM is believed to have a better scaling property than the first-generation MRAM. Figure 5.19a illustrates the general configuration.

STT-RAM is a good candidate for either cache or main memory.

PCRAM

Phase-change RAM (PCRAM) is the most mature of the new technologies, with an extensive technical literature ([RAOU09], [ZHOU09], [LEE10]).

PCRAM technology is based on a chalcogenide alloy material, which is similar to those commonly used in optical storage media (compact discs and digital versatile discs). The data storage capability is achieved from the resistance differences between an amorphous (high-resistance) and a crystalline (low-resistance) phase of the chalcogenide-based material. In SET operation, the phase change material is crystallized by applying an electrical pulse that heats a significant portion of the cell above its crystallization temperature. In RESET operation, a larger electrical current is applied and then abruptly cut off in order to melt and then quench the material, leaving it in the amorphous state. Figure 5.19b illustrates the general configuration.

PCRAM is a good candidate to replace or supplement DRAM for main memory.

ReRAM

ReRAM (also known as RRAM) works by creating resistance rather than directly storing charge. An electric current is applied to a material, changing the resistance of that material. The resistance state can then be measured and a 1 or 0 is read as the result. Much of the work done on ReRAM to date has focused on finding appropriate materials and measuring the resistance state of the cells. ReRAM designs are low voltage, endurance is far superior to flash memory, and the cells are much smaller—at least in theory. Figure 5.19c shows one ReRAM configuration.

ReRAM is a good candidate to replace or supplement both secondary storage and main memory.

