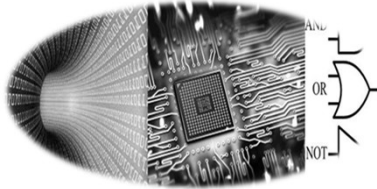


DIGITAL ELECTRONICS

Lecture 05: Karnaugh Map (K-Map)



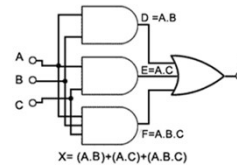
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Date: Jan-Jun, 2019

Why Minterm and Maxterm?

- Developing a digital logic circuit using logic gates is very straight forward when a Boolean Expression is given.

$$X = (A.B) + (A.C) + (A.B.C)$$



$$X = (A.B) + (A.C) + (A.B.C)$$

- Even for a large or complex BE the design is possible

$$f(A,B,C) = \overline{AB + ABC + BCD + \bar{E}(A+B) + A\bar{B}} F$$

IS IT POSSIBLE TO DESIGN A DIGITAL CIRCUIT FROM A TRUTH TABLE?





Why Minterm and Maxterm?

- The function of each logic gates could be possible to represent through its Truth Table.

Hence:

Designing a digital circuit from a Truth Table could be possible for a known or standard digital logic.

- Thus, if a Truth table is provided the gate could be designed.

	2-input OR gate	<table border="1"><tr><th>A</th><th>B</th><th>Y</th></tr><tr><td>0</td><td>0</td><td>0</td></tr><tr><td>0</td><td>1</td><td>1</td></tr><tr><td>1</td><td>0</td><td>1</td></tr><tr><td>1</td><td>1</td><td>1</td></tr></table>	A	B	Y	0	0	0	0	1	1	1	0	1	1	1	1	$Y = A + B$
A	B	Y																
0	0	0																
0	1	1																
1	0	1																
1	1	1																
	2-input NAND gate	<table border="1"><tr><th>A</th><th>B</th><th>Y</th></tr><tr><td>0</td><td>0</td><td>1</td></tr><tr><td>0</td><td>1</td><td>1</td></tr><tr><td>1</td><td>0</td><td>1</td></tr><tr><td>1</td><td>1</td><td>0</td></tr></table>	A	B	Y	0	0	1	0	1	1	1	0	1	1	1	0	$Y = \overline{A \cdot B}$
A	B	Y																
0	0	1																
0	1	1																
1	0	1																
1	1	0																
	2-input AND gate	<table border="1"><tr><th>A</th><th>B</th><th>Y</th></tr><tr><td>0</td><td>0</td><td>0</td></tr><tr><td>0</td><td>1</td><td>0</td></tr><tr><td>1</td><td>0</td><td>0</td></tr><tr><td>1</td><td>1</td><td>1</td></tr></table>	A	B	Y	0	0	0	0	1	0	1	0	0	1	1	1	$Y = A \cdot B$
A	B	Y																
0	0	0																
0	1	0																
1	0	0																
1	1	1																
	2-input EX-NOR gate	<table border="1"><tr><th>A</th><th>B</th><th>Y</th></tr><tr><td>0</td><td>0</td><td>1</td></tr><tr><td>0</td><td>1</td><td>0</td></tr><tr><td>1</td><td>0</td><td>0</td></tr><tr><td>1</td><td>1</td><td>1</td></tr></table>	A	B	Y	0	0	1	0	1	0	1	0	0	1	1	1	$Y = \overline{A \oplus B}$
A	B	Y																
0	0	1																
0	1	0																
1	0	0																
1	1	1																

Circuit Design from Truth Table

- But, if a unknown or non-standard or large truth table is provided then designing a digital circuit need Minterm and Maxterm based Boolean representation

A	B	C	X
0	0	0	1
0	0	1	1
0	1	0	1
0	1	1	0
1	0	0	0
1	0	1	1
1	1	0	0
1	1	1	1

Digital Ckt ?

Minterms or Maxterms from Truth Table

- From the truth table Minterms or Maxterms are found and accordingly the SSOP or SPOS are formed.
- Using the SSOP or SPOS, the digital circuits are formed.

A	B	C	X
0	0	0	1
0	0	1	1
0	1	0	1
0	1	1	0
1	0	0	0
1	0	1	1
1	1	0	0
1	1	1	1

$\overline{A} \overline{B} \overline{C}$
 Minterms (m_i)
 or
 Maxterms (M_i)
 $A + B + C$

Minterm and SSOP from Truth Table

- From the truth table Minterms or Maxterms are found and accordingly the SSOP or SPOS are formed.
- Using the SSOP or SPOS, the digital circuits are formed.

A	B	C	X	Minterms	Product Terms
0	0	0	1	m_0	$\overline{A} \overline{B} \overline{C}$
0	0	1	1	m_1	$\overline{A} \overline{B} C$
0	1	0	1	m_2	$\overline{A} B \overline{C}$
0	1	1	0	m_3	$\overline{A} B C$
1	0	0	0	m_4	$A \overline{B} \overline{C}$
1	0	1	1	m_5	$A \overline{B} C$
1	1	0	0	m_6	$A B \overline{C}$
1	1	1	1	m_7	$A B C$

Minterm and SSOP from Truth Table

- Check the 1s in the output (X) variable.
- Corresponding product terms will be there in SSOP.

