Tsinghua Berkeley Shenzhen Institute (TBSI) ME233 Advanced Control Systems II

Spring 2024

Homework #4

Assigned: April 8 (Monday) Due: April 16 (Tuesday)

1. Consider the system

$$\begin{bmatrix} x_1(k+1) \\ x_2(k+1) \end{bmatrix} = \begin{bmatrix} -0.08 & -1 \\ 0.7 & 0.1 \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + \begin{bmatrix} 0.34 \\ 0.3 \end{bmatrix} (u(k) + w(k))$$

$$y(k) = \begin{bmatrix} 0 & 3 \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + v(k)$$
(1)

where u(k) is a deterministic *known* input and x(0) is Gaussian and w(k) and w(k) are zero-mean stationary Gaussian sequences (i.e. **white**), with the following statistics:

•
$$x_o = E\{x(0)\} = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$$
, $\tilde{x}(0) = x(0) - x_0$ and $X_o = E\{\tilde{x}(0)\tilde{x}^T(0)\} = \begin{bmatrix} 0.1 & 0 \\ 0 & 0.1 \end{bmatrix}$

•
$$E\{w(k)\} = 0$$
, $\tilde{w}(k) = w(k) - m_w$, $E\{v(k)\} = 0$, and $\tilde{v}(k) = v(k)$.

$$\bullet \ E\left\{ \begin{bmatrix} \tilde{w}(k+j) \\ \tilde{v}(k+j) \end{bmatrix} \begin{bmatrix} \tilde{w}(k) & \tilde{v}(k) \end{bmatrix} \right\} = \begin{bmatrix} W & 0 \\ 0 & V \end{bmatrix} \delta(j) = \begin{bmatrix} 1 & 0 \\ 0 & 0.5 \end{bmatrix} \delta(j)$$

$$\bullet \ E\left\{ \begin{bmatrix} \tilde{w}(k) \\ \tilde{v}(k) \end{bmatrix} \tilde{x}^T(0) \right\} = 0$$

Notice that in this system, the control input u(k) and the input noise w(k) are injected at the same location, i.e. $B = B_w$. This is not necessarily the case in many systems.

(a) Assume that the control input u(k) is a given deterministic sequence. Obtain the state equations that define the **marginal** expected value of the state and the output equation for the **marginal** expected value of the output

$$m_x(k) = \begin{bmatrix} m_{x_1}(k) \\ m_{x_1}(k) \end{bmatrix} = \begin{bmatrix} E\{x_1(k)\} \\ E\{x_1(k)\} \end{bmatrix}$$

$$m_y(k) = E\{y(k)\}$$

(b) Define the **marginal** state and output estimation errors by

$$\tilde{X}(k) = x(k) - m_x(k)$$

 $\tilde{Y}(k) = y(k) - m_y(k)$

(Notice that we are using capital letters to distinguish the marginal estimation errors from the Kalman filter a-posteriori estimation errors, which will be written using lower case letters.)

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The state and output equations for the **marginal** state and output estimation errors can be written as

$$\tilde{X}(k+1) = A\tilde{X}(k) + Bw(k)
\tilde{Y}(k) = C\tilde{X}(k) + v(k)$$
(2)

i. The marginal state and output estimation covariances are

$$\Lambda_{\tilde{X}\tilde{X}}(k,j) = E\{\tilde{X}(k+j)\tilde{X}^T(k)\}$$

$$\Lambda_{\tilde{Y}\tilde{Y}}(k,j) = E\{\tilde{Y}(k+j)\tilde{Y}^T(k)\}$$

As discuss in class $\Lambda_{\tilde{X}\tilde{X}}(k,0)$ evolves according to the following Lyapunov equation (remember that $B=B_w$)

$$\Lambda_{\tilde{X}\tilde{X}}(k+1,0) = A\Lambda_{\tilde{X}\tilde{X}}(k,0)A^T + B_wWB_w^T \qquad \Lambda_{\tilde{X}\tilde{X}}(0,0) = X_o.$$
 (3)

Write a matlab code that will compute $\Lambda_{\tilde{X}\tilde{X}}(k,0)$, $\Lambda_{\tilde{Y}\tilde{Y}}(k,0)$ and

$$E\{\tilde{X}^T(k)\tilde{X}(k)\} = \operatorname{trace}(\Lambda_{\tilde{X}\tilde{X}}(k,0))$$

recursively forward in time, starting from k = 0 until k = 20.

