2-H Relations among <u>Filter Length N</u>, <u>Transition Band</u>, and Accuracy

- ◆ Suppose that we want:
 - ① passband ripple $\leq \delta_1$,
 - ② stopband ripple $\leq \delta_2$,
 - ③ width of transition band ≤ ΔF

(expressed by normalized frequency)

$$\Delta F = (f_1 - f_2)/f_s = (f_1 - f_2)T$$
 (f_s : sampling frequency, T : sampling interval)

Then, the estimated length N of the digital filter is:

$$N = \frac{2}{3} \frac{1}{\Delta F} \log_{10} \left(\frac{1}{10 \delta_1 \delta_2} \right)$$

- When there are two transition bands, $\Delta F = \min(\Delta F_1, \Delta F_2)$
- 犠牲 transition band 的 frequency response, 換取較高的 passband and stopband accuracies

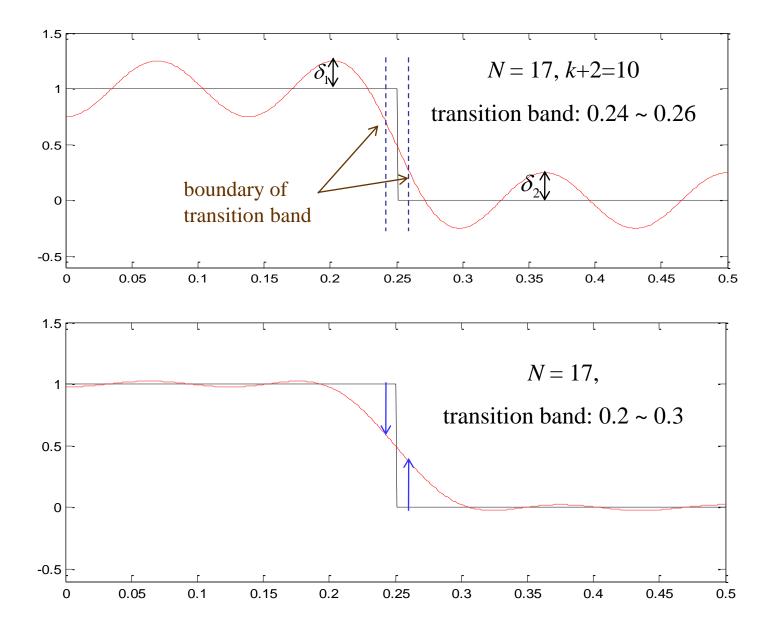
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$$N = \frac{2}{3} \frac{1}{\Delta F} \log_{10} \left(\frac{1}{10\delta_1 \delta_2} \right) \qquad \frac{3}{2} N \Delta F = \log_{10} \left(\frac{1}{10\delta_1 \delta_2} \right)$$

[Ref] F. Mintzer and L. Bede, "Practical design rules for optimum FIR bandpass digital filter", *IEEE Trans. ASSP*, vol. 27, no. 2, pp. 204-206, Apr. 1979.

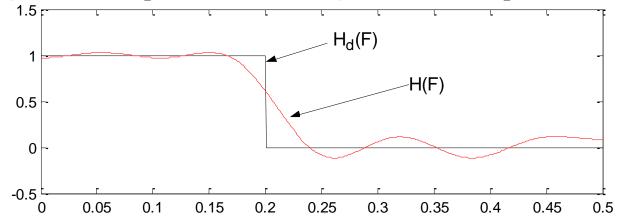
問題:假設 $\delta_1 = \delta_2 = \delta$, N 為固定,

當 ΔF 變為 A 倍時 δ 變為多少 ?

