

ESC201: INTRODUCTION TO ELECTRONICS

MODULE 2: ELEMENTS WITH MEMORY



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Capacitor and Inductor

$$v_C(t) = v_C(\infty) + (v_C(0^+) - v_C(\infty))e^{-\frac{t}{RC}}$$

Final
Voltage
(steady
state)

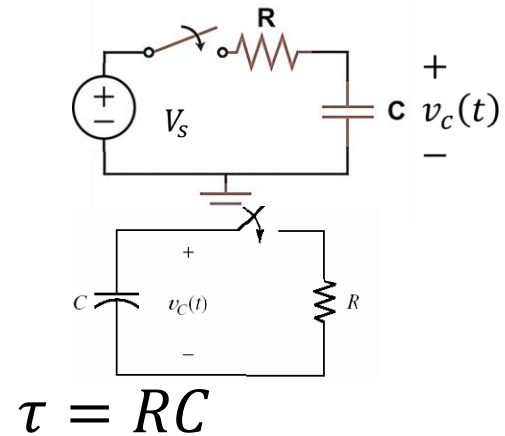
Initial
Voltage

Final
Voltage

Change in voltage

$$v_C(t) = V_S(1 - e^{-\frac{t}{\tau}})$$

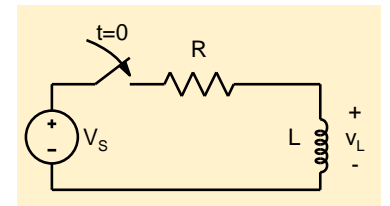
$$v_C(t) = V_i e^{-\frac{t}{\tau}}$$



$$i(t) = i(\infty) + (i(0^+) - i(\infty))e^{-\frac{R}{L}t}$$

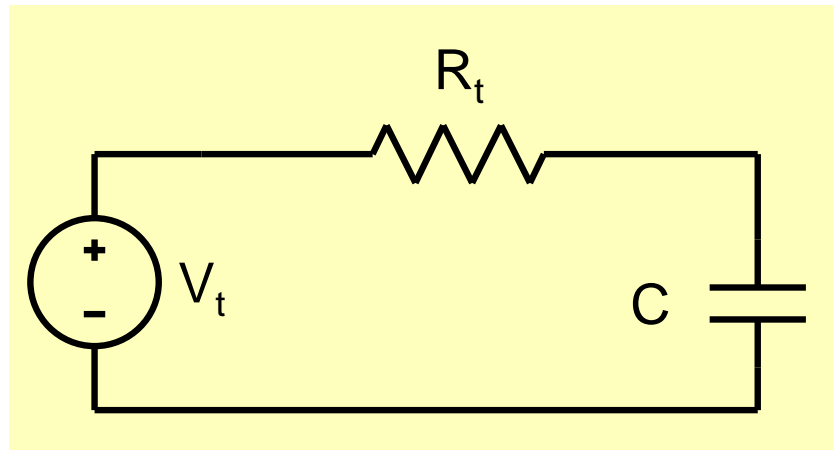
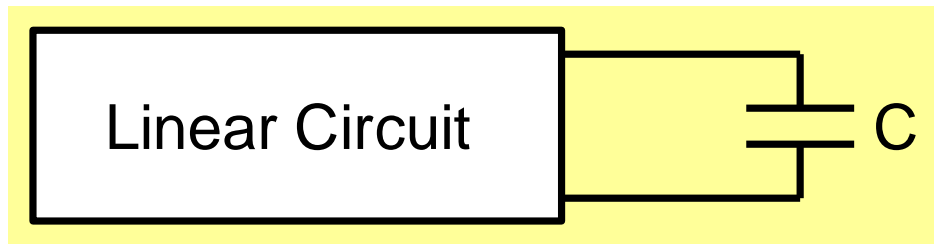
$$i(t) = \frac{V_S}{R} - \frac{V_S}{R}e^{-\frac{R}{L}t}$$

$$i(t) = i_0 e^{-\frac{R}{L}t}$$

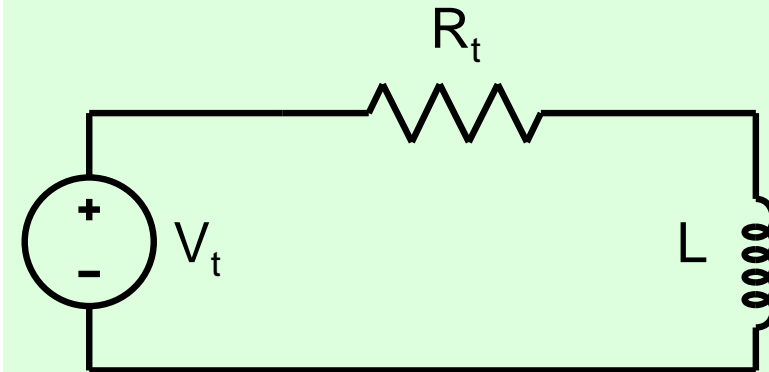
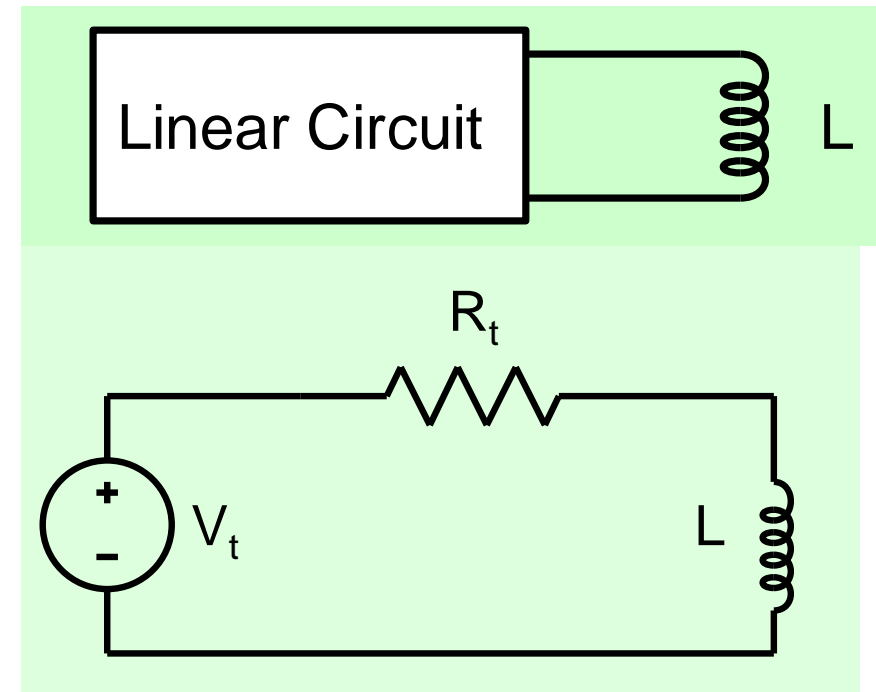


Circuit analysis with Inductor/Capacitor

Easy if the circuit contains a single L or C: Thevenin equivalent

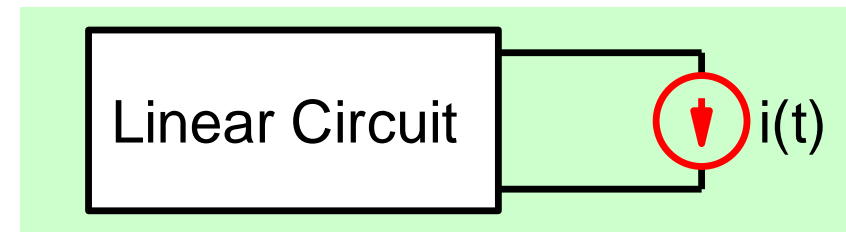
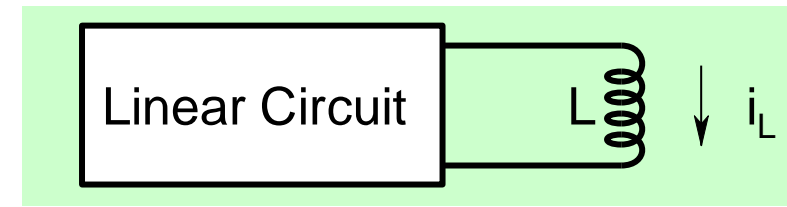
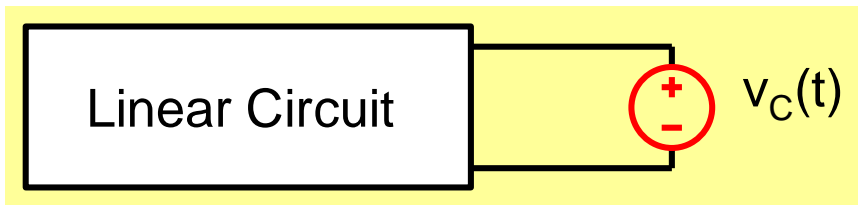
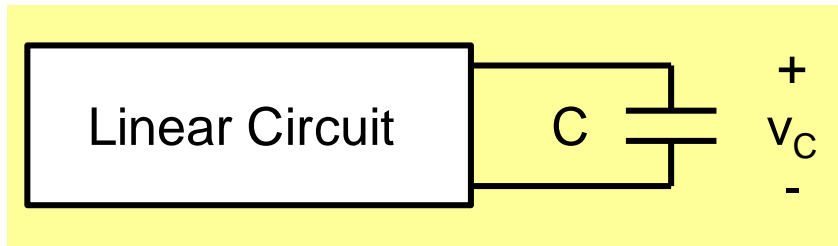


