

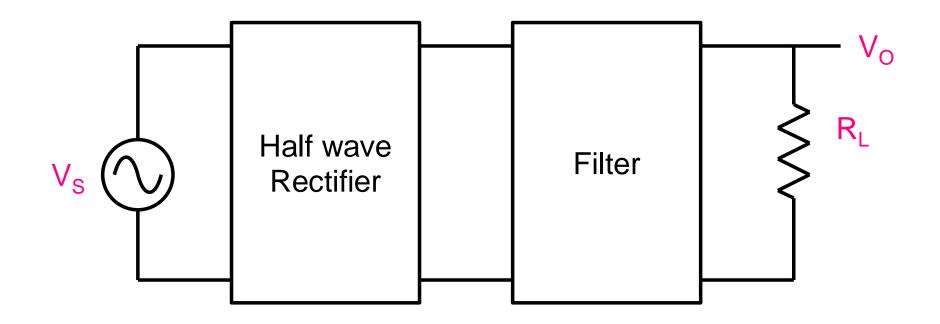
ESC201: Introduction to Electronics

MODULE 5: AMPLIFIERS



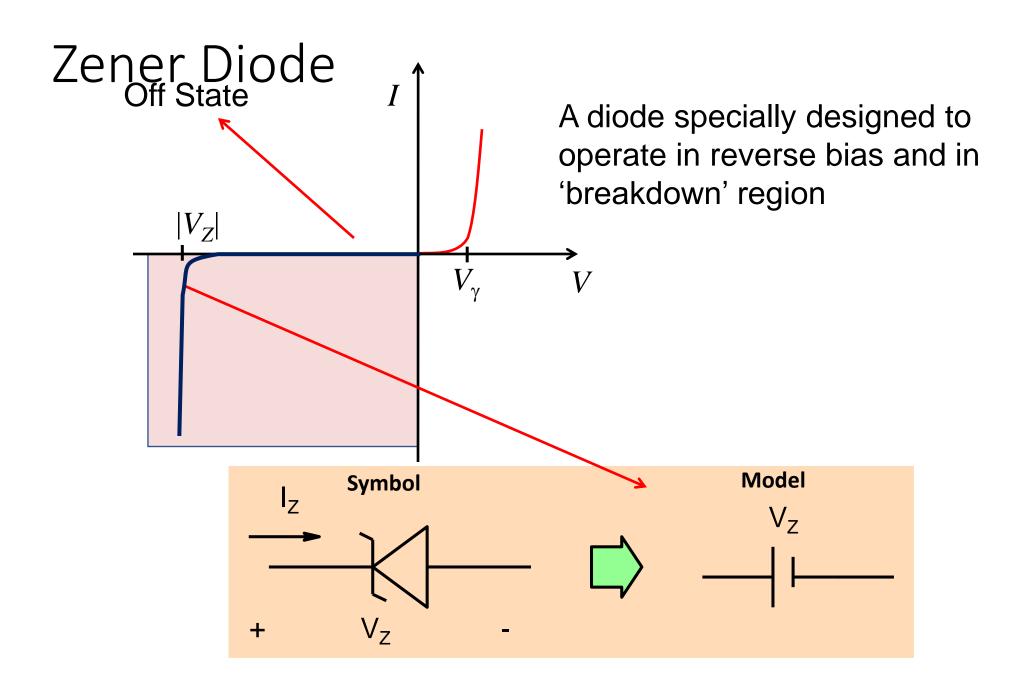
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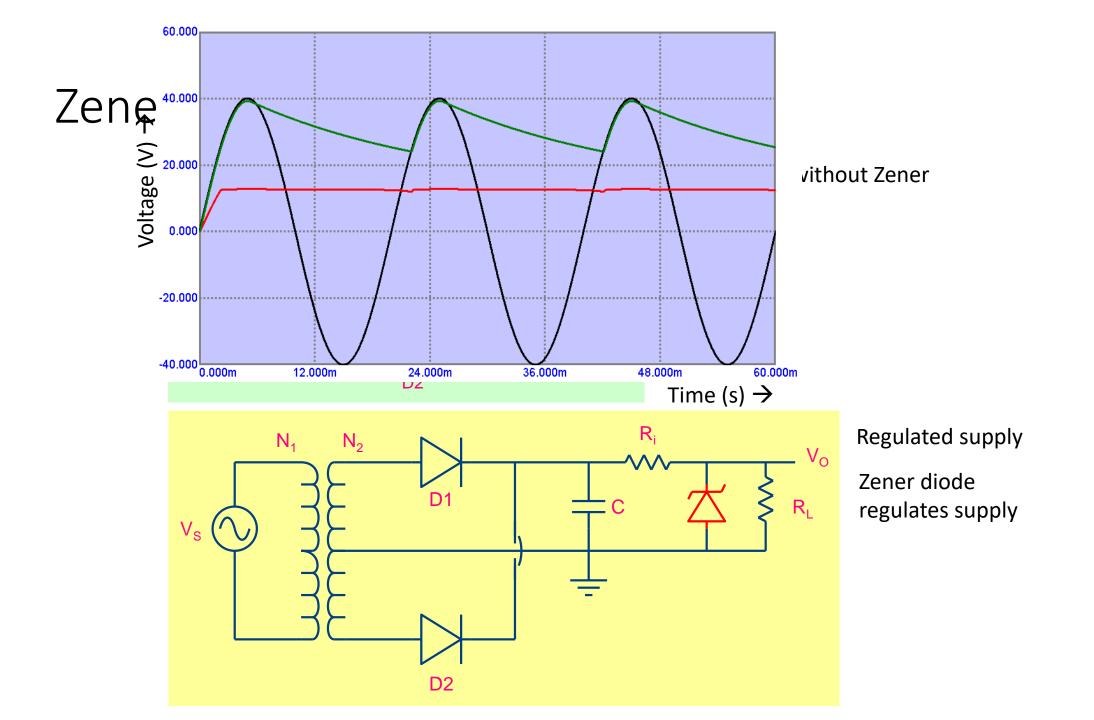
Power supply: block diagram