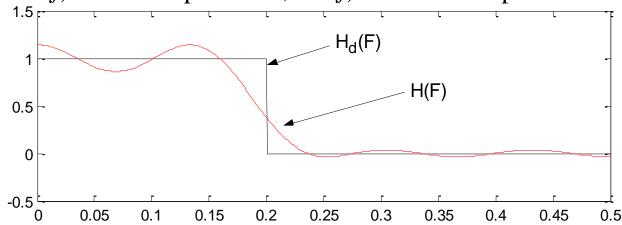


2-I Relations between Weight Functions and Accuracy

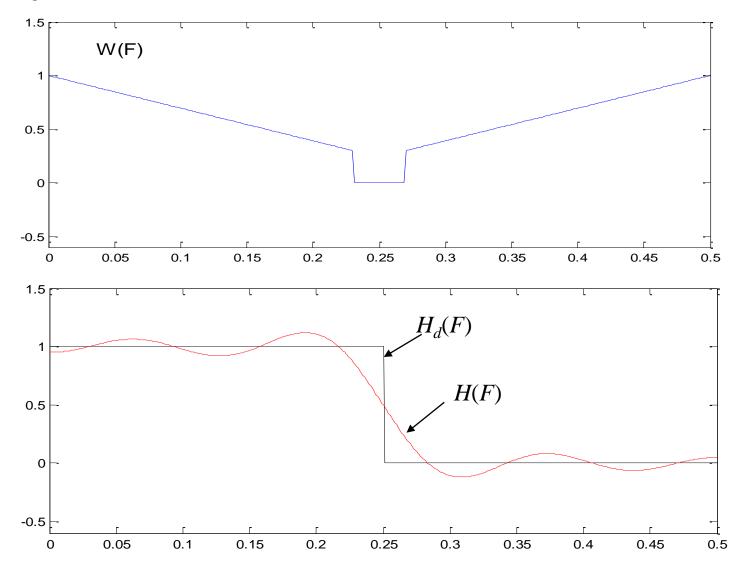
If we treat the passband more important than the stop band W(f) = 1 in the passband, 0 < W(f) < 1 in the stopband



If we treat the stop band more important than the pass band 0 < W(f) < 1 in the passband, W(f) = 1 in the stopband



Larger error near the transition band



2-J FIR Filter in MSE Sense with Weight Functions

$$R(F) = \sum_{n=0}^{k} s[n]\cos(2\pi nF)$$
 可對照 pages 47~49
$$MSE = \int_{-1/2}^{1/2} W(F) |R(F) - H_d(F)|^2 dF$$
 $F = f/f_s$

$$= \int_{-1/2}^{1/2} W(F) \left(\sum_{\tau=0}^{k} s[\tau]\cos(2\pi \tau F) - H_d(F)\right)^2 dF$$

$$\frac{\partial MSE}{\partial s[n]} = 2 \int_{-1/2}^{1/2} W(F)\cos(2\pi nF) \left(\sum_{\tau=0}^{k} s[\tau]\cos(2\pi \tau F) - H_d(F)\right) dF = 0$$

$$\sum_{\tau=0}^{k} s[\tau] \int_{-1/2}^{1/2} W(F)\cos(2\pi nF)\cos(2\pi \tau F) dF - \int_{-1/2}^{1/2} W(F) H_d(F)\cos(2\pi nF) dF = 0$$

$$n = 0 \sim k$$

問題:
$$\int_{-1/2}^{1/2} W(F) \cos(2\pi n F) \cos(2\pi \tau F) dF \neq 0 \text{ when } n \neq \tau$$
 (not orthogonal)

$$\sum_{\tau=0}^{k} s[\tau] \int_{-1/2}^{1/2} W(F) \cos(2\pi n F) \cos(2\pi \tau F) dF - \int_{-1/2}^{1/2} W(F) H_d(F) \cos(2\pi n F) dF = 0$$

$$\tau = 0 \sim k, n = 0 \sim k$$

可以表示成 $(k+1) \times (k+1)$ matrix operation

$$B[n,\tau] = \int_{-1/2}^{1/2} W(F) \cos(2\pi n F) \cos(2\pi \tau F) dF$$

$$C[n] = \int_{-1/2}^{1/2} W(F) H_d(F) \cos(2\pi n F) dF$$

Think: Is it possible to apply the transition band to the FIR filter in the MSE sense?

2-K Four Types of FIR Filter

$$h[n] = 0$$
 for $n < 0$ and $n \ge N$ 點數為 N

$$H(F) = \sum_{n=0}^{N-1} h[n] \exp(-j2\pi n F)$$

$$k = (N-1)/2$$

• Type 1:
$$R(F) = \sum_{n=0}^{K} s[n] \cos(2\pi n F)$$

 $\underline{h[n]} = \underline{h[N-1-n]}$ (even symmetric) and \underline{N} is odd.

• Type 2:
$$R(F) = \sum_{n=1}^{k+1/2} s[n] \cos(2\pi (n-1/2)F)$$

h[n] = h[N-1-n] (even symmetric) and N is even.

• Type 3:
$$R(F) = \sum_{n=1}^{k} s[n] \sin(2\pi n F)$$

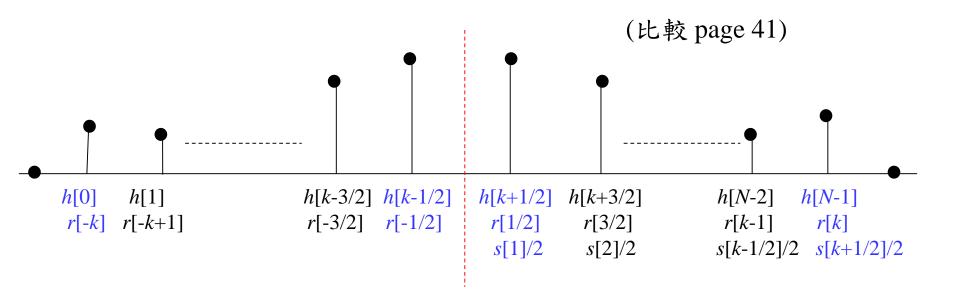
h[n] = -h[N-1-n] (odd symmetric) and N is odd.