A	B	C	X	Minterms	Product Terms
0	0	0	1	m_0	$\bar{A}\bar{B}\bar{C}$ ✓
0	0	1	1	m_1	$\bar{A}\bar{B}C$ ✓
0	1	0	1	m_2	$\bar{A}B\bar{C}$ ✓
0	1	1	0	m_3	$\bar{A}BC$ ✗
1	0	0	0	m_4	$A\bar{B}\bar{C}$ ✗
1	0	1	1	m_5	$A\bar{B}C$ ✓
1	1	0	0	m_6	$AB\bar{C}$ ✗
1	1	1	1	m_7	ABC ✓

$$f(A,B,C) = \bar{A}\bar{B}\bar{C} + \bar{A}\bar{B}C + \bar{A}B\bar{C} + A\bar{B}C + ABC$$

Maxterm and SPOS from Truth Table

- From the truth table Maxterms are found and accordingly the SPOS are formed.
- Using the SSOP or SPOS, the digital circuits are formed.

A	B	C	X	Maxterms	Sum Terms
0	0	0	1	m_0	$A+B+C$
0	0	1	1	m_1	$A+B+\bar{C}$
0	1	0	1	m_2	$A+\bar{B}+C$
0	1	1	0	m_3	$A+\bar{B}+\bar{C}$
1	0	0	0	m_4	$\bar{A}+B+C$
1	0	1	1	m_5	$\bar{A}+B+\bar{C}$
1	1	0	0	m_6	$\bar{A}+\bar{B}+C$
1	1	1	1	m_7	$\bar{A}+\bar{B}+\bar{C}$

Maxterm and SPOS from Truth Table

- Check the 1s in the output (X) variable.
- Corresponding product terms will be there in SPOS.

A	B	C	X	Minterms	Product Terms
0	0	0	1	M_0	$A+B+C$ ✗
0	0	1	1	M_1	$A+B+\bar{C}$ ✗
0	1	0	1	M_2	$A+\bar{B}+C$ ✗
0	1	1	0	M_3	$A+\bar{B}+\bar{C}$ ✓
1	0	0	0	M_4	$\bar{A}+B+C$ ✓
1	0	1	1	M_5	$\bar{A}+B+\bar{C}$ ✗
1	1	0	0	M_6	$\bar{A}+\bar{B}+C$ ✓
1	1	1	1	M_7	$\bar{A}+\bar{B}+\bar{C}$ ✗

$$f(A,B,C) = (A+\bar{B}+\bar{C})(\bar{A}+B+C)(\bar{A}+\bar{B}+C)$$

Maxterm and SPOS from Truth Table

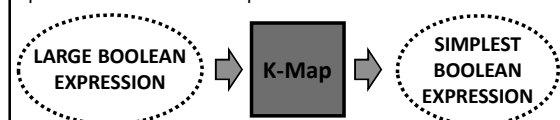
- From the truth table Minterms or Maxterms are found and accordingly the SSOP or SPOS are formed.
- Using the SSOP or SPOS, the digital circuits are formed.

A	B	C	X	Maxterms	Sum Terms
0	0	0	1	m_0	$A+B+C$
0	0	1	1	m_1	$A+B+\bar{C}$
0	1	0	1	m_2	$A+\bar{B}+C$
0	1	1	0	m_3	$A+\bar{B}+\bar{C}$
1	0	0	0	m_4	$\bar{A}+B+C$
1	0	1	1	m_5	$\bar{A}+B+\bar{C}$
1	1	0	0	m_6	$\bar{A}+\bar{B}+C$
1	1	1	1	m_7	$\bar{A}+\bar{B}+\bar{C}$

K-Map

Karnaugh Map (K-Map)

- Karnaugh Mapping is used to minimize the number of logic gates that are required in a digital circuit.
- This will replace Boolean reduction when the circuit is large.
- Write the Boolean equation in a SOP form first and then place each term on a map.



• Karnaugh Mapping

1. SOP-Based (BE should be in SSOP form)
2. POS-Based (BE should be in SPOS form)

Karnaugh Map (K-Map)

- The map is made up of a table of every possible SOP using the number of variables that are being used.
 - 2-variable function: 2×2 map
 - 3-variable function: 4×2 map
 - 4-variable function: 4×4 map
 - 5-variable function: 2×(4×4) map
 - 6-variable function: 4×(4×4) map
 - 7-variable function: K-Map is not used

K-Map SOP Minimization

Logic Simplification With Karnaugh Maps

- The logic simplification examples that we have done so far could have been performed with Boolean algebra about as quickly.
- Real world logic simplification problems call for larger Karnaugh maps so that we may do serious work.
- We will work some contrived examples in this section, leaving most of the real world applications for the Combinatorial Logic chapter.
- By contrived, we mean examples which illustrate techniques.
- This approach will develop the tools we need to transition to the more complex applications in the Combinatorial Logic chapter.

Karnaugh Map Representation of BE

- The K-Map could be represented either with SSOP i.e. either with Minterms (m_n).

