

Oscillator

Venkata Tejaswini Anangani*

For the circuit shown in Fig. 1.1, find the loop gain $L(s) = G(s)H(s)$, $L(j\omega)$, the frequency for zero loop phase, and R_2/R_1 for oscillation.

1. Draw the equivalent control system representation for the circuit in Fig. 1.1 as well as the small signal model.

Solution: See Figs. 1.2, 1.3 and 1.4. Oscillators do not include input signal.

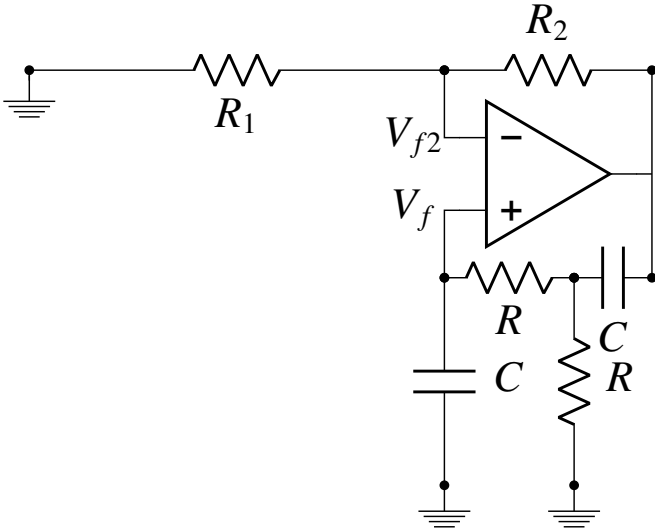


Fig. 1.1

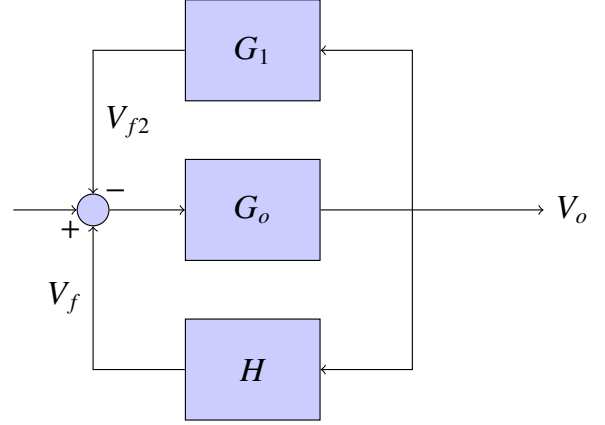


Fig. 1.2: Block diagram

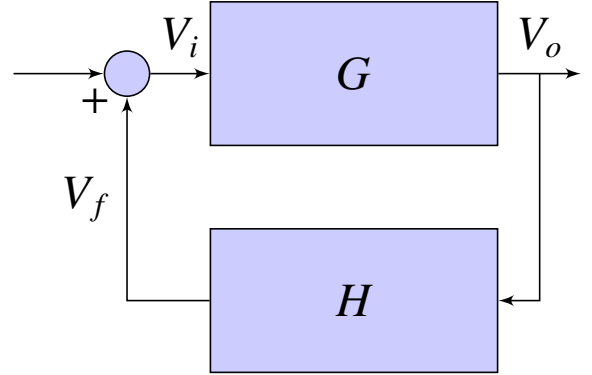


Fig. 1.3: Simplified equivalent block diagram

2. Draw the block diagram and circuit diagram for H .

Solution: See Figs. 2.5 and 2.6.

3. Find H .

Solution: In Fig. 2.6, let I_o be the current flowing from V_o . Then

$$I_o = \frac{V_o}{\frac{1}{sC} + R \parallel \left(R + \frac{1}{sC}\right)} \quad (3.1)$$

Using current division,

$$V_f = I_o \frac{R}{R + \left(R + \frac{1}{sC}\right)} \times \frac{1}{sC} \quad (3.2)$$

From (3.1) and (3.2),

$$\frac{V_f}{V_o} = \frac{R}{R + \left(R + \frac{1}{sC}\right)} \times \frac{1}{sC} \times \frac{1}{\frac{1}{sC} + R \parallel \left(R + \frac{1}{sC}\right)} \quad (3.3)$$

$$\Rightarrow H = \frac{1}{\left(3 + sRC + \frac{1}{sRC}\right)} \quad (3.4)$$

after simplification.

