

A  
Seminar Report  
On  
**PID Algorithm for Perfectly Grid Solving Robot**

*Submitted in partial fulfillment for the award of the degree of*  
**Bachelor of Technology**  
In  
**Computer Engineering**



(Session 2017-2018)

**Guide Name -**  
**Mr. Manish Sharma**  
**Name(s) of Seminar Coordinators:**  
**Mrs. Shalini Puri**

**Submitted by-**  
**Harshit Chhipa**  
**PCE/CE/14/042**  
**IV, A**

**DEPARTMENT OF COMPUTER ENGINEERING**  
**POORNIMA COLLEGE OF ENGINEERING, JAIPUR**  
**RAJASTHAN TECHNICAL UNIVERSITY, KOTA**  
**March, 2018**

## DECLARATION

I hereby declare that the work which is being presented in the seminar report entitled “**PID Algorithm for Perfectly Grid Solving Robot**” in partial fulfillment for the award of the Degree of Bachelor of Technology in Computer Engineering affiliated to **Rajasthan Technical University, Kota** and submitted to the Department of Computer Engineering of **Poornima College of Engineering, Jaipur**, is an authentic record of my own work carried out under the guidance of “**Mr. Manish Sharma**” during the session 2017 - 18.

I have not submitted the matter presented in this report anywhere for the award of any other degree of this or any other institute/university.

Signature of the Student with Name & Reg. No.:

Place: \_\_\_\_\_

Date: \_\_\_\_\_



# POORNIMA

## COLLEGE OF ENGINEERING

Department of Computer Engineering

---

### CERTIFICATE

This is to certify that the seminar report entitled “**PID Algorithm for Perfectly Grid Solving Robot**” has been submitted by “**Harshit Chhipa**” (41), student of IV year (VIII semester) Computer Engineering branch in the partial fulfillment of **Degree of Bachelor of Technology, Rajasthan Technical University, Kota** during the session 2017-18. The report has been found satisfactory and approved for submission.

**Guide/Supervisor:**

**HOD**

Signature :

Department of CE

Guide Name :

PCE

Designation :

Department : Computer Engineering

## ***Acknowledgement***

I wish to express my sincere thanks and gratitude to my guide Mr. Manish Sharma for his affectionate and undaunted guidance as well as for his morale boosting encouragement with worthy suggestions and support. We would also like to thank my Coordinator, Mrs. Shalini Puri (Asst. Prof.), for guiding and supporting me throughout my Seminar.

I will be failing in my duty, if we don't offer our gratitude towards our faculty members, Prof. Sunil Gupta who helped me time to time to understand my topic and various related areas.

I am also very grateful to our respected Campus Director Dr. Om Prakash Sharma & all those who helped me directly or indirectly in my endeavor.

Last but not least, acknowledgment will not be over without mentioning a word of thanks to all our friends & colleagues who helped us directly or indirectly in all the way through or project preparation.

## TABLE OF CONTENTS

<b>CHAPTER NO.</b>	<b>PARTICULARS</b>	<b>PAGE NO.</b>
	<b>Title Page</b>	i
	<b>Declaration</b>	ii
	<b>Certificate by the Department</b>	iii
	<b>Acknowledgment</b>	iv
	<b>Table of Contents</b>	v
	<b>List of Tables</b>	vi
	<b>List of Figures</b>	vii
	<b>List of Abbreviations</b>	viii
	<b>Abstract</b>	1
1	<b>Introduction</b> <b>1.1 Introduction</b> <b>1.2 History</b> <b>1.3 Overview</b>	2 3 4
2	<b>Literature Review</b>	5
3	<b>Your Proposed Work/Method/ Algorithmic Details</b>	9
5	<b>Conclusion</b>	26
<b>References</b>		27
<b>Your Paper (Sent or to be Sent)</b>		
<b>Acceptance Notification / Mail of Communication</b>		
<b>Plagiarism Report (contents must be at least 60% original)</b>		

### **LIST OF FIGURES**

<b>FIGURE NO.</b>	<b>TITLE</b>	<b>PAGE NO.</b>
1.	Linear classifier	17
2.	SVM graph	18
3.	Margin and VC dimension	18
4.	Maximal Margin in separable case	20
5.	Extension for non-separable case	22
6.	Multi Class classification	22
7.	Acceptance of Paper	31
8.	Plagiarism Report	32
9.	Plagiarism Report	32
10.	Plagiarism Report	33

### **PID Algorithm for Perfectly Grid Solving Robot**

---

#### **1. INTRODUCTION**

Line Tracking is one of the most useful and popular behaviour of mobile robots that has been acquiring increasing importance in the recent times. Line tracking has become the most convenient and reliable navigation technique by which autonomous mobile robots navigate in a controlled, usually indoor environment. The path of the robot is demarcated with a distinguishable line or track, which the robot uses to navigate. Commonly used examples of these lines include magnetic tape on a non-magnetic surface, a void or a cut on a flat surface, a reflective tape on a non-reflecting surface and vice-versa. In most of the situations, the robots are mounted with Hall-Effect sensors or Photo-interrupters, which are based on Infra-Red or visible light.

Traditionally, the approach to line tracking is usually considered in the digital domain where the outputs from the digital line sensors are given to a microcontroller which is programmed to run the motors of the robot in various speeds depending on where the robot is positioned with respect to the centre of the line [1] - [6]. The motor speed settings are selected by trial and error for various sensor readings. While this approach has the advantage of being very simple, it is usually not very convenient due to the following limitations.

1) Programming becomes increasing difficult when the speed has to be increased beyond 25 to 30 centimeters per second for a small sized indoor robot.

2) The addition of more sensors further increases programming complexity.

3) Re-tuning is needed when the line thickness has to be varied.

4) Program development time is very large and requires considerable experience in robotic programming to create reliable performance.

5) When digital sensors are used, the sensor state changes occur only when the deviation in robot position is greater than the gap between the sensors. The natural response

is to change the motor speed drastically, which results in the robot making the jerky movements. This is undesirable in many applications. All the above mentioned issues can be solved by using analog line sensors for detecting the line [7], [8] and by using automatic control techniques for manipulating the motor speeds [8-10].