$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + A\overline{B}\overline{C} + ABC$$

$$f(A,B,C) = m_3 + m_4 + m_5 + m_7 = \sum m(3,4,5,7)$$

- The K-Map could be represented either with SPOS i.e. with Maxterms (M_n).

$$f(A,B,C) = (A+\overline{B}+C)(\overline{A}+B+C)(A+\overline{B}+\overline{C})(A+B+C)$$

$$f(A,B,C) = M_2 + M_4 + M_5 + M_0 = \sum m(0,2,3,4)$$

Karnaugh Map Representation

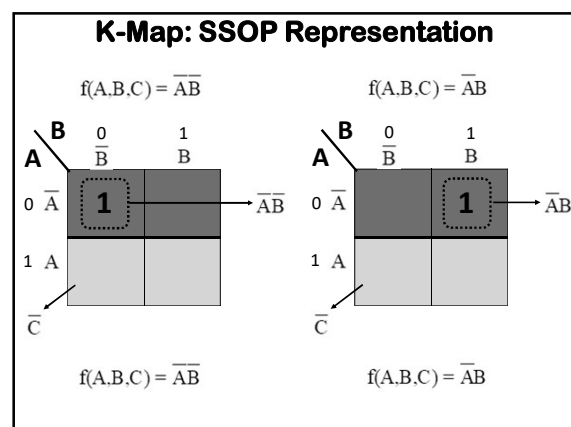
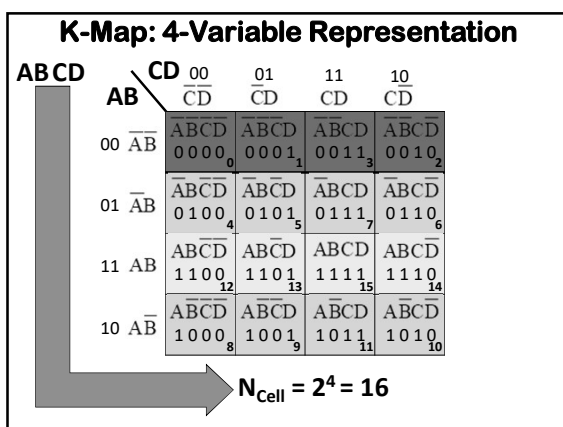
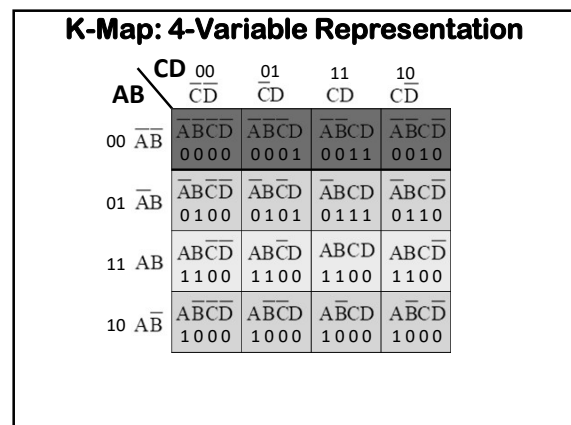
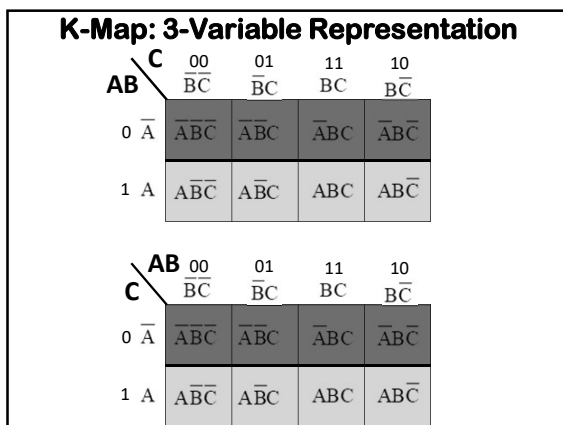
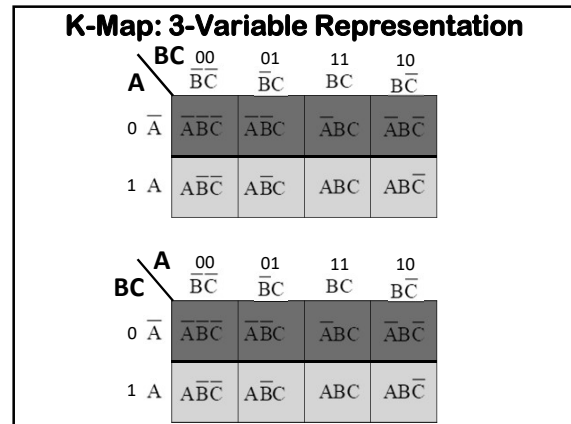
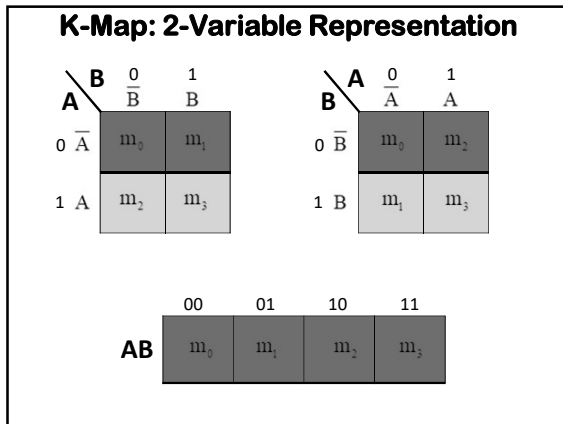
- The K-Map is represented by cells which represent each product term or min term in a SSOP.
- Total cells is given by
 - The 2-variable K-Map has: $N_{\text{cell}} = 2^2 = 4$
 - The 3-variable K-Map has: $N_{\text{cell}} = 2^3 = 8$
 - The 4-variable K-Map has: $N_{\text{cell}} = 2^4 = 16$
 - The 5-variable K-Map has: $N_{\text{cell}} = 2^5 = 32$
 - The 6-variable K-Map has: $N_{\text{cell}} = 2^6 = 64$