$$x(t) = x(\infty) + \{x(0^+) - x(\infty)\}e^{-\frac{t}{\tau}}$$



$$\tau = \frac{L}{R_{eq}} \text{ or } R_{eq}C$$

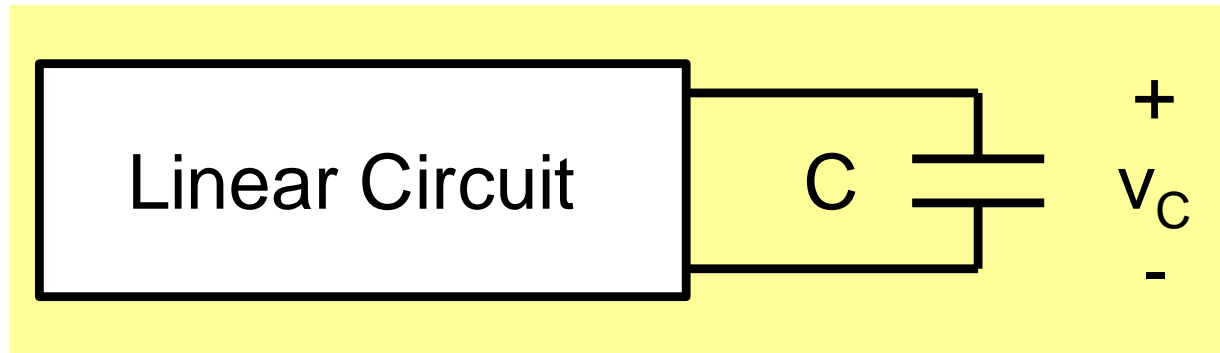
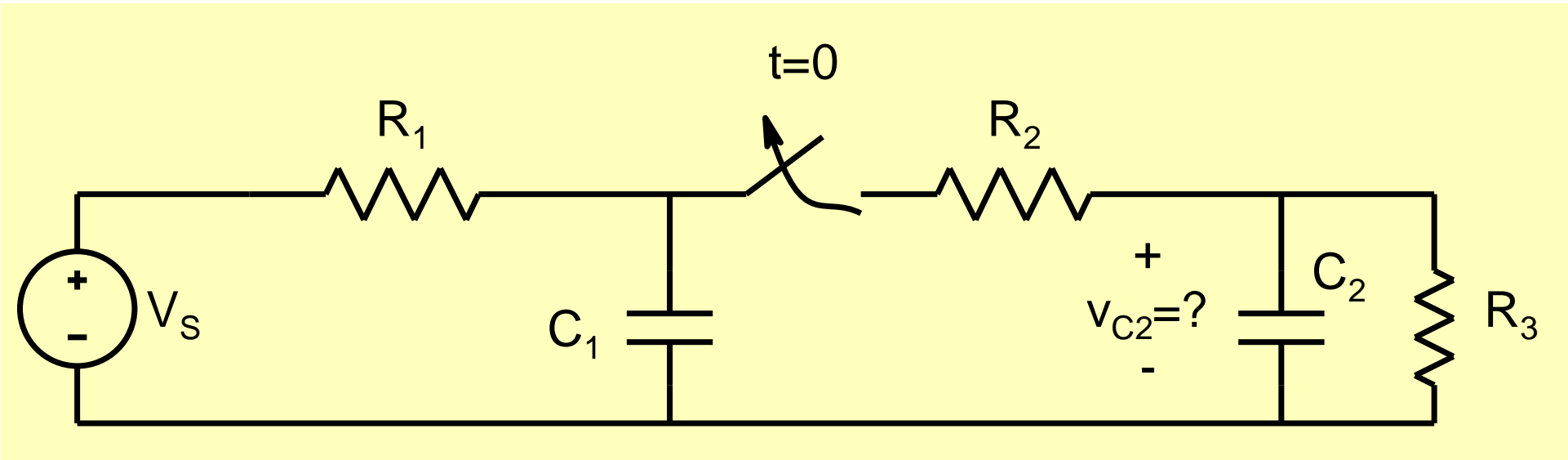
Voltages & currents inside the circuit?



Can Capacitance be Negative?

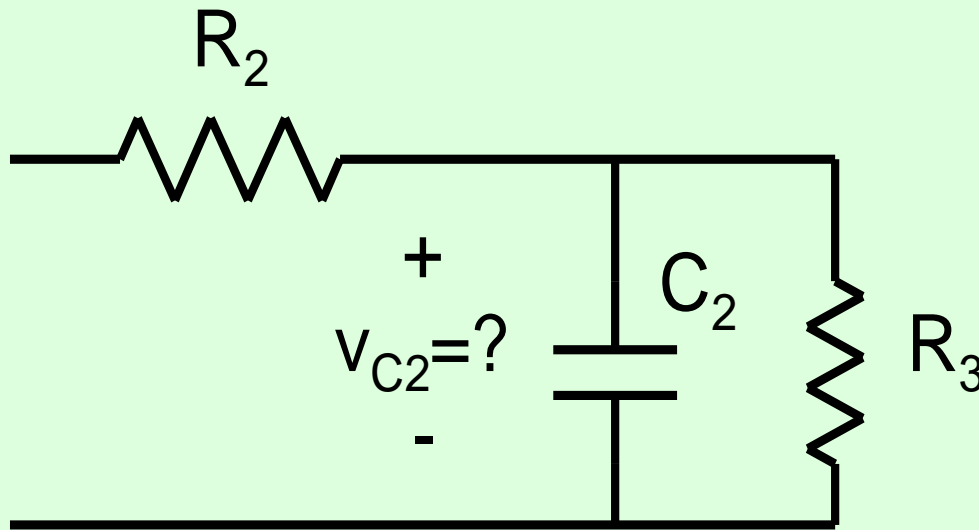
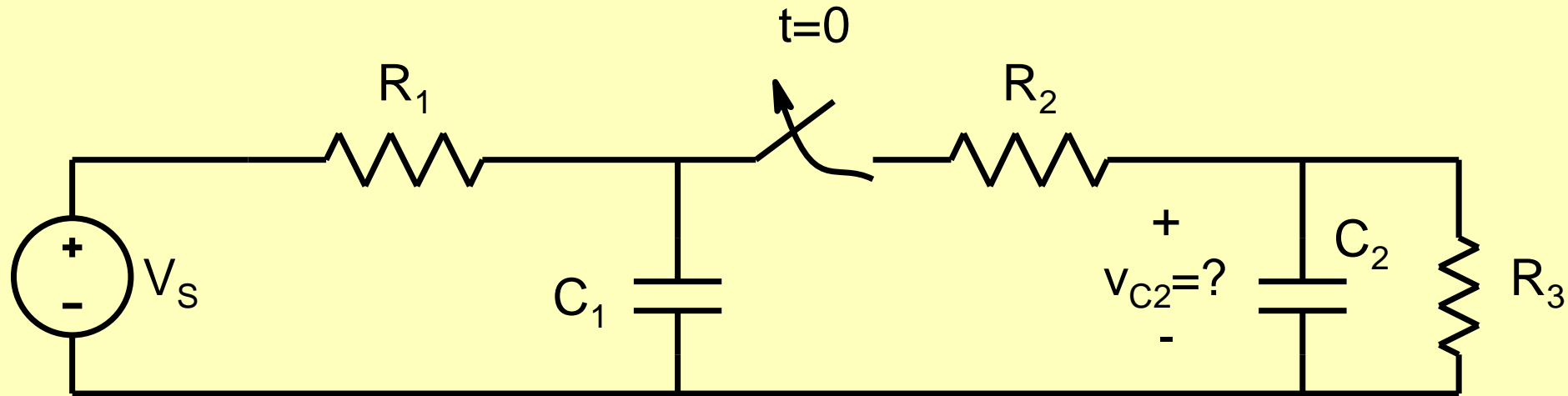
Why Intel is crazy about Negative Capacitance?

Can we solve this 2 capacitor problem using our present approach?



Circuit for $t > 0$

Can we solve this 2 capacitor problem using our present approach?