4. Find R_{11} and R_{22} from Fig. 2.6.

*The author is with the Department of Electrical Engineering, Indian Institute of Technology, Hyderabad 502285 India. All content in this manual is released under GNU GPL. Free and open source.

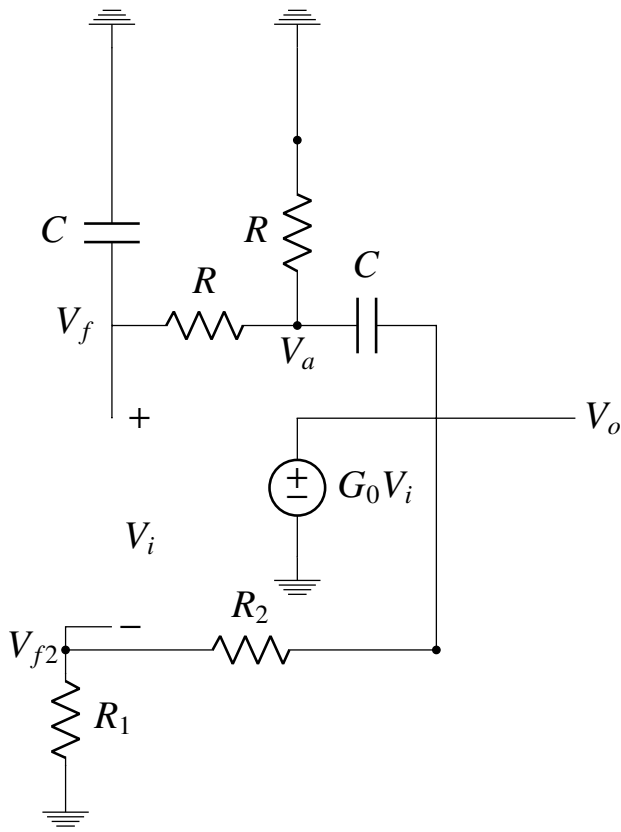


Fig. 1.4

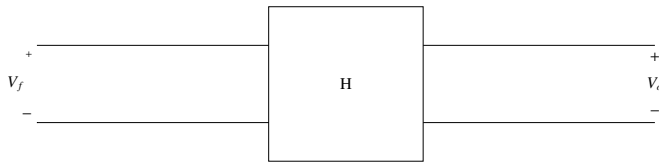


Fig. 2.5: Feedback block diagram

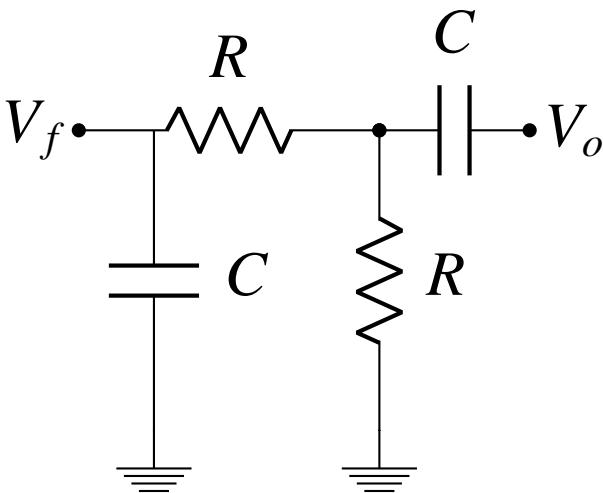


Fig. 2.6: Feedback circuit

Solution: Shorting V_o to ground,

$$R_{11} = \frac{1}{sC} \parallel \left(R + R \parallel \frac{1}{sC} \right) \quad (4.1)$$

Shorting V_f to ground,

$$R_{22} = \frac{1}{sC} + \frac{R}{2} \quad (4.2)$$

5. Draw the block diagram and circuit diagram for G .

Solution: See Figs. 5.1 and 5.2.

