# Robocup: Cooperative estimation and prediction of player and ball movement

Group Project

Department: IfA – Automatic Control Laboratory, ETH Zürich

> Supervising Professor: Prof. Dr. John Lygeros, ETH Zürich

Supervisor: M.S., Ph.D. Student Sean Summer, ETH Zürich

Zürich, March 2012

#### ABSTRACT

### **Abstract**

Hier kommt das Abstract ...

# **Contents**

1.	Introduction	5
2.	Simulation	6
	2.1. The Playing Field	6
	2.2. The Robots	
	2.3. The Ball	
	2.4. A Random Simulation	
3.	Kalman Filtering	8
	3.1. Linear Kalman Filter	8
	3.2. Extended Kalman Filter (EKF)	
4.	Estimation	11
	4.1. Estimate of the Ball	11
	4.2. Estimate of the Robots	
Α.	. Examples	12
	A.1. Kalman Filtering of Linear System	12
	A.2. Kalman Filtering of Nonlinear System	
	Literatur	

# **List of Figures**

2.1.	Playing field after initialization	6
3.1.	Example: Linear electrical circuit	8
3.2.	Example: Input signal, output signal, noisy output, and filtered output	9
3.3.	Example: Input signal, output signal, noisy output, and filtered output	10

# 1. Introduction

### 2. Simulation

One main aspect of this group work among the design of a Kalman filter was the construction of a graphical environment for our work, so we decided to implement a simulation of a Nao soccer match in MATLAB. Therefore we created three independent modules which build the framework for the simulation. The three parts contain functions concerning the playing field, the robots and the ball respectively. Since these parts are constructed in a modular fashion, we can change one module without influencing the other two.

#### 2.1. The Playing Field

All graphical features concerning the playing field are implemented by the function plot\_env(ball).m. As you can see the graphics of the ball are also part of this function. The reason why we didn't need a seperate function for the ball is that the appearance of the ball, contrary to this of the robots, always stays the same. The function plot\_env(ball).m itself is again subdivided in three functions, which draw the field, the ball and, as a neat add on, the scorecounter seperately. We mostly use built-in MATLAB commands for this task such as rectangle() or line(). The following short code excerpt shows for example how the center point of the playing field is drawn

```
% CenterCircle
draw_circle(0,0,Field.centerCircleRadius,'k',0);
```

where draw\_circle(x,y,r,color,filled) is a custom-build function to draw marker circles of the field, but also circles representing the robots and the ball. All functions are designed for fast calculations since we want as many frames as possible if the simulation is running. Figure 3.2 shows the playing field after the execution of plot\_env(ball).m with suitable parameters for ball

Figure 2.1.: Playing field after initialization.

#### 2.2. The Robots

Maybe the most important part of the simulation is the adequate depiction and behaviour of all eight robots. The latter task is split in three components: In a first step the robots are initialized, after that their new positions on the field, according to their motion equations, are computed and in the end we are adding measurement noise for our filtering task. The initialization of the robots is quite simple. The function dummy\_init().m just creates eight structs with the essential informations for every robot, i.e. its horizontal and vertical position, its direction and its team affiliation. Furthermore the function defines the robot's radius and its maximum possible angular change for one timestep as global variables. In a latter stage of development we dropped the simplification of a global eye and assumed that a location of a robot is only possible if it is in the sight of view of at least one robot. So in the second version robot\_init().m of this function, we additionally defined global variables for the robot's velocity, its distance of sight as well as its angle of sight. Once initialized we use the functions dummy\_step (Robot).m and robot\_step (Robot) .m respectively to compute the attributes of all robots for every timestep. The key issue of both functions however is the recalculation of the position, i.e. the motion of the robots

```
for i=1:8
    RobotStep(i).color = Robot(i).color;
    RobotStep(i).x = velocity(i) * cos(Robot(i).dir) + Robot(i).x;
    RobotStep(i).y = velocity(i) * sin(Robot(i).dir) + Robot(i).y;
    RobotStep(i).dir = d_omega(i) + Robot(i).dir;
end
```

