

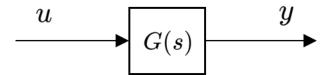
# **Output Feedback**

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## Plant G (or System)

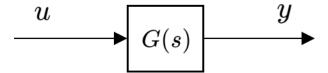
- Consider the plant *G* 
  - Input u(t)
  - Output y(t)
- So far, we have learnt about dynamics of plant *G*

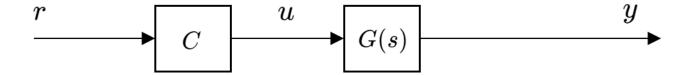


- Control problem
  - The output of this plant track a desired reference trajectory r(t)

## **Open Loop Control**

- The simplest solution to the tracking problem is to use a pre-compensator G
  - $-C = \frac{1}{G}$
  - Then y(t) = r(t)

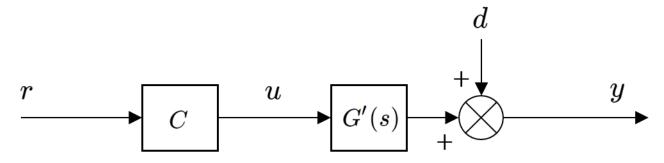




- Do you see any problems of this solution?
  - Model uncertainty
  - Disturbance

# **Open Loop Control**

- Consider the important practical issue of model uncertainty and disturbance.
  - It is always the case that the true system we wish to control will deviate from the nominal model used in control design
- Suppose uncertain plant G' and disturbance d(t)

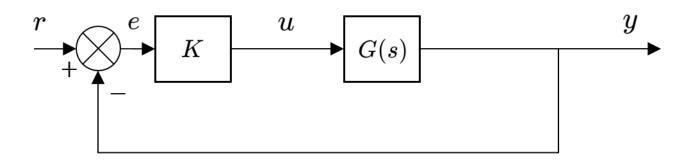


• Open loop controller  $C = \frac{1}{G}$ 

$$y(t) = rac{G'}{G} r(t) + d(t)$$

# **Closed Loop Control (= Negative Feedback Control)**

- An alternative solution is to purchase a sensor and use feedback control
  - Use a constant gain compensator K, that multiplies the measured tracking error e(t) = r(t) y(t)



$$u = Ke$$

$$e = r - y$$

$$y = Gu$$

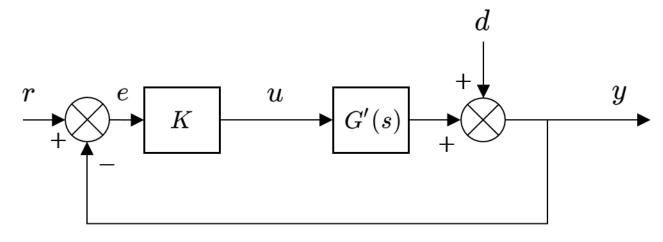
$$e = rac{1}{1 + GK} r$$

$$y = rac{GK}{1+GK} r$$

As 
$$|GK| o \infty, \; e(t) o 0$$
 and  $y(t) o r(t)$ 

## **Closed Loop Control**

- Feedback controller
  - Use of feedback with sufficiently high gain provides an approximate solution to the tracking problem even in the presence of system uncertainty and disturbance

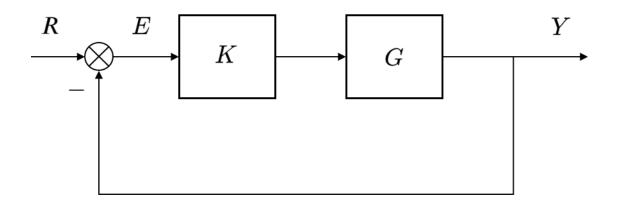


$$e = \frac{1}{1 + G'K}r - \frac{1}{1 + G'K}d$$

$$y = \frac{G'K}{1 + G'K}r + \frac{1}{1 + G'K}d$$

### **Transfer Function for Closed Loop System**

- Feedback changes the system transfer function
  - Change system dynamics
  - Change poles and zeros
  - Might change system stability