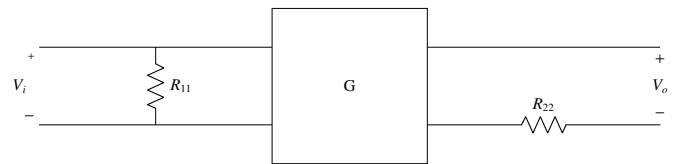


Fig. 5.1: Open loop block diagram

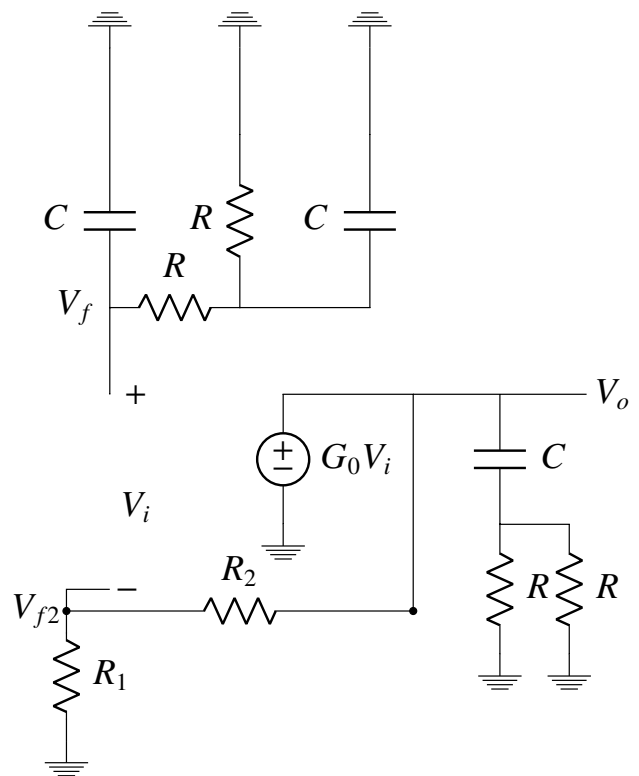


Fig. 5.2: Open loop circuit diagram

6. Find G .

Solution: From Fig.5.2

$$V_{f2} = \left(\frac{R_1}{R_1 + R_2} \right) V_o \quad (6.1)$$

$$G_1 = \frac{V_{f2}}{V_o} \quad (6.2)$$

$$\Rightarrow G_1 = \frac{R_1}{R_1 + R_2} \quad (6.3)$$

From Fig. 1.2, G_1 is the negative feedback factor and G_o is the gain of the op-amp. Therefore, equivalent G is given by

$$G = \frac{G_o}{1 + G_o G_1} \quad (6.4)$$

$$G = \frac{1}{\frac{1}{G_o} + G_1} \quad (6.5)$$

We assumed $G_o \rightarrow \infty$.

$$\Rightarrow G = \frac{1}{G_1} \quad (6.6)$$

From equation (6.3).

$$\Rightarrow G = \frac{R_1 + R_2}{R_1} = 1 + \frac{R_2}{R_1} \quad (6.7)$$

Hence verified with equation (8.5).

7. Find the feedback factor H .

Solution: The small signal model is shown in Fig. 1.4 Applying KCL at node V_f

$$\frac{V_f - 0}{\frac{1}{sC}} + \frac{V_f - V_a}{R} = 0 \quad (7.1)$$

$$V_f \left(sC + \frac{1}{R} \right) = \frac{V_a}{R} \quad (7.2)$$

$$V_a = V_f (sRC + 1) \quad (7.3)$$

Applying KCL at node V_a

$$\frac{V_a - V_f}{R} + \frac{V_a - 0}{R} + \frac{V_a - V_o}{\frac{1}{sC}} = 0 \quad (7.4)$$

$$V_a \left(\frac{2}{R} + sC \right) = \frac{V_f}{R} + V_o sC \quad (7.5)$$

Substitute V_a value from equation (7.3)

$$V_f (sRC + 1) \left(\frac{2}{R} + sC \right) = \frac{V_f}{R} + V_o sC \quad (7.6)$$

$$V_f \left(3 + sRC + \frac{1}{sRC} \right) = V_o \quad (7.7)$$

The feedback factor H is given by

$$H = \frac{V_f}{V_o} \quad (7.8)$$

$$\Rightarrow H = \frac{1}{\left(3 + sRC + \frac{1}{sRC} \right)} \quad (7.9)$$

8. Find the open loop gain G .

Solution: Let the closed loop gain, open-loop gain of op-amp connected in non-inverting configuration be T_0 and G_0 respectively. From Table ??

