

Control Systems

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CONTENTS

Abstract—This manual is an introduction to control systems based on GATE problems. Links to sample Python codes are available in the text.

Download python codes using

svn co <https://github.com/gadepall/school/trunk/control/codes>

1 FEEDBACK CIRCUITS

1.1. The CE BJT amplifier in Fig. 1.1 employs shunt–shunt feedback: Feedback resistor R_F senses the output Voltage V_o and provides a feedback current to the base node. ($R_f = 56k\Omega$, $R_C = 5.6k\Omega$, $R_S = 10k\Omega$)

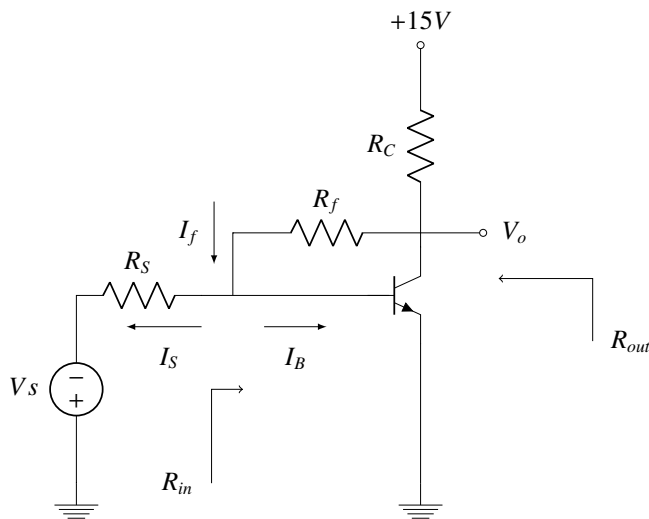


Fig. 1.1

1.2. If V_s has a zero dc component, find the dc collector current of the BJT. Assume the transistor $H = 100$.

Solution: Since, $V_E = 0$ and $V_S = V_{BE}$

$$I_S = \frac{V_{BE}}{R_S} = \frac{0.7}{10 * 10^3} \quad (1.2.1)$$

$$\Rightarrow I_S = 0.07mA \quad (1.2.2)$$

Applying KCL at feedback resistor output

$$-V_o + V_{BE} + I_f R_f = 0$$

$$(Since, I_f = I_B + I_S)$$

$$V_o = V_{BE} + (I_B + I_S) R_f$$

$$= 0.7 + (I_B + 0.07 * 10^{-3})(56 * 10^3)$$

$$\Rightarrow V_o = (56 * 10^3) I_B + 4.62 \quad (1.2.3)$$

Applying KCL at collector node

$$\frac{V_o - 15}{5.6 * 10^3} + I_C + I_f = 0$$

$$(Since, I_C = H I_B)$$

$$\frac{V_o - 15}{5.6 * 10^3} + H I_B + (I_B + I_S) = 0$$

$$\frac{V_o - 15}{5.6 * 10^3} + (100 + 1) I_B + (0.07 * 10^{-3}) = 0$$

$$\Rightarrow V_o = 14.608 - (565.5 * 10^3) I_B \quad (1.2.4)$$

Subtracting 1.2.3 from 1.2.4, we get,

$$I_B = 16.06\mu A \quad (1.2.5)$$

$$I_C = I_E = H I_B \quad (1.2.6)$$

$$Dc collector Current, I_C = 1.606mA \quad (1.2.7)$$

1.3. Find the small-signal equivalent circuit of the amplifier with the signal source represented by its Norton equivalent (as we usually do when the feedback connection at the input is shunt).

Solution: In fig 1.3

1.4. Find the G circuit and determine the value of G , R_i , and R_o .

Solution: G circuit in fig. 1.4

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Parameter	Description
R_{in}	Total Input Resistance
R_{out}	Total Output Resistance
r_o	Output resistance of NPN
R_f	Feedback resistance
R_l	Input resistance of G circuit
R_o	Output resistance of G circuit
R_{if}	Input resistance of Feedback
R_{of}	Output resistance of Feedback
R_s	Resistance of Current Source
R_L	Output Load Resistance
g_m	Trans conductance
I_C	Collector current
I_E	Emitter Current
I_B	Base Current

TABLE 1.2

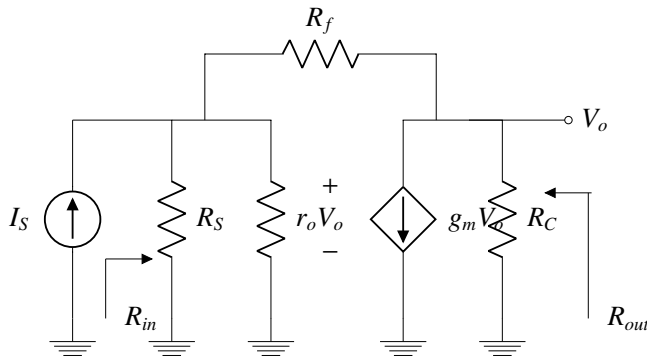


Fig. 1.3

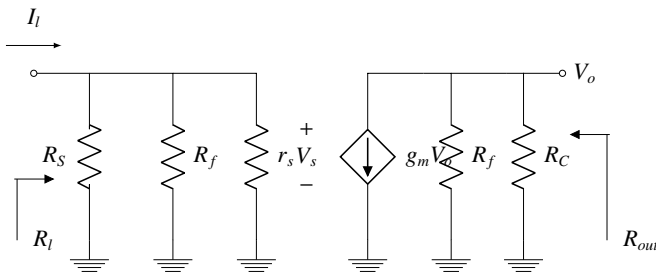


Fig. 1.4

$$g_m = \frac{I_C}{V_s} = \frac{1.606 \times 10^{-3}}{25 \times 10^{-3}} = 64 \text{ mA/V} \quad (1.4.1)$$

