

MIDDLE EAST TECHNICAL UNIVERSITY

Department of Electrical and Electronics Engineering

EE463 – Power Electronics Hardware Project

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# Design and Thermal Analysis of a DC-DC Buck Converter

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*Presented by Group:*

**“Buck ve Ötesi”**

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# Contents

<b>1</b>	<b>Component Selection and Revised Thermal Analysis</b>	<b>2</b>
1.1	Deviation from Initial Design (Component Substitution) . . . . .	2
1.2	Power Loss Calculation (IXFH80N65X2) . . . . .	2
1.2.1	Conduction Losses ( $P_{cond}$ ) and Effective Resistance . . . . .	2
1.2.2	Switching Losses ( $P_{sw}$ ) . . . . .	2
1.2.3	Total Power Dissipation . . . . .	3
1.3	Heatsink Selection and Cooling Strategy . . . . .	3
1.4	Conclusion on Main Switching Cooling . . . . .	3
1.5	Auxiliary Thermal Protection: Freewheeling Diode . . . . .	3
<b>A</b>	<b>Appendix A: Component Datasheet</b>	<b>4</b>

# 1 Component Selection and Revised Thermal Analysis

## 1.1 Deviation from Initial Design (Component Substitution)

The initial design specifications called for the **IXGH24N60C4D1** IGBT. However, due to a supply chain substitution, the **IXFH80N65X2** (Ultra-Junction X2-Class Power MOSFET) was received and implemented.

A comparative analysis confirmed that the substituted MOSFET offers superior characteristics for this application ( $V_{DSS} = 650V$ ,  $I_{D25} = 80A$ ,  $R_{thJC} = 0.14^\circ C/W$ ). Consequently, the thermal calculations were revised to account for the resistive nature of the MOSFET.

## 1.2 Power Loss Calculation (IXFH80N65X2)

Unlike IGBTs, which are modeled with a relatively constant voltage drop ( $V_{CE}$ ), the conduction loss of a MOSFET depends on its On-State Resistance ( $R_{DS(on)}$ ).

It is crucial to note that  $R_{DS(on)}$  **is not a constant value**; it varies significantly with the junction temperature ( $T_J$ ). MOSFETs exhibit a *Positive Temperature Coefficient*, meaning their resistance increases as they heat up. Relying solely on the datasheet's nominal resistance at  $25^\circ C$  would lead to a severe underestimation of power losses and potentially dangerous thermal runaway.

### 1.2.1 Conduction Losses ( $P_{cond}$ ) and Effective Resistance

To ensure a realistic and safe thermal design, an "Effective Hot Resistance" ( $R_{hot}$ ) was calculated. This value represents the device's resistance at the target operating temperature, rather than its cold state.

- **Nominal Resistance ( $R_{cold}$ ):** The datasheet specifies a maximum  $R_{DS(on)}$  of **38 mΩ** at  $T_J = 25^\circ C$ .
- **Thermal Adjustment Factor:** According to *Fig. 4:  $R_{DS(on)}$  Normalized vs. Junction Temperature* in the datasheet, the resistance increases by a factor of approximately **2.2** when the junction temperature approaches the design limit of  $125^\circ C$ .

Using this normalization factor, the effective resistance used for the power calculation is:

$$R_{hot} = R_{DS(on), 25^\circ C} \times 2.2 = 38m\Omega \times 2.2 \approx 83.6m\Omega \text{ (0.0836 } \Omega) \quad (1)$$

Consequently, the conduction power loss at the load current of 12A is:

$$P_{cond} = I_{load}^2 \times R_{hot} = 12^2 \times 0.0836 = 144 \times 0.0836 \approx \mathbf{12.04 \text{ W}} \quad (2)$$

### 1.2.2 Switching Losses ( $P_{sw}$ )

The switching losses are determined by the transition times provided in the datasheet ( $t_{on} \approx 56ns$ ,  $t_{off} \approx 81ns$ ) and the switching frequency ( $2kHz$ ).

$$P_{sw} \approx \frac{1}{2} V_{bus} I_{load} (t_{total}) f_{sw} \quad (3)$$

$$P_{sw} \approx 0.5 \times 400 \times 12 \times (137 \times 10^{-9}) \times 2000 \approx \mathbf{0.66 \text{ W}} \quad (4)$$

### 1.2.3 Total Power Dissipation

$$P_{total} = P_{cond} + P_{sw} = 12.04 \text{ W} + 0.66 \text{ W} \approx \mathbf{12.7 \text{ W}} \quad (5)$$

## 1.3 Heatsink Selection and Cooling Strategy

To maintain the junction temperature below the safety limit of  $110^\circ\text{C}$ , the thermal resistance requirement was calculated using the derived total power loss.