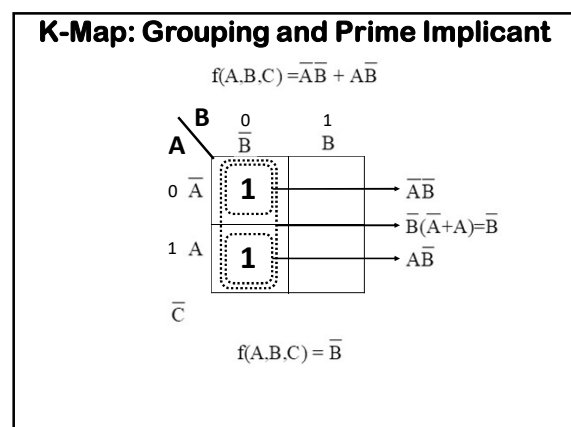
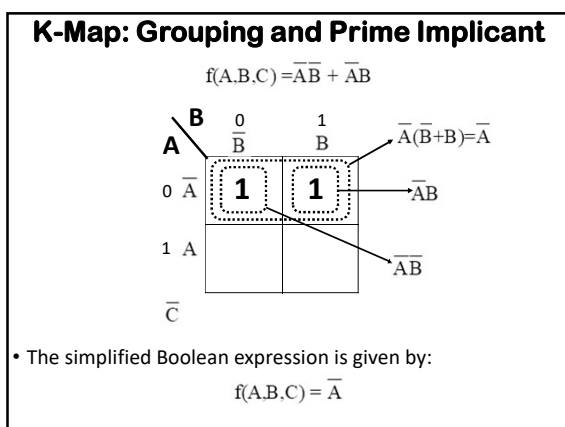
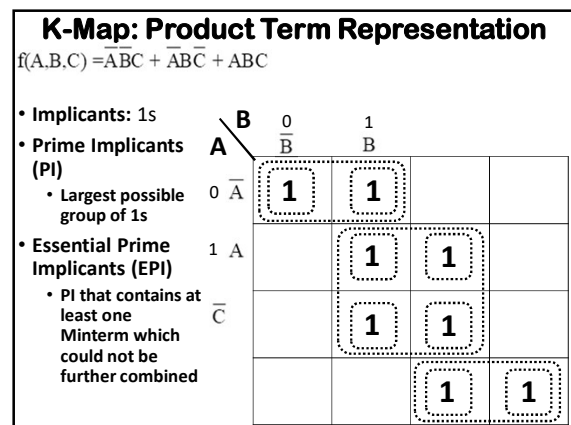
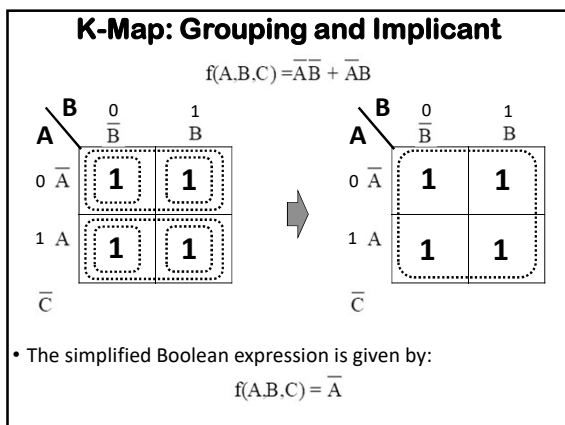
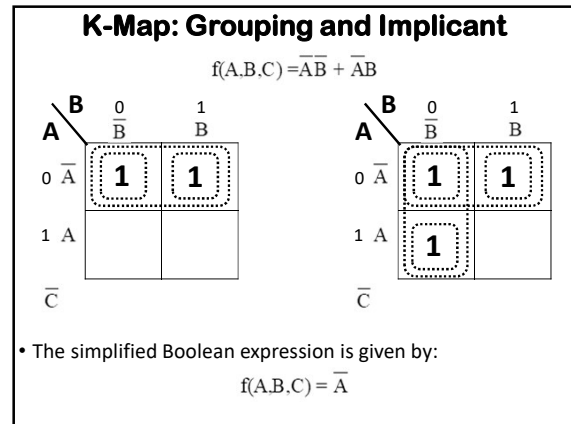
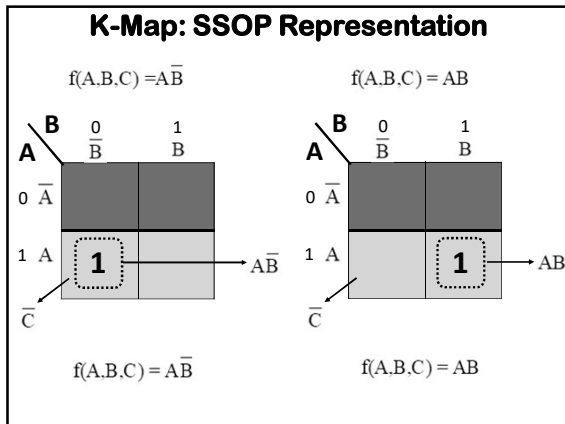
K-Map: 2-Variable Representation

