## Deriving System Equations & the State-Space Model for Cessna 172

## State-Space Representation and Transfer Function Derivation

State Equation:  $x' = Ax + B\mu$ Output Equation:  $y = Cx + D\mu$ 

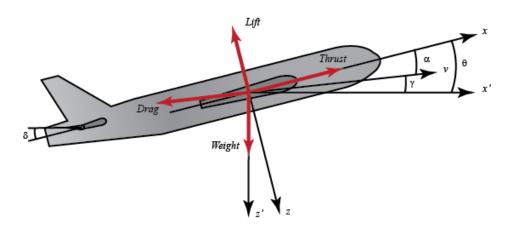


Figure 1: Basic coordinate axes and forces acting on an aircraft

Source: https://ctms.engin.umich.edu/CTMS/?example=AircraftPitch&section=SystemModeling

From the basic model shown in Figure 1, I was able to derive the following equations.

$$\dot{\alpha} = \mu \Omega \sigma \left[ -(C_L + C_D)\alpha + \frac{1}{\mu - C_L} q - (C_W \sin \lambda)\delta + C_L \right]$$

$$\dot{q} = \frac{\mu \Omega}{2I_{yy}} [[m - \eta(C_L + C_D)]\alpha + [C_M + \sigma C_m (1 - \mu C_L)]q + (\eta C_w \sin \lambda)\delta]$$

$$\dot{\theta} = \Omega q$$

Symbol	Description/Equation	Value	Units
$\alpha$	Angle of attack	N/A	rad
q	Pitch rate	N/A	rad/s
$\delta$	Elevator deflection angle	N/A	rad
$\mu$	$rac{ ho S ar c}{4m}$	0.00658	Constant
$\rho$	Density of air	1.204	$kg/m^3$
S	Platform area of the wing	16.165	$m^2$
$ar{c}$	Average chord length	1.472	m
m	Mass of the aircraft	1088.622	kg
$\Omega$	$\frac{2U}{\bar{c}}$	83.860	Constant
U	Equilibrium flight speed	62.733	m/s
$\sigma$	Normalization constant	0.04	Constant
$C_L$	Lift coefficient	1.60	Constant
$C_D$	Drag coefficient	0.0292	Constant
$C_M$	Pitch moment coefficient	-0.3	Constant
$C_W$	$\frac{2m}{\rho U^2 S}$	2.88	Constant
$\lambda$	Roll angle	0.05	rad
$\sigma$	$\frac{1}{1+\mu C_L}$	0.990	Constant
$I_{yy}$	$\frac{i_y}{m\bar{c}^2}$	0.773	Constant
$i_{yy}^{gg}$	Moment of inertia	1824.93	$kg \cdot m^2$
$\frac{\eta^{"}}{}$	$\mu\sigma C_M$	0.00195	Constant

Table 1: Cessna 172 model variables

After plugging in the values from Table 1 into the equations I was able to simplify the modeling equations.

$$\dot{\alpha} = -0.890\alpha - 0.343q - 1.57\delta$$
 
$$\dot{q} = -0.106\alpha - 0.117q - 0.002\delta$$
 
$$\dot{\theta} = 83.86q$$

Next, to find the transfer function I took the Laplace transform of the modeling equations seen above and performed basic algebra to rearrange them into  $P(s) = \frac{\Theta(s)}{\Delta(s)}$ .

$$sA(s) = -0.890A(s) - 0.343Q(s) - 1.57\Delta(s)$$
  
 $Q(s) = -0.106A(s) - 0.117Q(s) - 0.002\Delta(s)$   
 $\Theta(s) = 83.86Q(s)$ 

$$P(s) = \frac{\Theta(s)}{\Delta(s)} = \frac{-0.168s + 14.0886}{s^3 + 1.007s^2 + 0.0682s}$$

As the equations above are already in their state-variable form, they can easily be rewritten into the necessary matrices.

$$\begin{bmatrix} \dot{\alpha} \\ \dot{q} \\ \dot{\theta} \end{bmatrix} = \begin{bmatrix} -0.890 & -0.343 & 0 \\ -0.106 & -0.117 & 0 \\ 0 & 83.86 & 0 \end{bmatrix} \begin{bmatrix} \alpha \\ q \\ \theta \end{bmatrix} + \begin{bmatrix} -1.57 \\ -0.002 \\ 0 \end{bmatrix} [\delta]$$

$$y = \left[ egin{array}{ccc} 0 & 0 & 1 \end{array} 
ight] \left[ egin{array}{c} lpha \ q \ heta \end{array} 
ight]$$

## Closed-Loop System Response and PID Tuning

I then designed a pitch angle PID controller, considering actuator dynamics, and tuned it to a stable, responsive behavior.

$$Actuator(s) = \frac{1}{0.5s + 1}$$

$$P_{combined}(s) = P(s) \cdot Actuator(s)$$

By performing the PID tuning via adjusting the target bandwidth I was able to create a controller with stable closed-loop poles and a moderate rise time, settling time, and overshoot.

Table 2: Step Response Characteristics

Characteristic	Value	
Rise Time	1.4478 s	
Transient Time	14.5328  s	
Settling Time	$14.5328~\mathrm{s}$	
Minimum Value (Settling)	0.9197	
Maximum Value (Settling)	1.1966	
Overshoot	19.66~%	
Undershoot	0	
Peak Value	1.1966	
Peak Time	3.2599  s	

$$C(s) = K_p + \frac{K_i}{s} + K_d s$$

$$K_p = 0.00248$$
  
 $K_i = 1.91 \times 10^{-5}$   
 $K_d = 0.0805$ 

Table 3: Closed-Loop Poles

Closed-Loop Poles	Value
Pole 1	-20.0733
Pole 2	-0.4521 + 0.9819i
Pole 3	-0.4521 - 0.9819i
Pole 4	-0.0148 + 0.0032i
Pole 5	-0.0148 - 0.0032i

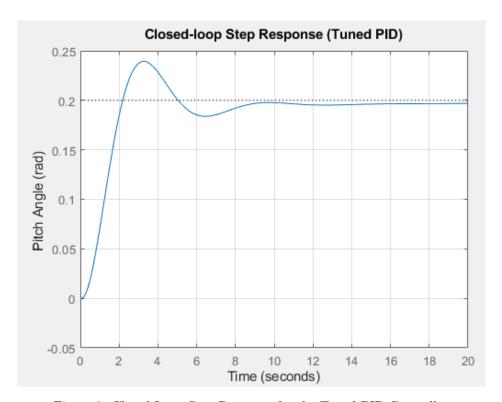


Figure 2: Closed-Loop Step Response for the Tuned PID Controller