**Thermal Parameters:**

- $R_{\theta JC}$  (**Junction-to-Case**):  $0.14 \text{ }^\circ\text{C/W}$  (Directly from datasheet).
- $R_{\theta CS}$  (**Case-to-Sink**):  $1.0 \text{ }^\circ\text{C/W}$  (Conservative estimate for the isolation pad).

**Calculation:**

$$R_{\theta SA} \leq \frac{T_{J,max} - T_{amb}}{P_{total}} - (R_{\theta JC} + R_{\theta CS}) \quad (6)$$

$$R_{\theta SA} \leq \frac{110 - 30}{12.7} - (0.14 + 1.0) \quad (7)$$

$$R_{\theta SA} \leq \frac{80}{12.7} - 1.14 \approx 6.30 - 1.14 = \mathbf{5.16 \text{ }^\circ\text{C/W}} \quad (8)$$

## 1.4 Conclusion on Main Switching Cooling

The calculation dictates that a heatsink with a thermal resistance of  $5.16 \text{ }^\circ\text{C/W}$  or lower is required.

Since the standard passive aluminum heatsinks in the laboratory inventory typically provide  $10\text{--}15 \text{ }^\circ\text{C/W}$  (which is insufficient for this load), the **Active Cooling Solution (Heatsink + Fan)** remains mandatory to ensure the MOSFET operates within its Safe Operating Area (SOA).

## 1.5 Auxiliary Thermal Protection: Freewheeling Diode

In addition to the main switching element, the thermal stability of the freewheeling diode was evaluated as a critical factor for the system's long-term reliability.

Although the theoretical power dissipation analysis suggested that the diode operates within its Safe Operating Area (SOA), practical considerations regarding ambient temperature fluctuations and continuous high-load operation necessitated a conservative approach.

To eliminate any risk of *thermal runaway* and to minimize the thermal stress on the junction, a dedicated **passive aluminum heatsink** was integrated onto the diode package. This preventive measure serves two key purposes:

1. **Derating:** It keeps the operational junction temperature ( $T_J$ ) significantly below the maximum rating, extending the component's lifespan.
2. **Transient Stability:** It provides a larger thermal mass to absorb transient heat spikes during rapid load changes.

Consequently, this auxiliary cooling solution ensures that the freewheeling diode remains thermally stable, safeguarding the overall integrity of the Buck Converter circuit.

## A Appendix A: Component Datasheet

The technical specifications for the substituted component, **IXFH80N65X2**, are attached below for reference.

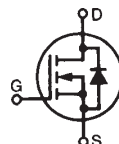


## X2-Class HiPerFET™ Power MOSFET

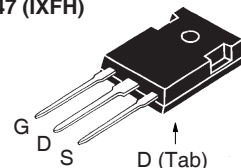
**IXFH80N65X2**  
**IXFK80N65X2**

$$\begin{aligned} V_{DSS} &= 650V \\ I_{D25} &= 80A \\ R_{DS(on)} &\leq 38m\Omega \end{aligned}$$

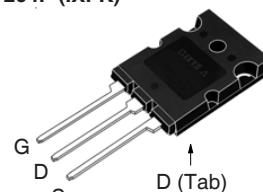
N-Channel Enhancement Mode  
Avalanche Rated  
Fast Intrinsic Diode



TO-247 (IXFH)



TO-264P (IXFK)