## 1.2 HISTORY OF SVM

- **Origins**

Continuous control, before PID controllers were fully understood and implemented, has one of its origins in the [centrifugal governor](#) which uses rotating weights to control a process. This had been invented by [Christian Huygens](#) in the 17th century to regulate the gap between [millstones](#) in [windmills](#) depending on the speed of rotation, and thereby compensate for the variable speed of grain feed.<sup>[3][4]</sup>

With the invention of the high pressure stationary steam engine there was a need for automatic speed control, and [James Watt](#)'s self-designed "conical pendulum" governor, a set of revolving steel balls attached to a vertical spindle by link arms, came to be an industry standard. This was based on the mill stone gap control concept.<sup>[5]</sup>

However, rotating governor speed control was still variable under conditions of varying load, where the shortcoming of what is now known as proportional control alone was evident. The error between the desired speed and the actual speed would increase with increasing load. In the 19th century the theoretical basis for the operation of governors was first described by [James Clerk Maxwell](#) in 1868 in his now-famous paper *On Governors*. He explored the mathematical basis for control stability, and progressed a good way towards a solution, but made an appeal for mathematicians to examine the problem.<sup>[6][5]</sup> The problem was examined further by [Edward Routh](#) in 1874, [Charles Sturm](#) and in 1895, [Adolf Hurwitz](#), who all contributed to the establishment of control stability criteria.<sup>[5]</sup> In practice, speed governors were further refined, notably by American scientist [Willard Gibbs](#), who in 1872 theoretically analysed Watt's conical pendulum governor.

About this time the invention of the [Whitehead torpedo](#) posed a control problem which required accurate control of the running depth. Use of a



depth pressure sensor alone proved inadequate, and a pendulum which measured the fore and aft pitch of the torpedo was combined with depth measurement to become the [pendulum-and-hydrostat control](#). Pressure control only provided a proportional control, which if the control gain was too high, would become unstable and go into overshoot, with considerable instability of depth-holding. The pendulum added what is now known as derivative control, which damped the oscillations by detecting the torpedo dive/climb angle and thereby the rate of change of depth.<sup>[7]</sup> This development (named by Whitehead as "The Secret" to give no clue to its action) was around 1868.<sup>[8]</sup>

Another early example of a PID-type controller was developed by [Elmer Sperry](#) in 1911 for ship-steering, though his work was intuitive rather than mathematically-based.<sup>[9]</sup>

However, it was not until 1922 that a formal control law for what we now call PID or three-term control was first developed using theoretical analysis, by [Russian American](#) engineer [Nicolas Minorsky](#).<sup>[10]</sup> Minorsky was researching and designing automatic ship steering for the US Navy and based his analysis on observations of a [helmsman](#). He noted the helmsman steered the ship based not only on the current course error, but also on past error, as well as the current rate of change;<sup>[11]</sup> this was then given a mathematical treatment by Minorsky.<sup>[5]</sup> His goal was stability, not general control, which simplified the problem significantly. While proportional control provided stability against small disturbances, it was insufficient for dealing with a steady disturbance, notably a stiff gale (due to [steady-state error](#)), which required adding the integral term. Finally, the derivative term was added to improve stability and control.

Trials were carried out on the [USS New Mexico](#), with the controller controlling the [angular velocity](#) (not angle) of the rudder. PI control yielded sustained yaw (angular error) of  $\pm 2^\circ$ . Adding the D element yielded a yaw error of  $\pm 1/6^\circ$ , better than most helmsmen could achieve.

The Navy ultimately did not adopt the system, due to resistance by personnel. Similar work was carried out and published by several others in the 1930s.

### 1.3 OVERVIEW

There are several methods for tuning a PID loop. The most effective methods generally involve the development of some form of process model, then choosing P, I, and D based on the dynamic model parameters. Manual tuning methods can be relatively time consuming, particularly for systems with long loop times.

The choice of method will depend largely on whether or not the loop can be taken "offline" for tuning, and on the response time of the system. If the system can be taken offline, the best tuning method often involves subjecting the system to a step change in input, measuring the output as a function of time, and using this response to determine the control parameters.

Choosing a tuning method

Method	Advantages	Disadvantages
Manual tuning	No math required; online.	Requires experienced personnel. <sup>[citation needed]</sup>
Ziegler-Nichols <sup>[b]</sup>	Proven method; online.	Process upset, some trial-and-error, very aggressive tuning. <sup>[citation needed]</sup>
Tyreus Luyben	Proven method; online.	Process upset, some trial-and-error, very aggressive tuning. <sup>[citation needed]</sup>
Software tools	Consistent tuning; online or offline - can employ computer-automated control system design ( <i>CAutoD</i> ) techniques; may include valve and sensor analysis; allows simulation before downloading; can support non-steady-state (NSS) tuning.	Some cost or training involved. <sup>[20]</sup>
Cohen-Coon	Good process models.	Some math; offline; only good for first-order processes. <sup>[citation needed]</sup>
Åström-Hägglund	Can be used for auto tuning; amplitude is minimum so this method has lowest process upset	The process itself is inherently oscillatory. <sup>[citation needed]</sup>

- Manual Testing-:** If the system must remain online, one tuning method is to first set  $K_p$  and  $K_i$  values to zero. Increase the  $K_p$  until the output of the loop oscillates, then the  $K_p$  should be set to approximately half of that value for a "quarter amplitude decay" type response. Then increase  $K_i$  until any offset is corrected in sufficient time for the process. However, too much  $K_i$  will cause instability. Finally, increase  $K_d$ , if required, until the loop is acceptably quick to reach its reference after a load disturbance. However, too much  $K_d$  will cause excessive response and overshoot. A fast PID loop tuning usually overshoots slightly to reach the setpoint more quickly; however, some systems cannot accept overshoot, in which case an overdamped closed-loop system is required, which will require a  $K_p$  setting significantly less than half that of the  $K_p$  setting that was causing oscillation.