$$r_s = \frac{H}{g_m} = \frac{100}{64 \times 10^{-3}} = 1.56 \text{ k}\Omega \quad (1.4.2)$$

$$\text{Gain, } G = \frac{V_o}{I_l} \quad (1.4.3)$$

$$G = \frac{V_s}{I_l} \left(\frac{V_o}{V_s} \right) \quad (1.4.4)$$

$$V_o = -g_m V_s (R_f \parallel R_C) \quad (1.4.5)$$

$$V_s = I_l (R_s \parallel R_f \parallel r_s) \quad (1.4.6)$$

Substituting V_o V_s in 1.4.4,

$$G = -g_m (R_f \parallel R_C) (R_s \parallel R_f \parallel r_s) \quad (1.4.7)$$

$$G = -429 \text{ k}\Omega \quad (1.4.8)$$

Input Resistance

$$R_l = (R_s \parallel R_f \parallel r_s) = 1.31 \text{ k}\Omega \quad (1.4.9)$$

$$R_o = (R_f \parallel R_C) \quad (1.4.10)$$

$$\text{Output Resistance, } R_o = 5.09 \text{ k}\Omega \quad (1.4.11)$$

1.5. Find H and hence AH and 1+AH.

Solution:

$$H = \frac{I_f}{V_o} = -\frac{1}{R_f} \quad (1.5.1)$$

$$\Rightarrow H = -17.85 \times 10^{-4} \quad (1.5.2)$$

$$GH = 7.662 \quad (1.5.3)$$

$$1 + GH = 8.66 \quad (1.5.4)$$

1.6. Find T, R_{if} and R_{of} and hence R_{in} and R_{out} .

Solution:

$$T = \frac{G}{1 + GH} \quad (1.6.1)$$

$$= -49.54 \text{ k}\Omega \quad (1.6.2)$$

$$R_{if} = \frac{R_l}{1 + GH} \quad (1.6.3)$$

$$= \frac{1.31 \times 10^3}{8.66} \quad (1.6.4)$$

$$= 151.27 \Omega \quad (1.6.5)$$

$$R_{of} = \frac{R_o}{1 + GH} \quad (1.6.6)$$

$$= \frac{5.09 \times 10^3}{8.66} \quad (1.6.7)$$

$$= 587.7 \Omega \quad (1.6.8)$$

$$R_{in} = \frac{1}{\frac{1}{R_{if}} - \frac{1}{R_s}} \quad (1.6.9)$$

$$= 153.2 \Omega \quad (1.6.10)$$

1.7. What voltage gain V_o/V_s is realized? How does this value compare to the ideal value obtained

if the loop gain is very large and thus the signal voltage at the base becomes almost zero (like what happens in an inverting op-amp circuit).

Solution:

$$\frac{V_o}{V_s} = \frac{V_o}{I_s R_s} \quad (1.7.1)$$

$$= \frac{T}{R_s} \quad (1.7.2)$$

$$\text{Since, } T = \frac{V_o}{I_s} \quad (1.7.3)$$

$$\frac{V_o}{V_s} = \frac{-49.54 * 10^3}{10 * 10^3} \quad (1.7.4)$$

$$= -4.95V/V \quad (1.7.5)$$

If the loop gain is very large, then the gain with feedback T is:

$$T = \frac{1}{H} \quad (1.7.6)$$

$$= \frac{1}{(-17.85 * 10^{-6})} \quad (1.7.7)$$

$$= -56k\Omega \quad (1.7.8)$$

\therefore the closed loop gain, $T = -R_f$

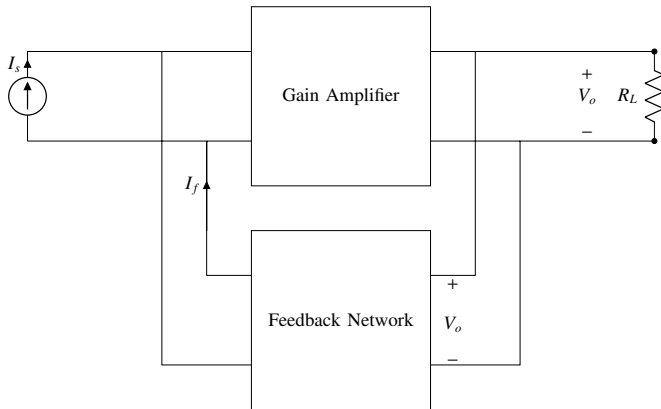


Fig. 1.7: Shunt-Shunt Amplifier Block Diagram

1.8. Verify your solution using spice

Solution: Doing operational Analysis on the Circuit 1.1

Table 1.2 is close to the numerical Calculation done above.

Parameter	Value
R_f	$56k\Omega$
R_s	$10k\Omega$
R_C	$5.6k\Omega$
I_s	$0.07mA$
I_B	$16.06\mu A$
I_C	$1.606mA$
I_E	$1.606mA$
g_m	$64mA/V$
r_s	$1.56k\Omega$
G	$-429k\Omega$
R_l	$1.31k\Omega$
R_o	$5.09k\Omega$
H	$-17.85 * 10^{-4}$
GH	7.662
$1 + GH$	8.66
T	$-49.54k\Omega$
R_{if}	$151.27k\Omega$
R_{of}	$587.7k\Omega$
R_{in}	153.2Ω

TABLE 1.7

Parameter	Value
I_C	$2.1mA$
I_E	$2.1mA$
I_B	$2.1\mu A$
I_s	$0.07mA$

TABLE 1.8