$$T_0 = \frac{G_0 (R_1 + R_2)}{(R_1 + R_2) + G_0 R_1} \quad (8.1)$$

$$T_0 = \frac{(R_1 + R_2)}{(R_1 + R_2)/G_0 + R_1} \quad (8.2)$$

Assuming $G_o \rightarrow \infty$

$$T_0 = 1 + \frac{R_2}{R_1} \quad (8.3)$$

The open loop gain of the circuit shown in Fig. 1.1 is equal to the closed loop gain of an op-amp connected in non-inverting configuration.

$$G = T_0 \quad (8.4)$$

$$\Rightarrow G = 1 + \frac{R_2}{R_1} \quad (8.5)$$

9. Find the loop gain $L(s)$.

Solution: The transfer function of the equivalent positive feedback circuit in Fig. 1.3 is

$$T = \frac{G}{1 - GH} \quad (9.1)$$

Therefore, loop gain is given by

$$L = GH \quad (9.2)$$

From equations (8.5) and (7.9)

$$L(s) = \left(1 + \frac{R_2}{R_1} \right) \left(\frac{1}{3 + sRC + \frac{1}{sRC}} \right) \quad (9.3)$$

$$\Rightarrow L(s) = \left(\frac{1 + \frac{R_2}{R_1}}{3 + sRC + \frac{1}{sRC}} \right) \quad (9.4)$$

10. Find the loop gain in terms of $j\omega$.

Solution: Substitute $s = j\omega$ in equation (9.4)

$$L(j\omega) = \left(\frac{1 + \frac{R_2}{R_1}}{3 + j\omega RC + \frac{1}{j\omega RC}} \right) \quad (10.1)$$

$$\Rightarrow L(j\omega) = \left(\frac{1 + \frac{R_2}{R_1}}{3 + j\left(\omega RC - \frac{1}{\omega RC}\right)} \right) \quad (10.2)$$

11. Find the frequency for zero loop phase.

Solution: The frequency at which loop phase will be zero (i.e. loop gain will be a real number). To obtain the required frequency, equate the imaginary part of the loop gain $L(j\omega)$ to zero.

$$j\left(\omega RC - \frac{1}{\omega RC}\right) = 0 \quad (11.1)$$

$$\omega^2 = \frac{1}{(RC)^2} \quad (11.2)$$

$$\Rightarrow \omega = \frac{1}{RC} \quad (11.3)$$

12. Find R_2/R_1 for oscillation.

Solution: For oscillations to start,

- the imaginary part of the loop gain should become zero.
- the loop gain must be at least equal to unity.

From equation (10.2)

$$\left(\frac{1 + \frac{R_2}{R_1}}{3 + j(0)} \right) \geq 1 \quad (12.1)$$

$$1 + \frac{R_2}{R_1} \geq 3 \quad (12.2)$$

$$\Rightarrow \frac{R_2}{R_1} \geq 2 \quad (12.3)$$

13. Find the amplitude and frequency for some arbitrary R,C values given in Table 13.

Solution: From equation (8.5)

$$G = 1 + \frac{R_2}{R_1} = 3 \quad (13.1)$$

From equation (7.9)

$$H = \frac{1}{3 + 0.25s + \frac{1}{0.25s}} \quad (13.2)$$

Parameter	Value
R	250Ω
C	$1mF$
R_2	$2k\Omega$
R_1	$1k\Omega$

TABLE 13

From equation (9.1)

$$T = \frac{3(0.0625s^2 + 0.75s + 1)}{0.0625s^2 + 1} \quad (13.3)$$

The following code plots the oscillating response of the system.