Comparison

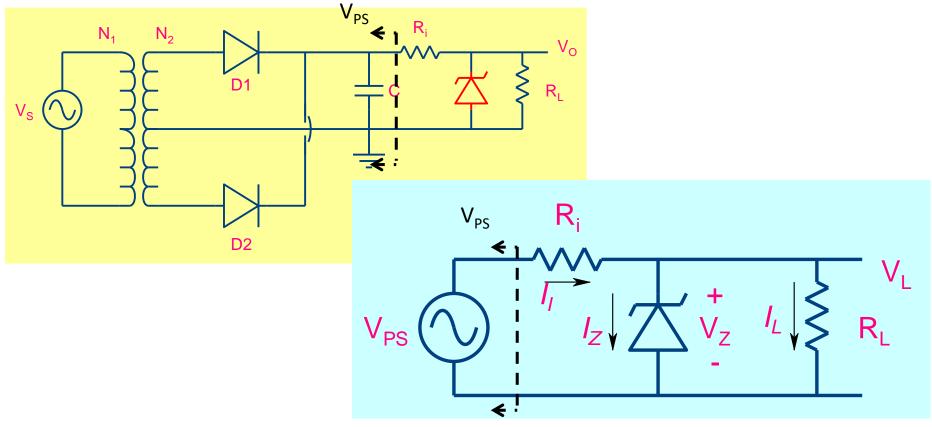
	Half Wave Rectifier	Full Wave Rectifier	Bridge Rectifier
Number of diodes	1	2	4
Ripple Voltage $V_{\rm r}$	$V_r \cong \frac{V_M}{fR_LC}$	$V_r \cong \frac{V_M}{2fR_LC}$	$V_r \cong \frac{V_M}{2fR_LC}$
Peak Diode Current i_{DMAX}	$\omega C \times \sqrt{2V_r V_M} + \frac{V_M}{R_L}$		$\omega C \times \sqrt{2V_r V_M} + \frac{V_M}{R_L}$
Peak Inverse Voltage <i>PIV</i>	$V_{ m M}$	$2V_{ m M}$ - $V_{ m \gamma}$	$V_{ m M}$ - $V_{ m \gamma}$