G = Gate      D = Drain  
S = Source    Tab = Drain

Symbol	Test Conditions	Maximum Ratings	
$V_{DSS}$	$T_J = 25^\circ\text{C}$ to $150^\circ\text{C}$	650	V
$V_{DGR}$	$T_J = 25^\circ\text{C}$ to $150^\circ\text{C}$ , $R_{GS} = 1M\Omega$	650	V
$V_{GSS}$	Continuous	$\pm 30$	V
$V_{GSM}$	Transient	$\pm 40$	V
$I_{D25}$	$T_C = 25^\circ\text{C}$	80	A
$I_{DM}$	$T_C = 25^\circ\text{C}$ , Pulse Width Limited by $T_{JM}$	160	A
$I_A$	$T_C = 25^\circ\text{C}$	20	A
$E_{AS}$	$T_C = 25^\circ\text{C}$	3	J
$dv/dt$	$I_S \leq I_{DM}$ , $V_{DD} \leq V_{DSS}$ , $T_J \leq 150^\circ\text{C}$	50	V/ns
$P_D$	$T_C = 25^\circ\text{C}$	890	W
$T_J$		-55 ... +150	$^\circ\text{C}$
$T_{JM}$		150	$^\circ\text{C}$
$T_{stg}$		-55 ... +150	$^\circ\text{C}$
$T_L$	Maximum Lead Temperature for Soldering	300	$^\circ\text{C}$
$T_{SOLD}$	1.6 mm (0.062in.) from Case for 10s	260	$^\circ\text{C}$
$M_d$	Mounting Torque	1.13 / 10	Nm/lb.in
Weight	TO-247	6	g
	TO-264P	10	g

### Features

- International Standard Packages
- Low  $R_{DS(ON)}$  and  $Q_G$
- Avalanche Rated
- Low Package Inductance

### Advantages

- High Power Density
- Easy to Mount
- Space Savings

### Applications

- Switch-Mode and Resonant-Mode Power Supplies
- DC-DC Converters
- PFC Circuits
- AC and DC Motor Drives
- Robotics and Servo Controls

Symbol	Test Conditions ( $T_J = 25^\circ\text{C}$ , Unless Otherwise Specified)	Characteristic Values		
		Min.	Typ.	Max.
$BV_{DSS}$	$V_{GS} = 0V$ , $I_D = 1mA$	650		V
$V_{GS(th)}$	$V_{DS} = V_{GS}$ , $I_D = 4mA$	3.5		5.0 V
$I_{GSS}$	$V_{GS} = \pm 30V$ , $V_{DS} = 0V$			$\pm 100$ nA
$I_{DSS}$	$V_{DS} = V_{DSS}$ , $V_{GS} = 0V$ $T_J = 125^\circ\text{C}$			50 $\mu\text{A}$ 3 mA
$R_{DS(on)}$	$V_{GS} = 10V$ , $I_D = 0.5 \cdot I_{D25}$ , Note 1			38 m $\Omega$

Symbol	Test Conditions (T <sub>J</sub> = 25°C, Unless Otherwise Specified)	Characteristic Values		
		Min.	Typ.	Max
g <sub>fs</sub>	V <sub>DS</sub> = 10V, I <sub>D</sub> = 0.5 • I <sub>D25</sub> , Note 1	33	55	S
R <sub>Gi</sub>	Gate Input Resistance		0.6	Ω
C <sub>iss</sub>	V <sub>GS</sub> = 0V, V <sub>DS</sub> = 25V, f = 1MHz		8300	pF
C <sub>oss</sub>			5010	pF
C <sub>rss</sub>			1.6	pF
Effective Output Capacitance				
C <sub>o(er)</sub>	Energy related	V <sub>GS</sub> = 0V V <sub>DS</sub> = 0.8 • V <sub>DSS</sub>	280	pF
C <sub>o(tr)</sub>	Time related		1160	pF
t <sub>d(on)</sub>	Resistive Switching Times V <sub>GS</sub> = 10V, V <sub>DS</sub> = 0.5 • V <sub>DSS</sub> , I <sub>D</sub> = 0.5 • I <sub>D25</sub> R <sub>G</sub> = 3Ω (External)		32	ns
t <sub>r</sub>			24	ns
t <sub>d(off)</sub>			70	ns
t <sub>f</sub>			11	ns
Q <sub>g(on)</sub>	V <sub>GS</sub> = 10V, V <sub>DS</sub> = 0.5 • V <sub>DSS</sub> , I <sub>D</sub> = 0.5 • I <sub>D25</sub>		140	nC
Q <sub>gs</sub>			50	nC
Q <sub>gd</sub>			40	nC
R <sub>thJC</sub>				0.14 °C/W
R <sub>thCS</sub>			0.21	°C/W
R <sub>thCS</sub>			0.15	°C/W