$$E = R - Y$$
 $Y = KGE$ 

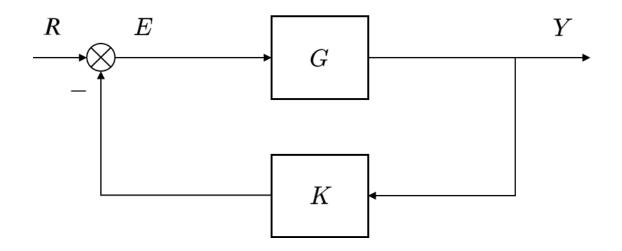
$$Y = KG(R - Y)$$
 $= KGR - KGY$ 

$$(1 + KG)Y = KGR$$

$$H = \frac{Y}{R} = \frac{KG}{1 + KG}$$

### **Transfer Function for Closed Loop System**

- Feedback changes the system transfer function
  - Change system dynamics
  - Change poles and zeros
  - Might change system stability



$$E = R - KY$$
 $Y = GE$ 

$$Y = G(R - KY)$$
 $= GR - KGY$ 

$$(1 + KG)Y = GR$$

$$H = \frac{Y}{R} = \frac{G}{1 + KG}$$

# **Transfer Function for Closed Loop System**

Example 1

$$G(s) = \frac{1}{s+1}$$
, pole at  $-1$ 

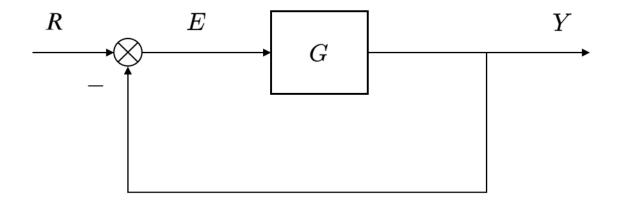
$$H(s) = rac{KG}{1 + KG} = rac{rac{K}{s+1}}{1 + rac{K}{s+1}} = rac{K}{s+1 + K}, ext{ new pole at } -(1 + K)$$

Example 2

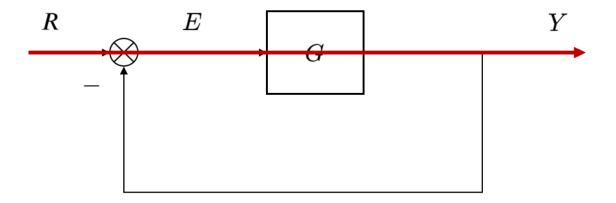
$$G(s) = rac{1}{s-1}, \quad ext{pole at} + 1, ext{unstable}$$

$$H(s)=rac{KG}{1+KG}=rac{rac{K}{s-1}}{1+rac{K}{s-1}}=rac{K}{s-1+K}, \quad ext{new pole at } (1-K)$$

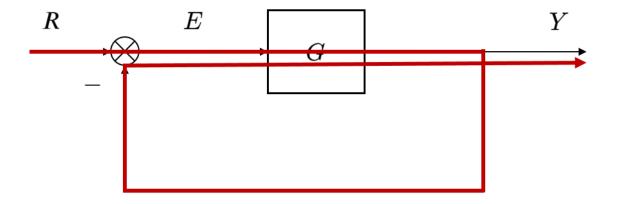
if k>1, the closed loop system H(s) becomes stable



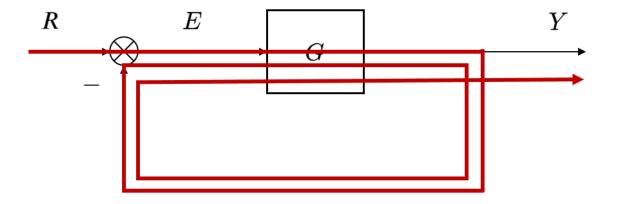






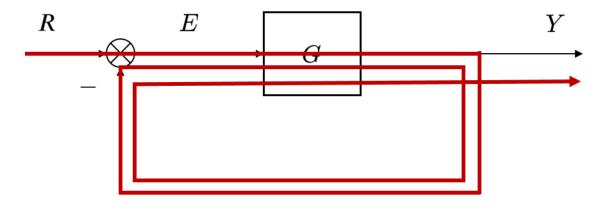








• Feedback = autoregressive = infinite length response



• If 
$$G = \frac{1}{s}$$

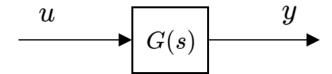
$$H = rac{Y}{R} = rac{1}{1+s} = 1 - s + s^2 - s^3 + \cdots$$

