| Zhou Yi, Signals and Systems Have Fun! | Properties | $\mathcal{F}\{x_{1}(t) * x_{2}(t)\} = X_{1}(j\omega)X_{2}(j\omega),$ $\mathcal{F}\{x_{1}(t)x_{2}(t)\} = \frac{1}{2\pi}X_{1}(j\omega) * X_{2}(j\omega)$ | 2.5 Sampling |
|---|--|---|--|
| 1 LTI | $\mathcal{FS}\{x(t)\} = c_k, \mathcal{FS}\{x(t-t_0)\} = c_k e^{-jk\omega_0 t_0}$ $\mathcal{FS}\{x(t)\} = c_k, \mathcal{FS}\{x^*(t)\} = c_{-k^*}$ | 2.3 DT FT $X(e^{j\omega}) = \sum_{n} x[n]e^{-j\omega n} = X(e^{j(\omega+2\pi)})$ (*) | Impulse |
| CT DE $d^{k}_{v(t)} \qquad M \qquad d^{k}_{x(t)}$ | $x(t) = \sum \frac{c_k + c_{-k}}{2} e^{jk\omega_0 t} + j \sum \frac{c_k - c_{-k}}{2j} e^{jk\omega_0 t}$ $TS(y(t))(t) = \sum c_j d_j - c_j d_j$ | $x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$ | |
| $\sum_{k=0}^{N} a_k \frac{d^k y(t)}{dt^k} = \sum_{k=0}^{M} b_k \frac{d^k x(t)}{dt^k}$ | $\mathcal{FS}\{x(t)y(t)\} = \sum_{i} c_{i} d_{k-i} = c_{k} * d_{k}$ $\mathcal{FS}\{x'(t)\} = jk\omega_{0}c_{k}$ | 21. 321. | $p(t) = \sum_{n=-\infty}^{+\infty} \delta(t - nT)$ |
| $\sum_{k=0}^{N} a_k x^k = a_N \prod_{i=1}^{p} (x - x_i)^{n_i}$ | $\mathcal{FS}\{y(t)\} = \int_{0}^{t} x(\tau) d\tau, (0 < t < T), y(t) = y(t + T)$ | Examples | $P(j\omega) = \frac{2\pi}{T} \sum_{k=-\infty}^{+\infty} \delta(\omega - k\omega_s), \omega_s = \frac{2\pi}{T}$ |
| $y_h(t) = \sum_{i=1}^{p} (\sum_{j=0}^{n_i-1} A_{i,j} t^j) e^{x_i t}$ | $y(t) = \sum_{k \neq 0} \frac{c_k}{jk\omega_0} (e^{jk\omega_0 t} - 1) + c_0 t$ | - | $x_p(t) = x(t)p(t), X_p(j\omega) = \frac{1}{T}\sum_k X(j(\omega - k\omega_s))$ (copy and paste) |
| If $x(t) = \sum_{k=0}^{m} C_k t^k e^{x_i t}$ | 2.1 DT FS | $\mathcal{F}\{[n \le N_1]\} = \frac{\sin \omega (N_1 + \frac{1}{2})}{\sin(\frac{\omega}{2})}$ | If $X(j\omega) = 0(\omega > \omega_M)$, No overlap in freq domain if |
| $y_p(t) = (\sum_{k=0}^m B_k t^k) t^{n_i} e^{x_i t}, n_i \ge 0$ Determine constants. | x[n] = x[n+N] $x[n] = \sum_{k=0}^{N-1} c_k e^{\frac{j2\pi kn}{N}}, 0 \le n < N$ | $\mathcal{F}\left\{\frac{\sin(Wn)}{\pi n}\right\} = [u(\omega + W) - u(\omega - W)](\text{Periodic}, T = 2\pi)$ | $\omega_s > 2\omega_M$ lowpass filter $\omega_c \in (\omega_M, \omega_s - \omega_M)$ to restore signal |
| DT DE | $x[n] = \sum_{k=0}^{N} c_k e^{-N} , 0 \le n < N$ $i2\pi kl$ | $x[n] = \begin{cases} 1, n \ge 0 \\ -1, n < 0 \end{cases}$ | |