### Source-Drain Diode

Symbol	Test Conditions ( $T_J = 25^\circ\text{C}$ , Unless Otherwise Specified)	Characteristic Values		
		Min.	Typ.	Max
$I_S$	$V_{GS} = 0\text{V}$			80 A
$I_{SM}$	Repetitive, pulse Width Limited by $T_{JM}$			320 A
$V_{SD}$	$I_F = I_S$ , $V_{GS} = 0\text{V}$ , Note 1			1.4 V
$t_{rr}$	$I_F = 40\text{A}$ , $-di/dt = 100\text{A}/\mu\text{s}$ $V_R = 100\text{V}$		200	ns
$Q_{RM}$			1.7	$\mu\text{C}$
$I_{RM}$			16.7	A

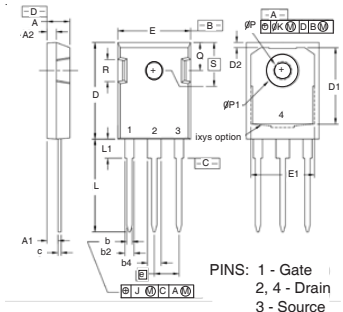
Note 1. Pulse test,  $t \leq 300\mu\text{s}$ , duty cycle,  $d \leq 2\%$ .

IXYS Reserves the Right to Change Limits, Test Conditions, and Dimensions.

IXYS MOSFETs and IGBTs are covered by one or more of the following U.S. patents:

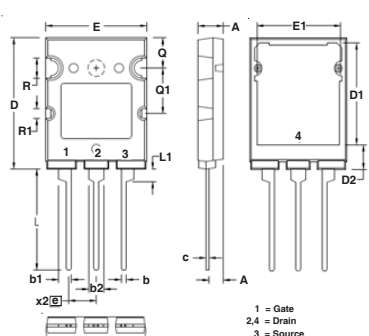
4,835,592	4,931,844	5,049,961	5,237,481	6,162,665	6,404,065B1	6,683,344	6,727,585	7,005,734B2	7,157,338B2
4,860,072	5,017,508	5,063,307	5,381,025	6,259,123B1	6,534,343	6,710,405B2	6,759,692	7,063,975B2	
4,881,106	5,034,796	5,187,117	5,486,715	6,306,728B1	6,583,505	6,710,463	6,771,478B2	7,071,537	

### TO-247 Outline



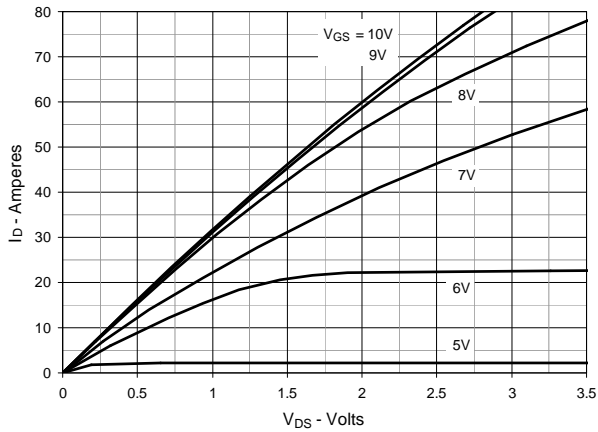
SYM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	.190	.205	4.83	5.21
A1	.090	.100	2.29	2.54
A2	.075	.085	1.91	2.16
b	.045	.055	1.14	1.40
b2	.075	.087	1.91	2.20
b4	.115	.126	2.92	3.20
C	.024	.031	0.61	0.80
D	.819	.840	20.80	21.34
D1	.650	.690	16.51	17.53
D2	.035	.050	0.89	1.27
E	.620	.635	15.75	16.13
E1	.545	.565	13.84	14.35
e	.215 BSC		5.45 BSC	
J	--	.010	--	0.25
K	--	.025	--	0.64
L	.780	.810	19.81	20.57
L1	.150	.170	3.81	4.32
øP	.140	.144	3.55	3.65
øP1	.275	.290	6.99	7.37
Q	.220	.244	5.59	6.20
R	.170	.190	4.32	4.83
S	.242 BSC		6.15 BSC	