Example

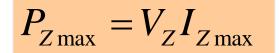
Design a voltage reference circuit

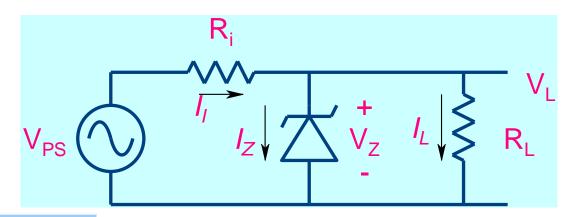


Design Problem: Determine R_i and Zener diode specifications such that output voltage is +12V, load current can vary between 0 to 0.1A. The input voltage may vary between 18 to 15.5V. I_{Zmax}/I_{Zmin} =10.

Example (continued)

Design Equations





$$I_{i} = \frac{V_{PS} - V_{Z}}{R_{i}} = I_{Z} + I_{L}$$

$$I_Z = \frac{V_{PS} - V_Z}{R_i} - I_L$$

$$I_{Z\max} = \frac{V_{PS\max} - V_{Z}}{R_i} - I_{L\min}$$

$$I_{Z\min} = \frac{V_{PS\min} - V_Z}{R_i} - I_{L\max}$$

$$\frac{I_{z\max}}{I_{z\min}} \cong 10$$

$$R_{i} = \frac{V_{PS \min} - 0.1V_{PS \max} - 0.9V_{Z}}{I_{L \max} - 0.1I_{L \min}}$$

Example (continued)

Design Problem: Determine R_i and zener diode specifications such that output voltage is +12V, load current can vary between 0 to 0.1A. The input voltage may vary between 18 to 15.5V.

$$R_{i} = \frac{V_{PS\min} - 0.1V_{PS\max} - 0.9V_{Z}}{I_{L\max} - 0.1I_{L\min}} = 29\Omega$$

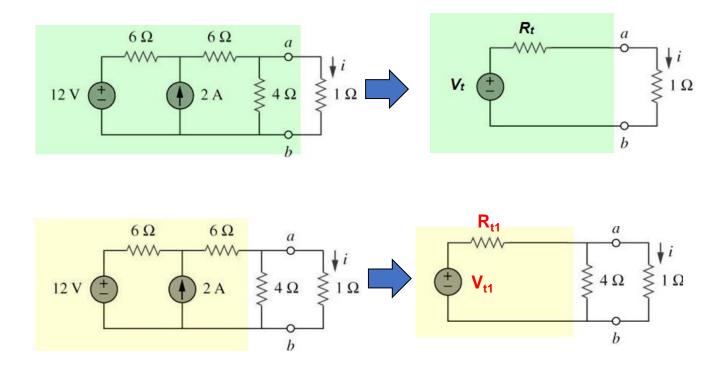
$$I_{Z \max} = \frac{V_{PS \max} - V_{Z}}{R_{i}} - I_{L \min} = 0.207A$$

$$I_{Z\min} = \frac{V_{PS\min} - V_{Z}}{R_{i}} - I_{L\max} = 0.0207$$

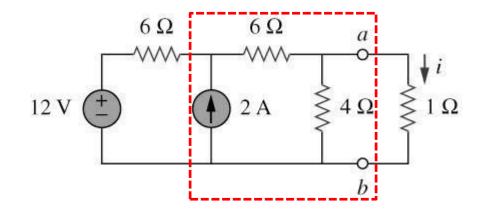
$$P_{Z\max} = V_Z I_{Z\max} = 2.48W$$

Abstractions

An abstract representation is a simplified representation that has appropriate level of detail for the problem being addressed.



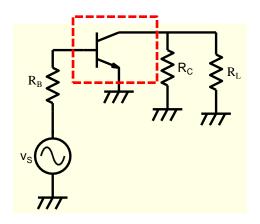
Limitation of Single Port Network



How do we build a simplified representation of only this portion of the circuit?