| $\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k]$ $\sum_{k=0}^{N} a_{N-k} x^k = a_0 \prod_{i=1}^{p} (x-x_i)^{n_i}$ | $c_k = \frac{1}{N} \sum_{l=0}^{N-1} x[l] e^{-\frac{j2\pi kl}{N}} = c_{k+N}$ | (| |
| $\sum_{k=0}^{N} a_{N-k} x^{k} = a_{0} \prod_{i=1}^{P} (x - x_{i})^{n_{i}}$ | P 1 | $\mathcal{F}\{x[n]\} = \frac{2}{1 - e^{-j\omega}}$ | |
| $y_h[n] = \sum_{i=1}^{p} (\sum_{j=0}^{n_i-1} A_{i,j} n^j) x_i^n$ | Examples | $\mathcal{F}\{u[n]\} = \frac{1}{1 - e^{-j\omega}} + \pi \sum_{k = -\infty}^{+\infty} \delta(\omega - 2k\pi)$ | |
| if $x[n] = \sum_{i=0}^{m} C_i n^i s^n$ | $x[n] = \begin{cases} 1, n <= N_1 \\ 0, N_1 < n <= \frac{N}{2} \end{cases} ,$ | $\mathcal{F}\{e^{j\omega_0 n}\} = \sum_{n=-\infty}^{+\infty} e^{-j(\omega-\omega_0)n} = 2\pi \sum_{n=-\infty}^{+\infty} \delta(\omega-\omega_0 - 2k\pi)$ | Zero-Order Holding |
| $y_p[n] = (\sum_{k=0}^m B_k n^k) n^{n_i} x_i^n, n_i \ge 0$ Special: $x_i = 1$ | $ (0, N_1 < n < \frac{\pi}{2} $ $ (\sin[(2\pi k/N)(N_1 + \frac{1}{\pi})] $ | $\mathcal{F}\{\binom{n+r-1}{n}a^nu[n]\} = (1 - ae^{-j\omega})^{-r}, a < 1$ | G |
| Unit Impulse Resp | $\mathcal{FS}\{x[n]\} = \begin{cases} \frac{\sin[(2\pi k/N)(N_1 + \frac{1}{2})]}{N\sin(2\pi k/2N)}, k \neq tN \\ \frac{2N_1 + 1}{N}, k = tN \end{cases}$ | D ' 1' C' 1 | |
| $h_p(t) = 0 		 (b - 1) 	 (b - 2)$ | | Periodic Signals | $\begin{array}{l} x_0(t) = \sum_{k=-\infty}^{+\infty} x(kT) p_0(t-kT) = x(t) p(t) * p_0(t) \\ p_0(t) = u(t) - u(t-T) \end{array}$ |
| Matching $\delta^{(k)}(t), \delta^{(k-1)}(t), \delta^{(k-2)}(t)$ (Only singularity part) | $\mathcal{FS}\{e^{j\omega_0 n}\} = \begin{cases} 1, k = m + tN \\ 0, otherwise \end{cases}, \omega_0 = \frac{2\pi m}{N}$ | $x[n] = x[n+N]$, Fourier Series $x[n] = \sum_{k=0}^{N-1} c_k e^{\frac{j2\pi kn}{N}}$ | $P(i\omega) = Te^{-\frac{j\omega T}{2}} Sa(\frac{\omega T}{2})$ |
| \int_{0-}^{0+} to determine coefficient | 2.2 CT FT $X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt$ | $c_k = \frac{1}{N} \sum x[l] e^{\frac{-j2\pi kl}{N}} = c_{k+N}$ | $X_0(j\omega) = \sum_{k=-\infty}^{+\infty} e^{-\frac{j\omega T}{2}} Sa(\frac{\omega T}{2}) X(j(\omega - k\omega_s))$ |
| $y(t) = \int_{\tau} x(\tau)h(t-\tau)$ | $x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$ | $\mathcal{F}\{x(t)\} = \sum_{k=0}^{N-1} c_k \mathcal{F}\{e^{\frac{j2\pi kn}{N}}\} = \sum_{k=-\infty}^{+\infty} 2\pi c_k \delta(\omega - \frac{2k\pi}{N})$ | $X_{0}(j\omega) = \sum_{k=-\infty}^{+\infty} e^{-\frac{j\omega T}{2}} Sa(\frac{\omega T}{2}) X(j(\omega - k\omega_{s}))$ $H_{r}(j\omega) = \frac{e^{\frac{j\omega T}{2}}}{Sa(\frac{\omega T}{2})}, \omega < \omega_{c} $ |