# Open Loop vs. Closed Loop

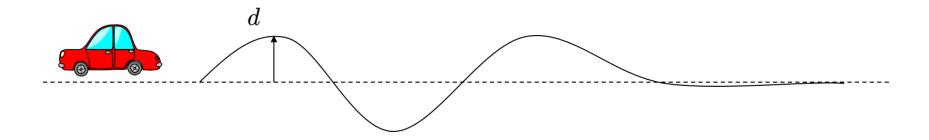


## Open Loop vs. Closed Loop

- Suppose that there is a car
  - input: force u(t)
  - output: velocity y(t)
  - transfer function G(s)

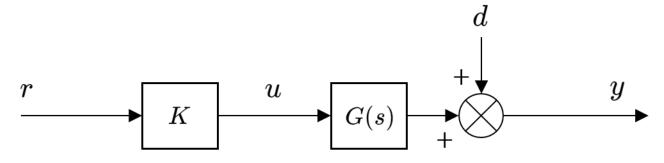


ullet Suppose that there is a car driving with a road wave disturbance d



#### **Open Loop**

- Reference velocity r(t) that is desired by a driver
- Assume the force produced by a engine is u = Kr



• The system input  $\boldsymbol{u}$ 

u = Kr

Calculating y

y = Gu + d = GKr + d

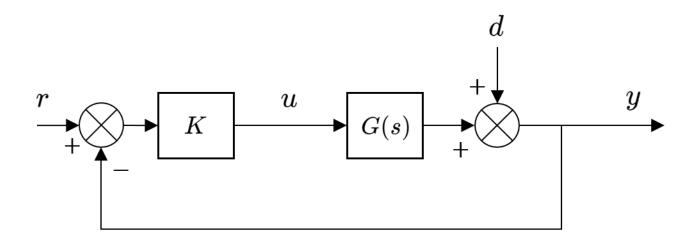
• Error

$$e = r - y = r - GKr - d$$

$$= r(1 - GK) - d$$

## **Closed Loop**

- Think about how we drive
  - We step on the gas or put the brake on based on the desired speed and the current one
  - We care about the difference r-y
  - The term of "Negative feedback" is coming from -y





# **Closed Loop**

- Now assume *K* is a controller
- The system input u

$$u = K(r - y)$$

Calculating y

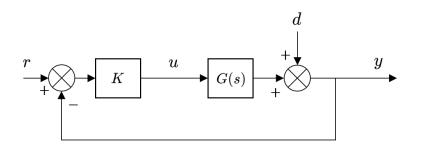
$$y = GK(r - y) + d$$
$$= KGr - KGy + d$$

$$(1 + KG)y = KGr + d$$

$$\therefore y = \frac{KGr + d}{1 + KG} = \frac{KG}{1 + KG}r + \frac{1}{1 + KG}d$$

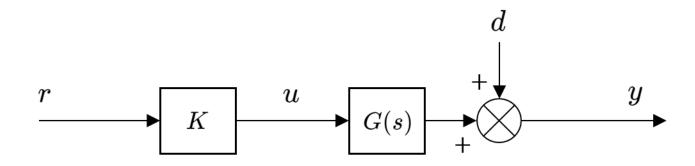
- Error
  - When K is large, the error becomes small

$$e = r - y = r - \frac{KGr + d}{1 + KG} = \frac{r - d}{1 + KG}$$



## **Model Uncertainty of Open Loop**

- Let us suppose that the predicted model is G(s) = 2, and actually G(s) = 1
- Let's design the K value when the desired output speed is  $100 \ km/h$