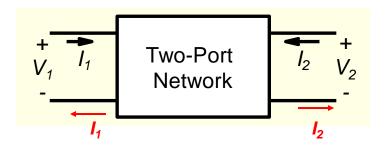
Thevenin's or Norton's Theorem are not Sufficient

Analysis of Elements Occurring In Circuits



How do we analyze circuits containing new components?

Two-Port Networks



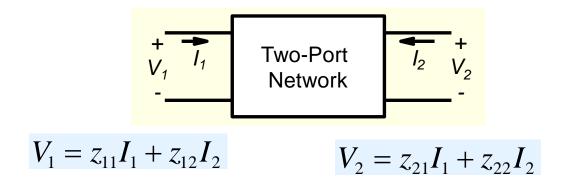
- Port: A pair of terminals through which a signal can enter/leave the network
- Constraints on analysis:
 - 1. Linear elements only (R,L,C, dependent sources,..)
 - 2. No independent sources or stored energy inside the network

No matter how complicated is the circuit inside the two-port network, it can be represented by only <u>four</u> elements!

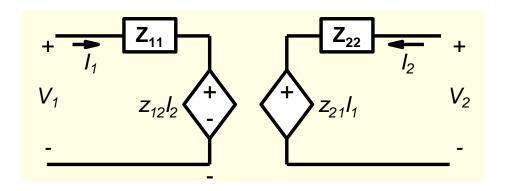
Popular Forms of Two Port Network

- Z (Impedance) Parameters
- Y (Admittance) Parameters
- H (Hybrid) Parameters
- G (Inverse Hybrid) Parameters

Z or Impedance Parameters



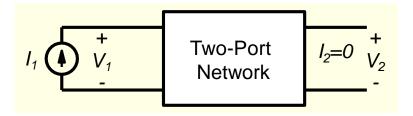
$$\begin{pmatrix} V_1 \\ V_2 \end{pmatrix} = \begin{pmatrix} z_{11} & z_{12} \\ z_{21} & z_{22} \end{pmatrix} \begin{pmatrix} I_1 \\ I_2 \end{pmatrix}$$



Z Parameter Determination

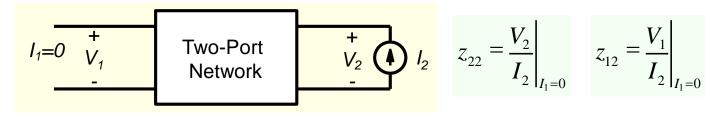
$$V_1 = z_{11}I_1 + z_{12}I_2$$

$$V_1 = z_{11}I_1 + z_{12}I_2 \qquad V_2 = z_{21}I_1 + z_{22}I_2$$



$$z_{11} = \frac{V_1}{I_1} \bigg|_{I_2 = 0}$$

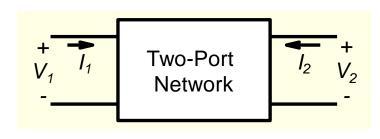
$$z_{12} = 0$$
 $z_{11} = \frac{V_1}{I_1} \Big|_{I_2 = 0}$ $z_{21} = \frac{V_2}{I_1} \Big|_{I_2 = 0}$



$$z_{22} = \frac{V_2}{I_2} \bigg|_{I_1 = 0}$$

$$Z_{12} = \frac{V_1}{I_2} \bigg|_{I_1 = 0}$$

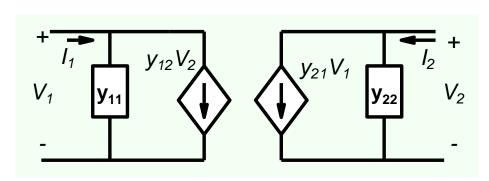
Y or Admittance Parameters



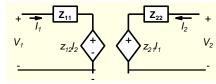
$$I_1 = y_{11}V_1 + y_{12}V_2$$

$$I_1 = y_{11}V_1 + y_{12}V_2$$
 $I_2 = y_{21}V_1 + y_{22}V_2$

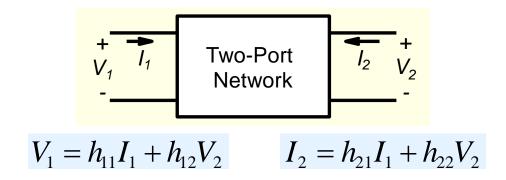
$$\begin{pmatrix} I_1 \\ I_2 \end{pmatrix} = \begin{pmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{pmatrix} \begin{pmatrix} V_1 \\ V_2 \end{pmatrix}$$