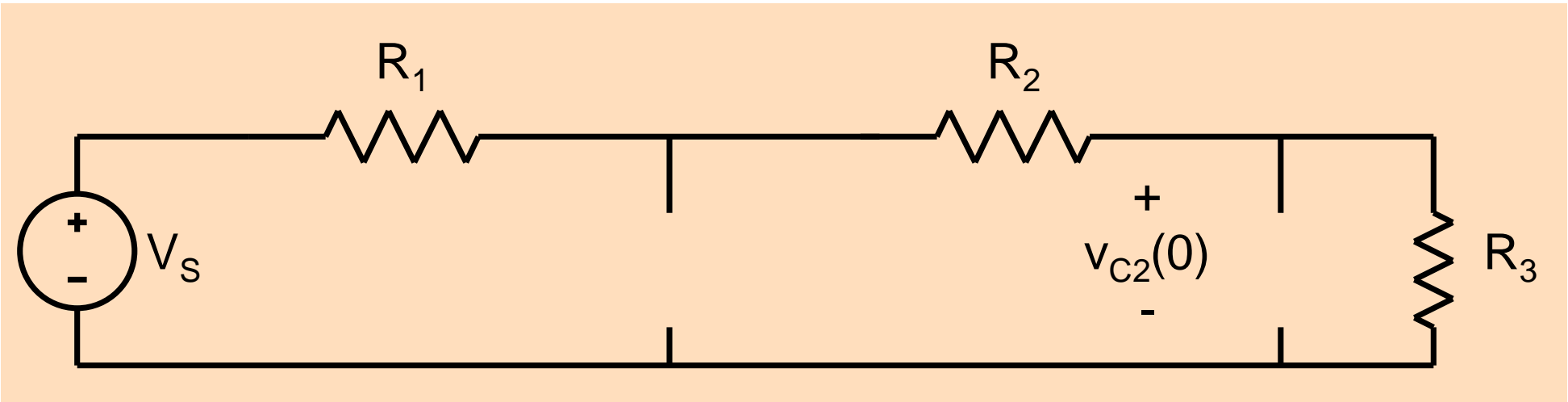
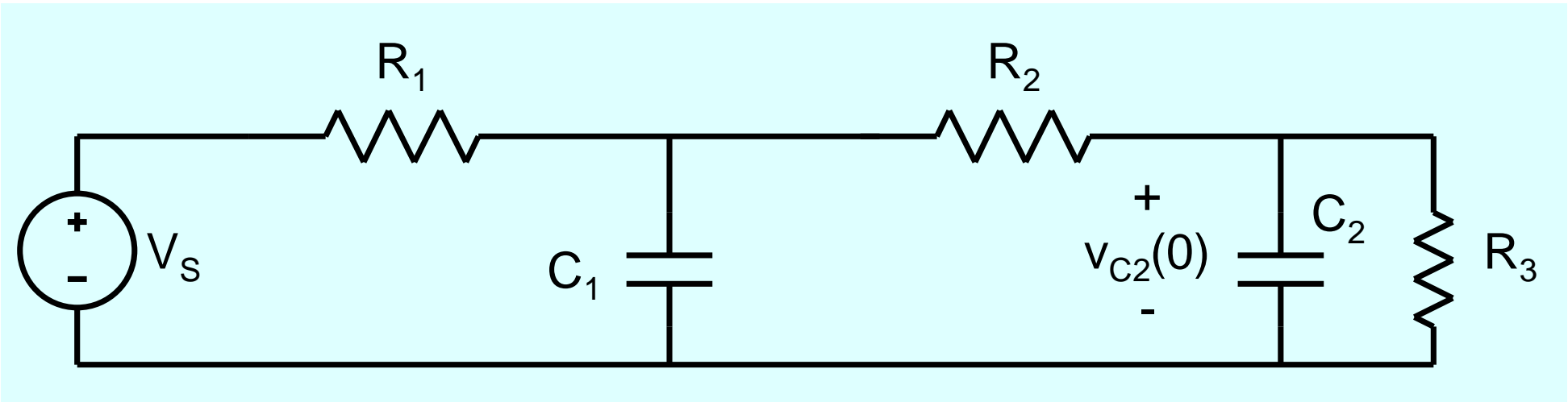


Circuit for $t > 0$

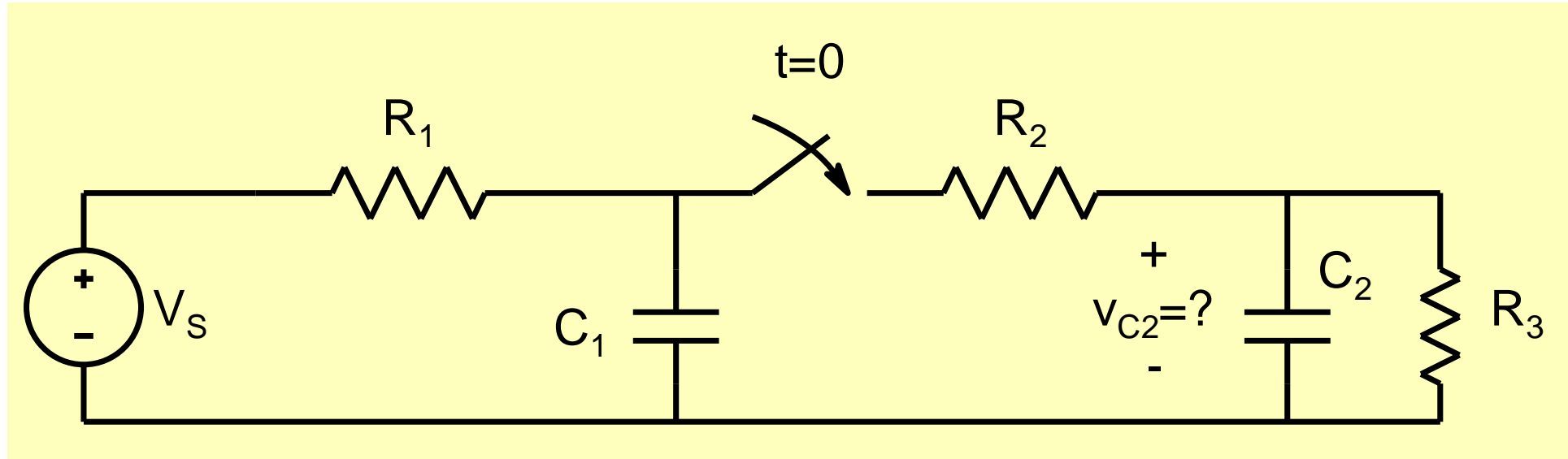
$$v_{c2}(0^+) = v_{c2}(0^-)$$

$$v_{c2}(t) = v_{c2}(\infty) + \{v_{c2}(0^+) - v_{c2}(\infty)\}e^{-\frac{t}{\tau}}$$

$$v_{c2}(0^+) = v_{c2}(0^-)$$



Will our approach work here?

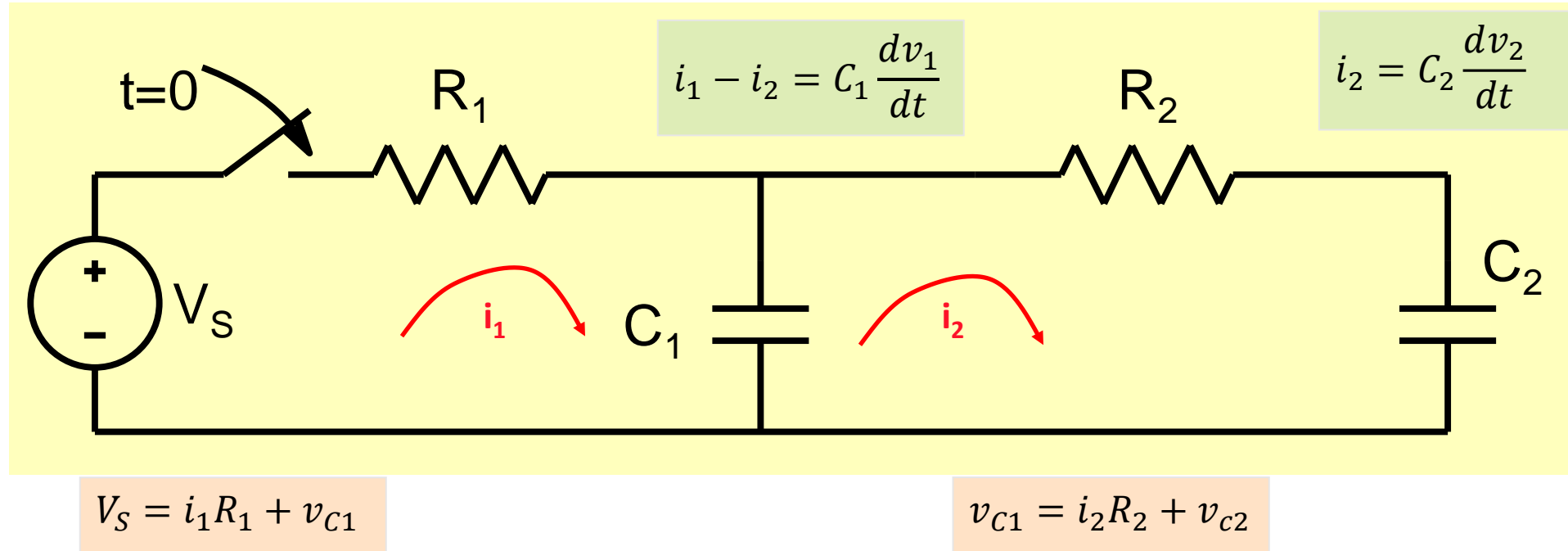


No, because circuit for $t > 0$ has two capacitances

As long as the circuit has single capacitor or inductor for the time interval for which the analysis is being carried out, the stated approach will work fine.