#### 2.3. The Ball

#### 2.4. A Random Simulation

# 3. Kalman Filtering

#### 3.1. Linear Kalman Filter

There exists a recursive Kalman filter algorithm for discrete time systems.[1] This ongoing Kalman filter cycle can be divided into two groups of equations

1. Time update equations

$$\hat{x}_{k}^{-} = A\hat{x}_{k-1} + Bu_{k-1}$$

$$P_{k}^{-} = AP_{k-1}A^{T} + Q$$
(3.1)

2. Measurement update equations

$$K_{k} = P_{k}^{-} H^{T} (H P_{k}^{-} H^{T} + R)^{-1}$$

$$\hat{x}_{k} = \hat{x}_{k-1}^{-} + K_{k} (z_{k} - H \hat{x}_{k}^{-})$$

$$P_{k} = (I - K_{k} H) P_{k}^{-}$$
(3.2)

To test this algorithm and to learn something about applying the Kalman filter on linear systems we created an example. A detailed description of the following example can be found in appendix A.1.

We consider a linear, timeinvariant model, given by the following circuit diagram in figure 3.1. First we added process noise and measurement noise to the system output.

```
 \begin{array}{l} \text{(-3,-1)}(9,5) \ \ (\text{-}2,4) \text{A} \ \ (\text{-}2,0) \text{B} \ \ (8,4) \text{C0} \ \ (8,0) \text{D} \ \ (3,4) \text{A2} \ \ (3,2) \text{M} \ \ (3,0) \text{B2} \ \ (6,4) \text{C2} \ \ (6,0) \text{D2} \\ \text{[intensity]abel} = i_L, \text{labeloffset} = -0.2] \text{(A)} \text{(A2)} L \ \ [\text{tension]abel} = U_1, \text{tension]abeloffset} = -1.2, \text{tensionoffset} = -0.8, \text{intensity]abel} = i_1, \text{dipole} = i_1, \text{dipole} = 0.8, \text{labeloffset} = -0.5] \text{(B2)} \text{(M)} R \\ \text{[tension]abel} = U_2, \text{tensionoffset} = -0.8, \text{labeloffset} = -1.2, \text{tensionoffset} = -0.8, \text{intensity]abel} = i_2, \text{dipole} = \text{convention} = \text{generator} \text{(C2)} \text{(D2)} C_2 \ \text{(B)} \text{(D)} \ \text{(A2)} \text{(C0)} \\ \text{[fillstyle} = \text{solid} \text{]} \text{(A)} 0.075 \ \text{[fillstyle} = \text{solid} \text{]} \text{(B)} 0.075 \ \text{[fillstyle} = \text{solid} \text{]} \text{(C0)} \text{(D)} U_{out} \\ \end{array}
```

Figure 3.1.: Example: Linear electrical circuit.

The aim is to get an estimation of the noisy output. Therefore we applied the Kalman

#### 3. Kalman Filtering

filter algorithm based on the equations 3.1. In the figure 3.2, above we can see the input signal and the ideal measurement of the output signal. That ideal measurement includes process noise, which can obviously not be filtered by the Kalman filtering algorithm. Below one can see the noisy measurement on the left side and the filtered output on the right side.

Figure 3.2.: Example: Input signal, output signal, noisy output, and filtered output.

#### 3.2. Extended Kalman Filter (EKF)

In the section above we discussed the usage of a Kalman filter on linear systems. Most systems, including the motion equations of our robots, however are nonlinear. Nevertheless linearizing our system around the current estimate still makes our Kalman filter useful and leads to the concept of the extended Kalman filter [1]. The recursive equations are quite similar to those of the linear Kalman filter

1. Time update equations

$$\hat{x}_{k}^{-} = f(\hat{x}_{k-1}, u_{k-1}, 0)$$

$$P_{k}^{-} = A_{k} P_{k-1} A_{k}^{T} + W_{k} Q_{k-1} W_{k}^{T}$$
(3.3)

