Control Systems

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2

CONTENTS

1 Feedback Voltage Amplifier: Series-Shunt 1

2 Feedback Current Amplifier: Shunt-Series

Abstract—This manual is an introduction to control systems in feedback circuits. Links to sample Python codes are available in the text.

Download python codes using

svn co https://github.com/gadepall/school/trunk/control/feedback/codes

- 1 FEEDBACK VOLTAGE AMPLIFIER: SERIES-SHUNT
- 1.1. Fig. 1.1.1 shows a non-inverting op-amp configuration with parameters described in Table 1.1. Draw the equivalent control system.

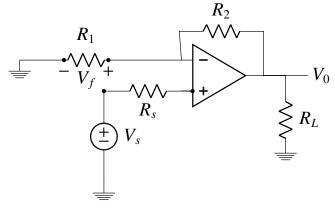


Fig. 1.1.1

Solution: See Fig. 1.1.2

1.2. Draw the small signal model for Fig. 1.1.1. **Solution:** The equivalent circuit of the amplifier is in Fig. 1.2

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Parameter	Value
input resistance	∞
output resistance	0
Input voltage	V_s
Output Voltage	V_o
Feeding resistance	R_1
Feedback resistance	R_2
Source resistance	R_s
load resistance	R_L

1

TABLE 1.1

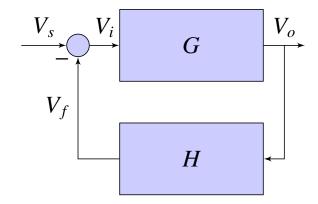


Fig. 1.1.2

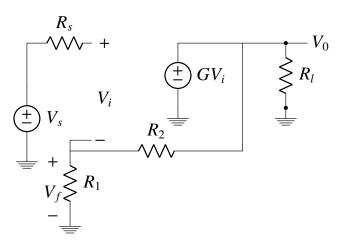


Fig. 1.2

1.3. Assuming that the operational amplifier has

infinite input resistance and zero output resistance, find the *feedback factor H*.

Solution: From Fig. 1.2,

$$V_0 = GV_i \tag{1.3.1}$$

$$V_i = V_s - V_f \tag{1.3.2}$$

$$V_f = \frac{R_1}{R_1 + R_2} V_o \tag{1.3.3}$$

assuming that the current through R_s is very small. Thus,

$$H = \frac{V_f}{V_o} = \frac{R_1}{R_1 + R_2} \tag{1.3.4}$$

1.4. Obtain the closed loop gain *T* and summarize your results through a Table.

Solution: Table 1.4 provides a summary.

$$T = \frac{V_0}{V_i} = \frac{G}{1 + GH} \tag{1.4.1}$$

$$=\frac{G(R_1+R_2)}{(R_1+R_2)+GR_1}$$
 (1.4.2)

Parame- ters	Definition	For given circuit
Open loop gain	G	G
Feedback factor	Н	$\frac{R_1}{R_1 + R_2}$
Loop gain	GH	$G^{\frac{R_1}{R_1+R_2}}$
Amount of feedback	1+GH	$1 + \frac{GR_1}{R_1 + R_2}$
Closed loop gain	<u>G</u> 1+ <i>GH</i>	$\frac{G(R_1 + R_2)}{R_1 + R_2 + GR_1}$

TABLE 1.4

1.5. Find the condition under which closed loop gain T is almost entirely determined by the feedback network.

Solution: If

$$GH \gg 1, \tag{1.5.1}$$

$$T \approx \frac{1}{H} = 1 + \frac{R_2}{R_1}$$
 (1.5.2)

1.6. If

$$G = 10^4 \tag{1.6.1}$$

$$T = 10,$$
 (1.6.2)

find H.

Solution: From Table 1.4

$$T = \frac{G}{1 + GH} = 10 \tag{1.6.3}$$

$$\implies H = 0.0999 \tag{1.6.4}$$

1.7. Gain Desensitivity: If G decreases by 20%, what is the corresponding decrease in T? Comment.

Solution: From From Table 1.4, Given

$$T = \frac{G}{1 + GH} \tag{1.7.1}$$

$$\implies dT = \frac{dG}{(1 + GH)^2} \tag{1.7.2}$$

$$\implies \frac{dT}{T} = \frac{1}{1 + GH} \frac{dG}{G}$$
 (1.7.3)

From the information available so far,

$$dG = 20\%, G = 10^4, H = 0.0999 \implies \frac{dT}{T} = 0.025\%$$
(1.7.4)

using the following code.

codes/ee18btech11005/ee18btech11005.py

Thus the closed loop gain is almost invariant to a relatively large (20%) variation in the open loop gain G. This is known as gain desensitivity.

- 2 FEEDBACK CURRENT AMPLIFIER: SHUNT-SERIES
- 2.1. Draw the equivalent control system for the feedback current amplifier shown in 2.1.4

Solution: See Fig. 2.1.5.