		B	
		0	1
A	0	$\overline{A}\overline{B}$ m_0	$\overline{A}B$ m_1
	1	$A\overline{B}$ m_2	AB m_3

		A	
		0	1
B	0	$\overline{B}\overline{A}$ m_0	$\overline{B}A$ m_1
	1	$B\overline{A}$ m_2	BA m_3

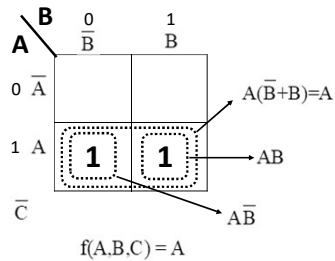
AB				
	00	01	10	11
0	$\overline{A}\overline{B}$ m_0	$\overline{A}B$ m_1	$A\overline{B}$ m_2	AB m_3



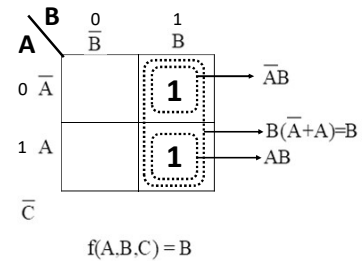


K-Map: Grouping and Prime Implicant

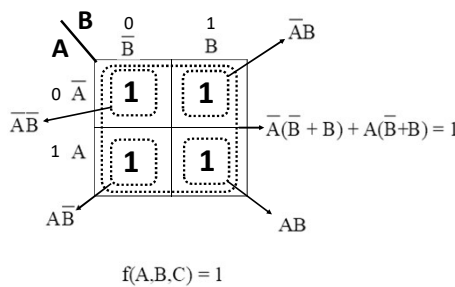
$$f(A,B,C) = \bar{A}\bar{B} + AB$$

**K-Map: Grouping and Prime Implicant**

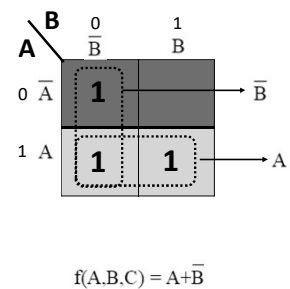
$$f(A,B,C) = \bar{A}\bar{B} + AB$$

**K-Map: Grouping and Prime Implicant**

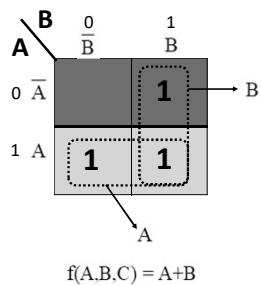
$$f(A,B,C) = \bar{A}\bar{B} + \bar{A}B + A\bar{B} + AB$$

**K-Map Simplification: Example 01**

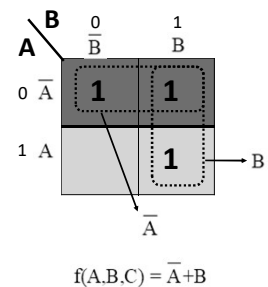
$$f(A,B,C) = \bar{A}\bar{B} + \bar{A}B + AB$$

**K-Map Simplification: Example 02**

$$f(A,B,C) = \bar{A}\bar{B} + \bar{A}B + AB$$

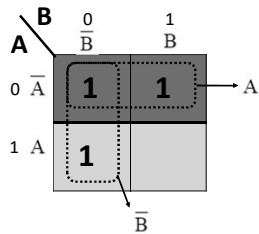
**K-Map Simplification: Example 02**

$$f(A,B,C) = \bar{A}\bar{B} + \bar{A}B + AB$$



K-Map Simplification: Example 02

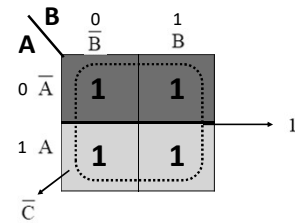
$$f(A,B,C) = \overline{A}\overline{B} + \overline{A}B + A\overline{B}$$



$$f(A,B,C) = A + \overline{B}$$

K-Map Simplification: Example 03

$$f(A,B,C) = \overline{A}\overline{B} + \overline{A}B + A\overline{B} + AB$$



$$f(A,B,C) = 1$$

2 Variables Karnaugh Map

	\overline{B}	B
\overline{A}	0	1
A	2	3

Notice that the map is going false to true, left to right and top to bottom

The upper right hand cell is $\overline{A}B$ if $X = \overline{A}B$ then put an X in that cell