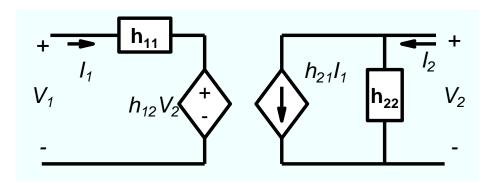
For comparison



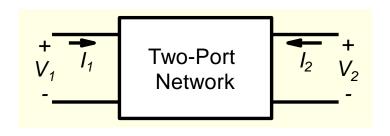
H or Hybrid Parameters



$$\begin{pmatrix} V_1 \\ I_2 \end{pmatrix} = \begin{pmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{pmatrix} \begin{pmatrix} I_1 \\ V_2 \end{pmatrix}$$



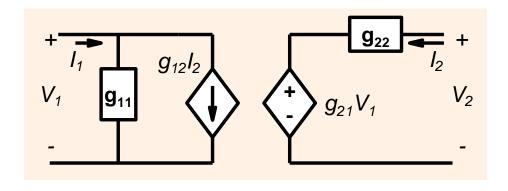
G or Inverse Hybrid Parameters



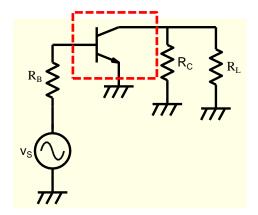
$$I_1 = g_{11}V_1 + g_{12}I_2$$

$$I_1 = g_{11}V_1 + g_{12}I_2$$
 $V_2 = g_{21}V_1 + g_{22}I_2$

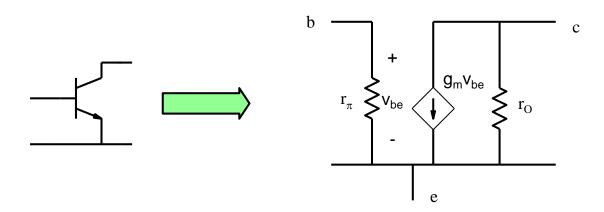
$$\begin{pmatrix} I_1 \\ V_2 \end{pmatrix} = \begin{pmatrix} g_{11} & g_{12} \\ g_{21} & g_{22} \end{pmatrix} \begin{pmatrix} V_1 \\ I_2 \end{pmatrix}$$



Representation of Complex Elements Within Circuits

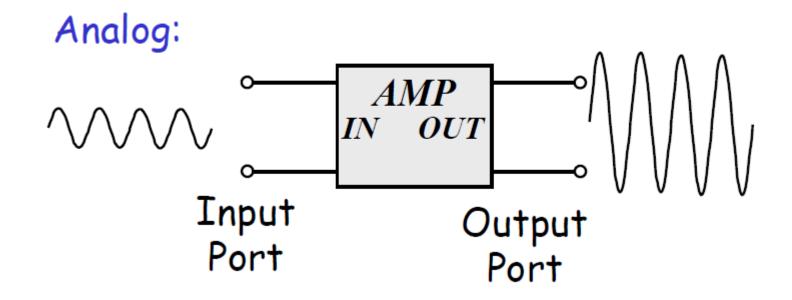


Two port network allows transistor representation in terms of familiar elements.



Why amplify?