- ii. Plot $E\{\tilde{X}^T(k)\tilde{X}(k)\}$ and $\Lambda_{\tilde{Y}\tilde{Y}}(k,0)$ as a function of k.
- iii. Since that the input noise w(k) and measurement noise v(k) are white, the marginal estimation error dynamics in Eq. (2) will converge to a stationary system. As a consequence,

$$\lim_{k \to \infty} \Lambda_{\tilde{X}\tilde{X}}(k,0) = \bar{\Lambda}_{\tilde{X}\tilde{X}}(0).$$

where the steady state marginal state estimation covariance, $\bar{\Lambda}_{\tilde{X}\tilde{X}}(0)$, is the solution of the following algebraic Lyapunov equation

$$A\bar{\Lambda}_{\tilde{X}\tilde{X}}(0)A^T - \bar{\Lambda}_{\tilde{X}\tilde{X}}(0) = -B_w W B_w^T. \tag{4}$$

Use the matlab function dlyap, to solve the Lyapunov equation (4) and compute the steady state marginal state estimation covariance $\bar{\Lambda}_{\tilde{X}\tilde{X}}(0)$ and its trace. Then compute the steady state marginal output estimation covariance:

$$\begin{split} \bar{\Lambda}_{\tilde{X}\tilde{X}}(0) &= \lim_{k \to \infty} \Lambda_{\tilde{X}\tilde{X}}(k,0) = \lim_{k \to \infty} E\{\tilde{X}(k)\tilde{X}^T(k)\} \\ \operatorname{trace}(\bar{\Lambda}_{\tilde{X}\tilde{X}}(0)) &= \lim_{k \to \infty} E\{\tilde{X}^T(k)\tilde{X}(k)\} \\ \bar{\Lambda}_{\tilde{Y}\tilde{Y}}(0) &= \lim_{k \to \infty} = \Lambda_{\tilde{Y}\tilde{Y}}(k,0) = \lim_{k \to \infty} E\{\tilde{Y}^2(k)\} \end{split}$$

Also use matlab to compute the differences

$$\Delta \operatorname{trace}(\Lambda_{\tilde{X}\tilde{X}}(20,0)) = \operatorname{trace}(\bar{\Lambda}_{\tilde{X}\tilde{X}}(0)) - \operatorname{trace}(\bar{\Lambda}_{\tilde{X}\tilde{X}}(20))$$

and

$$\Delta \bar{\Lambda}_{\tilde{Y}\tilde{Y}}(20,0) = \bar{\Lambda}_{\tilde{Y}\tilde{Y}}(0) - \Lambda_{\tilde{Y}\tilde{Y}}(20,0)$$

to check if the system has reached stationarity by k = 20.

- (c) Run the matlab code that executes steps 1(b)i 1(b)iii, only that in this case use V = 10.
- (d) Run the matlab code that executes steps 1(b)i 1(b)iii, only that in this case use V = 0.01.
- 2. Consider again the system in Eq. (1). We will now analyze the response of a Kalman filter to estimate the state of the system using a sequence of output measurements $\{y(k)\}$ for $k = 0, 1, 2, \cdots$.

The a-posteriori estimation structure of the Kalman filter can be written as follows

$$\tilde{y}^{o}(k) = y(k) - C\hat{x}^{o}(k)
\hat{x}(k) = \hat{x}^{o}(k) + F(k)\tilde{y}^{o}(k)
F(k) = M(k)C^{T}[AM(k)C^{T} + V]^{-1}$$
(6)

$$Z(k) = M(k) - M(k)C^{T}[AM(k)C^{T} + V]^{-1}CM(k)$$
(6)

$$\hat{x}^{o}(k+1) = A\hat{x}(k) + Bu(k)
M(k+1) = AZ(k)A^{T} + B_{w}W(k)B_{w}^{T}$$
(7)

where the a-posteriori and a-priori state estimates and estimation errors are define as follows. Given $Y_k = [y(0), y(1), \cdots y(k)]$,

$$\begin{array}{rcl} \hat{x}(k) & = & E\{x(k)|Y_k\} & & \tilde{x}(k) = x(k) - \hat{x}(k) \\ \hat{x}^o(k) & = & E\{x(k)|Y_{k-1}\} & & \tilde{x}^o(k) = x(k) - \hat{x}^o(k) \end{array}$$