codes/ee18btech11047/ee18btech11047.py

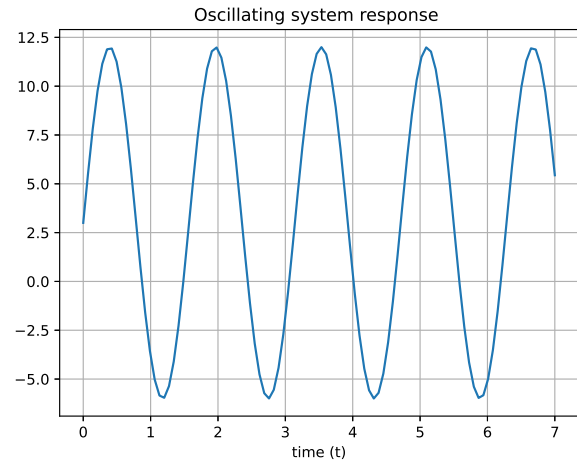


Fig. 13

Amplitude: From Fig. 13 $V_{\text{peak-peak}}$ is

$$V_{p-p} = 11.929 - (-5.957) = 17.886 \quad (13.4)$$

$$V_{\text{max}} = \frac{V_{p-p}}{2} = 8.943 \quad (13.5)$$

Frequency: From equation (11.3)

$$\omega = \frac{1}{RC} = 4 \text{ rad/sec} \quad (13.6)$$

$$f = \frac{\omega}{2\pi} = 0.636 \text{ Hz} \quad (13.7)$$

14. Verify the amplitude and frequency using spice simulation.

Solution: The following readme file provides

necessary instructions to simulate the circuit in spice.

```
codes/ee18btech11047/spice/README
```

The following netlist simulates the given circuit.

```
codes/ee18btech11047/spice/ee18btech11047.
net
```

The following code plots the output from the oscillator spice simulation which is shown in Fig. 14.1.

```
codes/ee18btech11047/spice/
ee18btech11047_spice.py
```

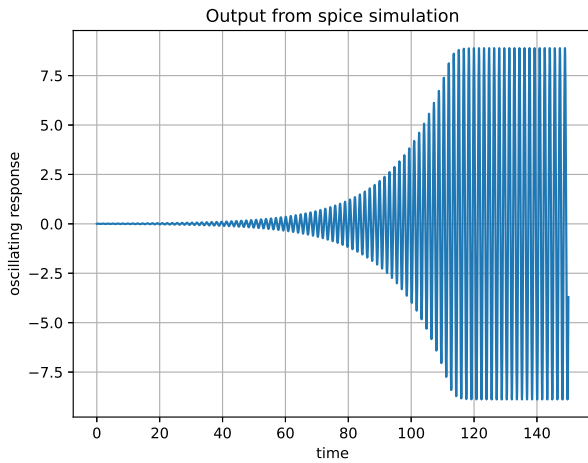


Fig. 14.1

The following code plots a part of the spice output from which we can observe a clear sinusoidal output shown in Fig. 14.2.

```
codes/ee18btech11047/spice/
ee18btech11047_spice2.py
```

Amplitude: From Fig. 14.2 $V(\text{peak-peak})$ is

$$V_{p-p} = 8.89 - (-8.89) = 17.78 \quad (14.1)$$

$$V_{max} = \frac{V_{p-p}}{2} = 8.89 \quad (14.2)$$

Frequency: From Fig. 14.2 time period is calculated by any two end points of one cycle,

$$T = 120.344 - (-118.734) = 1.61 \text{ sec} \quad (14.3)$$

$$f = \frac{1}{T} = 0.621 \text{ Hz} \quad (14.4)$$

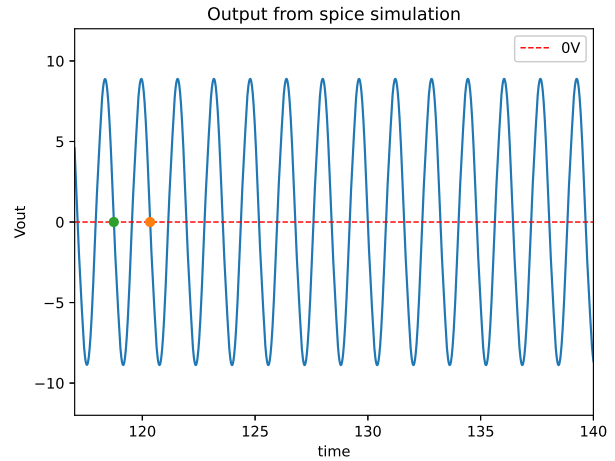


Fig. 14.2

Hence, the amplitude and frequency are verified through the spice simulation.