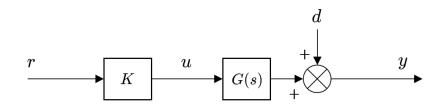


In the open loop model, y is

$$y_{\text{model}} = G_{\text{model}}Kr + d = 2Kr + d$$

# **Model Uncertainty of Open Loop**

• In the open loop model, y is



$$y_{\text{model}} = G_{\text{model}}Kr + d = 2Kr + d$$

• In order for y to approach 100 reference input

$$K = 0.5$$

Then the actual y is

$$y_{\text{true}} = G_{\text{true}}Kr + d = Kr + d$$

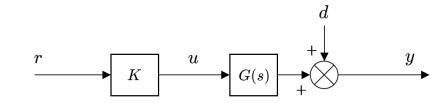
The model and true errors are

$$e_{\text{model}} = r - y_{\text{model}} = r(1 - G_{\text{model}}K) - d = r(1 - 2K) - d$$

$$e_{\text{true}} = r - y_{\text{true}} = r(1 - G_{\text{true}}K) - d = r(1 - K) - d$$

## **Model Uncertainty of Open Loop**

Discrepancy from model uncertainty

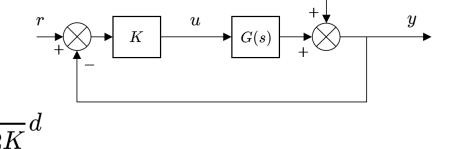


$$y_{\text{model}} - y_{\text{true}} = Kr$$

- Stability
  - The open loop system cannot change the poles of system
- Uncertainty, Low Robustness
  - Predicting models incorrectly has a critical impact on speed
- Disturbance rejection
  - Disturbance directly affects the system

# **Model Uncertainty of Closed Loop**

• In the closed loop model, y is



$$y_{
m model} = rac{KG_{
m model}r + d}{1 + KG_{
m model}} = rac{2Kr + d}{1 + 2K} = rac{2K}{1 + 2K}r + rac{1}{1 + 2K}d$$

In order for y to approach 100 reference input, the larger K is better

$$K = 100$$

Then the actual y is

$$y_{ ext{true}} = rac{KG_{ ext{true}}r + d}{1 + KG_{ ext{true}}} = rac{Kr + d}{1 + K} = rac{K}{1 + K}r + rac{1}{1 + K}d$$

• The model and true errors are

$$e_{\text{model}} = r - y_{\text{model}} = \frac{r - d}{1 + KG_{\text{model}}} = \frac{r - d}{1 + 2K} = \frac{1}{1 + 2K}r - \frac{1}{1 + 2K}d$$

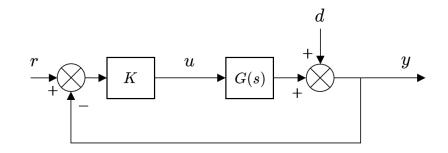
$$e_{\text{true}} = r - y_{\text{true}} = \frac{r - d}{1 + KG_{\text{true}}} = \frac{r - d}{1 + K} = \frac{1}{1 + K}r - \frac{1}{1 + K}d$$

## **Model Uncertainty of Closed Loop**

• Discrepancy from model uncertainty (assume d=0) is

$$y_{
m model} - y_{
m true} = rac{K}{2K^2 + 3K + 1} r ~pprox ~0$$

- Stability
  - The closed loop system can change the poles of system
- Uncertainty, Robustness
  - Model uncertainty has a reduced impact on speed
- Disturbance rejection
  - Disturbance little affects the system