| 2 Fourier Analysis CT FS | (γ 2π J−∞ | K-0 - K- W - 11 | $H_r(j\omega) = \frac{1}{Sa(\frac{\omega T}{2})}, \omega < \omega_c $ |
| $x(t) = x(t+T), \omega_0 = \frac{2\pi}{T}$ | Examples | CT FS and DT FT | |
| $c_k = \frac{1}{T} \int_T x(t) e^{-jk\omega_0 t} dt$ | $\mathcal{F}\{x(t) = a[u(t + \frac{\tau}{2}) - u(t - \frac{\tau}{2})]\} = a\tau Sa(\frac{\omega\tau}{2})$ | $X(e^{j\omega}) = \sum_{n} x[n]e^{-j\omega n} = X(e^{j(\omega + 2\pi)})$ | |
| $\tilde{x}(t) = \sum_{k=-\infty}^{+\infty} c_k e^{jk\omega_0 t} = \begin{cases} x(t) \\ \frac{x(t+)+x(t-)}{2} \end{cases}$ | $\mathcal{F}\{x(t) = e^{-\alpha t} u(t), (\alpha > 0)\} = \frac{1}{\alpha + i\omega}$ | $x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$ | |
| $\sum_{k=-\infty}^{\infty} c_k c_k c_k = -\left(\frac{x(t+)+x(t-)}{2}\right)$ | $\mathcal{F}\{\delta(t)\}=1, \mathcal{F}\{1\}=2\pi\delta(\omega)$ | $c_k = \frac{1}{T} \int_T x(t) e^{-jk\omega_0 t} dt$ | |
| | $\mathcal{F}\{e^{j\omega_0t}\} = 2\pi\delta(\omega - \omega_0)$ | $x(t) = \sum_{k=-\infty}^{+\infty} c_k e^{jk\omega_0 t}$ | Linear Interpolation |
| Real Signals | $\mathcal{F}\{\operatorname{sgn}(t)\} = \frac{2}{j\omega}$ | $x[n] = c_{-n} \to x(t) = X(e^{j\omega_0 t})$ | 1 |
| When $c_k *= c_{-k}$ | Periodic Signals | D (* | M |
| $x(t) = \frac{a_0}{2} + \sum_{k=1}^{\infty} (a_k \cos(k\omega_0 t) + b_k \sin(k\omega_0 t))$ | | Properties | $x_1(t) = \sum_{n=-\infty}^{+\infty} x(kT)p_1(t-kT), p_1(t) = (1-\frac{ t }{T})[u(t+T) - u(t-T)]$ |
| $\sum_{k=1}^{\infty}$ | $x(t) = \sum_{k=-\infty}^{+\infty} c_k e^{jk\omega_0 t}$ | $x_k[n] = \begin{cases} x[n/k], n/k \in \mathbb{Z} \\ 0, otherwise \end{cases}$ | $u(t-T)]$ $P_1(j\omega) = TSa^2(\frac{\omega T}{2})$ |
| $=A_0 + \sum_{k=1}^{\infty} A_k \cos(k\omega_0 t + \phi_k)$ | $\mathcal{F}\{x(t)\} = \int_{-\infty}^{+\infty} \sum_{k=-\infty}^{+\infty} c_k e^{jk\omega_0 t} e^{-j\omega t} dt$ | $\mathcal{F}\{x_k[n]\} = X(e^{jk\omega})$ | $X_1(j\omega) = \sum_{k=-\infty}^{+\infty} Sa^2(\frac{\omega T}{2})X(j(\omega - k\omega_s))$ |
| | $=2\pi\sum_{k=-\infty}^{+\infty}c_k\delta(\omega-k\omega_0)$ | $\mathcal{F}\{x_{k}[n]\} = X(e^{j\omega})$ $\mathcal{F}\{x[n] - x[n-1]\} = (1 - e^{-j\omega})X(e^{jk\omega})$ | $H_r(j\omega) = Sa^{-2}(\frac{\omega T}{2}), \omega < \omega_c $ |
