

# Digital Communication IC Design

## Homework #3

(Due on 2024/10/30)

Please upload your codes and reports (with simulation results and explanations) to the NTUCool.

Total points: 130 points

1. Consider one channel power delay profile listed in the following table.

Table 1: Channel power delay profile

Tap	Power	Delay ( $\tau_i$ )
0	0 dB	0 ns
1	-3 dB	10 ns
2	-4 dB	40 ns
3	-8 dB	80 ns
4	-15 dB	100 ns

- (a) Please list the power ratio of 5 taps in linear scale. (5%)
- (b) Please calculate the mean excess delay. (5%)
- (c) Please calculate the RMS excess delay ( $\tau_{RMS}$ ). (5%)
2. Please generate complex Gaussian random variable ( $Z=X+jY$ ) with zero mean and unit variance (i.e. the variance of both  $X$  and  $Y$  is 0.5) to obtain  $\gamma_i$ ,  $i = 0, 1, \dots, 4$ .
- (a) Assume that both  $h_i(t)$  and  $\tau_i(t)$  are time-invariant so that  $h_i(t) = h_i$  and  $\tau_i(t) = \tau_i$ .  $g_i = \alpha_i * \gamma_i$ , where  $\alpha_i$  is the path **magnitude** gain (not power) from Question 1. Now normalize  $g_i$  so that  $h_i = K g_i$  and  $\sum_{i=0}^4 |h_i|^2 = 1$ , which can keep the same signal energy under the multipath fading channel effect. Please write down your  $h_i$ ,  $i = 0, 1, \dots, 4$ . (10%)
- (b) Given that the channel impulse response is written as  $\sum_{i=0}^4 h_i \delta(t - \tau_i)$ . Now, use Fourier transform to derive its frequency domain channel response. Write down the equation. (10%)
- (c) Plot the magnitude and phase of channel frequency response in the range of -100 MHz and 100 MHz. (10%)
- (d) If the coherence bandwidth is defined as  $B_c = 1/(5\tau_{RMS})$ , where  $\tau_{RMS}$  comes from Q1. Please calculate the coherence bandwidth. (5%) How many subcarriers should be allocated in a 100MHz bandwidth if we want to produce a flat-fading channel in each subcarrier of an OFDM systems? (5%)

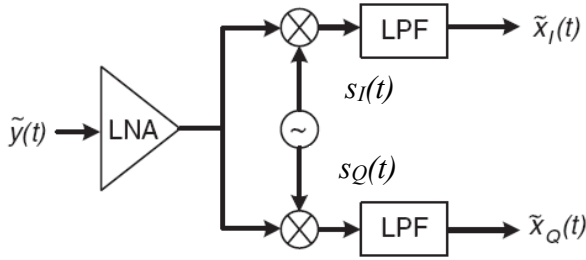
3. If the transmitted RF signal is denoted as

$$\tilde{y}(t) = \text{Re}\{(x_I(t) + jx_Q(t))(C(t) + jS(t))\}$$

Ideally,  $C(t) = \cos(\omega_c t)$  and  $S(t) = \sin(\omega_c t)$ . Assume at the **transmitter side**,

IQ imbalance exists, and thus  $C(t) = \left(1 + \frac{\varepsilon}{2}\right) \cos(\omega_c t + \frac{\varnothing}{2})$  and  $S(t) = \left(1 - \frac{\varepsilon}{2}\right) \sin(\omega_c t - \frac{\varnothing}{2})$ . Please demodulate the input signal  $\tilde{y}(t)$  and express  $\tilde{x}_I(t)$  and

$\tilde{x}_Q(t)$  in terms of  $x_I(t)$  and  $x_Q(t)$  as well as the IQ imbalance parameters with the assumption of perfect demodulation, i.e.  $s_I(t) = \cos(\omega_c t)$  and  $s_Q(t) = -\sin(\omega_c t)$ . (15%).



4. Given the OFDM systems. The whole bandwidth is 100MHz and is partitioned into 64 subcarriers. The sampling frequency ( $f_s$ ) is also 100MHz. Cyclic prefix (CP) of 16 samples is used in the guard interval.

- Please randomly generate BPSK data  $X_k$  for each subcarrier  $k$ ,  $k = 0, 1, \dots, 63$ .
- Obtain the time-domain signal  $x[n]$  for  $n = 0, 1, \dots, 63$ . Then, insert CP, which is indicated by  $x[n]$  for  $n = -16, -15, \dots, -1$ . Draw the real part and imaginary part of the baseband transmitted waveform (10%)
- Note that the discrete baseband equivalent channel in Q2(a) can be described

by  $h[n] = \sum_{i=0}^4 h_i \delta[n - n_{\tau_i}]$ , where  $n_{\tau_i} = \frac{\tau_i}{T_s}$  denote the normalized delay

from Table 1 and  $T_s$  is the sampling period. Assume the OFDM signal generated in Q4(b) passes through the channel. The received signal  $y[n]$  is expressed as  $y[n] = h[n] \otimes x[n]$ . Draw the real part and imaginary part of the baseband received signal  $y[n]$  for  $n = -16, -15, \dots, 63, 64, \dots$ . Note that you need to draw the entire output sequence of  $y[n]$  considering the length of channel impulse response  $h[n]$ . (10%)

- Now, remove CP and transform  $y[n]$ , for  $n = 0, 1, \dots, 63$ , into frequency

domain to obtain the received frequency domain data  $Y_k$ . Calculate  $H_k = \frac{Y_k}{X_k}$ .

Draw the magnitude and phase of  $H_k$ . (10%)

- (e) Compare Q4(d) and Q2(c). How to interpret the results given that one of their X-axes is expressed by Hertz and other is expressed by index? (10%)
- (f) Let's observe the results to see what will happen if the CP is not inserted. Now, generate  $z[n] = h[n] \otimes x[n]$  for  $n = 0, 1, \dots, 63$ . Transform  $z[n]$  into frequency domain to obtain the received frequency domain data  $Z_k$ . Calculate  $H'_k = \frac{Z_k}{X_k}$ . Draw the magnitude and phase of  $H'_k$ . (10%)
- (g) Compare Q4(d) and Q4(f). Please explain why  $H'_k$  can not acquire the proper channel frequency response. (10%)