### TO-264P Outline

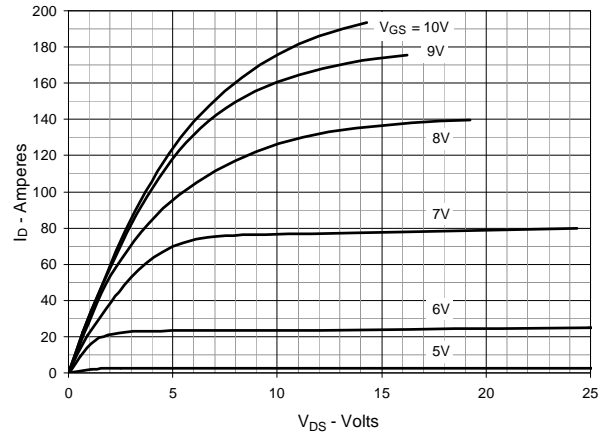


SYM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	.185	.209	4.70	5.30
A1	.102	.118	2.60	3.00
b	.035	.049	0.90	1.25
b1	.091	.106	2.30	2.70
b2	.110	.126	2.80	3.20
c	.020	.033	0.50	0.85
D	1.012	1.035	25.70	26.30
D1	.783	.799	19.90	20.30
D2	.185	.205	4.70	5.20
E	.776	.799	19.70	20.30
E1	.661	.677	16.80	17.20
e	.215 BSC		5.46 BSC	
L	.768	.807	19.50	20.50
L1	.091	.106	2.30	2.70
Q	.228	.244	5.80	6.20
Q1	.346	.362	8.80	9.20
øR	.150	.165	3.80	4.20
øR1	.071	.087	1.80	2.20

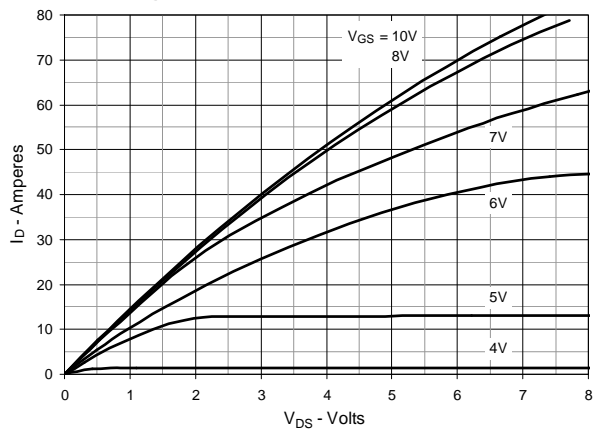
**Fig. 1. Output Characteristics @  $T_J = 25^\circ\text{C}$**



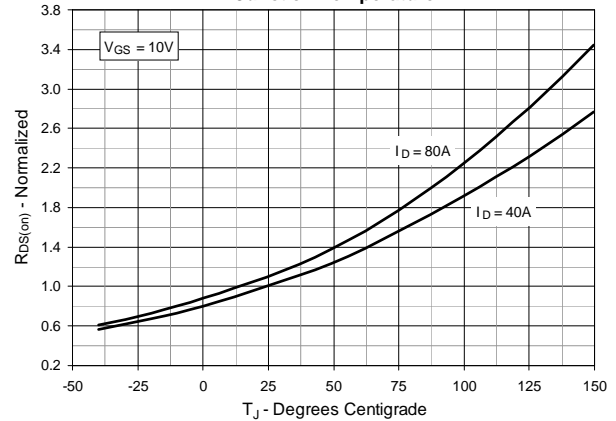
**Fig. 2. Extended Output Characteristics @  $T_J = 25^\circ\text{C}$**



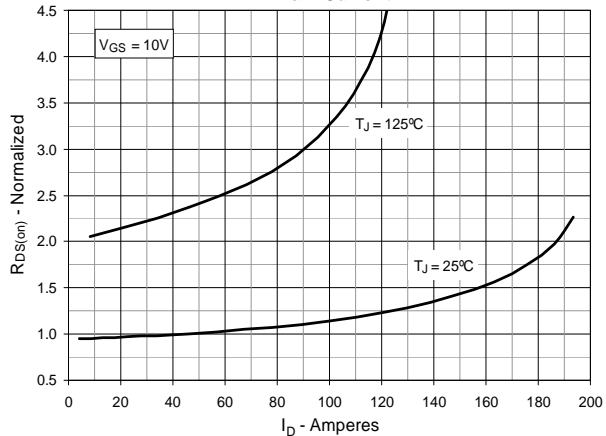
**Fig. 3. Output Characteristics @  $T_J = 125^\circ\text{C}$**