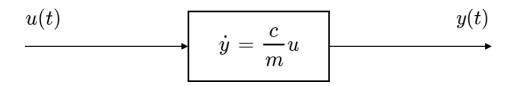


# **PID Control**

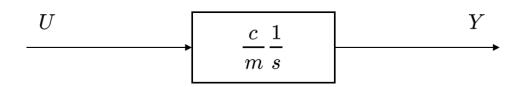


#### **Car Model**

- For the car model
  - velocity y
  - input force u
- In a block diagram



• In a Laplace transform



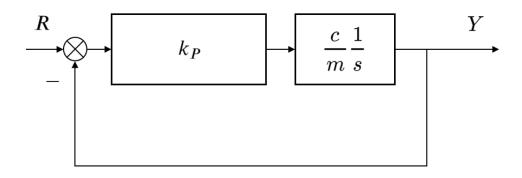
• We want to achieve

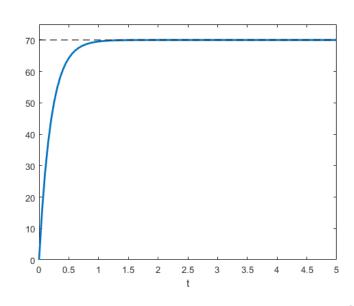
$$y 
ightarrow r$$
 as  $t 
ightarrow \infty \; (e = r - y 
ightarrow 0)$ 

#### **P Control**

- The proportional term produces an output value that is proportional to the current error value.
  - Small error yields small control signals
  - Nice and smooth
  - So-called proportional regulation (P regulator)

$$u = k_P e$$



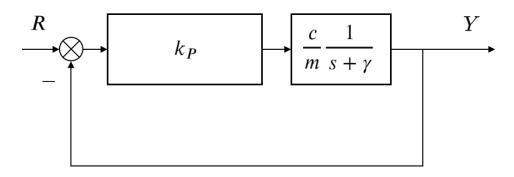


## **Model Uncertainty**

• Caveat: the "real" model is augmented to include a wind resistance term:

$$\dot{y} = \frac{c}{m}u - \gamma y$$

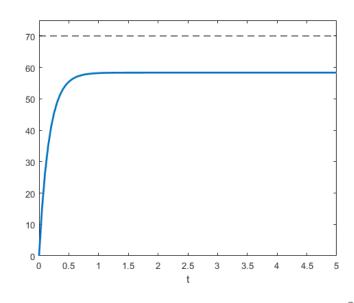
$$u = k_P e = k_P (r - y)$$



At steady-state

$$\dot{y} = 0 = \frac{c}{m}u - \gamma y = \frac{c}{m}k_P(r-y) - \gamma y$$

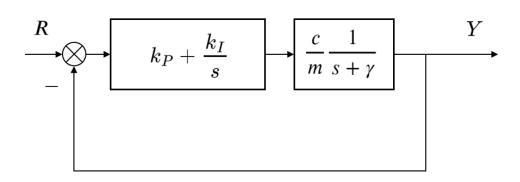
$$\implies y = \frac{ck_P}{ck_P + m\gamma}r$$

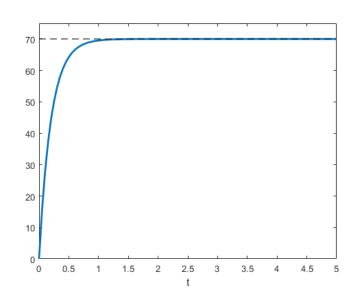


#### **PI Control**

- The integral controller is the sum of the instantaneous error over time and gives the accumulated offset that should have been corrected previously
  - Stability
  - Tracking
  - Robustness

$$u(t) = k_P \, e(t) + k_I \int_0^t e( au) d au$$

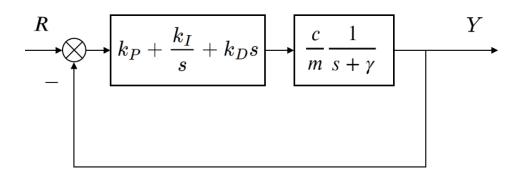




#### **PID Control**

- PID: by far the most used low-level controller
  - P: contributes to stability, medium-rate responsiveness
  - I: tracking and disturbance rejection, slow-rate responsive, may cause oscillations
  - D: fast-rate responsiveness, sensitive to noise

