

Control Systems

Varum SM*

Consider the positive-feedback circuit shown in Fig. 0.

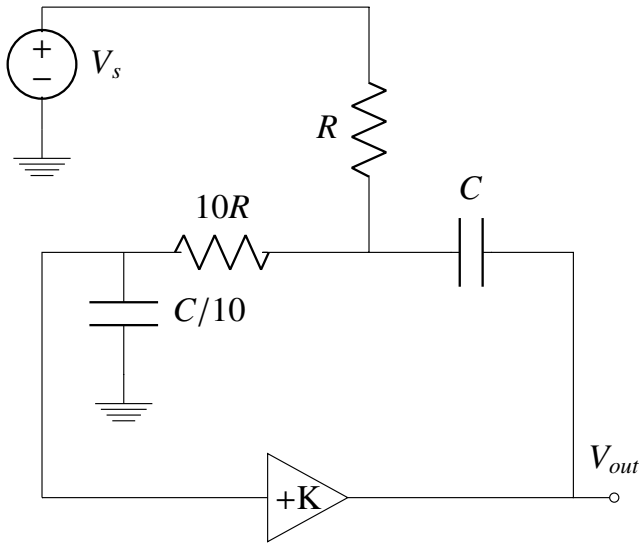


Fig. 0: Positive Feedback Circuit

- Find the loop transmission $L(s)$ and the characteristic equation, find the expressions for resulting pole frequency ω_o and Q factor?
- Sketch a Pole-Zero plot for varying K.
- For what value of K do the poles coincide? For what value of K does the response becomes maximally flat? For what value of K does the circuit oscillate?

Assume the amplifier has frequency-independent gain, infinite input impedance, and zero output impedance.

1. Compute $G(s)$

Solution: :

- For Gain $G(s)$ greater than 1, the gain block(K) in Fig:0 is built using LM741 op-amp.

$$G(s) = \frac{v_o}{v_i} = 1 + \frac{R_2}{R_1} \quad (1.1)$$

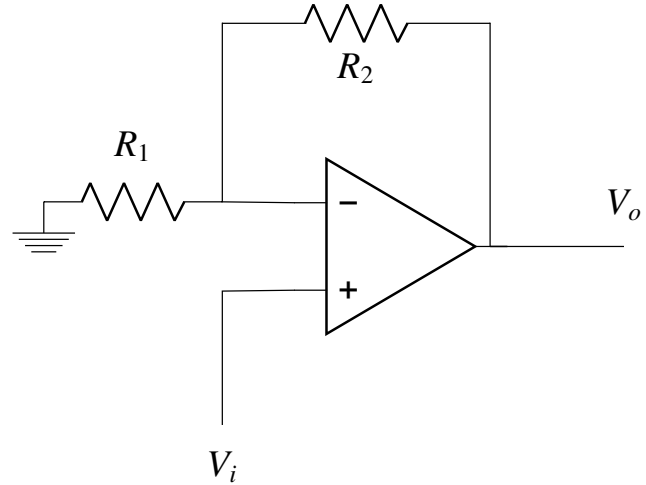


Fig. 1: LM741 op-amp

- For Gain $G(s)$ less than 1, the gain block(K) in Fig: 0 is built using the voltage divider.

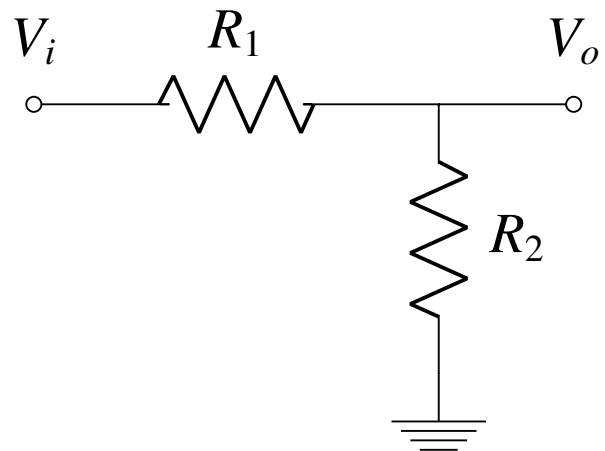


Fig. 1: Voltage Divider

$$G(s) = \frac{v_o}{v_i} = \frac{R_2}{R_1 + R_2} \quad (1.2)$$

2. Compute $H(s)$

Solution: :