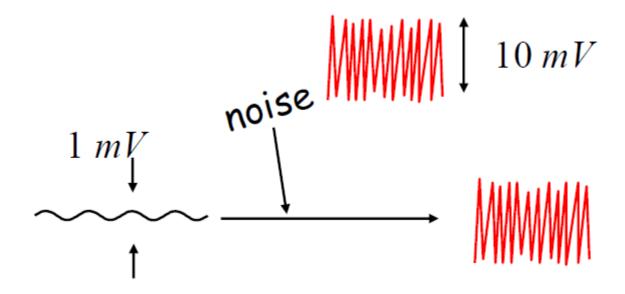
Key to analog and digital processing



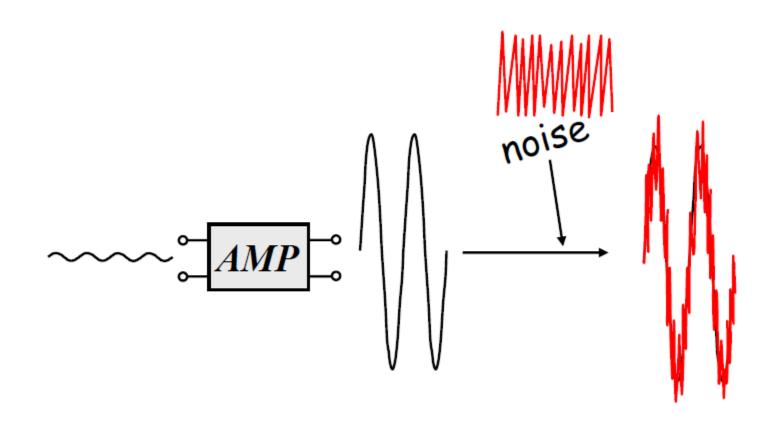
Active Device: supplies power

Noise tolerance

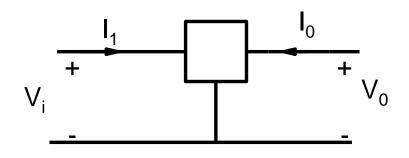
Amplification is the key to noise tolerance during communication