	\overline{B}	B
\overline{A}		1
A		

This show the expression true when $A = 0$ and $B = 0$

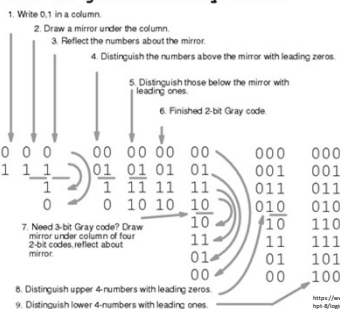
3 Variables Karnaugh Map

	$\overline{B}\overline{C}$	$\overline{B}C$	BC	$B\overline{C}$
\overline{A}				
A				

	00	01	11	10
\overline{A}				
A				

Generating Gray Code

- If we sketch our own Karnaugh maps, we need to generate Gray code for any size map that we may use. This is how we generate Gray code of any size.

How to generate Gray code.**3 Variables Karnaugh Map**

Gray Code	0	1
	\overline{C}	C
00	$\overline{A}\overline{B}$	
01	$\overline{A}B$	
11	AB	
10	$A\overline{B}$	

2 Variables Karnaugh Map: Grouping

If $X = \overline{A}\overline{B} + A\overline{B}$ then
put an X in both of
these cells

	\overline{B}	B
\overline{A}	1	
A	1	

From Boolean reduction we know that $\overline{A}\overline{B} + A\overline{B} = \overline{B}$

From the Karnaugh map we
can circle adjacent cell and
find that $X = \overline{B}$

	\overline{B}	B
\overline{A}	1	
A	1	

3 Variables Karnaugh Map (cont'd)

$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + A\overline{B}\overline{C} + ABC$$

Gray Code		0	1
	\overline{C}	C	
00	$\overline{A}\overline{B}$	1	1
01	$\overline{A}B$		
11	AB		
10	$A\overline{B}$	1	1

One
simplification
could be
 $X = \overline{A}\overline{B} + A\overline{B}$

3 Variables Karnaugh Map (cont'd)

$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + A\overline{B}\overline{C} + ABC$$

Gray Code		0	1
	\overline{C}	C	
00	$\overline{A}\overline{B}$	1	1
01	$\overline{A}B$		
11	AB		
10	$A\overline{B}$	1	1

Another
simplification
could be
 $X = \overline{B}\overline{C} + \overline{B}C$
A Karnaugh
Map does wrap
around

3 Variables Karnaugh Map (cont'd)

$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + A\overline{B}\overline{C} + ABC$$

Gray Code		0	1
	\overline{C}	C	
00	$\overline{A}\overline{B}$	1	1
01	$\overline{A}B$		
11	AB		
10	$A\overline{B}$	1	1

The Best
simplification
would be
 $X = \overline{B}$

3 Variables Karnaugh Map

$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C}$$

$$\text{Out} = \overline{A}\overline{B}C + \overline{A}B\overline{C}$$

BC	00	01	11	10
0	1	1		
1				

$$\text{Out} = \overline{A}\overline{B}$$

3 Variables Karnaugh Map

$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + \overline{A}BC + \overline{A}\overline{B}\overline{C}$$

$$\text{Out} = \overline{A}\overline{B}C + \overline{A}B\overline{C} + \overline{A}BC + \overline{A}\overline{B}\overline{C}$$

BC	00	01	11	10
0	1	1	1	1
1				

$$\text{Out} = \overline{A}$$

3 Variables Karnaugh Map

$$f(A,B,C) = \bar{A}\bar{B}C + \bar{A}BC + A\bar{B}C + ABC$$

$$\text{Out} = \bar{A}\bar{B}C + \bar{A}BC + A\bar{B}C + ABC$$

A \ BC	00	01	11	10
0		1	1	
1		1	1	

$$\text{Out} = C$$

3 Variables Karnaugh Map

$$f(A,B,C) = \bar{A}\bar{B}\bar{C} + \bar{A}\bar{B}C + \bar{A}BC + \bar{A}B\bar{C} + ABC + AB\bar{C}$$

$$\text{Out} = \bar{A}\bar{B}\bar{C} + \bar{A}\bar{B}C + \bar{A}BC + \bar{A}B\bar{C} + ABC + AB\bar{C}$$

A \ BC	00	01	11	10
0	1	1	1	1
1			1	1

$$\text{Out} = \bar{A} + B$$

3 Variables Karnaugh Map

$$f(A,B,C) = \bar{A}BC + \bar{A}B\bar{C} + ABC + AB\bar{C}$$

$$\text{Out} = \bar{A}BC + \bar{A}B\bar{C} + ABC + AB\bar{C}$$

A \ BC	00	01	11	10
0			1	1
1			1	1

$$\text{Out} = B$$

3 Variables Karnaugh Map

$$f(A,B,C) = \bar{A}\bar{B}\bar{C} + \bar{A}\bar{B}C + \bar{A}B\bar{C} + AB\bar{C}$$

$$\text{Out} = \bar{A}\bar{B}\bar{C} + \bar{A}\bar{B}C + \bar{A}B\bar{C} + AB\bar{C}$$

