

Fourier Transform Spectrometry



主要讲授：傅里叶变换光谱测量原理和傅里变换
红外波段测量法和可见光傅里叶变换测量方法。

研究重点对象：可见光傅里叶变换光谱测量方法。

邮箱： Ftsouc@126.com 密码： **ftsouc**

参考书： *Fourier-Transform spectroscopy Instrumentation Engineering* (作者： Vidi Saptari) .



Chapter 1

General Introduction(概述)

Principal Contents

- 1.1 Background of Fourier Transform Spectrometry
- 1.2 Review of mathematical background knowledge (**Fourier Transform, DTFT, DFT, FFT**)
- 1.3 Review of superposition of light waves



■ 1.1 Background of Fourier Transform Spectrometry

FTSs have been used in many infrared and near-infrared applications over the last couple of decades.

multiplex (Fellgett's), **throughput** (Jacquinot's) and **wavenumber** (Connes') advantages

The ultraviolet-visible FTS is more sensitive to the position-tracking accuracy and the quality of the interferometer alignment, the translation stage system etc. because of their short wavelengths.



■ 1.1 Background of Fourier Transform Spectrometry

1.1.1 History of Fourier Transform Spectrometry

- (1) Michelson in 1891, Ruben and Wood interferogram in 1911, but no computer is available.**
- (2) 1950s, FFT 1965(N^2 to $N \ln N$)**
- (3) High-resolution FTS is applied in space shuttle in 1985**
- (4) In recent years, Imaging Fourier Transform Spectrometer (IFTS), Nonsanning FTS**



■ 1.1 Background of Fourier Transform Spectrometry

1.1.2 Current position-tracking methods of FTS

- (1) Spectral folding technique.**
- (2) Uniform time-sampling method**
- (3) Subdivision of the He-Ne laser fringes by electronic or optical methods**



■ 1.2 Review of mathematical background knowledge

● 1.2.1 Fourier Transform

A continuous time signal $x(t)$ with finite energy

$$E_N = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

Can be represented in the frequency domain

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt \quad \omega = 2\pi f$$

$X(\omega)$ is Fourier Transform expression of $x(t)$

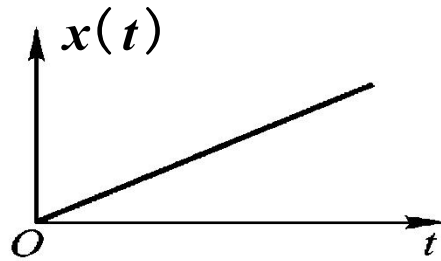
1.2 Review of mathematical background knowledge

1.2.2 Discrete Time Fourier Transform(DTFT)

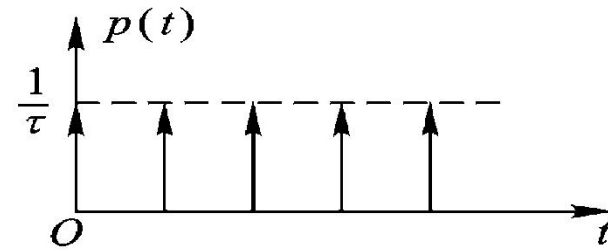
1. Sampling Process (采样过程)

$$x^*(t) = x(t) \cdot p(t)$$

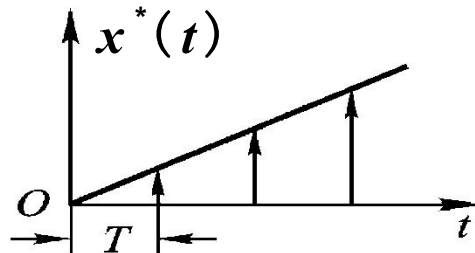
$$p(t) = \frac{1}{\tau} \quad \left(-\frac{1}{2}\tau + kT \leq t \leq \frac{1}{2}\tau + kT\right)$$



(a)



(b)



(c)