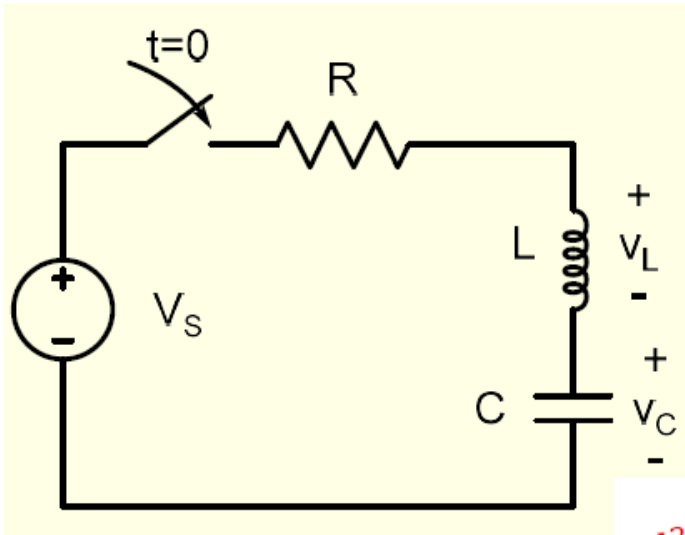
Circuits with Multiple Capacitors

What happens when there is more than one storage element?



$$R_1 R_2 C_1 C_2 \frac{d^2 v_{C2}}{dt^2} + (R_1 C_1 + R_1 C_2 + R_2 C_2) \frac{dv_{C2}}{dt} + v_{C2} = V_S$$

Series RLC circuit



$$V_S = I \times R + L \frac{dI}{dt} + V_C \quad I = C \frac{dV_C}{dt}$$

$$V_S = C \frac{dV_C}{dt} \times R + LC \frac{d^2V_C}{dt^2} + V_C$$

$$\frac{d^2V_C}{dt^2} + \frac{R}{L} \times \frac{dV_C}{dt} + \frac{V_C}{LC} = \frac{V_S}{LC}$$

$$V_C(t) = A \times e^{st} \quad As^2e^{st} + \frac{R}{L}Ase^{st} + \frac{A}{LC}e^{st} = 0$$

$$s^2 + \frac{R}{L}s + \frac{1}{LC} = 0 \quad s = -\frac{R}{2L} \pm \sqrt{\frac{R^2}{4L^2} - \frac{1}{LC}}$$

$$\omega_o = \frac{1}{\sqrt{L \times C}}$$

$$Q = \frac{\omega_o L}{R}$$

$$\frac{s}{\omega_o} = -\frac{1}{2Q} \pm \sqrt{\frac{1}{4Q^2} - 1}$$

How to find constants?

$$V_C(t) = V_S + A \times e^{s_1 t} + B \times e^{s_2 t}$$

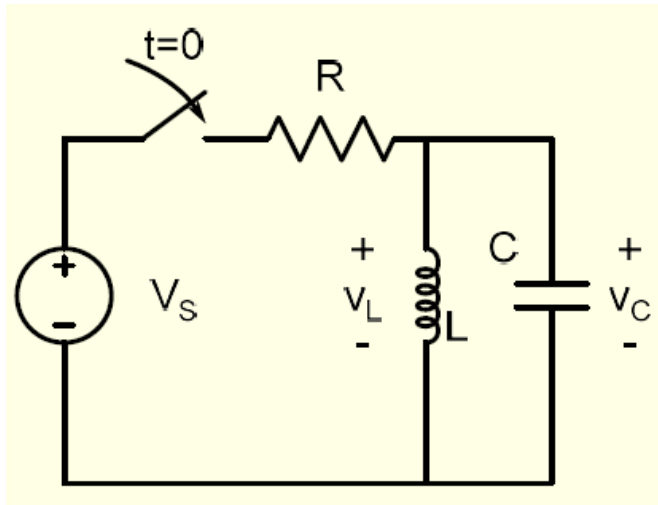
$$\omega_o = \frac{1}{\sqrt{L \times C}} \quad Q = \frac{\omega_o L}{R} \quad \frac{s}{\omega_o} = -\frac{1}{2Q} \pm \sqrt{\frac{1}{4Q^2} - 1}$$

$$V_C(0) = V_S + A + B = 0 \quad (1)$$

$$I_C(t) = C \frac{dV_C(t)}{dt} = C\{V_S + As_1 \times e^{s_1 t} + Bs_2 \times e^{s_2 t}\}$$

$$I_C(0) = 0 = C\{V_S + As_1 + Bs_2\} \quad (2)$$

Parallel RLC Circuit



$$V_S = I \times R + V_L \quad I = C \frac{dV_C}{dt} + I_L \quad V_L = L \frac{dI_L}{dt}$$

$$\frac{V_S}{R} = LC \frac{d^2 I_L}{dt^2} + \frac{L}{R} \times \frac{dV_L}{dt} + I_L$$

$$\frac{d^2 I_L}{dt^2} + \frac{1}{RC} \times \frac{dI_L}{dt} + \frac{I_L}{LC} = \frac{V_S}{RLC}$$

$$I_L(t) = A \times e^{st}$$

$$As^2 e^{st} + \frac{1}{RC} A s e^{st} + \frac{A}{LC} e^{st} = 0$$

$$s^2 + \frac{1}{RC} s + \frac{1}{LC} = 0 \quad s = -\frac{1}{2RC} \pm \sqrt{\frac{1}{4R^2C^2} - \frac{1}{LC}}$$

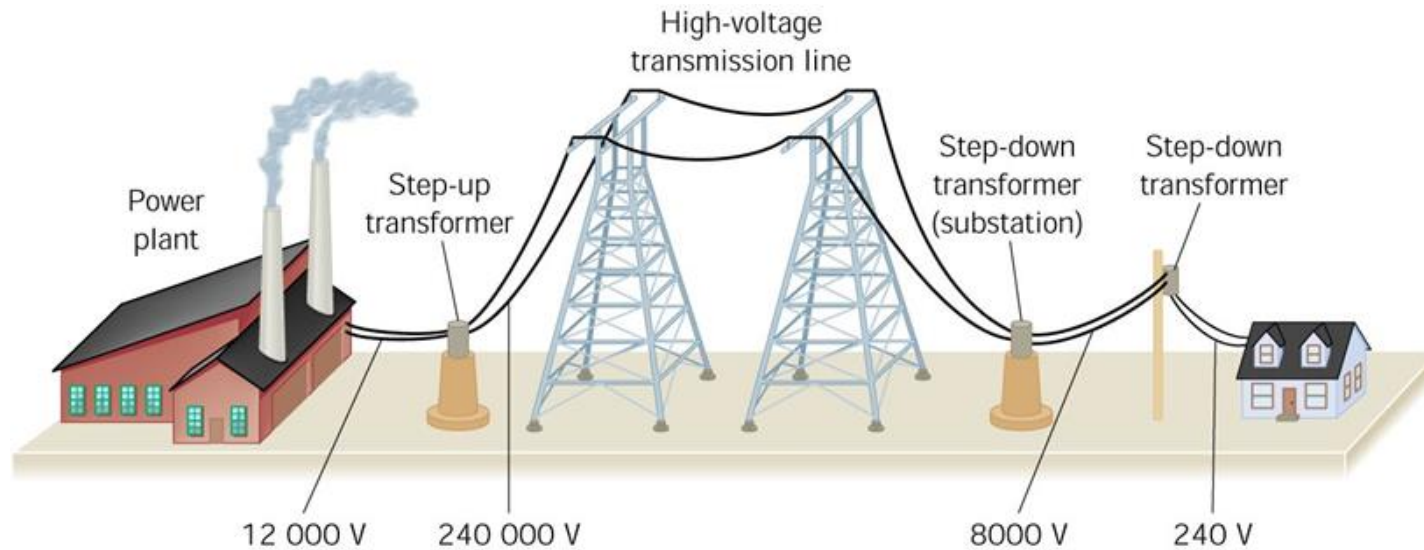
$$\omega_o = \frac{1}{\sqrt{L \times C}} \quad Q = \frac{R}{\omega_o \times L} \quad \frac{s}{\omega_o} = -\frac{1}{2Q} \pm \sqrt{\frac{1}{4Q^2} - 1}$$

Sinusoidal Signals

- Appear in many practical applications
 - Electric power is distributed by sinusoidal currents and voltages

$$p = v \times i$$
$$2.2KW = 2.2KV \times 1A$$
$$2.2KW = 220V \times 10A$$

$$Loss = i^2 R_{wire}$$



Communication

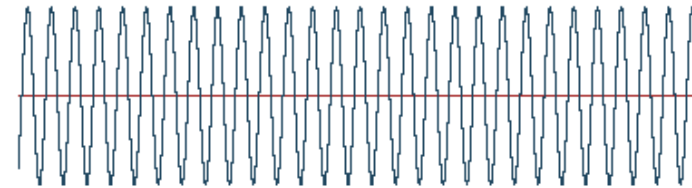


20 Hz -20KHz

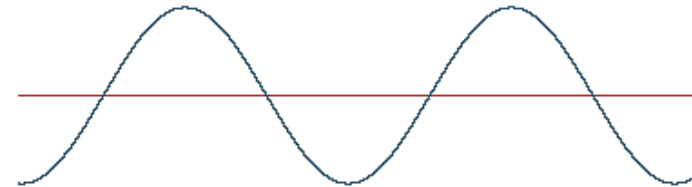
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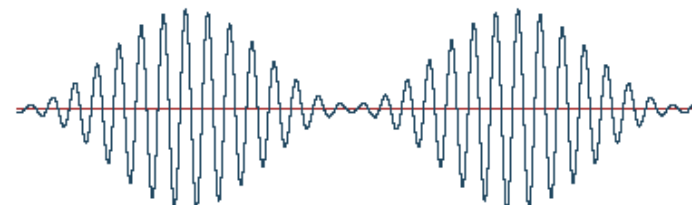
Carrier



Modulating Wave



Modulated Result

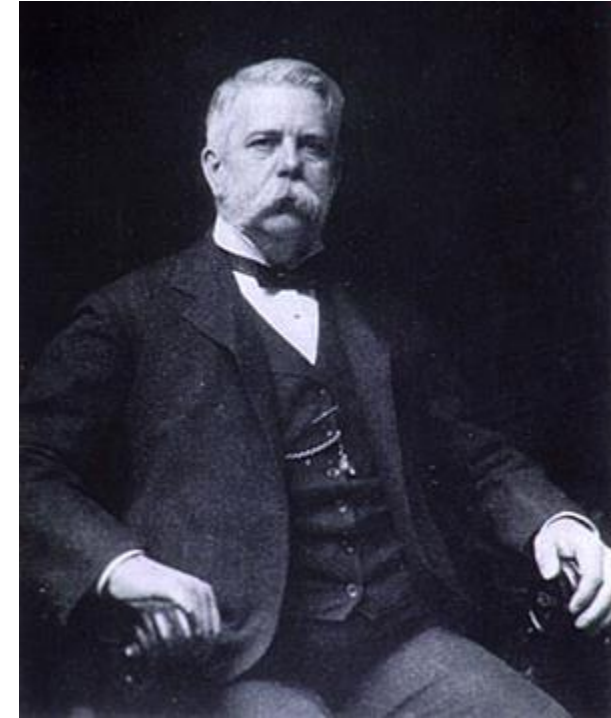


War of currents: AC vs DC



Tell Westinghouse to stick to air brakes. He knows all about them.