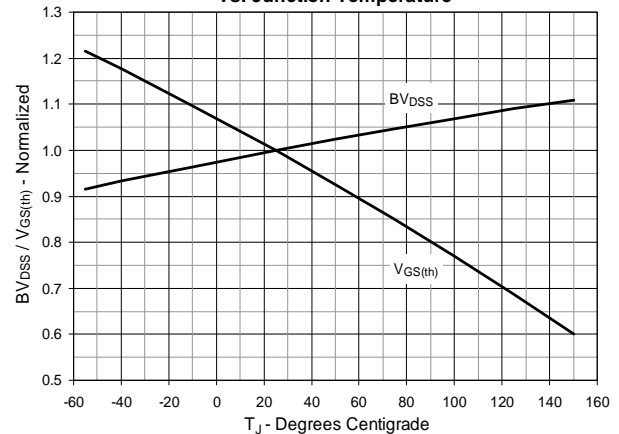
**Fig. 4.  $R_{DS(on)}$  Normalized to  $I_D = 40\text{A}$  Value vs. Junction Temperature**



**Fig. 5.  $R_{DS(on)}$  Normalized to  $I_D = 40\text{A}$  Value vs. Drain Current**

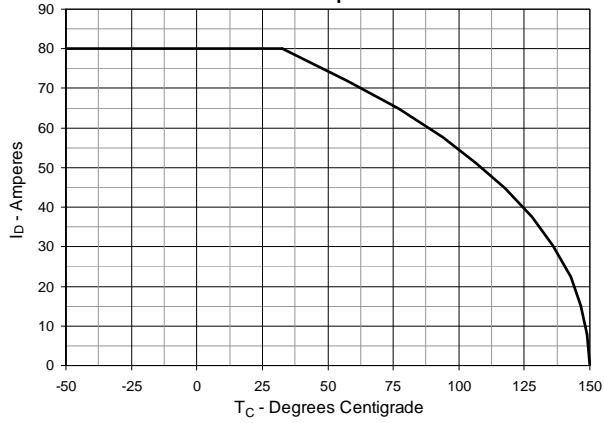


**Fig. 6. Normalized Breakdown & Threshold Voltages vs. Junction Temperature**

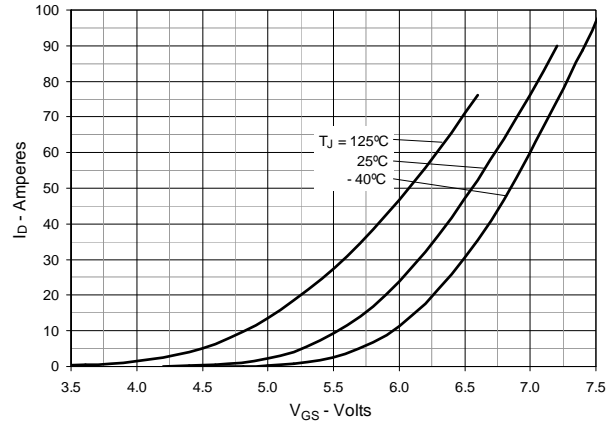




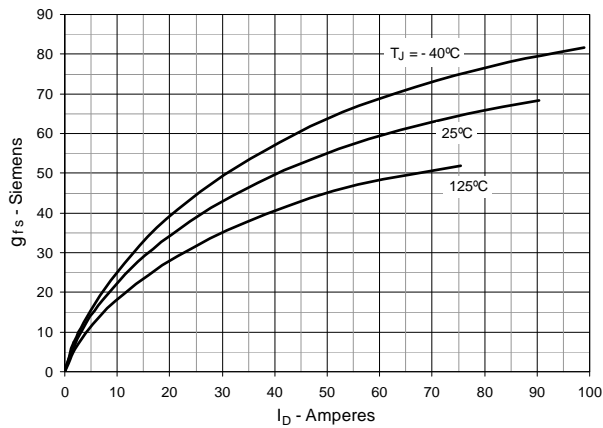
**Fig. 7. Maximum Drain Current vs. Case Temperature**