**Effects of increasing a parameter independently<sup>[21][22]</sup>**

Parameter	Rise time	Overshoot	Settling time	Steady-state error	Stability
$K_p$	Decrease	Increase	Small change	Decrease	Degrade
$K_i$	Decrease	Increase	Increase	Eliminate	Degrade
$K_d$	Minor change	Decrease	Decrease	No effect in theory	Improve if $K_d$ small

- **Ziegler–Nichols method :-** Another heuristic tuning method is formally known as the [Ziegler–Nichols method](#), introduced by [John G. Ziegler](#) and [Nathaniel B. Nichols](#) in the 1940s. As in the method above, the  $K_p$  and  $K_d$  gains are first set to zero. The proportional gain is increased until it reaches the ultimate gain,  $K_u$ , at which the output of the loop starts to oscillate.  $K_u$  and the oscillation period  $T_u$  are used to set the gains as follows:

**Ziegler–Nichols method**

Control Type	$K_p$	$K_i$	$K_d$
<b>P</b>	$0.50K_u$	—	—
<b>PI</b>	$0.45K_u$	$0.54K_u/T_u$	—
<b>PID</b>	$0.60K_u$	$1.2K_u/T_u$	$3K_uT_u/40$

These gains apply to the ideal, parallel form of the PID controller. When applied to the standard PID form, the integral and derivative time parameters  $T_i$  and  $T_d$  are only dependent on the oscillation period  $T_u$ . Please see the section "[Alternative nomenclature and PID forms](#)".

### **LITERATURE REVIEW**

---

- The context can be extracted from the information about user's geographical location, state of the device and the surrounding environment.
- Context can help find a proper time for an interruption to get the input from users as well as reduce the number of inputs.
- We are utilizing the GPS present in Smartphone to locate the user.
- Once the user sets the destination point, he is relieved from peeking into the mobile phone for direction.
- Finger Sleeve will take over the task of navigation and continue to operate till destination has been reached.
- Previous research by Thomas J. Pingel and Keith C. Clarke has indicated that wearable computer system has had a positive impact on navigation system.
- With the smart device development, technology has shrunk into a size of a coin, which helps computation to be almost invisible.
- Factors thought to be influencing outdoor navigation have been explored by Bruce Thomas et al., but the proposed navigation system is for a user who is traveling by foot.
- Our approach is an illustrative example of calm computing, where interaction with the device is almost invisible and device proactively decides on what driving decisions user may take in immediate future.
- Here retrieving context based on geographical location is a major factor. With the availability of 3G (3rd generation) as well as 4G (4th generation) communication technology data rates have been significantly more.
- B.H. Thomas and W. Piekarski have discussed glove based user interaction techniques, which are very specific to the outdoor Augmented Reality (AR)

- Application and practically not feasible to be used in day-to-day activities because of the bulky nature of the system.
- 

## CHAPTER 3

### PROPOSED WORK

---

PROPORTIONAL TERM:

- ⊙ The proportional term is given by:

$$P_{out} = K_p e(t)$$

Where

$P_{out}$ : Proportional term of output.

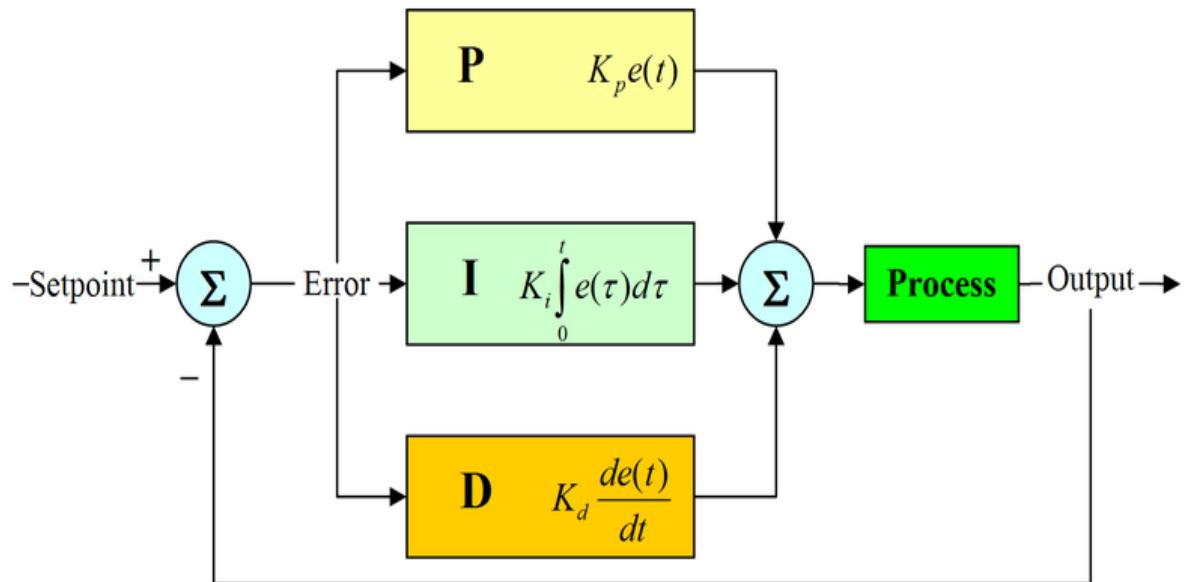
$K_p$ : Proportional gain, a tuning parameter.

PV: Process value (or process variable), the measured value.

$e$ : Error = SP – PV.

$t$ : Time or instantaneous time (the present)

- ⊙ The PID controller calculation involves three separate parameters, and is accordingly sometimes called three-term control: the proportional, the integral and derivative values, denoted  $P$ ,  $I$ , and  $D$ .
- ⊙  $P$  depends on the *present* error,  $I$  on the accumulation of *past* errors, and  $D$  is a prediction of *future* errors, based on current rate of change.
- ⊙ The weighted sum of these three actions is used to adjust the process variable.