2.2. For the feedback current amplifier shown in 2.1.4, draw the Small-Signal Model. Neglect the Early effect in Q_1 and Q_2 .

Solution: See Fig. 2.2.

While drawing a Small-Signal Model, we ground all constant voltage sources and open all constant current sources. All Small-Signal paramters are obtained from DC-Analysis of the circuit. Neglecting Early effect, in Small-Signal Analysis a N-MOSFET is modelled as a Current Source with value of current equal to $g_m v_{gs}$ flowing from Drain to Source. Whereas

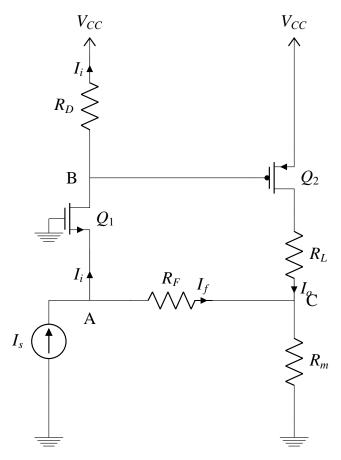
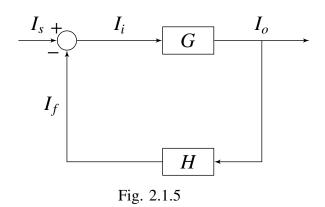


Fig. 2.1.4



a P-MOSFET is modelled as a Current Source with value of current equal to $g_m v_{sg}$ flowing from Source to Drain.

2.3. Write all the node/loop equations using KCL/KVL.

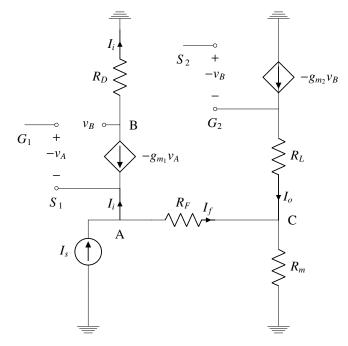


Fig. 2.2: Small Signal Model

Solution: From Figs. 2.1.4 and 2.2,

$$I_i = \frac{v_B}{R_D} \tag{2.3.1}$$

$$I_o = -g_{m2} v_B (2.3.2)$$

$$v_C + v_B = -I_o R_L (2.3.3)$$

$$v_C = \left(I_o - I_f\right) R_M \tag{2.3.4}$$

2.4. Find the Expression for the Open-Loop Gain G.

Solution: From (2.3.1) and (2.3.2),

$$G = \frac{I_o}{I_i} = -g_{m2}R_D \tag{2.4.1}$$

2.5. Find the Expression of the Feedback Factor *H*. **Solution:**

$$H = \frac{I_f}{I_o},$$
 (2.5.1)

From (2.3.3) and (2.3.4),

$$(I_o + I_f)R_M + v_B = -I_o R_L$$
 (2.5.2)

$$\implies (I_o + I_f)R_M + I_iR_D = -I_oR_L \qquad (2.5.3)$$

from (2.3.1)

$$\implies (1+H)R_M + \frac{R_D}{G} = -R_L \qquad (2.5.4)$$

dividing by I_o

$$\implies H = -1 - \frac{1}{R_M} \left(R_L - \frac{1}{g_{m2}} \right)$$
 (2.5.5)

upon substituting from (2.4.1) and simplifying. I_o is fed to a current divider formed by R_M and R_F . R_F is a Large Resistance compared to Input resistance of Amplifier and so most of the current flows through it leaving a small current as input to Amplifier. Hence the voltage at point 'A' is very small and is considered, $v_A \simeq 0$. So R_F and R_M are parallel and Voltage Drop across them is same.

$$(I_o + I_f)R_M \simeq -I_f R_F$$
 (2.5.6)

$$\frac{I_f}{I_o} \simeq -\frac{R_M}{R_F + R_M} \tag{2.5.7}$$

So, the Feedback Factor,

$$H \equiv \frac{I_f}{I_o} \simeq -\frac{R_M}{R_F + R_M} \tag{2.5.8}$$

2.6. Find the Expression for the Closed-Loop Gain $T = \frac{I_o}{I_c}$.

Solution: From (??) and (2.5.8),

$$T = \frac{I_o}{I_s} = \frac{G}{1 + GH} \tag{2.6.1}$$

$$= -\frac{g_{m_2}R_D}{1 + g_{m_2}R_D/\left(1 + \frac{R_F}{R_M}\right)}$$
 (2.6.2)

$$= -\frac{g_{m_2}R_D}{1 + g_{m_2}R_D / \left(1 + \frac{R_F}{R_M}\right)}$$
 (2.6.2)

$$\implies T = -\frac{g_{m_2}R_D}{1 + g_{m_2}R_D / \left(1 + \frac{R_F}{R_M}\right)}$$
 (2.6.3)