- Consider the voltage at the end of resistor R

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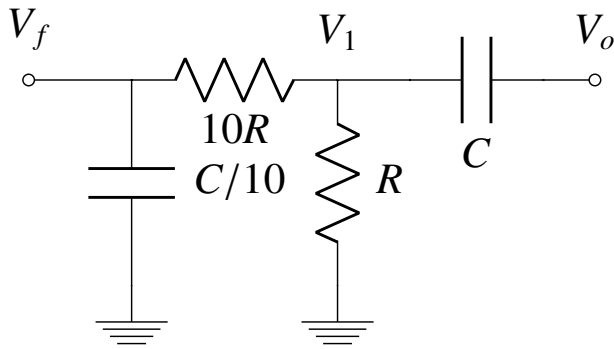


Fig. 2

to be V_1 . Apply KCL at nodes.

$$\frac{V_f - V_1}{10R} + \frac{V_f - 0}{\frac{10}{sC}} = 0 \implies V_1 = V_f(sCR + 1) \quad (2.1)$$

$$\frac{V_1 - V_f}{10R} + \frac{V_1 - 0}{R} + \frac{V_1 - V_o}{\frac{1}{sC}} = 0 \quad (2.2)$$

- Substitute V_1 from Eq: 2.1 in Eq: 2.2

$$H(s) = \frac{V_f}{V_o} = \frac{1}{sRC + \frac{1}{sRC} + 2.1} \quad (2.3)$$

3. Find $L(s)$.

Solution: To obtain the loop transmission $L(s)$,

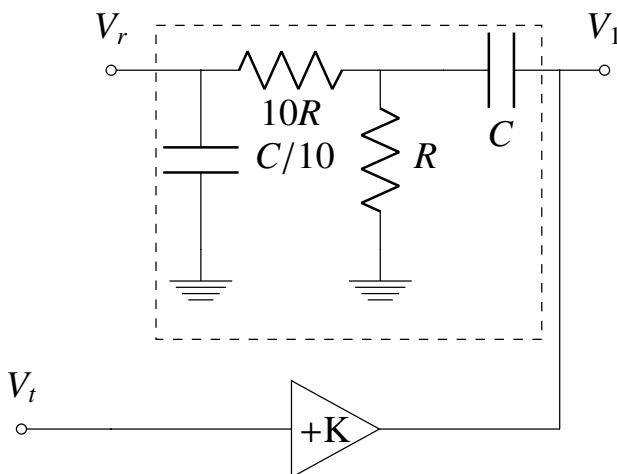


Fig. 3

The loop transmission is given by

$$L(s) = -\frac{V_r(s)}{V_t(s)} = -KH(s) \quad (3.1)$$

where $H(s)$ is the transfer function of the two-port RC network shown inside the broken-line

box in Figure: 3.

$$L(s) = \frac{-s(\frac{K}{CR})}{s^2 + s(\frac{2.1}{CR}) + (\frac{1}{CR})^2} \quad (3.2)$$

The characteristic equation is

$$1 + L(s) = 0 \quad (3.3)$$

$$s^2 + s(\frac{2.1 - K}{CR}) + (\frac{1}{CR})^2 = 0 \quad (3.4)$$

The standard characteristic equation of a second order network can be written as

$$s^2 + \frac{\omega_o}{Q}s + \omega_o^2 = 0 \quad (3.5)$$

ω_o is called pole frequency, Q is called pole Qfactor. By comparing the Eq:3.4 with the standard characteristic equation Eq:3.5

$$\omega_o = \frac{1}{RC}; Q = \frac{1}{2.1 - K} \quad (3.6)$$

4. Equivalent control system model of Fig. 3

Solution:

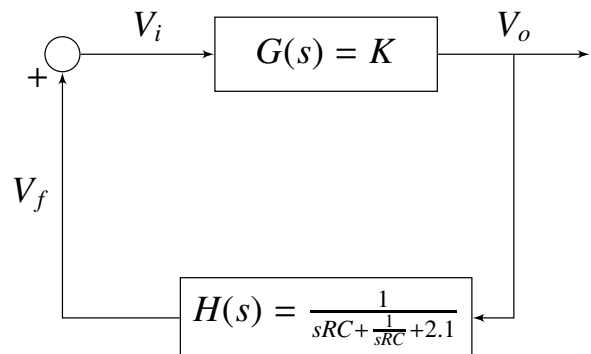


Fig. 4: Positive Feedback Circuit

Closed Loop gain

$$T = \frac{k(s^2 + s(\frac{2.1}{RC}) + (\frac{1}{RC})^2)}{s^2 + s(\frac{2.1-K}{CR}) + (\frac{1}{CR})^2} \quad (4.1)$$