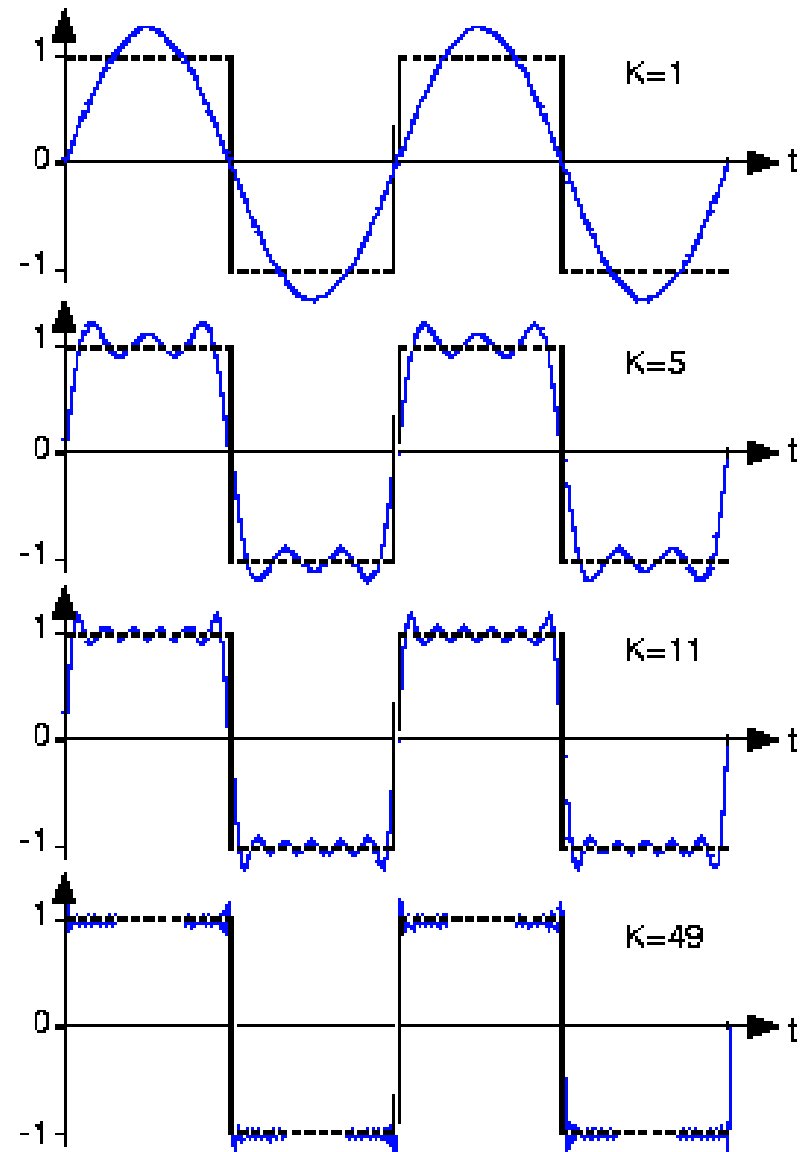
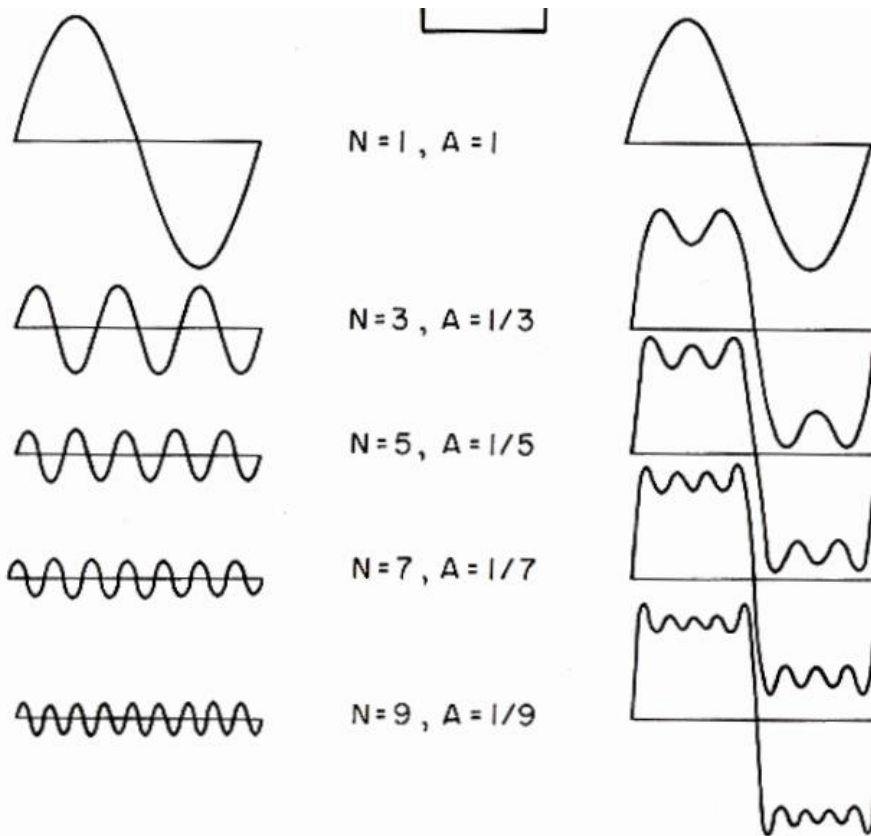
-Thomas Edison



George Westinghouse
formed an alliance with
Nikola Tesla

Fourier Analysis

$$A \sin\left(\frac{N\pi t}{T}\right) \rightarrow \frac{1}{N} \sin\left(\frac{N\pi t}{T}\right)$$



$$f(t) = \frac{4}{\pi} \sum_{n=1,3,5}^{\infty} \frac{1}{n} \sin\left(\frac{n\pi t}{T}\right)$$

Sinusoids through Circuits

Sinusoids have following interesting property

- Derivative is a sinusoid
- Integral is a sinusoid

$$\frac{d}{dt} \sin \omega t = \cos \omega t = \sin \left(\omega t + \frac{\pi}{2} \right)$$

$$i_c = C \frac{dv_c}{dt}$$

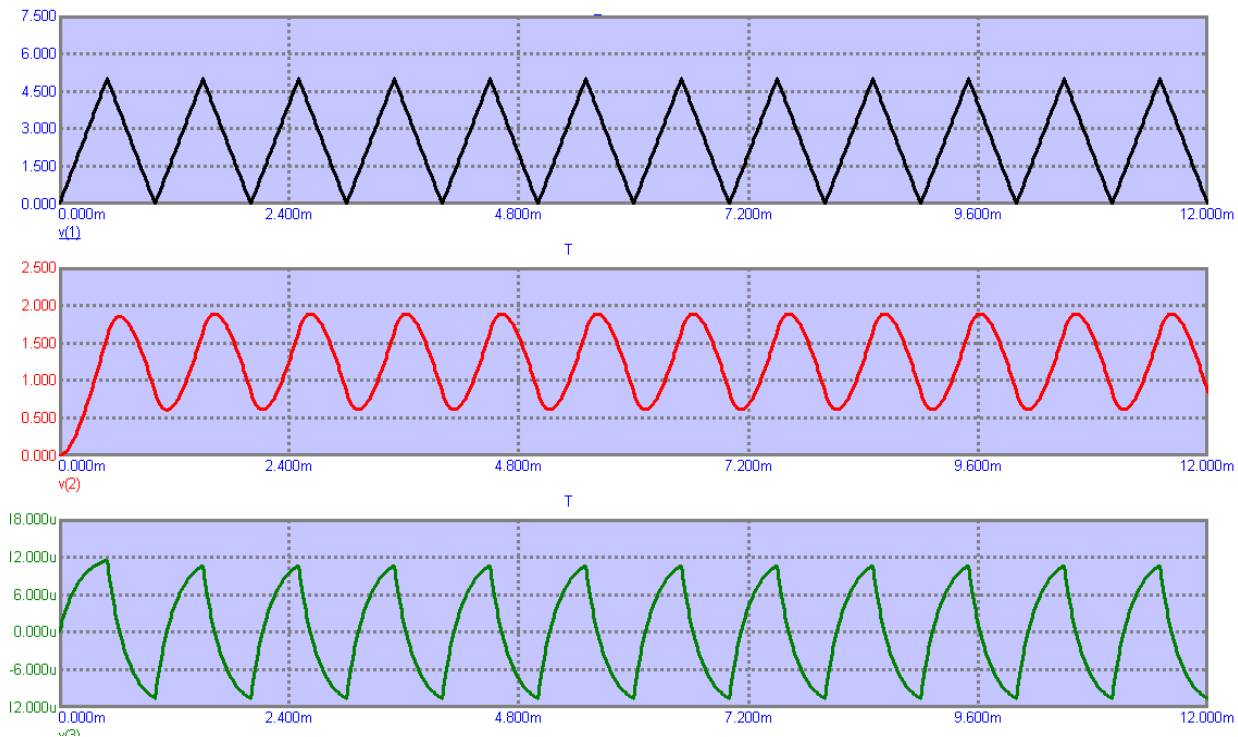
$$\int \sin \omega t \, dt = -\cos \omega t = \sin \left(\omega t - \frac{\pi}{2} \right)$$

$$v = L \frac{di}{dt}$$

So as a sinusoidal signal goes through a circuit, it remains a sinusoid of the same frequency.