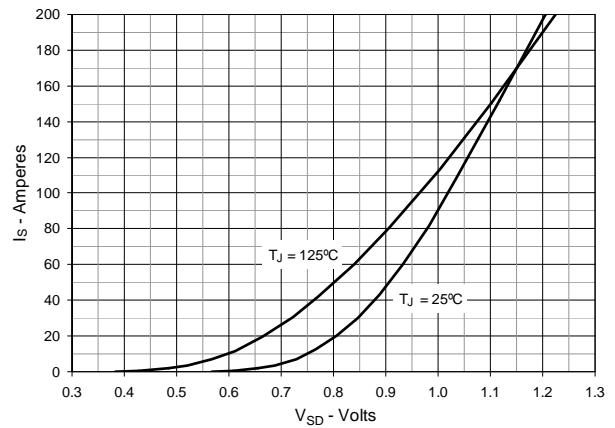
**Fig. 8. Input Admittance**



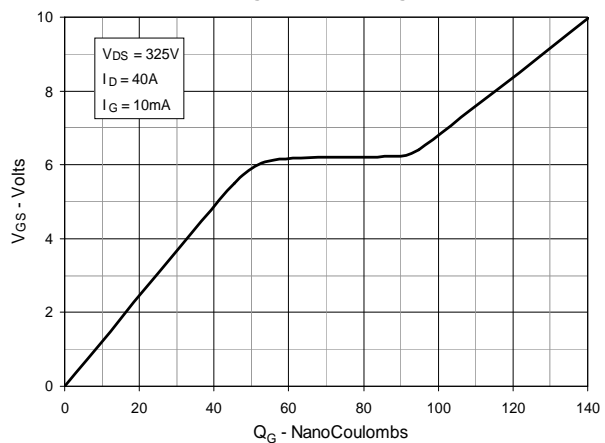
**Fig. 9. Transconductance**



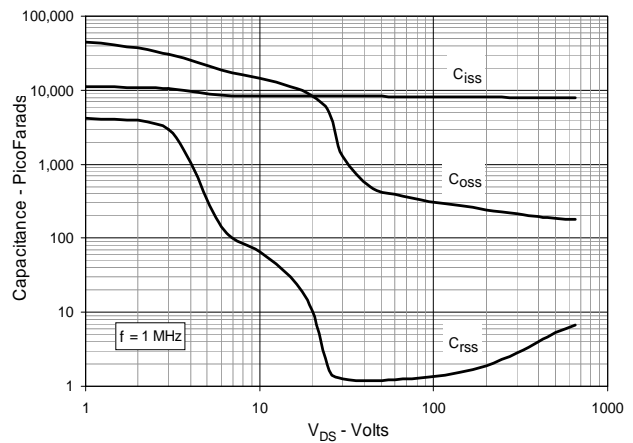
**Fig. 10. Forward Voltage Drop of Intrinsic Diode**



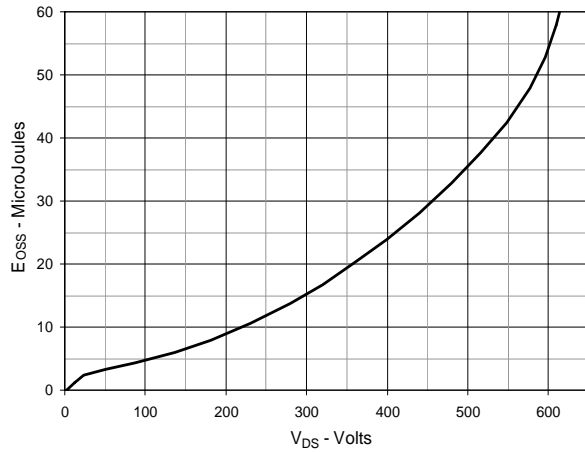
**Fig. 11. Gate Charge**



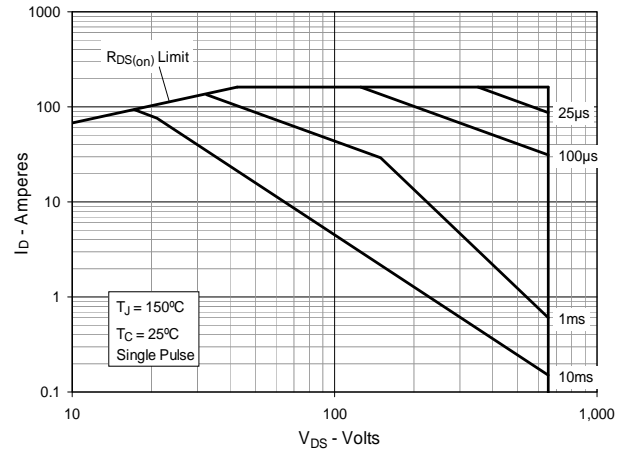
**Fig. 12. Capacitance**



**Fig. 13. Output Capacitance Stored Energy**



**Fig. 14. Forward-Bias Safe Operating Area**



**Fig. 15. Maximum Transient Thermal Impedance**