5. Sketch the Normalised closed loop gain of T for various Q values

Solution:

The following is code for the plot

```
codes/ee18btech11030/ee18btech11030.py
```

From Figure:5 ,

- It is observed that maximally flat response is obtained when $Q = 0.71$

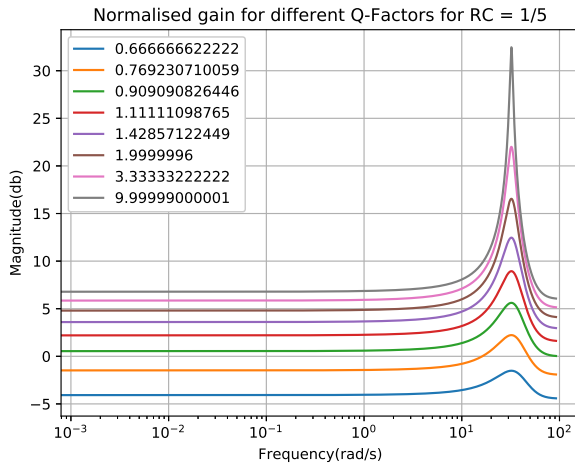


Fig. 5

- It will be seen that response of the feedback amplifier under consideration shows almost no peaking for $Q \leq 0.71$

6. Sketch a Pole-Zero Plot to Eq:4.1 for a varying K

Solution:

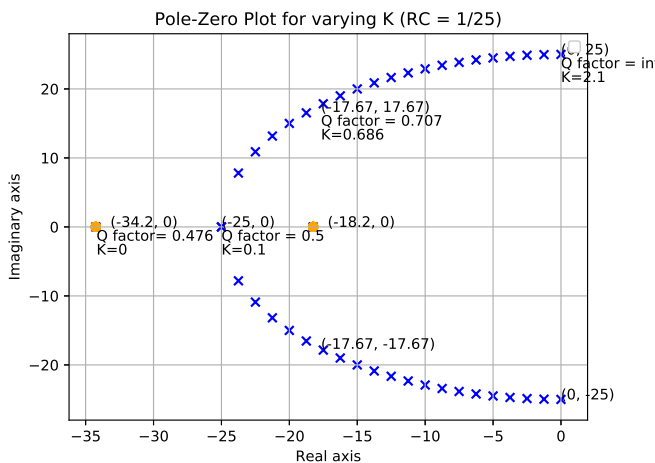


Fig. 6

The following is code for the plot

```
codes/ee18btech11030/ee18btech11030_1.py
```

From Figure : 6

- For $K = 0$, the poles have $Q = 0.476$ and therefore located on negative real axis.
- As K increases poles are brought closer together and eventually coincide at $K = 0.1$ and $Q = 0.5$

- Further increase in K results in poles becoming complex conjugate
- Maximally flat response is obtained when $Q = 0.707$, which results when $K = 0.686$. In this case poles are at 45° .
- Oscillating response is obtained when poles are completely imaginary when $Q = \infty$ which results when $K = 2.1$

Q-Factor	K	Requirement
0.5	0.1	Poles are coincident
0.707	0.686	Maximally flat response
∞	2.1	Oscillatory response

TABLE 6

7. Verify the response in time domain using parameters in Table : 8 for $K = 2.13$

Solution:

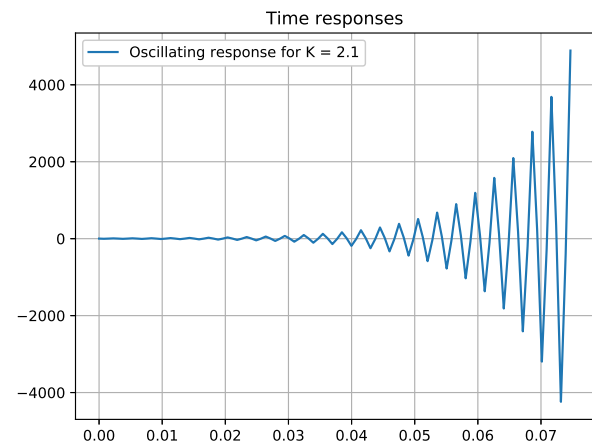


Fig. 7

The following is code for the plot

```
codes/ee18btech11030/ee18btech11030_2.py
```

8. Choose the appropriate values of R and C to simulate the circuit for $K = 2.1$

Solution:

9. Verify the response using the spice model

Solution:

- Figure 9 is the spice simulated output for $K = 2.1$ using Table 8 parameters.
- The following is the netlist for simulated circuit.

```
spice/ee18btech11030/ee18btech11030.net
```