True for any Linear time-invariant system.

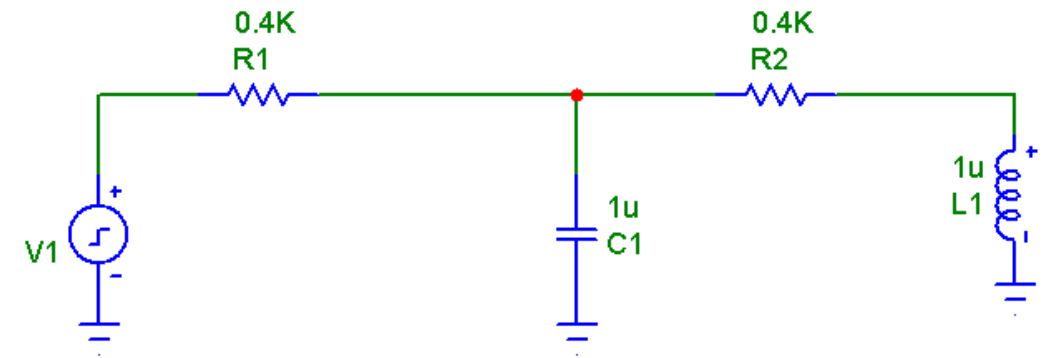
Triangular signal



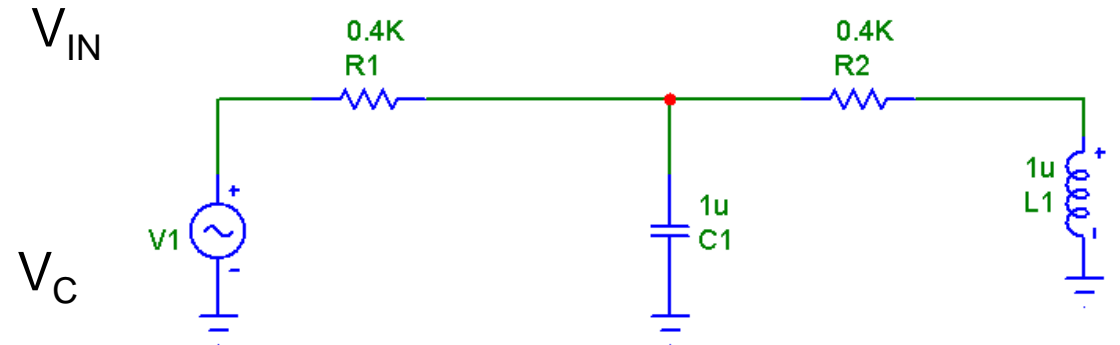
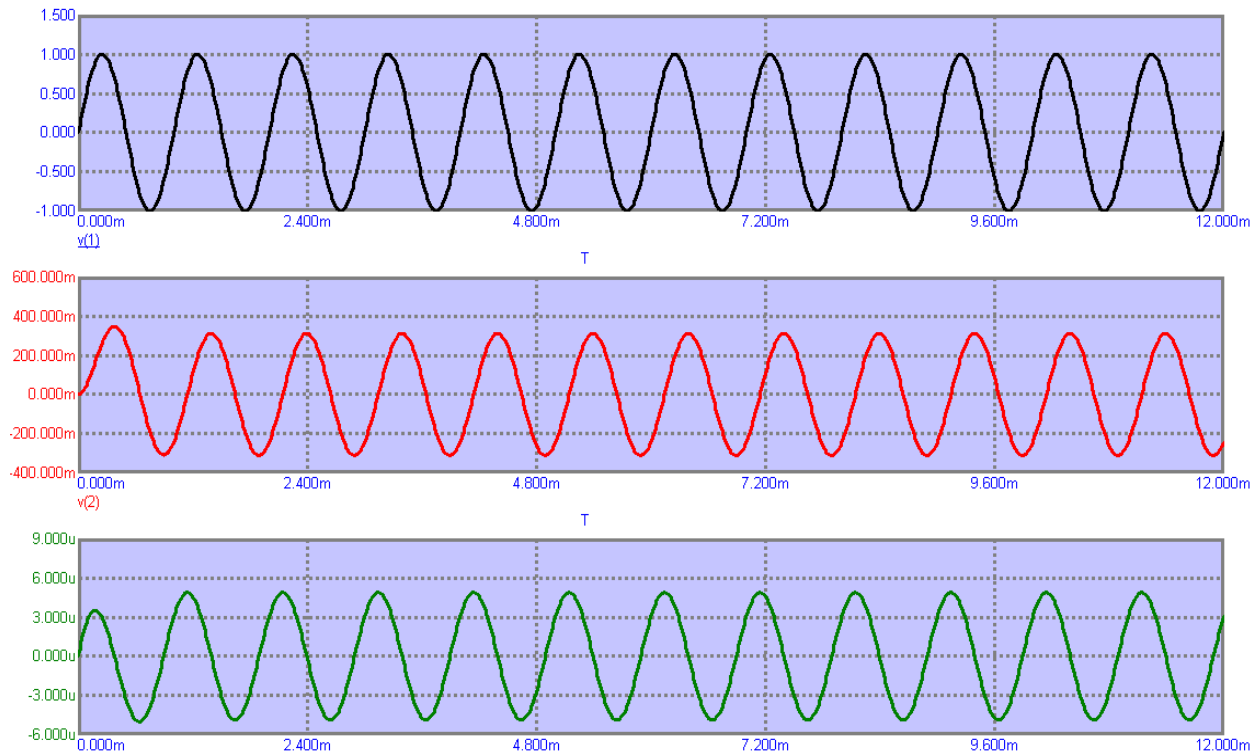
V_{IN}

V_C

V_L



Sinusoidal signal

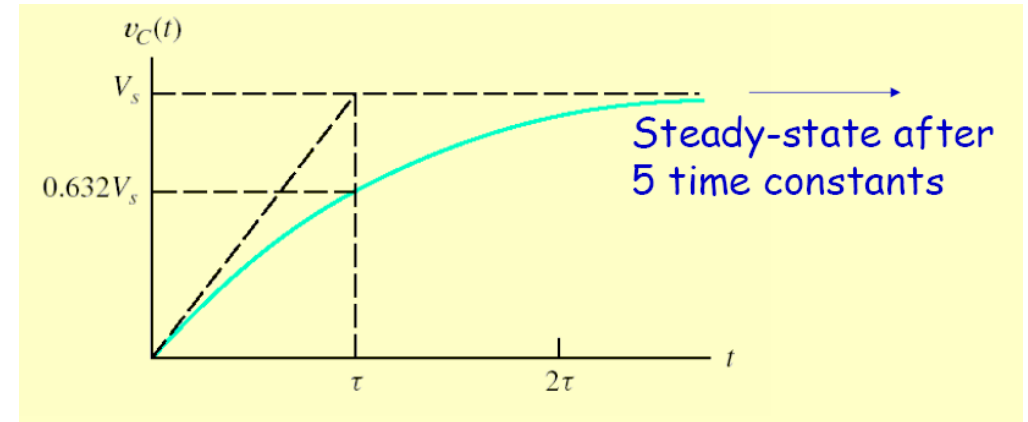
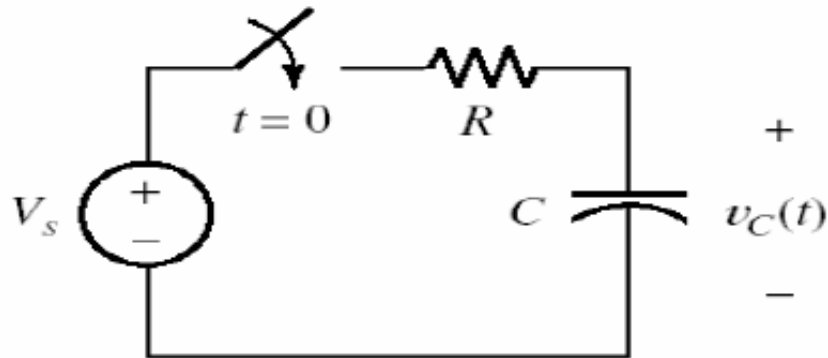