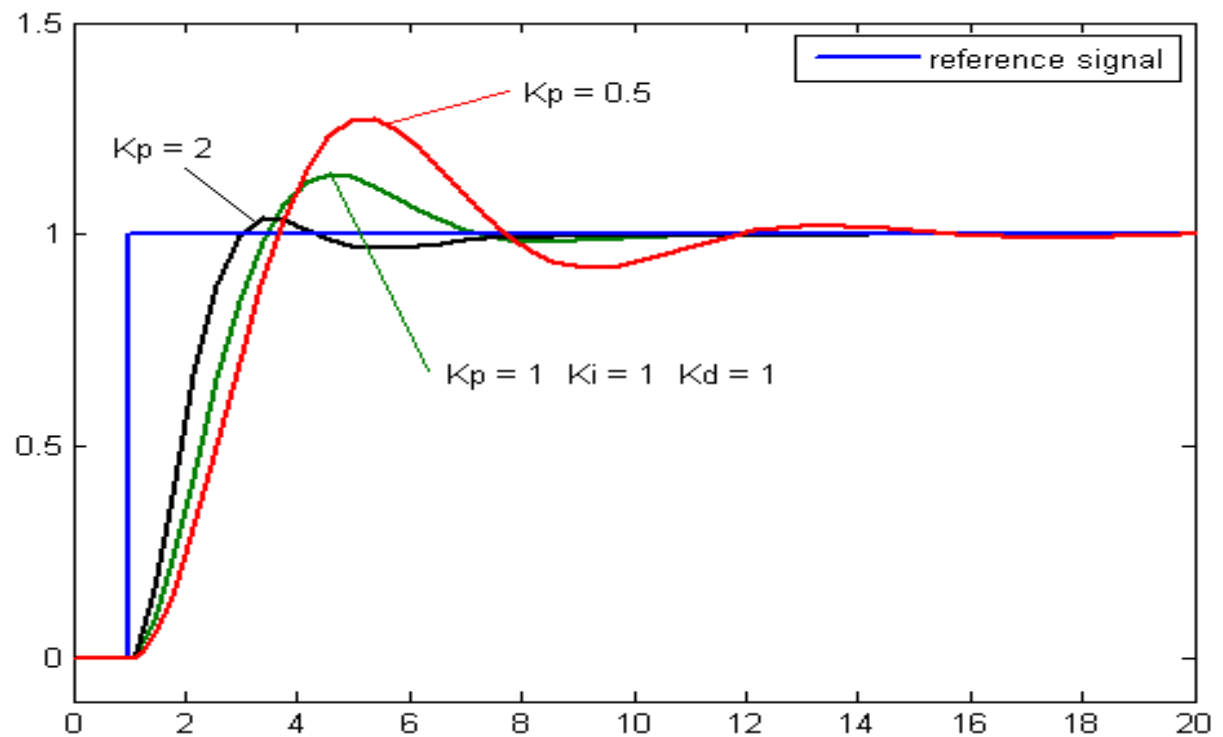


- By tuning the three constants in the PID controller algorithm, the controller can provide control action designed for specific process requirements.
- This section describes the parallel or non-interacting form of the PID controller.
- $MV(t) = P_{out} + I_{out} + D_{out}$

Where

$P_{out}$ ,  $I_{out}$ , and  $D_{out}$  are the contributions to the output from the PID controller from each of the three terms.

### Plot of PV vs. time, for three values of $K_p$ ( $K_i$ and $K_d$ held constant)



### **Integral term:**

- ⦿ The contribution from the integral term is proportional to both the magnitude of the error and the duration of the error.
- ⦿ Summing the instantaneous error over time (integrating the error) gives the accumulated offset that should have been corrected previously.
- ⦿ The accumulated error is then multiplied by the integral gain and added to the controller output.

The magnitude of the contribution of the integral term to the overall control action is determined by the integral gain,  $K_i$ .

The integral term is given by:



$$I_{\text{out}} = K_i \int_0^t e(\tau) d\tau$$

where

$I_{\text{out}}$ : Integral term of output

$K_i$ : Integral gain, a tuning parameter

$SP$ : Set point, the desired value

$PV$ : Process value (or process variable), the measured value

$e$ : Error =  $SP - PV$

$t$ : Time or instantaneous time (the present)

The integral term (when added to the proportional term) accelerates the movement of the process towards setpoint and eliminates the residual steady-state error that occurs with a proportional only controller.

However, since the integral term is responding to accumulated errors from the past, it can cause the present value to overshoot the setpoint value.

- **Derivative term: -**

The rate of change of the process error is calculated by determining the slope of the error over time and multiplying this rate of change by the derivative gain  $K_d$ .

The derivative term is given by:

$$D_{\text{out}} = K_d \frac{d}{dt} e(t)$$

where

$D_{\text{out}}$ : Derivative term of output

$K_d$ : Derivative gain, a tuning parameter

$SP$ : Setpoint, the desired value

$PV$ : Process value (or process variable), the measured value

$e$ : Error =  $SP - PV$

- ⦿ The derivative term slows the rate of change of the controller output and this effect is most noticeable close to the controller setpoint.
- ⦿ Hence, derivative control is used to reduce the magnitude of the overshoot produced by the integral component and improve the combined controller-process stability.

## **FINAL OUTPUT:-**

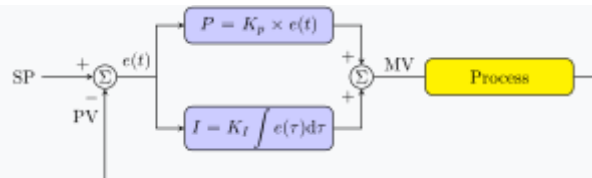
The proportional, integral, and derivative terms are summed to calculate the output of the PID controller. Defining  $u(t)$  as the controller output, the final form of the PID algorithm is:

$$u(t) = MV(t) = K_p e(t) + K_i \int_0^t e(\tau) d\tau + K_d \frac{d}{dt} e(t)$$

## Overshooting from known disturbances

For example, a PID loop is used to control the temperature of an electric resistance furnace where the system has stabilized. Now when the door is opened and something cold is put into the furnace the temperature drops below the setpoint. The integral function of the controller tends to compensate for error by introducing another error in the positive direction. This overshoot can be avoided by freezing of the integral function after the opening of the door for the time the control loop typically needs to reheat the furnace.

## PI controller[\[edit\]](#)



Basic block of a PI controller

A **PI Controller** (proportional-integral controller) is a special case of the PID controller in which the derivative (D) of the error is not used.

The controller output is given by

$$K_p e + K_i \int e dt$$

where  $e$  is the error or deviation of actual measured value (**PV**) from the setpoint (**SP**).

A PI controller can be modelled easily in software such as [Simulink](#) or [Xcos](#) using a "flow chart" box involving [Laplace](#) operators:

where

$K_p$  = proportional gain

$K_i$  = integral gain Setting a value for is often a trade off between decreasing overshoot and increasing settling time.

The lack of derivative action may make the system more steady in the steady state in the case of noisy data. This is because derivative action is more sensitive to higher-frequency terms in the inputs.

Without derivative action, a PI-controlled system is less responsive to real (non-noise) and relatively fast alterations in state and so the system will be slower to reach setpoint and slower to respond to perturbations than a well-tuned PID system may be.