$$u(t) = k_P \, e(t) + k_I \int_0^t e( au) d au + k_D rac{de(t)}{dt}$$

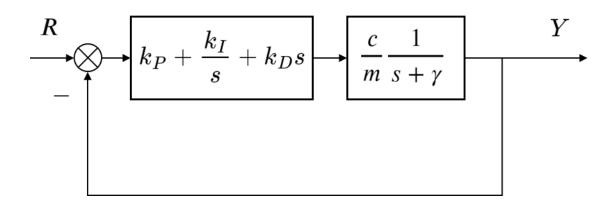


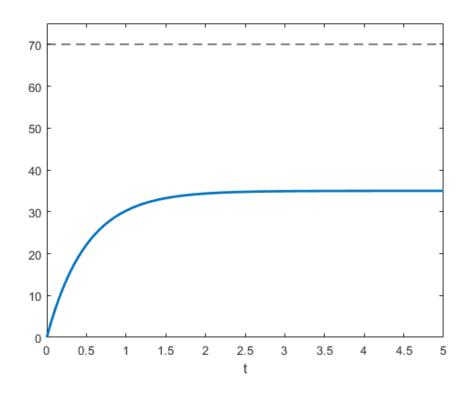
• Feedback has a remarkable ability to fight uncertainty in model parameters!

#### **PID Control**

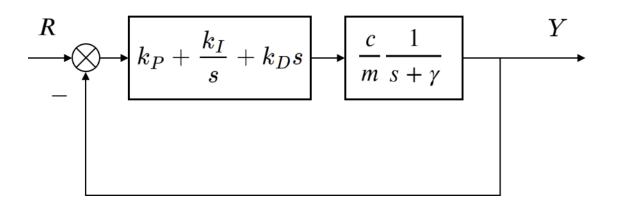
- Feedback has a remarkable ability to fight uncertainty in model parameters!
- The goal of this problem is to show how each of the term,  $k_P$ ,  $k_I$  and  $k_D$  contributes to obtaining the common goals of:
  - Fast rise time
  - Minimal overshot
  - Zero steady-state error