V_L

Voltage everywhere in the circuit is sinusoidal

Transient & forced response

- Split solution into two components:
 - Transient (time-dependent component)
 - Forced (steady-state)



$$v_C(t) = V_s - V_s e^{-t/\tau}$$

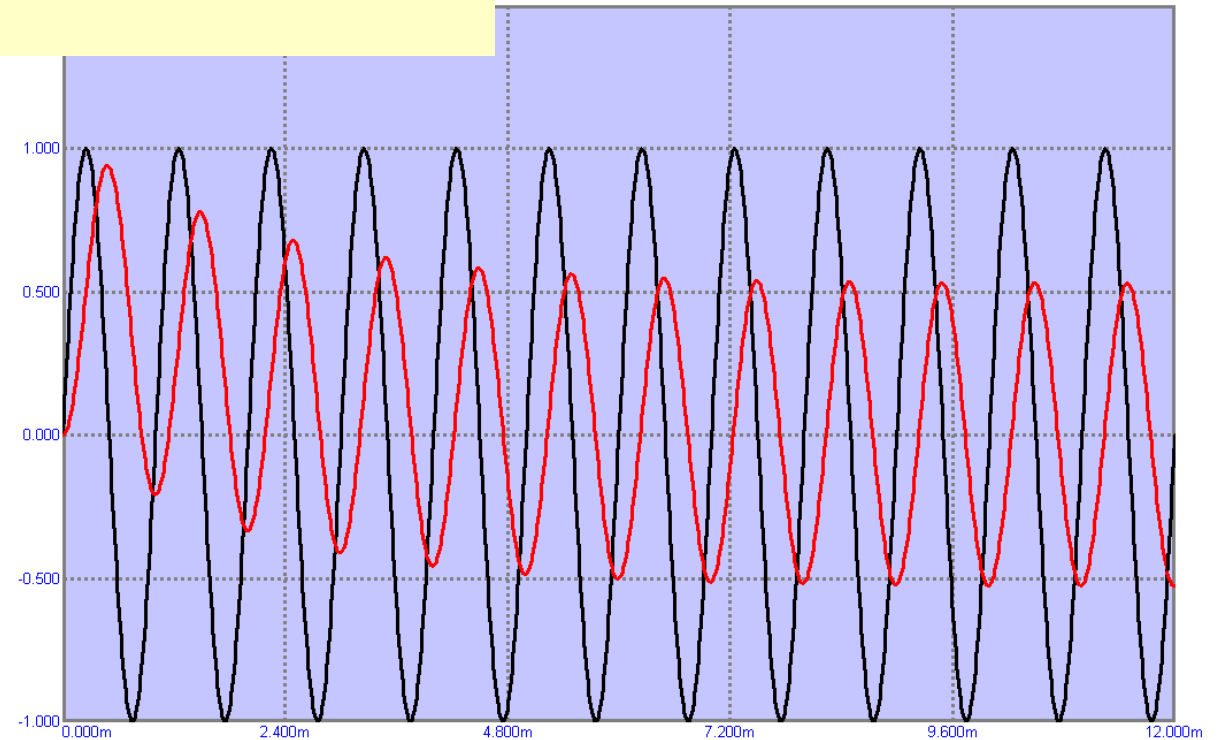
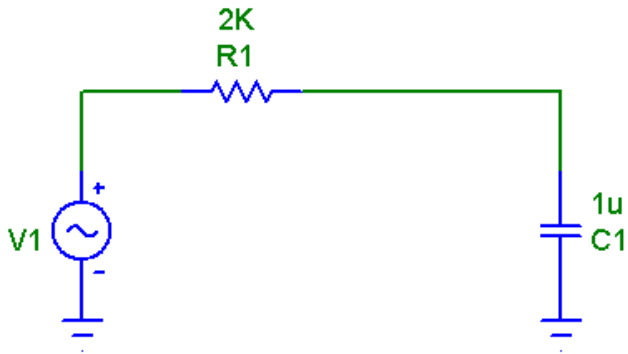
Steady-state or forced response

Transient response

Sinusoidal steady state

Sinusoidal Steady-State

- Whenever the forced input to the circuit is sinusoidal the response will be sinusoidal
- If the input persists, the response will persist and we call it steady-state response



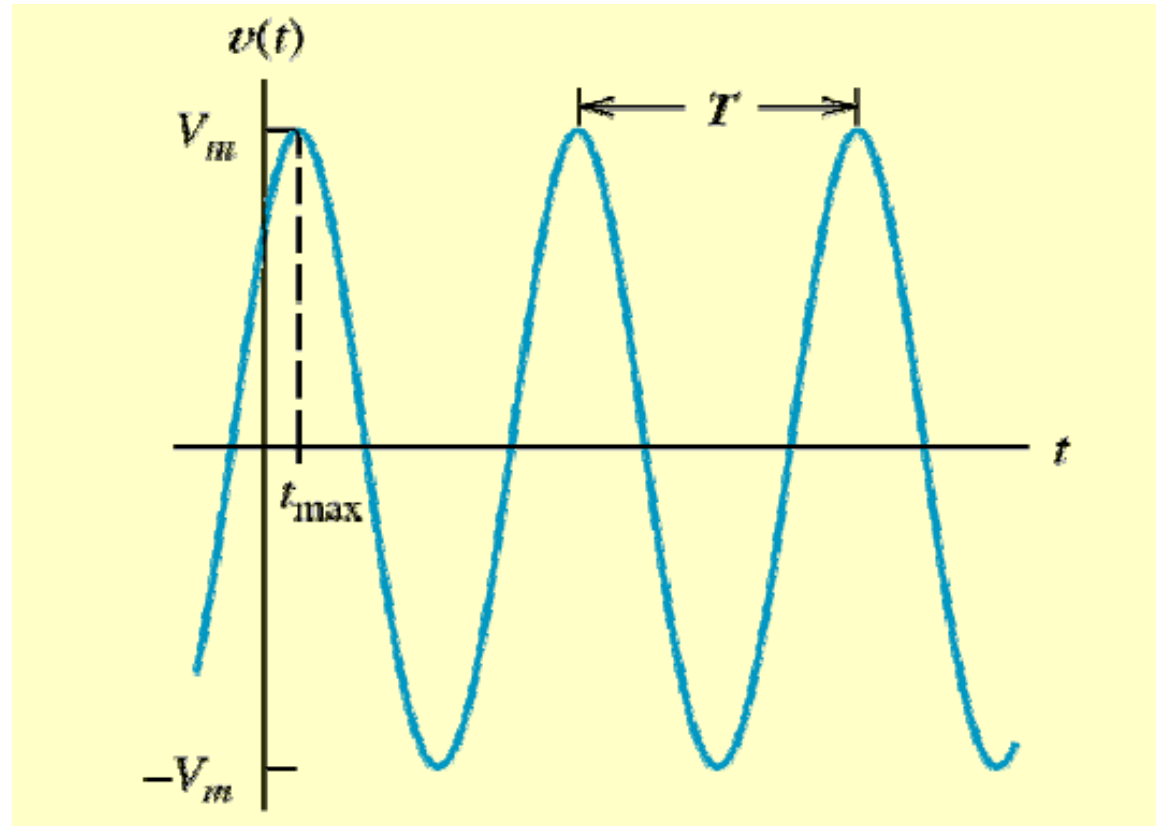
Preliminaries: Representation of Sinusoidal Signals

Canonical Form

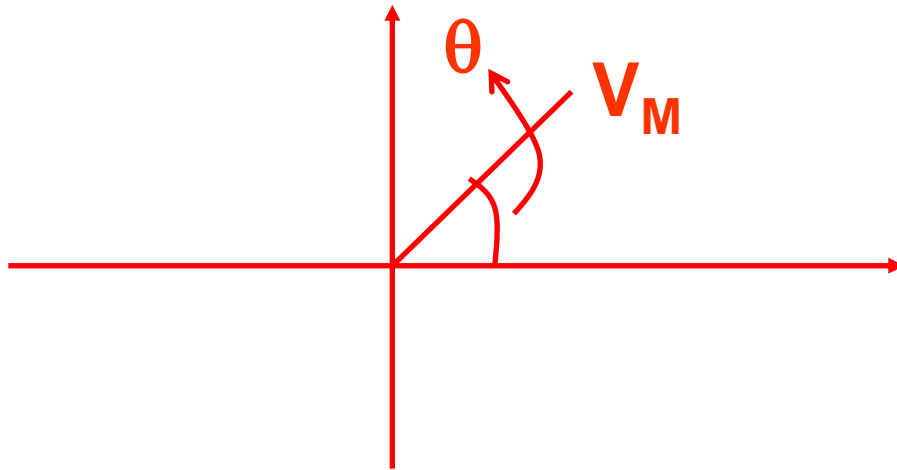
$$v(t) = V_m \cos(\omega t + \theta)$$

peak
value

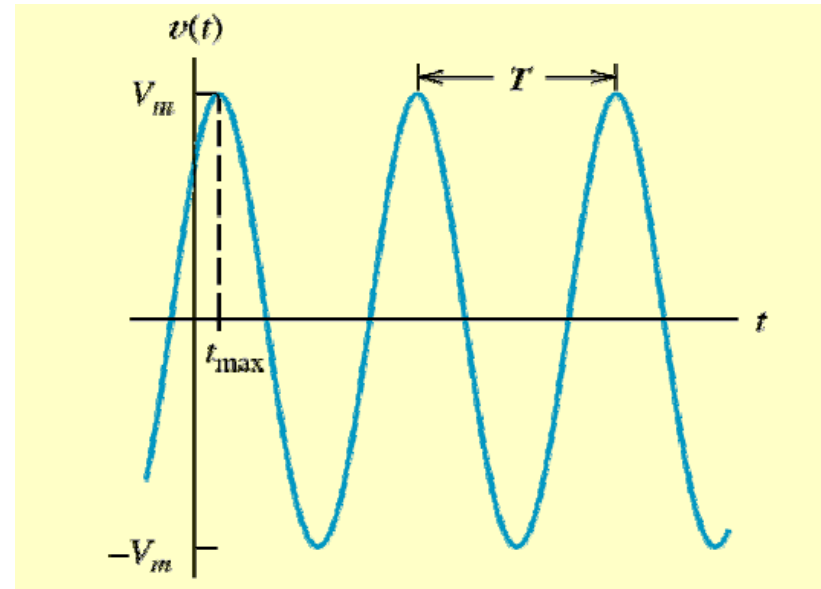
phase



Representation of Sinusoidal Signals



$$v(t) = V_m \cos(\omega t + \theta)$$



ω is the **angular frequency** in radians per second

T is the **period**, where $f = \frac{1}{T}$ is the **frequency**

$$\omega = \frac{2\pi}{T}$$

$$\omega = 2\pi f$$

θ is the **phase angle**

Representation of Sinusoidal Signals...