## Basing proportional action on PV

Most commercial control systems offer the *option* of also basing the proportional action solely on the process variable. This means that only the integral action responds to changes in the setpoint. The modification to the algorithm does not affect the way the controller responds to process disturbances. Basing proportional action on PV eliminates the instant and possibly very large change in output caused by a sudden change to the setpoint. Depending on the process and tuning this may be beneficial to the response to a setpoint step.

$$MV(t) = K_p \left( -PV(t) + \frac{1}{T_i} \int_0^t e(\tau) d\tau - T_d \frac{d}{dt} PV(t) \right)$$

----

## Laplace form of the PID controller

Sometimes it is useful to write the PID regulator in [Laplace transform](#) form:

$$G(s) = K_p + \frac{K_i}{s} + K_d s = \frac{K_d s^2 + K_p s + K_i}{s}$$

Having the PID controller written in Laplace form and having the [transfer function](#) of the controlled system makes it easy to determine the closed-loop transfer function of the system.

## PID pole zero cancellation

The PID equation can be written in this form:

$$G(s) = K_d \frac{s^2 + \frac{K_p}{K_d} s + \frac{K_i}{K_d}}{s}$$

When this form is used it is easy to determine the closed loop transfer function.

$$H(s) = \frac{1}{s^2 + 2\zeta\omega_0 s + \omega_0^2}$$

$$\frac{K_p}{K_d} = 2\zeta\omega_0$$

Then

$$G(s)H(s) = \frac{K_d}{s}$$

While this appears to be very useful to remove unstable poles, it is in reality not the case. The closed loop transfer function from disturbance to output still contains the unstable poles.

### Series/interacting form [\[ edit \]](#)

Another representation of the PID controller is the series, or *interacting* form

$$G(s) = K_c \frac{(\tau_i s + 1)}{\tau_i s} (\tau_d s + 1)$$

where the parameters are related to the parameters of the standard form through

$$K_p = K_c \cdot \alpha, T_i = \tau_i \cdot \alpha, \text{ and}$$

$$T_d = \frac{\tau_d}{\alpha}$$

with

$$\alpha = 1 + \frac{\tau_d}{T_i}$$

This form essentially consists of a PD and PI controller in series, and it made early (analog) controllers easier to build. When the controllers later became digital, many kept using the interacting form.

#### Discrete implementation [\[edit\]](#)

The analysis for designing a digital implementation of a PID controller in a [microcontroller](#) (MCU) or [FPGA](#) device requires the standard form of the PID controller to be *discretized*.<sup>[34]</sup> Approximations for first-order derivatives are made by backward [finite differences](#). The integral term is discretised, with a sampling time  $\Delta t$ , as follows,

$$\int_0^{t_k} e(\tau) d\tau = \sum_{i=1}^k e(t_i) \Delta t$$

The derivative term is approximated as,

$$\frac{de(t_k)}{dt} = \frac{e(t_k) - e(t_{k-1})}{\Delta t}$$

Thus, a *velocity algorithm* for implementation of the discretized PID controller in a MCU is obtained by differentiating  $u(t)$ , using the numerical definitions of the first and second derivative and solving for  $u(t_k)$  and finally obtaining:

$$u(t_k) = u(t_{k-1}) + K_p \left[ \left( 1 + \frac{\Delta t}{T_i} + \frac{T_d}{\Delta t} \right) e(t_k) + \left( -1 - \frac{2T_d}{\Delta t} \right) e(t_{k-1}) + \frac{T_d}{\Delta t} e(t_{k-2}) \right]$$

## CHAPTER 4

### CONCLUSION

Here is a simple software loop that implements a PID algorithm:

```
previous_error = 0
integral = 0
loop:
    error = setpoint - measured_value
    integral = integral + error * dt
    derivative = (error - previous_error) / dt
    output = Kp * error + Ki * integral + Kd * derivative
    previous_error = error
    wait(dt)
    goto loop
```

In this example, two variables that will be maintained within the loop are initialized to zero, then the loop begins. The current *error* is calculated by subtracting the *measured\_value* (the process variable, or PV) from the current *setpoint* (SP). Then, *integral* and *derivative* values are calculated, and these and the *error* are combined with three preset gain terms – the proportional gain, the integral gain and the derivative gain – to derive an *output* value. In the real world, this is D-to-A converted and passed into the process under control as the manipulated variable (MV). The current error is stored elsewhere for re-use in the next differentiation, the program then waits until dt seconds have passed since start, and the loop begins again, reading in new values for the PV and the setpoint and calculating a new value for the error.

## REFERENCES

- [1] K.E. Årzèn. A simple event-based PID controller. In Proceedings of 14th World Congress of IFAC. Beijing, China, 1999. Monika Lagwal, Neha Bhardwaj, “Load balancing in Cloud Computing using Genetic Algorithm” in International Conference on Intelligent Computing and Control Systems ICICCS 2017.
- [2] K. J. Åström, T. Hägglund. Revisiting the Ziegler–Nichols step response method for PID control. *Journal of Process Control* 14:635–650, 2004. Priyanka Gonnade, Sonali Bodkhe ,” An Efficient load balancing using Genetic algorithm in Hierarchical structured distributed system” in *International Journal of Advanced Computer Research* (ISSN (print): 2249-7277 ISSN (online): 2277-7970) Volume-2 Number-4 Issue-6 December-2012.
- [3] K. J. Åström, T. Hägglund. *Advanced PID Control*. ISA Press, Research Triangle Park, USA, 2006. Parveen Kumar , Er.Mandeep Kaur, ”LOAD BALANCING IN CLOUD USING ACO AND GENETIC ALGORITHM” in *International Journal of Scientific Research Engineering & Technology (IJSRET)*, ISSN 2278 – 0882 Volume 4, Issue 7, July 2015.
- [4] R. S. Barbosa, J. A. Tenreiro Machado, I. M. Ferreira. Tuning of PID controllers based on Bode’s ideal transfer function. *Nonlinear Dynamics* 38:305–321, 2004.
- [5] M. Beschi, S. Dormido, J. Sanchez, A. Visioli. Characterization of symmetric send-ondelta PI controllers. *Journal of Process Control*, 2012.
- [6] T. L. Blevins. PID advances in industrial control. In *Proceedings IFAC Conference on Advances in PID Control*. Brescia, I, 2012.
- [7] J. Garrido, F. Vázquez, F. Morilla. Centralized multivariable control by simplified decoupling. *Journal of Process Control* 22:1044–1062, 2012.
- [8] E. Grassi, K. Tsakalis. PID controller tuning by frequency loop-shaping: Application to diffusion furnace temperature control. *IEEE Transactions on Control Systems Technologies* 8:842–847, 2000.
- [9] C. Grimholt, S. Skogestad. Optimal PI-control and verification of the SIMC tuning rule. In *Proceedings IFAC Conference on Advances in PID Control*. Brescia, I, 2012. [10] J. L. Guzman, K. J. Åström, S. Dormido, et al. Interactive learning modules for PID control. *IEEE Control Systems Magazine* 28:118–134, 2008