(d)

Fig.1.1 Sampling Process



1.2 Review of mathematical background knowledge

2. Ideal Sampling Process (理想采样过程)

- To simplify the description of Sampling Process, Ideal Sampling Switch(理想采样开关) is introduced
- Carrier signal $p(t)$ can approximately be equal to the following ideal pulse sequence when $\tau \rightarrow 0$
- $$\delta_T(t) = \sum_{k=0}^{\infty} \delta(t - kT) \quad (1-1)$$



1.2 Review of mathematical background knowledge

Suppose $x(t) = 0$ when $t < 0$

Then the sampling signal can be described as

$$x^*(t) = x(t) \bullet \delta_T(t) = \sum_{k=0}^{+\infty} x(t) \bullet \delta(t - kT) \quad (1-2)$$

此时，采样过程如图 *Fig.1.1(c)* 所示。

理想采样开关的输出是一个 **理想**脉冲序列。



■ 1.2 Review of mathematical background knowledge

3. Discrete Time Fourier Transform (DTFT)

$$x^*(t) = x(t) \cdot \delta_T(t) = \sum_{k=0}^{+\infty} x(t) \cdot \delta(t-kT) \quad (1-3)$$

Considering $x(t)=0$ when t is bigger enough ($\geq NT$), then

$$\therefore X^*(\omega) = \int_{-\infty}^{\infty} x^*(t) e^{-j\omega t} dt \approx \sum_{n=0}^{N-1} x(nT) e^{-jn\omega T}$$

4. Shannon sampling theorem

显然 $\delta_T(t)$ 是周期函数，故可以展成如Fourier级数

$$\delta_T(t) = \sum_{n=-\infty}^{+\infty} C_n e^{jn\omega_s t} \quad (1-4)$$

其中 $C_n = \frac{1}{T}$



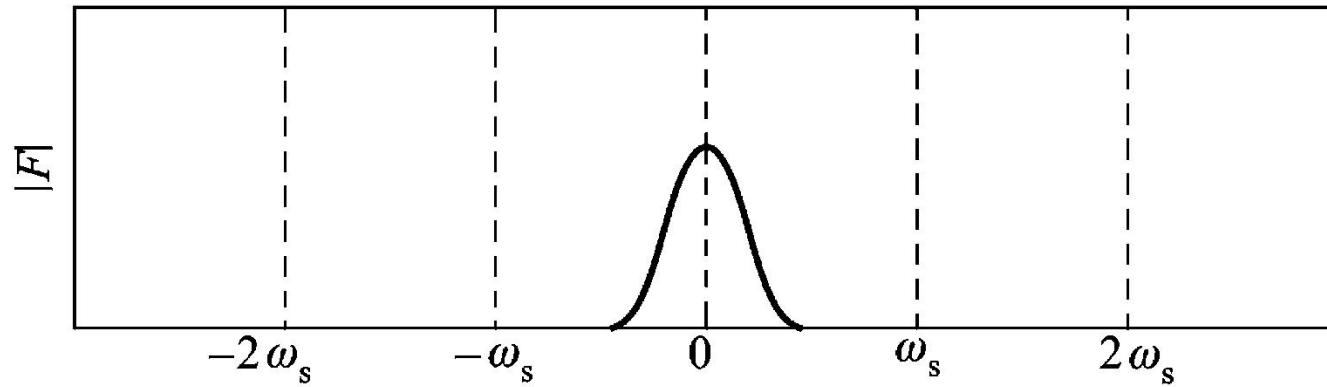
1.2 Review of mathematical background knowledge

则有 $x^*(t) = \frac{1}{T} \sum_{n=-\infty}^{+\infty} x(t) e^{jn\omega_s t}$ (1-5)