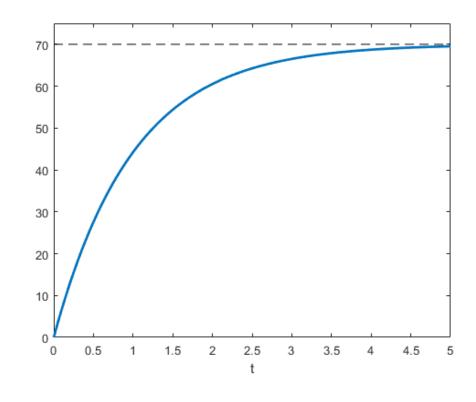
$$k_P = 1, k_I = 0, k_D = 0$$



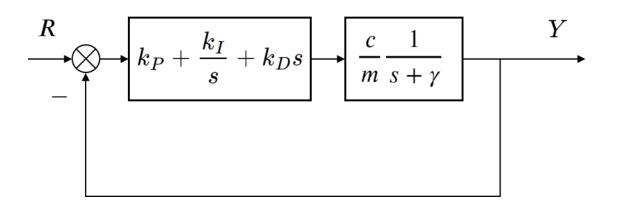


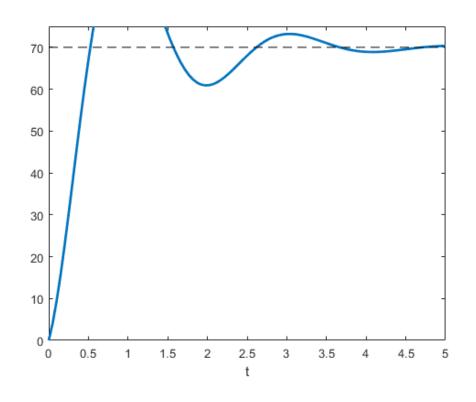
$$k_P = 1, k_I = 1, k_D = 0$$



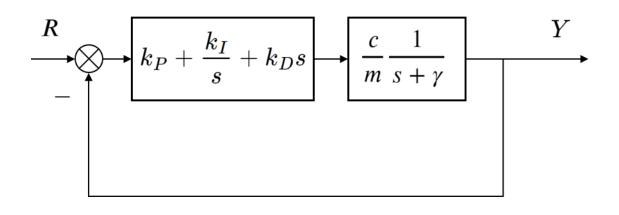


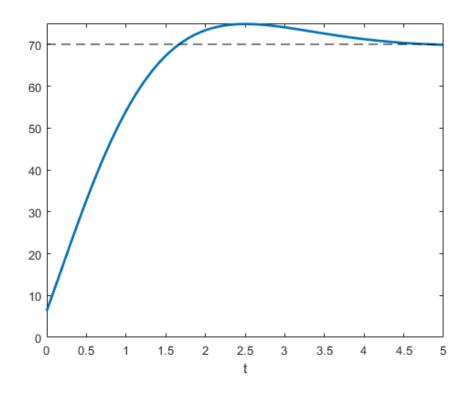
$$k_P = 1, k_I = 10, k_D = 0$$





$$k_P = 1, k_I = 2, k_D = 0.1$$





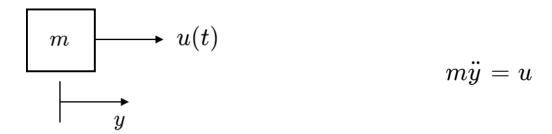
### **General Tips for Designing a PID Controller**

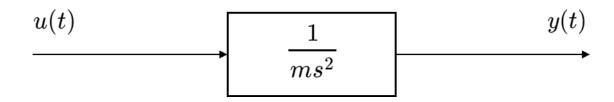
- When you are designing a PID controller for a given system, follow the steps shown below to obtain a
  desired response.
  - Obtain an open-loop response and determine what needs to be improved
  - Add a proportional control to improve the rise time
  - Add a derivative control to reduce the overshoot
  - Add an integral control to reduce the steady-state error
  - Adjust each of the gains  $k_P$ ,  $k_I$  and  $k_D$  until you obtain a desired overall response.

• Lastly, please keep in mind that you do not need to implement all three controllers (proportional, derivative, and integral) into a single system, if not necessary. For example, if a PI controller meets the given requirements (like the above example), then you don't need to implement a derivative controller on the system. Keep the controller as simple as possible.

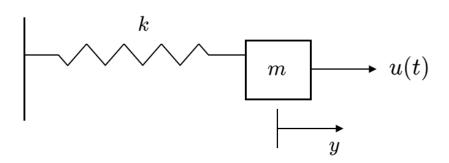


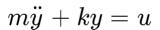
Mass

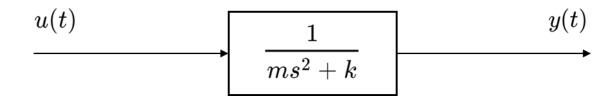




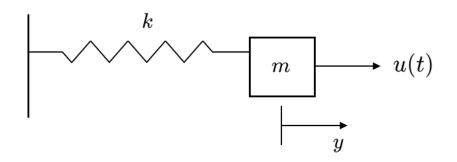
Mass and spring

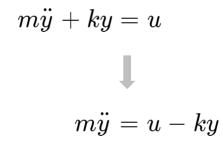


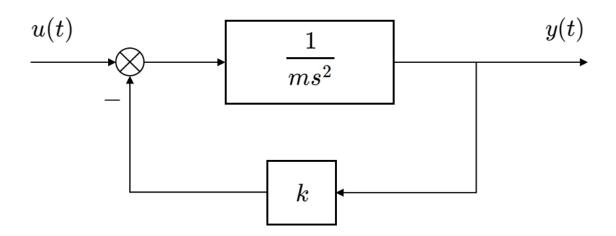




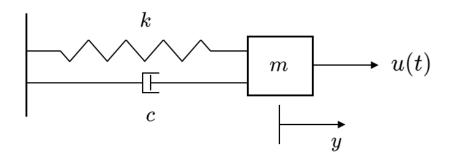
Mass and spring



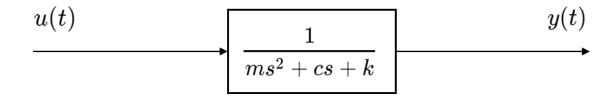




Mass, spring and damper



$$m\ddot{y} + c\dot{y} + ky = u$$



Mass, spring and damper

$$m\ddot{y} + c\dot{y} + ky = u$$
  $\Longrightarrow$   $m\ddot{y} = u - ky - c\dot{y}$ 