2. Measurement update equations

$$K_{k} = P_{k}^{-} H_{k}^{T} (H_{k} P_{k}^{-} H_{k}^{T} + V_{k} R_{k} V_{k}^{T})^{-1}$$

$$\hat{x}_{k} = \hat{x}_{k-1}^{-} + K_{k} (z_{k} - h(\hat{x}_{k}^{-}, 0))$$

$$P_{k} = (I - K_{k} H_{k}) P_{k}^{-}$$
(3.4)

The function  $f(\hat{x}_k, u_k, w_k)$  contains the system's nonlinear dynamics where  $w_k$  represents the current process noise and  $h(\hat{x}_k, v_k)$  denotes the nonlinear state-to-output relationship with the measurement noise  $v_k$ . Furthermore the matrices  $A_k$ ,  $H_k$ ,  $W_k$  and  $V_k$  are linearizations, i.e. partial drivatives, of their respective functions at time step k

#### 3. Kalman Filtering

$$A_{i,j} = \frac{\partial f_i}{\partial x_j} (\hat{x}_{k-1}, u_{k-1}, 0)$$

$$H_{i,j} = \frac{\partial h_i}{\partial x_j} (\hat{x}_{k-1}, u_{k-1}, 0)$$

$$W_{i,j} = \frac{\partial f_i}{\partial w_j} (\tilde{x}_k, 0)$$

$$V_{i,j} = \frac{\partial h_i}{\partial v_j} (\tilde{x}_k, 0)$$
(3.5)

We dropped the fact that all matrices should have a subscript k and that they are allowed be different at each time step. Again, before applying the theory to our simulation, we created a generic example of a nonlinear system. Further details can be looked up in section A.2. We did essentially the same as we did for the linear system: We simulated the ideal system, then added process and measurement noise and in the last step checked whether we could get rid of our artifically added measurement noise. The results for a sample run on MATLAB are shown in figure 3.3 below

Figure 3.3.: Example: Input signal, output signal, noisy output, and filtered output.

The two pictures above show the sinusoidal input on the left and the ideal output, i.e. without process and measurement noise, on the right. Thereunder we can see the plots of the noisy signal and the signal after it has been filtered by the extended Kalman filter. We can conclude that the extended Kalman filter too provides an acceptable performance if our measurement noise is not too big.

# 4. Estimation

- 4.1. Estimate of the Ball
- 4.2. Estimate of the Robots

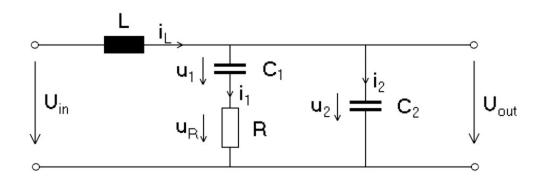
# A. Examples

## A.1. Kalman Filtering of Linear System

### Group project - Linear model

### Daniel Gilgen, David Lehnen, Matthias Roggo, Fabio Marti March 12, 2012

We are considering a linear, timeinvariant model, given by the following circuit diagram



This is obviously a system of order 3 with the states  $x_1 = i_L$ ,  $x_2 = u_1$  and  $x_3 = u_2$ , the input  $u = U_{\text{in}}$  and the output  $y = U_{\text{out}}$ . The system's equations are given by

$$L \cdot \frac{di_L}{dt} = u_L = U_{\text{in}} - u_2$$

$$\implies \frac{di_L}{dt} = \frac{U_{\text{in}}}{L} - \frac{u_2}{L}$$

$$C_1 \cdot \frac{du_1}{dt} = i_1 = u_R \cdot R = (u_2 - u_1) \cdot R$$

$$\implies \frac{du_1}{dt} = \frac{R}{C_1} \cdot u_2 - \frac{R}{C_1} \cdot u_1$$

$$C_2 \cdot \frac{du_2}{dt} = i_2 = i_L - i_1 = i_L - (u_2 - u_1) \cdot R$$

$$\implies \frac{du_2}{dt} = \frac{i_L}{C_2} - \frac{R}{C_2} \cdot u_2 + \frac{R}{C_2} \cdot u_1$$