| $ c_k ^2 = c_{-k} ^2 = \frac{a_k^2 + b_k^2}{4}$ | Parseval Theorem | $\mathcal{F}\{nx[n]\} = j\frac{\mathrm{d}X(e^{j\omega})}{\mathrm{d}\omega}$ | |
| _ | $\int_{-\infty}^{+\infty} x(t) ^2 dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) ^2 d\omega$ | $\mathcal{F}\{\sum_{k\leq n} x[k]\} = \frac{X(e^{j\omega})}{(1-e^{-j\omega})} + \pi X(e^{j0}) \sum \delta(\omega - 2k\pi)$ | |
| Examples | | $\mathcal{F}\{x[n]y[n]\} = \frac{1}{2\pi} \int_{2\pi} X(e^{j\theta}) Y(e^{j(\omega-\theta)}) d\theta$ | |
| $x(t) = a\left[u(t + \frac{\tau}{2}) - u(t - \frac{\tau}{2})\right]$ | Properties | 2.4 DFT | |
| $c_0 = a \frac{\tau}{T}$ | $\mathcal{F}\{x(t)\} = X(j\omega), \mathcal{F}\{X(jt)\} = 2\pi x(j^2\omega)$ | $X[k] = X(e^{\frac{j2\pi k}{N}})$ (sampling DT FT) | Processing |
| $c_k = \frac{a\tau}{T} Sa(\frac{k\omega_0 \tau}{2})$ | $\mathcal{F}\{x(t-\tau)\} = X(j\omega)e^{-j\omega\tau}$ | Distortionless | |
| $Sa(t) = \begin{cases} \frac{\sin(t)}{t}, t \neq 0\\ 1, t = 0 \end{cases}$ | $\mathcal{F}\{x(\alpha t)\} = \frac{1}{ \alpha } X(\frac{j\omega}{\alpha})$ | $Y(j\omega) = X(j\omega)H(j\omega) = AX(j\omega)e^{-j\omega\tau}$ $Y(e^{j\omega}) = X(e^{j\omega})H(e^{j\omega}) = AX(e^{j\omega})e^{-j\omega n_0}$ | $y(t) = x(t) * h(t), y_d[n] = x_d[n] * h_d[n], x_d[n] = x(nT), y_d[n] = x_d[n] * h_d[n], y_d[n] * h_d[n], y_d[n] = x_d[n] * h_d[n], y_d[n] * h$ |
| (1,t=0 | $\mathcal{F}\{x^*(t)\} = \dot{X}^*(-j\omega)$ $\mathcal{F}\{x'(t)\} = j\omega X(j\omega), \mathcal{F}\{tx(t)\} = j\frac{\mathrm{d}X(j\omega)}{\mathrm{d}\omega}$ | I(e,) - V(e,) II(e,) - VV(e,)e, | y(nT) |
| Dawaayal Thaawam | | Group Delay | $Y_d(e^{j\Omega}) = X_d(e^{j\Omega})H_d(e^{j\Omega})$ (also band limited) |
| Parseval Theorem | $\mathcal{F}\{\int_{-\infty}^{t} x(\tau) d\tau\} = \frac{X(j\omega)}{j\omega} + \pi X(0)\delta(\omega)$ | • • | $X_d(e^{j\Omega}) = X_p(\frac{j\Omega}{T}) = \frac{1}{T} \sum_{k=-\infty}^{+\infty} X(j\frac{j(\Omega - 2k\pi)}{T})$ |
| $\frac{1}{T} \int_{T} x(t) ^{2} dt = \sum_{k} c_{k} ^{2} = A_{0}^{2} + \frac{1}{2} \sum_{k \ge 1} A_{k}^{2}$ | $\mathcal{F}\left\{-\frac{x(t)}{it} + \pi x(0)\delta(t)\right\} = \int_{-\infty}^{\omega} X(j\eta) d\eta$ | $\tau(\omega) = -\frac{\mathrm{d}}{\mathrm{d}\omega}\phi(H(j\omega))$ | $H_d(e^{j\Omega}) = H(\frac{j\Omega}{T}), -\pi \le \Omega < \pi$ |