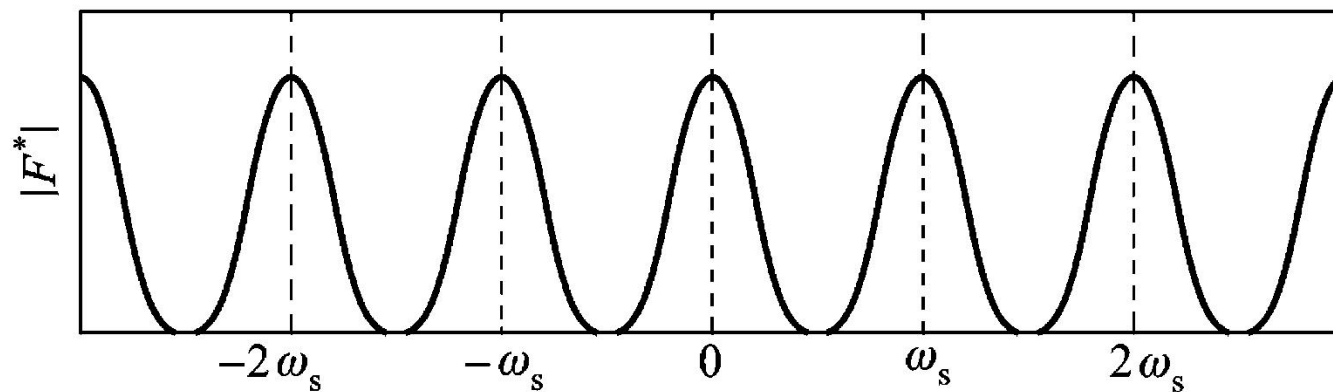
和

$$X^*(j\omega) = \frac{1}{T} \sum_{n=-\infty}^{\infty} X(j\omega + jn\omega_s) \quad (1-6)$$

1.2 Review of mathematical background knowledge



(a) 连续信号的频谱



(b) 理想采样信号的频谱

Fig.1.2 连续信号和采样信号的频谱

香农 (Shannon) 采样定理

- 若存在一个理想的低通滤波器，其频率特性如图Fig.1.3所示，便可以将采样信号完全恢复成原连续信号。由此可得如下著名的 香农 (**Shannon**) 采样定理：

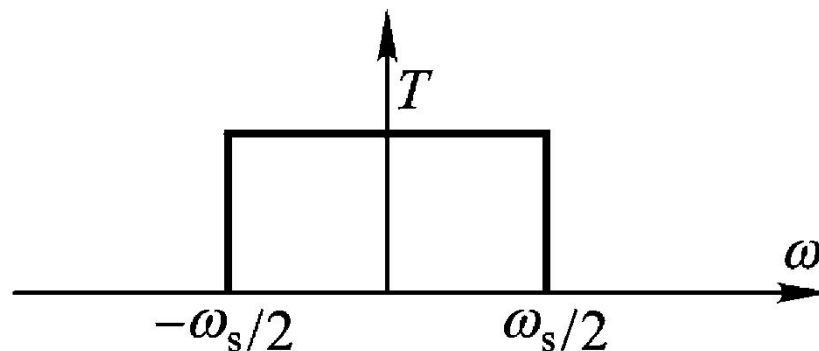


Fig.1.3



1.2 Review of mathematical background knowledge

如果采样频率 ω_s 满足以下条件

$$\omega_s \geq \omega_{max} \quad (1-7)$$

式中 ω_{max} 为连续信号频谱的上限频率

则经采样得到的脉冲序列可以无失真地恢复为原连续信号。



1.2 Review of mathematical background knowledge

注意：

上述香农采样定理要求满足以下两个条件：

- ① 频谱的上限频率是有限的；
- ② 存在一个理想的低通滤波器。但可以证明理想的低通滤波器在物理上是不可实现的，在实际应用中只能用非理想的低通滤波器来代替理想的低通滤波器；



■ 1.2 Review of mathematical background knowledge

• 1.2.3 Discrete Fourier Transform (DFT)