Parameter	Value
R_1	$10k\Omega$
R_2	$11.3k\Omega$
R	$10k\Omega$
$10R$	$100k\Omega$
$C/10$	$1.6nF$
C	$16nF$

TABLE 8

- The following is code for generating output

```
spice/ee18btech11030/ee18btech11030.py
```

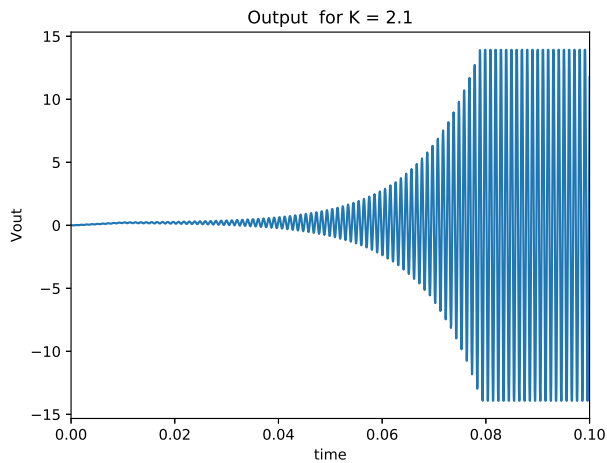


Fig. 9

10. Choose the appropriate values of R and C to simulate the circuit for $K = 0.1$ and $K = 0.686$

Solution: :

Parameter	$K = 0.686$	$K = 0.1$
R_1	$4.14k\Omega$	$1k\Omega$
R_2	$6.86k\Omega$	$9k\Omega$
R	$10k\Omega$	$10k\Omega$
$10R$	$100k\Omega$	$100k\Omega$
$C/10$	$1.6nF$	$1.6nF$
C	$16nF$	$16nF$

TABLE 10

11. Verify the response using the spice model

Solution:

- Figure 11 is the spice simulated output for $K = 0.686$ using Table 10 parameters.

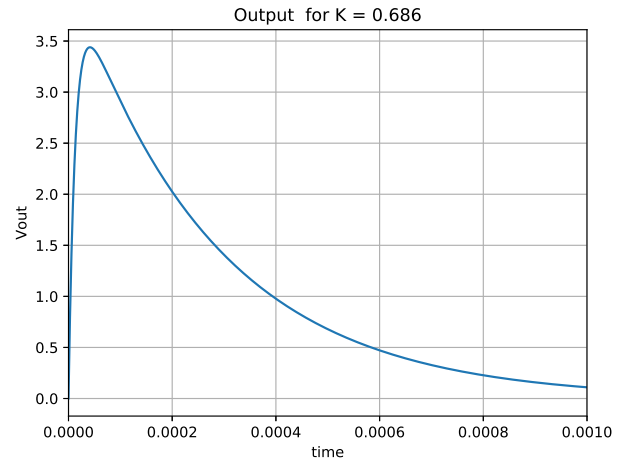


Fig. 11

- The following is the netlist for simulated circuit.

```
spice/ee18btech11030/ee18btech11030_1.
net
```

- The following is code for generating output

```
spice/ee18btech11030/ee18btech11030_1.
py
```

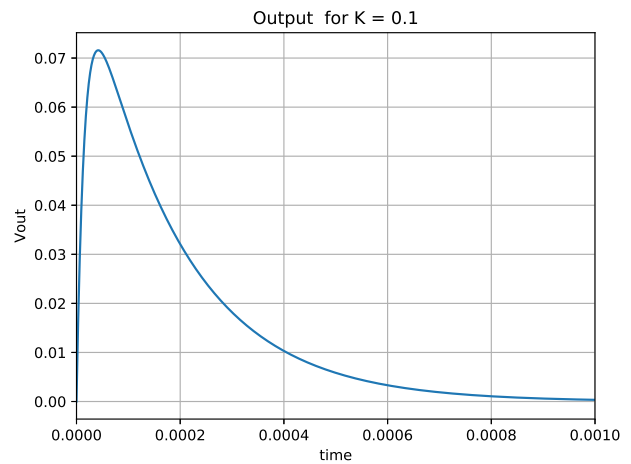


Fig. 11

- Figure 11 is the spice simulated output for $K = 0.1$ using Table 10 parameters.
- The following is the netlist for simulated circuit.

```
spice/ee18btech11030/ee18btech11030_2.
net
```

- The following is code for generating output

```
spice/ee18btech11030/ee18btech11030_2.  
py
```