A \ BC	00	01	11	10
0	1			1
1	1			1

$$\text{Out} = \bar{C}$$

3 Variables Karnaugh Map

$$f(A,B,C) = \bar{A}\bar{B}\bar{C} + \bar{A}\bar{B}C + \bar{A}BC + \bar{A}B\bar{C} + A\bar{B}\bar{C} + ABC$$

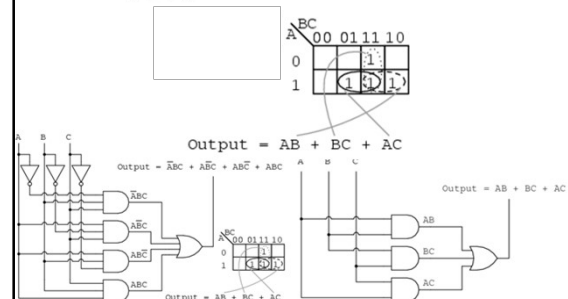
$$\text{Out} = \bar{A}\bar{B}\bar{C} + \bar{A}\bar{B}C + \bar{A}BC + \bar{A}B\bar{C} + A\bar{B}\bar{C} + ABC$$

A \ BC	00	01	11	10
0	1	1	1	1
1	1			1

$$\text{Out} = \bar{A} + \bar{C}$$

3 Variables Karnaugh Map

$$f(A,B,C) = \bar{A}BC + \bar{A}\bar{B}C + A\bar{B}\bar{C} + ABC$$



3 Variables Karnaugh Map

$f(A,B,C) = \bar{A}BC + A\bar{B}C + AB\bar{C} + ABC$

	BC	00	01	11	10
A	0			1	
	1	1	1	1	1

Output = $AB + BC + AC$

$\bar{A}BC + A\bar{B}C + AB\bar{C} + ABC$
 Factoring BC out of 1st and 4th terms
 $BC(\bar{A} + A) + A\bar{B}C + AB\bar{C}$
 Applying identity $A + \bar{A} = 1$
 $BC(1) + A\bar{B}C + AB\bar{C}$
 $BC + A\bar{B}C + AB\bar{C}$
 Applying identity $1A = A$
 $BC + A\bar{B}C + AB\bar{C}$
 Factoring B out of 1st and 3rd terms
 $B(C + A\bar{C}) + AB\bar{C}$
 Applying rule $A + \bar{A}B = A + B$ to the C + $A\bar{C}$ term
 $B(C + A) + AB\bar{C}$
 Distributing terms
 $BC + AB + A\bar{B}C$
 Factoring A out of 2nd and 3rd terms
 $BC + A(B + \bar{B}C)$
 Applying rule $A + \bar{A}B = A + B$ to the B + $\bar{B}C$ term
 $BC + A(B + C)$
 Distributing terms
 $BC + AB + AC$
 or
 Simplified result
 $AB + BC + AC$

On a 3 Variables Karnaugh Map

- One cell requires 3 Variables
- Two adjacent cells require 2 variables
- Four adjacent cells require 1 variable
- Eight adjacent cells is a 1

4 Variables Karnaugh Map

$f(A,B,C,D) = \bar{A}\bar{B}CD + \bar{A}BCD + A\bar{B}C\bar{D} + ABC\bar{D} + AB\bar{C}D + \bar{A}BC\bar{D}$

Gray Code 00 01 11 10

		$\bar{C}\bar{D}$	$\bar{C}D$	CD	$C\bar{D}$
00	$\bar{A}\bar{B}$			1	
01	$\bar{A}B$			1	
11	AB		1	1	
10	$A\bar{B}$		1	1	

Now try it with Boolean reductions

$X = ABD + \bar{A}BC + CD$

4 Variables Karnaugh Map

$f(A,B,C,D) = \bar{A}\bar{B}CD + \bar{A}BCD + A\bar{B}C\bar{D} + ABC\bar{D} + AB\bar{C}D + \bar{A}BC\bar{D}$

Gray Code 00 01 11 10

		$\bar{C}\bar{D}$	$\bar{C}D$	CD	$C\bar{D}$
00	$\bar{A}\bar{B}$			1	
01	$\bar{A}B$			1	
11	AB		1	1	
10	$A\bar{B}$		1	1	

Now try it with Boolean reductions

$f(A,B,C,D) = ABD + CD + \bar{A}BC$

3 Variables K-Map: Example 01

$f(A,B,C) = \bar{A}BC + A\bar{B}C + A\bar{B}\bar{C} + A\bar{B}C$

	BC	00	01	11	10
		$\bar{B}\bar{C}$	$\bar{B}C$	BC	$B\bar{C}$
A	0 \bar{A}			1	1
	1 A	1	1		