This leads to the following state space representation

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 0 & -\frac{1}{L} \\ 0 & -\frac{R}{C_1} & \frac{R}{C_1} \\ \frac{1}{C_2} & \frac{R}{C_2} & -\frac{R}{C_2} \end{bmatrix} \cdot \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} \frac{1}{L} \\ 0 \\ 0 \end{bmatrix} \cdot u$$

$$y = \left[ \begin{array}{ccc} 0 & 0 & 1 \end{array} \right] \cdot \left[ \begin{array}{c} x_1 \\ x_2 \\ x_3 \end{array} \right]$$

What we have now is a continuous time state space model of the form

$$\dot{x}(t) = \bar{A}x(t) + \bar{B}u(t)$$
$$y(t) = \bar{C}x(t)$$

Now we can assume, that all electrical devices are not ideal or that there are external disturbances like fluctuations in temperature or air moisture, such that their behaviour is not ideal. These factors will lead to process noise w(t). Furthermore we will measure the output voltage  $U_{\text{out}}$  with a voltmeter, which will not measure the values exactly or which has an inappropriate resolution. This will add some measurement noise v(t) to our model. Our new state space representation will be

$$\dot{x}(t) = \bar{A}x(t) + \bar{B}u(t) + w(t)$$
$$y(t) = \bar{C}x(t) + v(t)$$

with

$$E[w(t)w(\tau)] = Q_e \delta(t - \tau)$$
$$E[v(t)v(\tau)] = R_e \delta(t - \tau)$$

so w(t) and v(t) are white Gaussian noise processes. Since we assume that our model is quite reliable and there are only few external disturbances, the process noise will be much smaller that the measurement noise. Reasonable choices for  $Q_e$  and  $R_e$  could be

$$Q_e = \begin{bmatrix} 10^{-4} & 0 & 0 \\ 0 & 10^{-4} & 0 \\ 0 & 0 & 10^{-4} \end{bmatrix}, \qquad R_e = 10^{-1}$$

The last step is to convert this system into a discrete time linear system. The state space representation for such a system is given by

$$x_{k+1} = Ax_k + Bu_k + w_k$$
$$y_k = Cx_k + v_k$$

The matrices  $Q_e$  and  $R_e$  will stay the same. All other matrices are given by

$$A = e^{\bar{A}T}, \qquad B = \int_0^T e^{\bar{A}(T-\tau)} \bar{B} \, d\tau, \qquad C = \bar{C}$$

where T represents the sampling rate of our discrete time model. By choosing appropriate values for L,  $C_1$ ,  $C_2$  and R, we finally have enough information to build a Kalman filter for this linear system.

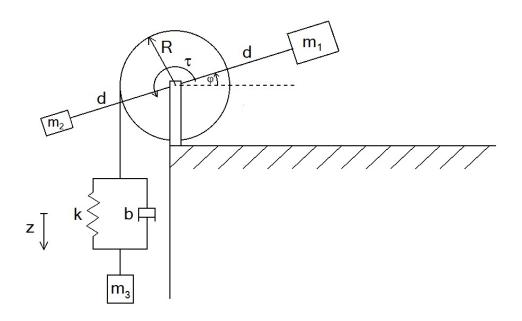
#### A. Examples

## A.2. Kalman Filtering of Nonlinear System

### Group project - Nonlinear model

### Daniel Gilgen, David Lehnen, Matthias Roggo, Fabio Marti April 22, 2012

Consider the following mechanical system



We assume now that the wheel doesn't have a mass, that  $m_1$  has a moment of inertia of  $J_1 = c_1 m_1$  and that  $m_2$  has  $J_2 = c_2 m_2$  with respect to the wheel's hub. Both masses are modelled as dots with distance d to the hub. Furthermore we are assuming that the force of the damper is given by  $F_b = b \cdot \Delta z$ . Using Newton's laws we obtain the equations

$$m_3 \ddot{z} = m_3 g + k(\varphi R - z) + b(\dot{\varphi} R - \dot{z})$$
  
$$(J_1 + J_2) \ddot{\varphi} = \tau + g d \cos(\varphi) (m_2 - m_1) + k(z - \varphi R) + b(\dot{z} - \dot{\varphi} R)$$