Noise tolerance

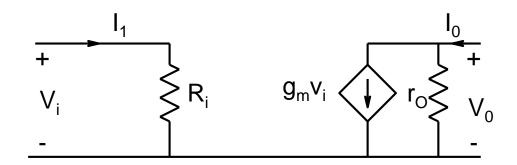


Simple voltage amplifier



$$V_o = G V_i$$
Gain
 $G > 1$

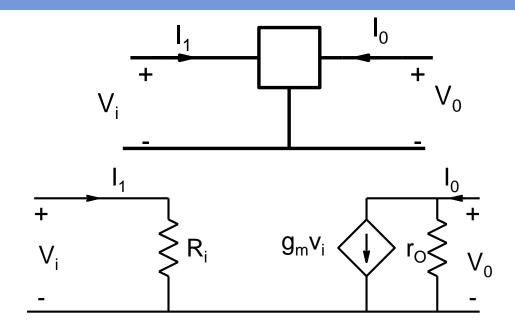
• Equivalent representation



Note: V_i depends on I_1 but does not depend on I_o

 V_o depends on I_1 and I_o

Voltage amplifier parameters



$$I_o = g_m V_i + g_o V_o$$

$$R_i = \frac{V_i}{I_i}$$

Input resistance

$$g_m = \frac{I_o}{V_i} \bigg|_{V_o = 0}$$

Trans-conductance

$$g_o = \frac{1}{r_o} = \frac{I_o}{V_o} \bigg|_{V_i = 0}$$

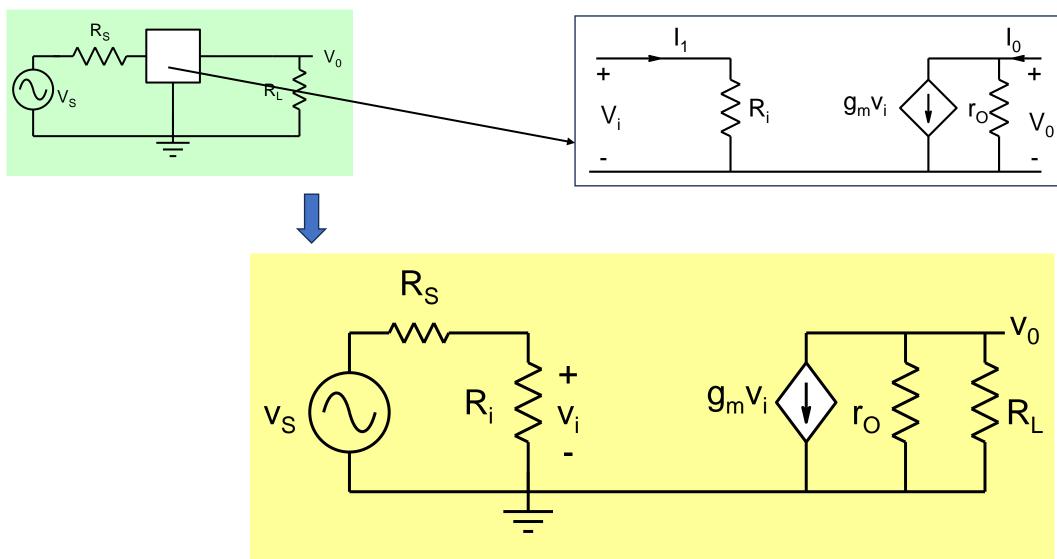
Output conductance

Large

Large

Small

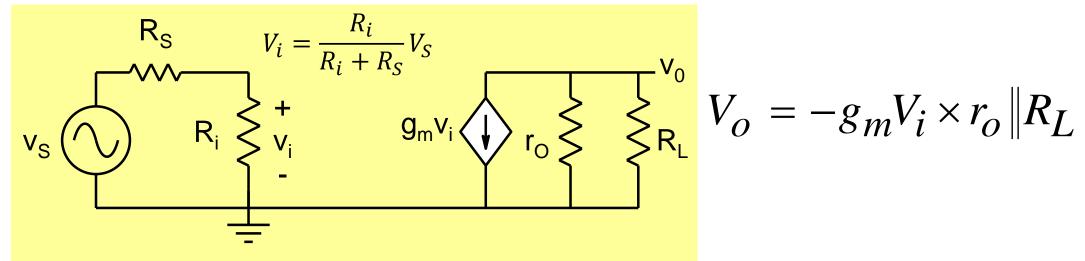
Amplifier circuits



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Voltage gain



$$V_o = -g_m V_i \times r_o \| R_L$$

$$A_V = \frac{V_O}{V_S} = -g_m r_O \times \frac{R_L}{r_O + R_L} \times \frac{R_i}{R_i + R_S}$$

Necessary Condition for Voltage Amplification

$$|A_V| \le g_m \times r_0$$
$$g_m \times r_0 > 1$$

$$R_i$$
 Large

$$g_m$$
 Large

$$r_o$$
 Large

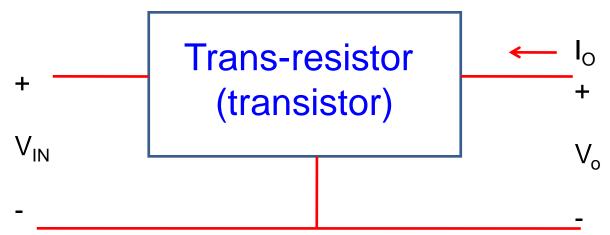
High voltage gain

$$g_m r_o >> 1$$

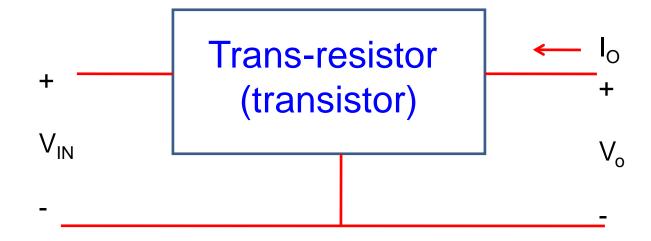
$$g_m >> \frac{1}{r_o} = g_o$$

Trans-conductance >> Output Conductance Trans-resistance << Output resistance

i.e. current I_O is much more sensitive to V_{IN} than V_O



High voltage gain



i.e. current I_O is much more sensitive to V_{IN} than V_O

- Can be used for voltage amplification
- Can be used as a switch
- Implement logic

• . . .