For the sequence of data
 $x(n) = [x(0), x(1), x(2), \dots, x(N-1)]$

DFT is Defined as follow.

$$X_s(\omega) = \sum_{n=0}^{N-1} x(t) e^{-jnT\omega}$$



1.2 Review of mathematical background knowledge

1.2.3 Fast Fourier Transform Algorithm(FFT)

FFT是快速的离散傅里叶变换的计算方法，
它要求采样数据长度为 2^N



1.3 Review of superposition of light waves

1. Superposition of waves of the same frequency

A solution of the differential wave equation can be written in the form

$$E(x, t) = E_0 \sin[\omega t + kx + \varepsilon] = E_0 \sin[\omega t + \alpha(x, \varepsilon)] \quad (1-9)$$

in which E_0 is the amplitude of the electric field, i.e. the of amplitude of harmonic disturbance propagation along the positive X -axis, ε is the *initial phase angle*



■ 1.3 Review of superposition of light waves

Suppose then that there are two such waves

$$E_1(x, t) = E_{01} \sin[\omega t + \alpha_1] \quad (1-10)$$

$$E_2(x, t) = E_{02} \sin[\omega t + \alpha_2] \quad (1-11)$$

each with the same frequency, coexisting in space.

The resultant disturbance is the linear superposition of these waves

$$E = E_1 + E_2$$



■ 1.3 Review of superposition of light waves

$$E = E_0 \cos \alpha \sin \omega t + E_0 \sin \alpha \cos \omega t$$

$$E = E_0 \sin(\omega t + \alpha) \quad (1-12)$$

where

$$E_0^2 = E_{01}^2 + E_{02}^2 + 2E_{01}E_{02} \cos(\alpha_1 - \alpha_2) \quad (1-13)$$

$$\tan \alpha = \frac{E_{01} \sin \alpha_{01} + E_{02} \sin \alpha_{02}}{E_{01} \cos \alpha_{01} + E_{02} \cos \alpha_{02}}$$



1.3 Review of superposition of light waves

The superposition of Many Waves

$$E = E_0 \cos \alpha \sin \omega t + E_0 \sin \alpha \cos \omega t$$

where $E = E_0 \sin(\omega t + \alpha)$ (1-12)

$$E_0^2 = \sum_{i=1}^N E_{0i}^2 + 2 \sum_{i=1}^N \sum_{j>i}^N E_{0i} E_{0j} \cos(\alpha_j - \alpha_i) \quad (1-13)$$

$$\tan \alpha = \frac{\sum_{i=1}^N E_{0i} \sin \alpha_{0i}}{\sum_{i=1}^N E_{0i} \cos \alpha_{0i}}$$



■ 1.3 Review of superposition of light waves

$$E_0^2 = \sum_{i=1}^N E_{0i}^2 + 2 \sum_{i=1}^N \sum_{j>i}^N E_{0i} E_{0j} \cos(\alpha_j - \alpha_i) \quad (1-13)$$

Random Sources:

$$E_0^2 = \sum_{i=1}^N E_{0i}^2 = N E_{01}^2 \quad (1-14)$$

Coherent Sources:

$$\begin{aligned} E_0^2 &= \sum_{i=1}^N E_{0i}^2 + 2 \sum_{i=1}^N \sum_{j>i}^N E_{0i} E_{0j} \\ &= \left(\sum_{i=1}^N E_{0i} \right)^2 = N^2 E_{01}^2 \end{aligned} \quad (1-15)$$



■ 1.3 Review of superposition of light waves

2. Addition of waves of different frequency

Suppose then that there are two such waves

$$E_1(x, t) = E_{01} \sin[\omega_1 t + \alpha_1] \quad (1-16)$$

$$E_2(x, t) = E_{02} \sin[\omega_2 t + \alpha_2] \quad (1-17)$$

each with the same frequency, coexisting in space.

The resultant disturbance is the linear superposition of these waves

$$E = E_1 + E_2$$