# PAPER

## SUPPORT VETOR MACHINE

Asst. Prof. Pratibha Shakya  
CS Department  
Poornima College of Engineering  
Sitapura, Jaipur, Rajasthan, India

Harshit Chhipa  
CS Department  
Poornima College of Engineering  
Sitapura, Jaipur, Rajasthan, India  
2014pcecsheemant@poornima.org

**1. ABSTRACT:** - Classification is one of the most important tasks for different application such as text categorization, tone recognition, image classification, micro-array gene expression, proteins structure predictions, data Classification etc. Most of the existing supervised classification methods are based on traditional statistics, which can provide ideal results when sample size is tending to infinity. However, only finite samples can be acquired in practice. In this paper, a novel learning method, Support Vector Machine (SVM), is applied on different data (Diabetes data, Heart Data, Satellite Data and Shuttle data) which have two or multi class. SVM, a powerful machine method developed from statistical learning and has made significant achievement in some field. Introduced in the early 90's, they led to an explosion of interest in machine learning. The foundations of SVM have been developed by Vapnik and are gaining popularity in field of machine learning due to many attractive features and promising

empirical performance. SVM method does not suffer the limitations of data dimensionality and limited samples [1] & [2].

In our experiment, the support vectors, which are critical for classification, are obtained by learning from the training samples. In this paper we have shown the comparative results using different kernel functions for all data samples.

**2. INTRODUCTION :-** The Support Vector Machine (SVM) was first proposed by Vapnik and has since pulled in a high level of enthusiasm for the machine learning research group. A few late investigations have revealed that the SVM (bolster vector machines) for the most part are fit for conveying higher execution regarding characterization exactness than the other information order calculations. Sims have been utilized in an extensive variety of true issues, for example, content

arrangement, written by hand digit acknowledgment, tone acknowledgment, picture order and question recognition, miniaturized scale cluster quality articulation information investigation, information grouping. It has been demonstrated that Sims is reliably better than other regulated learning techniques. Be that as it may, for some datasets, the execution of SVM is extremely delicate to how the cost parameter and bit parameters are set. Subsequently, the client regularly needs to lead broad cross approval keeping in mind the end goal to make sense of the ideal parameter setting. This procedure is regularly alluded to as model choice. One functional issue with demonstrate determination is that this procedure is exceptionally tedious. We have explored different avenues regarding various parameters related with the utilization of the SVM calculation that can affect the outcomes. These parameters incorporate decision of piece works, the standard deviation of the Gaussian part, relative weights related with slack factors to represent the non-uniform circulation of marked information, and the quantity of preparing illustrations.

For instance, we have taken four distinct applications informational collection, for example, diabetes information, heart information and satellite information which all have diverse highlights, classes, number of preparing information and diverse number of testing information.

This paper is sorted out as takes after. In next area, we present some related foundation including some essential ideas of SVM, portion work determination, and model choice

(parameters choice) of SVM. In Section 3, we detail all trials comes about. At long last, we have a few conclusions and highlight course in Section 4.

**3. OVERVIEW OF SVM :-** SVMs are set of related administered learning techniques utilized for grouping and relapse [2]. They have a place with a group of summed up straight characterization. A unique property of SVM is , SVM at the same time limit the experimental grouping mistake and augment the geometric edge. So SVM called Maximum Margin

Classifiers. SVM depends on the Structural hazard Minimization (SRM). SVM outline vector to a higher dimensional space where a maximal isolating hyperplane is developed. Two parallel hyperplanes are built on each side of the hyperplane that different the information. The isolating hyperplane is the hyperplane that augment the separation between the two parallel hyperplanes. A suspicion is made that the bigger the edge or separation between these parallel hyperplanes the better the speculation blunder of the classifier will be [2].

We consider information purposes of the shape

$\{(x_1, y_1), (x_2, y_2), (x_3, y_3), (x_4, y_4) \dots \dots (x_n, y_n)\}$ .

Where  $y_n = 1/-1$  , a consistent indicating the class to which that point  $x_n$  has a place.  $n$  = number of test. Every  $x_n$  is  $p$ -dimensional genuine vector. The scaling is critical to prepare for variable (traits) with bigger variance. We can see this Training information



, by methods for the separating (or seperating) hyperplane , which takes

$$w \cdot x + b = 0 \quad \text{----- (1)}$$

Where  $b$  is scalar and  $w$  is  $p$ -dimensional Vector. The vector  $w$  directs opposite toward the isolating hyperplane . Including the balance parameter  $b$  enables us to build the edge. Missing of  $b$ , the hyperplane is forced to go through the inception , limiting the arrangement. As we are fascinating in the greatest edge , we are intrigued SVM and the parallel hyperplanes. Parallel hyperplanes can be depicted by condition

$$w \cdot x + b = 1$$

$$w \cdot x + b = -1$$

In the event that the preparation information are directly distinct, we can choose these hyperplanes so that there are no focuses amongst them and afterward attempt to boost their separation. By geometry, We discover the separation between the hyperplane is  $2/\|w\|$  . So we need to limit  $\|w\|$  . To energize information focuses, we have to guarantee that for all  $I$  either

$$w \cdot x_i - b \geq 1 \text{ or } w \cdot x_i - b \leq -1$$

This can be composed as

$$y_i (w \cdot x_i - b) \geq 1, \quad 1 \leq I \leq n \quad \text{--- (2)}$$

Figure.1 Maximum edge hyperplanes for a SVM prepared with tests from two classes Samples along the hyperplanes are called Support Vectors (SVs). An isolating hyperplane with the biggest edge characterized by  $M = 2/\|w\|$  that is indicates bolster vectors

implies preparing information guides storage rooms toward it. Which fulfill?

$$y_j [w^T \cdot x_j + b] = 1, \quad I=1 \quad \text{----- (3)}$$

Ideal Canonical Hyperplane (OCH) is an accepted Hyperplane having a maximum margin. For every one of the information, OCH ought to fulfill the accompanying Constraint ;

$$y_i [w^T \cdot x_i + b] \geq 1; \quad I=1,2,\dots,1 \quad \text{----- (4)}$$

#### 4 KERNEL SELECTION OF SVM -

Preparing vectors  $x_i$  are mapped into a higher (might be interminable) dimensional space by the capacity  $\Phi$ . At that point SVM finds a direct isolating hyperplane with the maximal edge in this higher measurement space . $C > 0$  is the reformative nature parameter of the mistake term.