• Type 4:
$$R(F) = \sum_{n=1}^{k+1/2} s[n] \sin(2\pi (n-1/2)F)$$

h[n] = -h[N-1-n] (odd symmetric) and N is even.

$$k = (N-1)/2$$

• Type 2: When h[n] = h[N-1-n] and N is even: (even symmetric)



當
$$R(F) = \sum_{n=-k}^{k} r[n] \exp(-j2\pi nF)$$

$$R(F) = e^{j2\pi F k} H(F)$$

$$R(F) = \sum_{n=1/2}^{k} \{r[n] \exp(-j2\pi n F) + r[-n] \exp(j2\pi n F)\}$$

$$= \sum_{n=1/2}^{k} r[n] \{\exp(-j2\pi n F) + \exp(j2\pi n F)\} = \sum_{n=1/2}^{k} 2r[n] \cos(2\pi n F)$$

$$R(F) = \sum_{n=1}^{k+1/2} s[n] \cos(2\pi (n-1/2)F)$$

$$n_{(new)} = n_{(old)} + \frac{1}{2}$$

$$n_{(old)} = n_{(new)} - \frac{1}{2}$$

$$n_{(new)} = n_{(old)} + \frac{1}{2}$$
 $n_{(old)} = n_{(new)} - \frac{1}{2}$

$$s[n] = 2r[n-1/2]$$
 $n = 1, 2, ..., k+1/2$

設計出 s[n] 之後

$$r[n] = s[n+1/2]/2, h[n] = r[n-k],$$

Design Method for Type 2

$$R(F) = \sum_{n=1}^{k+1/2} s[n] \cos(2\pi (n-1/2)F)$$

由於 n 和 n+1 兩項相加可得

$$\cos(2\pi(n-1/2)F) + \cos(2\pi(n+1/2)F) = 2\cos(\pi F)\cos(2\pi nF)$$

所以可以「判斷」R(F)能被改寫成

$$R(F) = \cos(\pi F) \sum_{n=0}^{k_1} s_1[n] \cos(2\pi n F)$$

求 $s_1[n]$ 和 s[n] 之間的關係

$$R(F) = \sum_{n=0}^{k_1} s_1[n] \cos(\pi F) \cos(2\pi nF)$$

$$= \sum_{n=0}^{k_1} \frac{1}{2} s_1[n] \cos(2\pi (n-1/2)F) + \sum_{n=0}^{k_1} \frac{1}{2} s_1[n] \cos(2\pi (n+1/2)F)$$

$$= \sum_{n=0}^{k_1} \frac{1}{2} s_1[n] \cos(2\pi (n-1/2)F) + \sum_{n=1}^{k_1+1} \frac{1}{2} s_1[n-1] \cos(2\pi (n-1/2)F)$$

$$R(F) = \sum_{n=0}^{k_1} \frac{1}{2} s_1[n] \cos(2\pi(n-1/2)F) + \sum_{n=1}^{k_1+1} \frac{1}{2} s_1[n-1] \cos(2\pi(n-1/2)F)$$

$$R(F) = \frac{1}{2} s_1[0] \cos(\pi F) + \sum_{n=1}^{k_1} \frac{1}{2} (s_1[n] + s_1[n-1]) \cos(2\pi (n-1/2)F)$$
$$+ \frac{1}{2} s_1[k_1] \cos(2\pi (k_1 + 1/2)F)$$

比較係數可得
$$s[1] = s_1[0] + \frac{1}{2}s_1[1]$$

$$s[n] = \frac{1}{2}(s_1[n] + s_1[n-1]) \quad \text{for } n = 2, 3,, k-1/2$$

$$s[k+1/2] = \frac{1}{2}s_1[k-1/2]$$

$$err(F) = [H_d(F) - R(F)]W(F)$$

$$= \left[H_d(F) - \cos(\pi F) \sum_{n=0}^{k-1/2} s_1[n] \cos(2\pi nF)\right]W(F)$$

$$= \left[\sec(\pi F)H_d(F) - \sum_{n=0}^{k-1/2} s_1[n] \cos(2\pi nF)\right] \cos(\pi F)W(F)$$

只需將 pages 53-56 的方法當中, $H_d(F)$ 換成 $\sec(\pi F)H_d(F)$ W(F) 換成 $\cos(\pi F)W(F)$ k 換成 k-1/2=N/2-1 注意 $s_1[n]$ 和 s[n] 之間的關係即可