$$5 \sin(4\pi t - 60^\circ)$$

$$= 5 \cos(4\pi t - 60^\circ - 90^\circ)$$

Amplitude = 5 ;
Phase = -150°

$$v(t) = V_m \cos(\omega t + \theta)$$

$$\sin(z) = \cos(z - 90^\circ)$$

Phase in radians: $360^\circ = 2\pi$ $\theta = \frac{-150}{360} \times 2\pi = -2.618$ radians

$$\omega = 4\pi \text{ rad/s}$$

$$\omega = \frac{2\pi}{T} = 4\pi \Rightarrow T = 0.5s$$

$$f = \frac{1}{T} = 2\text{Hz}$$

Preliminaries:

Find the phase difference between the two currents

$$i_1 = 4 \sin(377t + 25^\circ)$$

Canonical Form $x(t) = x_m \cos(\omega t + \theta)$

$$i_2 = -5 \cos(377t - 40^\circ)$$

$$i_1 = 4 \cos(377t + 25^\circ - 90^\circ)$$

$$\theta_1 = -65^\circ$$

$$\theta_1 - \theta_2 = -205^\circ$$

$$i_2 = 5 \cos(377t - 40^\circ + 180^\circ)$$

$$\theta_2 = 140^\circ$$

$$\sin(\omega t \pm 180^\circ) = -\sin \omega t$$

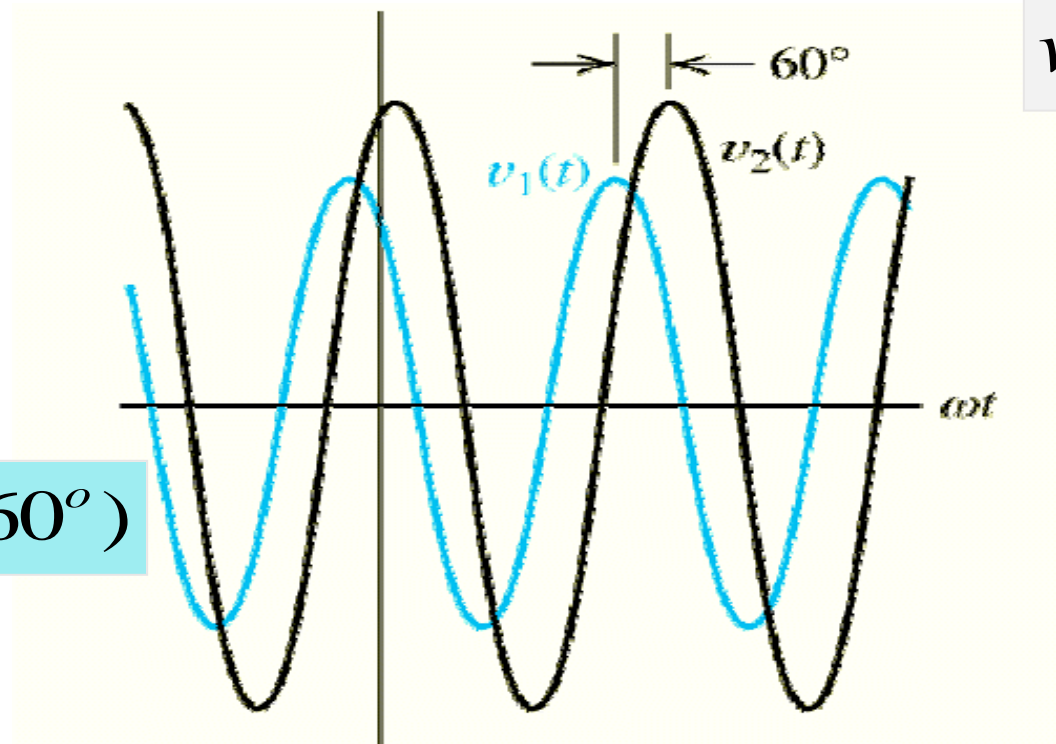
$$\cos(\omega t \pm 180^\circ) = -\cos \omega t$$

$$\sin(\omega t \pm 90^\circ) = \pm \cos \omega t$$

$$\cos(\omega t \pm 90^\circ) = \mp \sin \omega t$$

Phase relationship

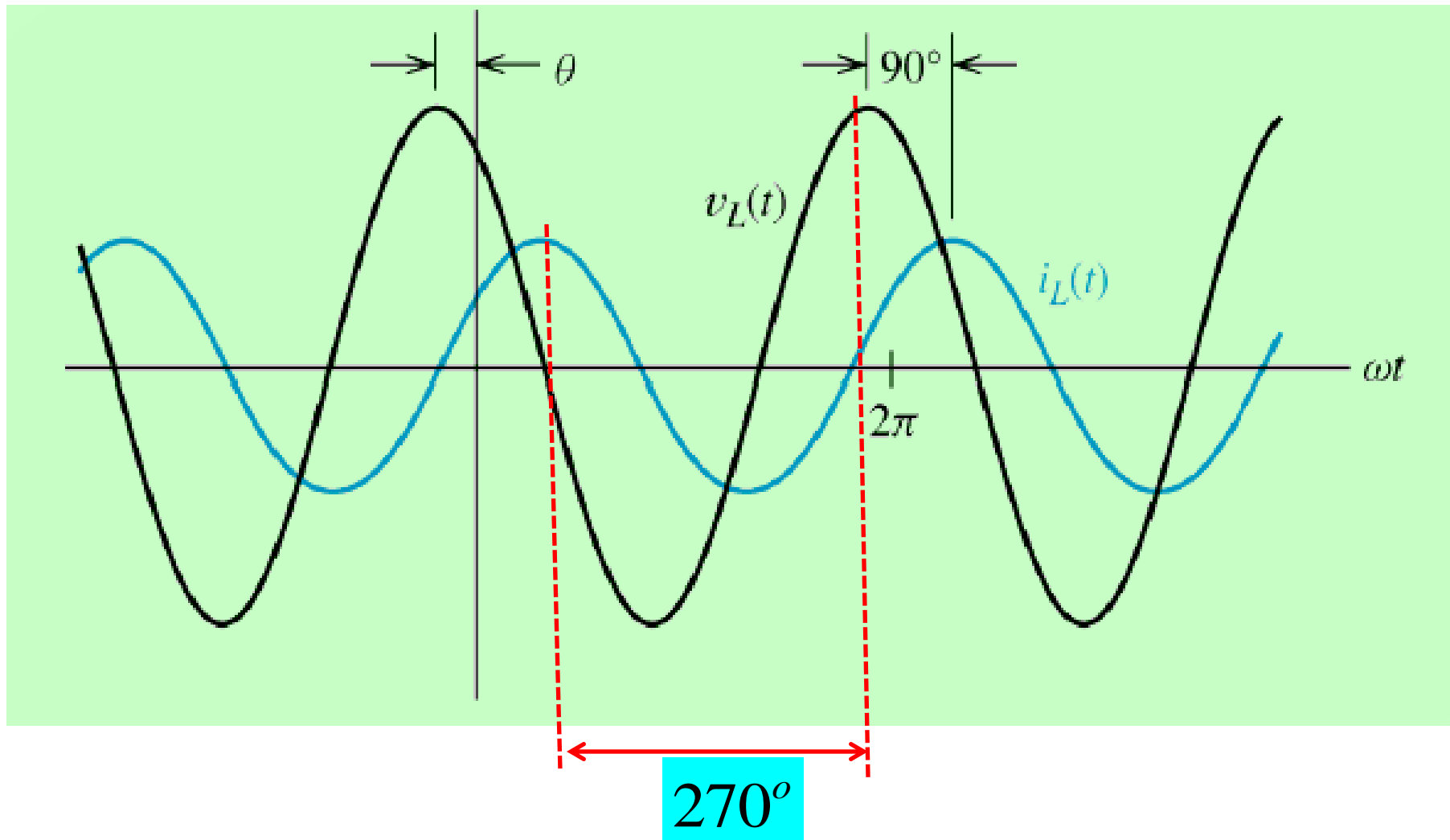
$$v_1(t) = v_{1m} \cos(\omega t + 60^\circ)$$



$$v_2(t) = v_{2m} \cos(\omega t)$$

The peaks of $v_1(t)$ occur 60° before the peaks of $v_2(t)$.

In other words, $v_1(t)$ leads $v_2(t)$ by 60° .



Voltage leads current by 90° or lags current by 270° ?

Phase difference is usually considered between -180° to 180°

Add or subtract 360° to bring the phase between -180° to 180°

$$i_1 = 4 \cos(377t - 65^\circ)$$

$$i_2 = 5 \cos(377t + 140^\circ)$$

Does i_2 lead i_1 ?

$$\theta_1 - \theta_2 = -205^\circ$$

$$\theta_1 - \theta_2 = -205^\circ + 360^\circ = 155^\circ$$

i_1 leads i_2 by 155°