Besides,  $K(x_i, x_j) \equiv \Phi(x_i)^T \Phi(x_j)$  is known as the bit function[2]. There are numerous portion capacities in SVM, so how to choose a decent part so as to locate the ideal hyperplane, a double lagrangian (Ld) must be expanded as for nonnegative  $\alpha_i$  (I .e.  $\alpha_i$  must be in the nonnegative quadrant) and concerning the equity requirements as take after

$$\alpha_i \geq 0, \quad i = 1,2,\dots, \dots,1$$

$$\sum \alpha_i y_i = 0 \quad i=1$$

work is additionally an examination issue.However, for general purposes, there are some prevalent part works [2] and [3]:

- Linear part:  $K(x_i, x_j) = x_i^T x_j$ .
- Polynomial part:

$$K(x_i, x_j) = (\gamma x_i^T x_j + r)^d, \gamma > 0$$

- RBF kernel :

$$K(x_i, x_j) = \exp(-\gamma \|x_i - x_j\|^2), \gamma > 0$$

- Sigmoid kernel:

$$K(x_i, x_j) = \tanh(\gamma x_i^T x_j + r)$$

Here,  $\gamma$ ,  $r$  and  $d$  are part parameters. In these mainstream bit capacities, RBF is the fundamental portion work as a result of following reasons [2]:

1. The RBF kernel nonlinearly maps samples into a higher dimensional space unlike to linear kernel.
2. The RBF kernel has less hyperparameters than the polynomial kernel.

The RBF kernel has less numerical difficulties.

**5. MODEL SELECTION OF SVM:-** Model determination is additionally a critical issue in SVM. As of late, SVM have demonstrated great execution in information grouping. Its prosperity relies upon the tuning of a few parameters which influence the speculation blunder. We regularly call this parameter tuning system as the model choice. In the event that you utilize the straight SVM, you just need to tune the cost parameter  $C$ . Tragically, straight SVM are regularly connected to directly divisible issues.

Numerous issues are non-straightly distinguishable. For instance, Satellite information and Shuttle information are not directly distinguishable. Along these lines, we regularly apply nonlinear portion to tackle order issues, so we have to choose the cost parameter ( $C$ ) and piece parameters ( $\gamma$ ,  $d$ ) [4] and [5].

We normally utilize the matrix look strategy in cross approval to choose the best parameter set. At that point apply this parameter set to the preparation dataset and after that get the classifier. From that point forward, utilize the classifier to arrange the testing dataset to get the speculation precision.

## 6. SVM ADVANTAGES:-

- SVM's are great when we have no clue on the information.
- Works well with even unstructured and semi organized information like content, Images and trees.
- The portion trap is genuine quality of SVM. With a fitting piece work, we can take care of any perplexing issue.
- Unlike in neural systems, SVM isn't illuminated for neighborhood optima.
- It scales moderately well to high dimensional information.
- SVM models have speculation by and by, the danger of overfitting is less in SVM.

## 7. SVM DISADVANTAGES

- Choosing a "decent" bit work isn't simple.
- Long preparing time for vast datasets.
- Difficult to comprehend and decipher the last model, variable weights and individual effect.
- Since the last model isn't so natural to see, we can't do little alignments to the model thus it's hard to join our business rationale.

## 8. SVM APPLICATION

- Protein Structure Prediction
- Intrusion Detection
- Handwriting Recognition
- Detecting Steganography in advanced pictures
- Breast Cancer Diagnosis.

## 9. CONCLUSION

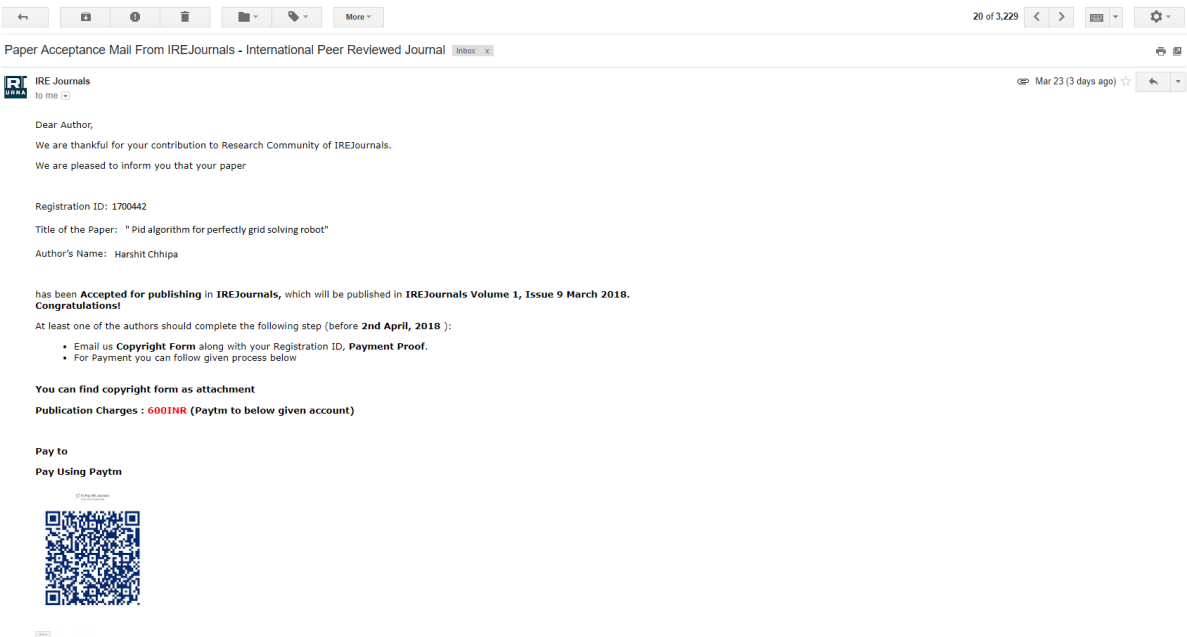
- Many programming apparatuses are accessible for SVM usage.