Design Method for Type 3

$$R(F) = \sum_{n=1}^{k} s[n] \sin(2\pi n F)$$

由於 n-1 和 n+1 兩項相減可得

$$\sin(2\pi(n+1)F) - \sin(2\pi(n-1)F) = 2\sin(2\pi F)\cos(2\pi nF)$$

所以「判斷」可將 R(F) 改寫為

$$R(F) = \sin(2\pi F) \sum_{n=0}^{k_1} s_1[n] \cos(2\pi n F)$$

$$R(F) = \frac{1}{2} \sum_{n=0}^{k_1} s_1[n] \sin(2\pi (n+1)F) - \frac{1}{2} \sum_{n=0}^{k_1} s_1[n] \sin(2\pi (n-1)F)$$

$$= \frac{1}{2} \sum_{n=0}^{k_1} s_1[n] \sin(2\pi (n+1)F) + \frac{s_1[0]}{2} \sin(2\pi F) - \frac{1}{2} \sum_{n=2}^{k_1} s_1[n] \sin(2\pi (n-1)F)$$

$$= \frac{1}{2} \sum_{n=1}^{k_1+1} s_1[n-1] \sin(2\pi nF) + \frac{s_1[0]}{2} \sin(2\pi F) - \frac{1}{2} \sum_{n=1}^{k_1-1} s_1[n+1] \sin(2\pi nF)$$

$$R(F) = \frac{s_1[0]}{2} \sin(2\pi F) + \frac{1}{2} (s_1[0] - s_1[2]) \sin(2\pi F)$$

$$+ \frac{1}{2} \sum_{n=2}^{k_1-1} (s_1[n-1] - s_1[n+1]) \sin(2\pi n F)$$

$$+ \frac{1}{2} s_1[k_1 - 1] \sin(2\pi k_1 F) + \frac{1}{2} s_1[k_1] \sin(2\pi (k_1 + 1) F)$$
令 $k_1 = k - 1$, 比較係數可得
$$s[1] = s_1[0] - \frac{1}{2} s_1[2]$$

$$s[n] = \frac{1}{2} s_1[n-1] - \frac{1}{2} s_1[n+1] \qquad \text{for } n = 2, 3,, k-2$$

$$s[k-1] = \frac{1}{2} s_1[k-2]$$

$$s[k] = \frac{1}{2} s_1[k-1]$$

$$err(F) = [H_d(F) - R(F)]W(F)$$

$$= \left[H_d(F) - \sin(2\pi F) \sum_{n=0}^{k-1} s_1[n] \cos(2\pi nF) \right] W(F)$$

$$= \left[\csc(2\pi F) H_d(F) - \sum_{n=0}^{k-1} s_1[n] \cos(2\pi nF) \right] \sin(2\pi F) W(F)$$
將 pages 53-56 的方法當中, $H_d(F)$ 换成 $\csc(2\pi F) H_d(F)$

$$W(F)$$
 换成 $\sin(2\pi F) W(F)$

$$k$$
 换成 $k-1$
注意 $s_1[n]$ 和 $s[n]$ 之間的關係即可

Think: Design Method for Type 4

附錄三: Matlab 寫程式需注意的地方

一、各種程式語言寫程式共通的原則

- (1) 能夠不在迴圈內做的運算,則移到迴圈外,以節省運算時間
- (2) 寫一部分即測試,不要全部寫完再測試(縮小範圍比較容易 debug)
- (3) 先測試簡單的例子,成功後再測試複雜的例子

二、Matlab 寫程式特有的技巧

- (1) 迴圈能避免就儘量避免
- (2) 儘可能使用 Matrix 及 Vector operation

Example: 由 1 加 到 100,用 Matlab 一行就可以了 sum([1:100])

完全不需迴圈

三、一些重要的 Matlab 指令

- (1) function: 放在第一行,可以將整個程式函式化
- (2) tic, toc: 計算時間 tic 為開始計時, toc 為顯示時間
- (3) find: 找尋一個 vector 當中不等於 0 的entry 的位置 範例: find([1 0 0 1]) = [1, 4] find(abs([-5:5])<=2) = [4, 5, 6, 7, 8] (因為 abs([-5:5])<=2 = [0 0 0 1 1 1 1 1 1 0 0 0])
- (4) ': Hermitian (transpose + conjugation) '. ': transpose
- (5) imread: 讀圖, image, imshow, imagesc: 將圖顯示出來,(註: 較老的 Matlab 版本 imread 要和 double 並用A=double(imread('Lena.bmp'));
- (6) imwrite: 製做圖檔

- (7) xlsread: 由 Excel 檔讀取資料
- (8) xlswrite: 將資料寫成 Excel 檔
- (9) aviread: 讀取 video 檔
- (10) dlmread: 讀取 *.txt 或其他類型檔案的資料
- (11) dlmwrite: 將資料寫成 *.txt 或其他類型檔案