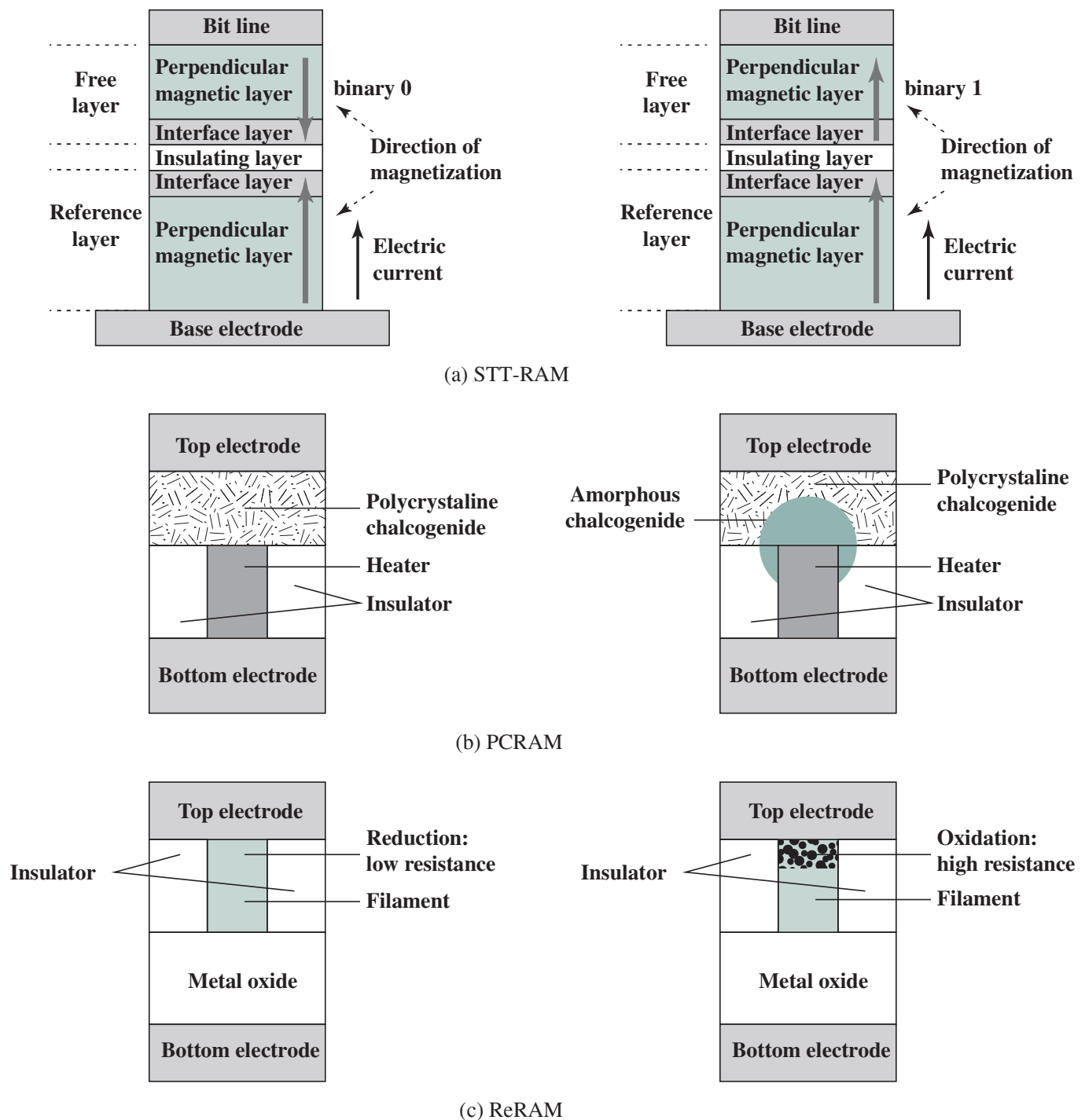


Figure 5.19 Nonvolatile RAM Technologies

5.6 KEY TERMS, REVIEW QUESTIONS, AND PROBLEMS

Key Terms

bank group double data rate DRAM (DDR DRAM) dynamic RAM (DRAM)	electrically erasable programmable ROM (EEPROM) erasable programmable ROM (EPROM)	error correcting code (ECC) error correction flash memory Hamming code hard failure
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magnetic RAM (MRAM) NAND flash memory nonvolatile memory NOR flash memory phase-change RAM (PCRAM) programmable ROM (PROM) random access memory (RAM)	read-mostly memory read-only memory (ROM) resistive RAM (ReRAM) semiconductor memory single-error-correcting (SEC) code single-error-correcting, double-error-detecting (SEC-DED) code	soft error spin-transfer torque RAM (STT-RAM) static RAM (SRAM) synchronous DRAM (SDRAM) syndrome volatile memory
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Review Questions

- 5.1 What are the key properties of semiconductor memory?
- 5.2 What are two interpretations of the term *random-access memory*?
- 5.3 What is the difference between DRAM and SRAM in terms of application?
- 5.4 What is the difference between DRAM and SRAM in terms of characteristics such as speed, size, and cost?
- 5.5 Explain why one type of RAM is considered to be analog and the other digital.
- 5.6 What are some applications for ROM?
- 5.7 What are the differences among EPROM, EEPROM, and flash memory?
- 5.8 Explain the function of each pin in Figure 5.4b.
- 5.9 What is a parity bit?
- 5.10 How is the syndrome for the Hamming code interpreted?
- 5.11 How does SDRAM differ from ordinary DRAM?
- 5.12 What is DDR RAM?
- 5.13 What is the difference between NAND and NOR flash memory?
- 5.14 List and briefly define three newer nonvolatile solid-state memory technologies.

Problems

- 5.1 Suggest reasons why RAMs traditionally have been organized as only one bit per chip whereas ROMs are usually organized with multiple bits per chip.
- 5.2 Consider a dynamic RAM that must be given a refresh cycle 64 times per ms. Each refresh operation requires 150 ns; a memory cycle requires 250 ns. What percentage of the memory's total operating time must be given to refreshes?
- 5.3 Figure 5.20 shows a simplified timing diagram for a DRAM read operation over a bus. The access time is considered to last from t_1 to t_2 . Then there is a recharge time, lasting from t_2 to t_3 , during which the DRAM chips will have to recharge before the processor can access them again.
 - a. Assume that the access time is 60 ns and the recharge time is 40 ns. What is the memory cycle time? What is the maximum data rate this DRAM can sustain, assuming a 1-bit output?
 - b. Constructing a 32-bit wide memory system using these chips yields what data transfer rate?
- 5.4 Figure 5.6 indicates how to construct a module of chips that can store 1 MB based on a group of four 256-Kbyte chips. Let's say this module of chips is packaged as a single 1-MB chip, where the word size is 1 byte. Give a high-level chip diagram of how to construct an 8-MB computer memory using eight 1-MB chips. Be sure to show the address lines in your diagram and what the address lines are used for.