Their respective estimation error covariances are

$$Z(k) = E\{\tilde{x}(k)\tilde{x}^T(k)\}$$

$$M(k) = E\{\tilde{x}^o(k)\tilde{x}^{oT}(k)\}$$

As discussed in class, the update equations for Z(k) and M(k), Eqs. (6) and (7), can be combined into a single **Riccati** equation for updating M(k) (remember that $B = B_w$):

$$M(k+1) = AM(k)A^{T} + B_{w}WB_{w}^{T} - AM(k)C^{T}[CM(k)C^{T} + V]^{-1}CM(k)A^{T}, \quad M(0) = X_{o}$$
(8)

Finally, the innovation covariance, also known as the a-posteriori output error covariance, is given by

$$\Lambda_{\tilde{v}^o\tilde{v}^o}(k,0) = E\{\tilde{v}^o(k)^2\} = CM(k)C^T + V.$$

- (a) Write a matlab program that will compute:
 - M(k)
 - Z(k)
 - $E\{\tilde{x}^T(k)\tilde{x}(k)\} = \text{trace}(Z(k))$
 - $\Lambda_{\tilde{y}^o\tilde{y}^o}(k,0)$

recursively forward in time, starting from k = 0 and until k = 10. (Remember to reset V = 0.5.)

- (b) Plot trace(Z(k)) and $\Lambda_{\tilde{y}^o\tilde{y}^o}(k,0)$ as a function of k. Check if these quantities appear to be converging to steady state values. If they don't you may need to extend the simulation beyond k = 10.
- (c) Run the matlab code that executes steps 2a and 2b, only that in this case use V=10.
- (d) Run the matlab code that executes steps 2a and 2b, only that in this case use V=0.01.
- 3. Discuss the performance of the Kalman filter in Problem 2 as a function the magnitude of the measurement noise intensities V = 10, V = 0.5 and V = 0.01 relative to the constant magnitude of the input noise intensity W = 1. Do so by comparing
 - The trace of the Kalman filter a-posteriori state estimation error covariance,

$$E\{\tilde{x}^T(k)\tilde{x}(k)\} = \operatorname{trace}(Z(k))$$

with the trace of the marginal state estimation error covariance

$$E\{\tilde{X}^T(k)\tilde{X}(k)\} = \operatorname{trace}(\Lambda_{\tilde{X}\tilde{X}}(k,0))$$

in Problem 1

for the same value of output noise intensity V.

4. Kalman filter with correlated input and measurement noise:

Consider the discrete-time system given by

$$x(k+1) = Ax(k) + Bu(k) + w(k)$$
(9)

$$y(k) = Cx(k) + v(k) \tag{10}$$

where $E\{x(0)\} = x_o$, $E\{w(k)\} = 0$, $E\{v(k)\} = 0$, $E\{(x(0) - x_o)(x(0) - x_o)^T\} = X_o$, $E\{(x(0) - x_o)w^T(k)\} = 0$, $E\{(x(0) - x_o)v^T(k)\} = 0$, and

$$E\left\{\begin{bmatrix} w(k) \\ v(k) \end{bmatrix} \begin{bmatrix} w(j)^T & v(j)^T \end{bmatrix}\right\} = \begin{bmatrix} W & S \\ S^T & V \end{bmatrix} \delta(k-j)$$

where $V \in \mathbb{R}^{m \times m}$ is positive definite matrix. The a-priori Kalman filter for this system can be written as

$$\hat{x}^{o}(k+1) = A\hat{x}^{o}(k) + Bu(k) + L(k)[y(k) - C\hat{x}^{o}(k)]$$
(11)

$$L(k) = [AM(k)C^{T} + S][CM(k)C^{T} + V]^{-1}$$
(12)

$$M(k+1) = AM(k)A^{T} + W - [AM(k)C^{T} + S][CM(k)C^{T} + V]^{-1}[CM(k)A^{T} + S^{T}]$$
(13)

with initial conditions $\hat{x}^o(0) = x_o$ and $M(0) = X_o$.

Derive Eqs. (11)–(13) using previously-derived results in Kalman filtering and noticing that Eqs. (9)–(10) can be written as

$$x(k+1) = A'x(k) + Bu(k) + w'(k) + SV^{-1}y(k)$$
,

where $A' = A - SV^{-1}C$ and

$$E\left\{\begin{bmatrix} w^{'}(k) \\ v(k) \end{bmatrix} \begin{bmatrix} w^{'}(j)^T & v(j)^T \end{bmatrix} \right\} = \begin{bmatrix} W^{'} & 0 \\ 0 & V \end{bmatrix} \, \delta(k-j), \qquad \quad W^{'} = W - SV^{-1}S^T \; .$$