- SVMs are better than average for content characterization.
- SVMs are great thinking optimistically direct separator. The portion trap makes SVMs non-direct learning calculations.
- Choosing a fitting piece is the key for good SVM and picking the correct bit work isn't simple.
- We should be understanding while at the same time constructing SVMs on extensive datasets. They set aside a great deal of time for preparing.

## REFERENCES

- [1] Boser, B. E., I. Guyon, and V. Vapnik (1992). A training algorithm for optimal margin classifiers . In Proceedings of the Fifth Annual Workshop on Computational Learning Theory, pages. 144 -152. ACM
- [2] V. Vapnik. The Nature of Statistical Learning Theory. NY: Springer-Verlag. 1995.

## ACCEPTANCE



Plagiarism Scan Report	
Summary	
Report Generated Date	23 Mar, 2018
Plagiarism Status	100% Unique
Total Words	546
Total Characters	3176
Any Ignore Url Used	

## Content Checked For Plagiarism:

Vectors (SVs). An isolating hyperplane with the biggest edge characterized by  $M = 2/|w|$  that indicates bolster vectors implies preparing information guides storage rooms toward it. Which fulfill?

$$y_i(wT \cdot x_i + b) = 1, i = 1 \dots l \quad (3)$$

Ideal Canonical Hyperplane (OCH) is an accepted Hyperplane having a maximum margin.

For every one of the information, OCH ought to fulfill the accompanying Constraint ;

$$y_i(wT \cdot x_i + b) \geq 1, i = 1, 2, \dots, l \quad (4)$$

## 4 KERNEL SELECTION OF SVM -

Preparing vectors  $x_i$  are mapped into a higher (might be interminable) dimensional space by the capacity  $\Phi$ . At that point SVM finds a direct isolating hyperplane with the maximal edge in this higher measurement space.  $C > 0$  is the reformative nature parameter of the mistake term.

Besides,  $K(x_i, x_j) = \Phi(x_i)^T \Phi(x_j)$  is known as the bit function[2]. There are numerous portion capacities in SVM, so how to choose a decent part so as to locate the ideal hyperplane, a double lagrangian (Ld) must be expanded as for nonnegative  $\alpha_i$  (i.e.  $\alpha_i$  must be in the nonnegative quadrant) and concerning the equity requirements as take after

$$\alpha_i \geq 0, i = 1, 2, \dots, l$$

$$\sum \alpha_i y_i = 0$$

work is additionally an examination issue. However, for general purposes, there are some prevalent part works [2] and [3]:

- Linear part:  $K(x_i, x_j) = x_i^T x_j$ .
- Polynomial part:  $K(x_i, x_j) = (\gamma x_i^T x_j + r)^d, \gamma > 0$
- RBF kernel:  $K(x_i, x_j) = \exp(-\gamma \|x_i - x_j\|^2), \gamma > 0$
- Sigmoid kernel:  $K(x_i, x_j) = \tanh(\gamma x_i^T x_j + r)$

Plagiarism Scan Report	
Summary	
Report Generated Date	23 Mar, 2018
Plagiarism Status	71% Unique
Total Words	945
Total Characters	6003
Any Ignore Url Used	

## Content Checked For Plagiarism:

1. ABSTRACT:- Classification is one of the most important tasks for different application such as text categorization, tone recognition, image classification, micro-array gene expression, proteins structure predictions, data Classification etc. Most of the existing supervised classification methods are based on traditional statistics, which can provide ideal results when sample size is tending to infinity. However, only finite samples can be acquired in practice. In this paper, a novel learning method, Support Vector Machine (SVM), is applied on different data (Diabetes data, Heart Data, Satellite Data and Shuttle data) which have two or multi class. SVM, a powerful machine method developed from statistical learning and has made significant achievement in some field. Introduced in the early 90's, they led to an explosion of interest in machine learning. The foundations of SVM have been developed by Vapnik and are gaining popularity in field of machine learning due to many attractive features and promising empirical performance. SVM method does not suffer the limitations of data dimensionality and limited samples [1] & [2]. In our experiment, the support vectors, which are critical for classification, are obtained by learning from the training samples. In this paper we have shown the comparative results using different kernel functions for all data samples.

2. INTRODUCTION :- The Support Vector Machine (SVM) was first proposed by Vapnik and has since pulled in a high level of enthusiasm for the machine learning research group. A few late investigations have revealed that the SVM (bolster vector machines) for the most part are fit for conveying higher execution regarding characterization exactness than the other information order calculations. Sims have been utilized in an extensive variety of true issues, for example, content arrangement, written by hand digit acknowledgment, tone acknowledgment, picture order and question recognition, miniaturized scale cluster quality articulation information investigation, information grouping. It has been demonstrated that Sims is reliably better than other regulated learning techniques. Be that as it may, for some datasets, the execution of SVM is extremely delicate to how the cost parameter and bit parameters are set. Subsequently, the client regularly needs to lead broad cross approval keenness in mind the end goal to make sense of the ideal parameter settings. This procedure

Plagiarism Scan Report	
Summary	
Report Generated Date	23 Mar, 2018
Plagiarism Status	80% Unique
Total Words	266
Total Characters	1680
Any Ignore Uri Used	

#### Content Checked For Plagiarism:

- The portion trap is genuine quality of SVM. With a fitting piece work, we can take care of any perplexing issue.
- Unlike in neural systems, SVM isn't illuminated for neighborhood optima.
- It scales moderately well to high dimensional information.
- SVM models have speculation by and by, the danger of overfitting is less in SVM.

#### 7. SVM DISADVANTAGES

- Choosing a "decent" bit work isn't simple.
- Long preparing time for vast datasets.
- Difficult to comprehend and decipher the last model, variable weights and individual effect.
- Since the last model isn't so natural to see, we can't do little alignments to the model thus it's hard to join our business rationale.

#### 8. SVM APPLICATION

- Protein Structure Prediction
- Intrusion Detection
- Handwriting Recognition
- Detecting Steganography in advanced pictures
- Breast Cancer Diagnosis.

#### 9. CONCLUSION

- Many programming apparatuses are accessible for SVM usage.
- SVMs are better than average for content characterization.
- SVMs are great thinking optimistically direct separator. The portion trap makes SVMs non-