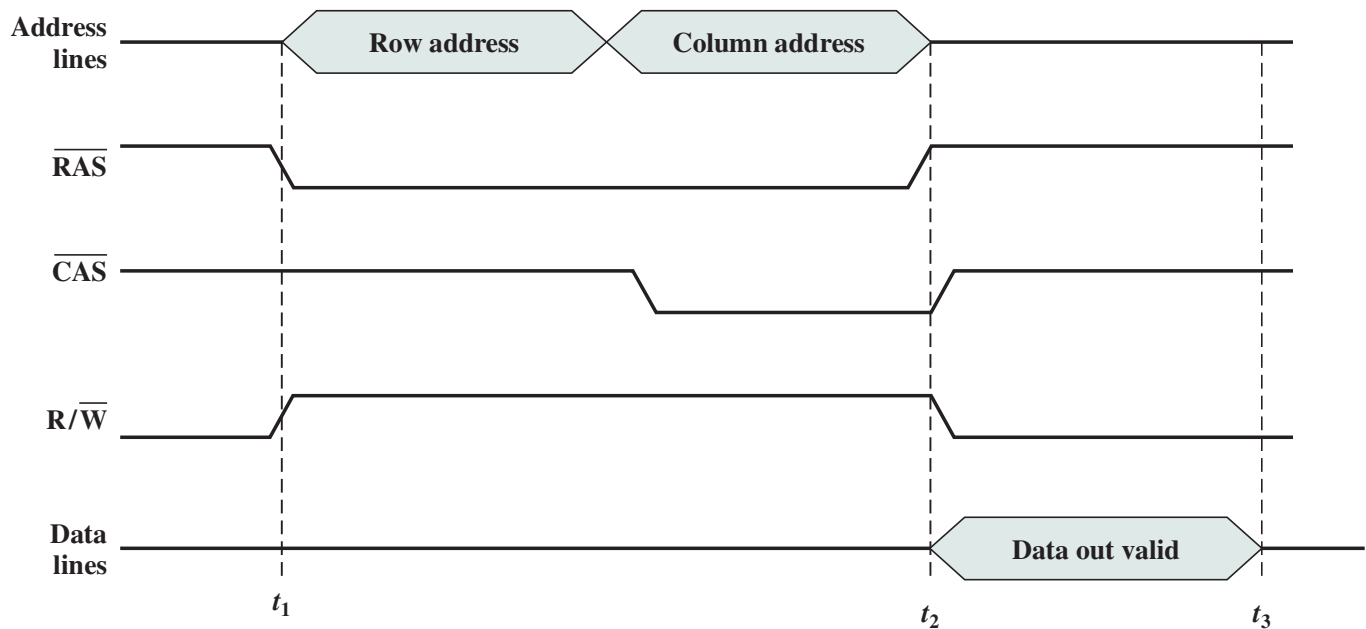
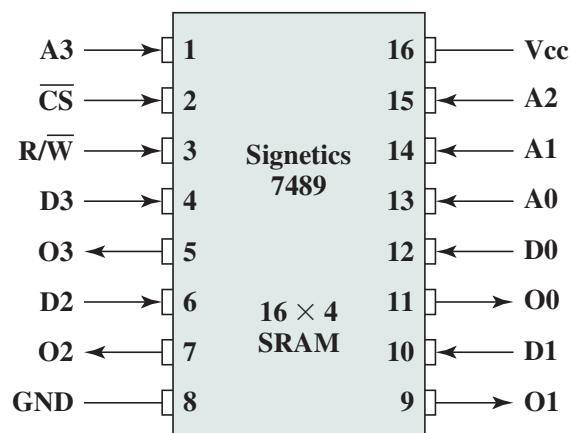


Figure 5.20 Simplified DRAM Read Timing

- 5.5** On a typical Intel 8086-based system, connected via system bus to DRAM memory, for a read operation, $\overline{\text{RAS}}$ is activated by the trailing edge of the Address Enable signal (Figure C.1 in Appendix C). However, due to propagation and other delays, $\overline{\text{RAS}}$ does not go active until 50 ns after Address Enable returns to a low. Assume the latter occurs in the middle of the second half of state T_1 (somewhat earlier than in Figure C.1). Data are read by the processor at the end of T_3 . For timely presentation to the processor, however, data must be provided 60 ns earlier by memory. This interval accounts for propagation delays along the data paths (from memory to processor) and processor data hold time requirements. Assume a clocking rate of 10 MHz.
- How fast (access time) should the DRAMs be if no wait states are to be inserted?
 - How many wait states do we have to insert per memory read operation if the access time of the DRAMs is 150 ns?
- 5.6** The memory of a particular microcomputer is built from $64\text{K} \times 1$ DRAMs. According to the data sheet, the cell array of the DRAM is organized into 256 rows. Each row must be refreshed at least once every 4 ms. Suppose we refresh the memory on a strictly periodic basis.
- What is the time period between successive refresh requests?
 - How long a refresh address counter do we need?
- 5.7** Figure 5.21 shows one of the early SRAMs, the 16×4 Signetics 7489 chip, which stores 16 4-bit words.
- List the mode of operation of the chip for each $\overline{\text{CS}}$ input pulse shown in Figure 5.21c.
 - List the memory contents of word locations 0 through 6 after pulse n.
 - What is the state of the output data leads for the input pulses h through m?
- 5.8** Design a 16-bit memory of total capacity 8192 bits using SRAM chips of size 64×1 bit. Give the array configuration of the chips on the memory board showing all required input and output signals for assigning this memory to the lowest address space. The design should allow for both byte and 16-bit word accesses.
- 5.9** A common unit of measure for failure rates of electronic components is the **Failure unit** (FIT), expressed as a rate of failures per billion device hours. Another well known but less used measure is **mean time between failures (MTBF)**, which is the average time of operation of a particular component until it fails. Consider a 1 MB memory of a 16-bit microprocessor with $256\text{K} \times 1$ DRAMs. Calculate its MTBF assuming 2000 FITS for each DRAM.