By defining the states  $x_1 = z$ ,  $x_2 = \dot{z}$ ,  $x_3 = \varphi$  and  $x_4 = \dot{\varphi}$ , the input  $u = \tau$  and the output  $y = \dot{\varphi}$  we find the following differential equations

$$\begin{array}{rcl} \dot{x}_1 & = & x_2 \\ \\ \dot{x}_2 & = & g + \frac{k}{m_3}(x_3R - x_1) + \frac{b}{m_3}(x_4R - x_2) \\ \\ \dot{x}_3 & = & x_4 \\ \\ \dot{x}_4 & = & \frac{1}{c_1m_1 + c_2m_2}\left(u + gd\cos(x_3)(m_2 - m_1) + k(x_1 - x_3R) + b(x_2 - x_4R)\right) \end{array}$$

$$y = x_4$$

In a next step we transform our continuous time system into a discrete time system. We use a simple Euler step for this task

$$\dot{x} \approx \frac{x_{k+1} - x_k}{T}$$

where T is the sampling period. The discrete system's equations now look as follows

$$\begin{array}{lll} x_{1,k+1} & = & x_{1,k} + T \cdot x_{2,k} \\ \\ x_{2,k+1} & = & x_{2,k} + T \cdot \left(g + \frac{k}{m_3}(x_{3,k}R - x_{1,k}) + \frac{b}{m_3}(x_{4,k}R - x_{2,k})\right) \\ \\ x_{3,k+1} & = & x_{3,k} + T \cdot x_{4,k} \\ \\ x_{4,k+1} & = & x_{4,k} + T \cdot \left(\frac{1}{c_1m_1 + c_2m_2}(u_k + gd\cos(x_{3,k})(m_2 - m_1) + k(x_{1,k} - x_{3,k}R) + b(x_{2,k} - x_{4,k}R))\right) \\ \\ y_k & = & x_{4,k} \end{array}$$

Since we are dealing with nonideal effects, caused by model uncertainties or measurement errors, we add process noise  $w_k$  and measurement noise  $v_k$  to our model. We will get equations of the form

$$x_{k+1} = f(x_k, u_k, w_k)$$
$$y_k = h(x_k, v_k)$$

Note that  $x_{k+1}$  as well as  $w_k$  and  $v_k$  are column vectors. In our nonlinear model,  $w_k$  and  $v_k$  are realizations of white Gaussian noise processes

$$E[w_k w_n] = W_k Q_k W_k^T \delta[k-n]$$

$$E[v_k v_n] = V_k R_k V_k^T \delta[k-n]$$

where  $W_k = \frac{\partial f}{\partial w_k}(x_{k-1}, u_{k-1}, 0)$  and  $V_k = \frac{\partial h}{\partial v_k}(x_k, 0)$ . Note that  $Q_k$  and  $R_k$  could change every timestep k. Possible values for them are

$$Q_k = \begin{bmatrix} 10^{-6} & 0 & 0 & 0\\ 0 & 10^{-6} & 0 & 0\\ 0 & 0 & 10^{-6} & 0\\ 0 & 0 & 0 & 10^{-6} \end{bmatrix}, \qquad R_k = 10^{-2}$$

We have a full description of our nonlinear system and are now able to implement an extended Kalman filter by choosing reasonable values for the system's parameters.

# **Bibliography**

- [1] G. Welch and G. Bishop, "An Introduction to the Kalman Filter," UNC-Chapel Hill, 2006.
- [2] RoboCup Technical Committee, "RoboCup Standard Platform League (Nao) Rule Book," RoboCup, 2011.