(a) Pin layout

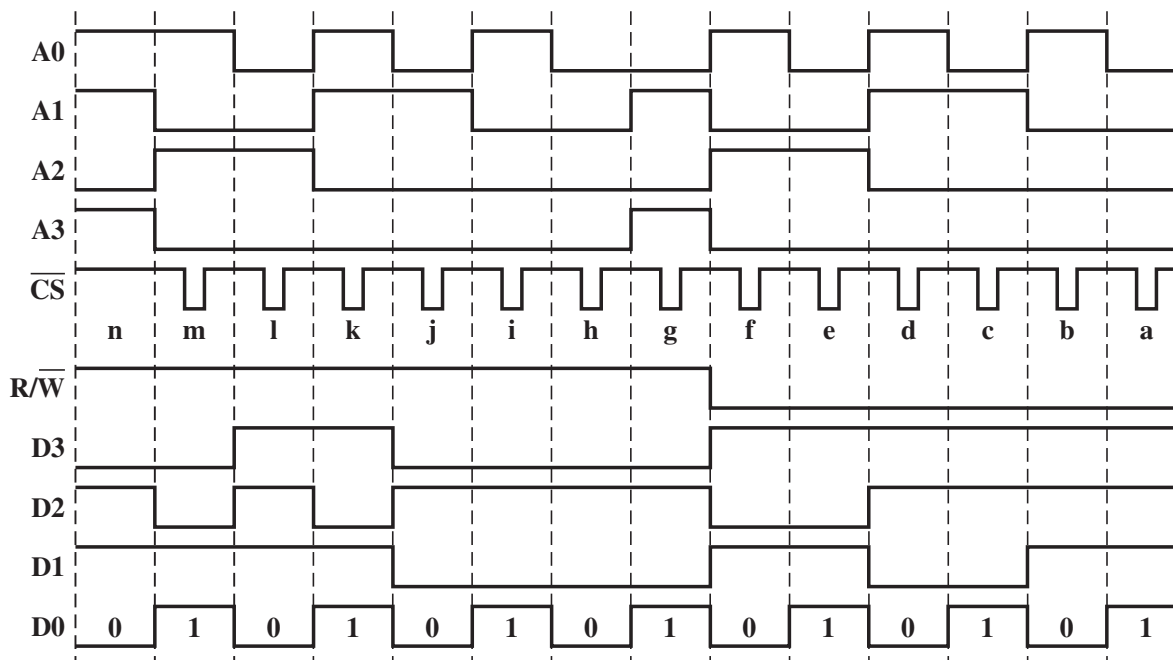
Operating Mode	Inputs			Outputs
	$\overline{\text{CS}}$	$\text{R}/\overline{\text{W}}$	D_n	O_n
Write	L	L	L	L
	L	L	H	H
Read	L	H	X	Data
Inhibit writing	H	L	L	H
	H	L	H	L
Store - disable outputs	H	H	X	H

H = high voltage level

L = low voltage level

X = don't care

(b) Truth table



(c) Pulse train

Figure 5.21 The Signetics 7489 SRAM

- 5.10** For the Hamming code shown in Figure 5.10, show what happens when a check bit rather than a data bit is in error?
- 5.11** Suppose an 8-bit data word stored in memory is 11000010. Using the Hamming algorithm, determine what check bits would be stored in memory with the data word. Show how you got your answer.
- 5.12** For the 8-bit word 00111001, the check bits stored with it would be 0111. Suppose when the word is read from memory, the check bits are calculated to be 1101. What is the data word that was read from memory?
- 5.13** How many check bits are needed if the Hamming error correction code is used to detect single bit errors in a 1024-bit data word?
- 5.14** Develop an SEC code for a 16-bit data word. Generate the code for the data word 0101000000111001. Show that the code will correctly identify an error in data bit 5.