$\bar{A}B$
 AB
 $f(A,B,C) = \bar{A}B + AB$
 $f(A,B,C) = A \oplus B$

3 Variables K-Map: Example 01

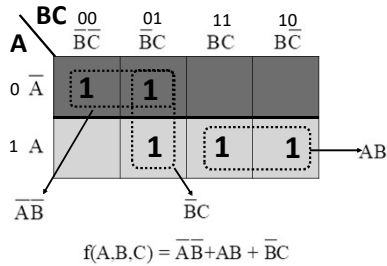
$f(A,B,C) = \bar{A}\bar{B}C + \bar{A}B\bar{C} + ABC + A\bar{B}C$

	BC	00	01	11	10
		$\bar{B}\bar{C}$	$\bar{B}C$	BC	$B\bar{C}$
A	0 \bar{A}	1	1		
	1 A			1	1

$\bar{A}B$
 AB
 $f(A,B,C) = \bar{A}B + AB$
 $f(A,B,C) = A \oplus B$

3 Variables K-Map: Example 01

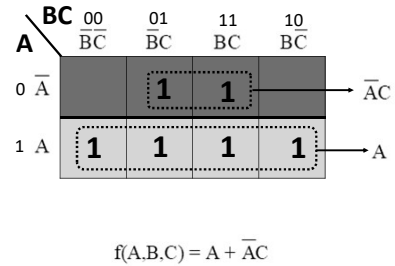
$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + A\overline{B}\overline{C} + ABC + A\overline{B}C$$



Ref: T15

3 Variables K-Map: Example 01

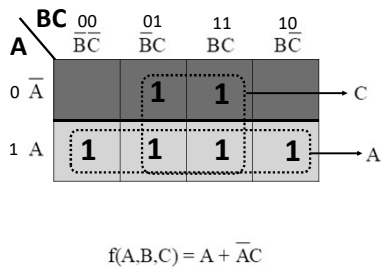
$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + A\overline{B}\overline{C} + A\overline{B}C + ABC + A\overline{B}C$$



Ref: T15

3 Variables K-Map: Example 01

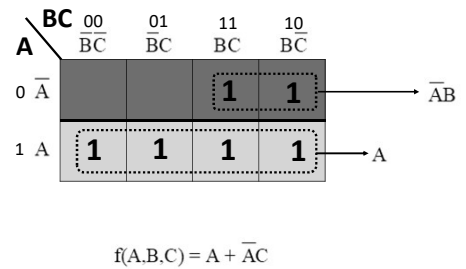
$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + A\overline{B}\overline{C} + A\overline{B}C + ABC + A\overline{B}C$$



Ref: T15

3 Variables K-Map: Example 01

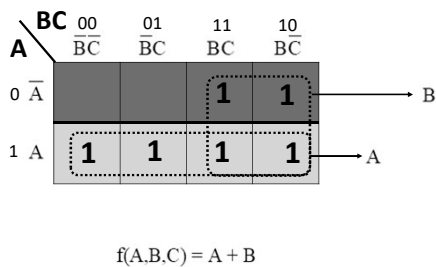
$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + A\overline{B}\overline{C} + A\overline{B}C + ABC + A\overline{B}C$$



Ref: T15

3 Variables K-Map: Example 01

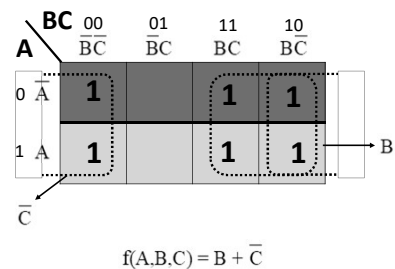
$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + A\overline{B}\overline{C} + A\overline{B}C + ABC + A\overline{B}C$$



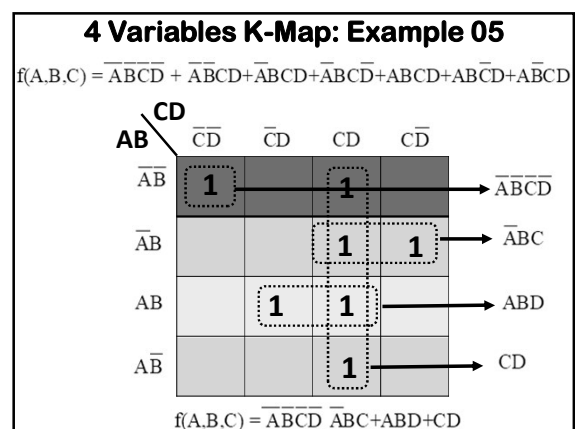
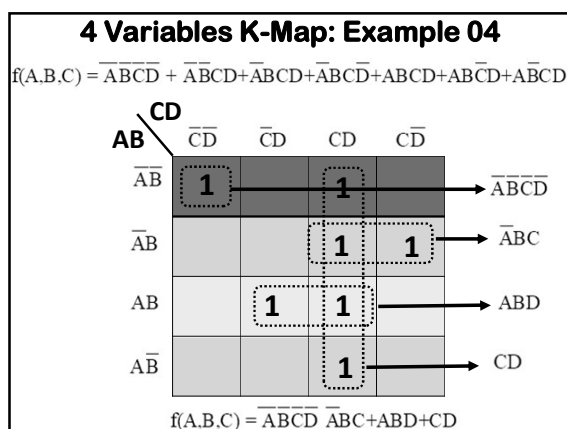
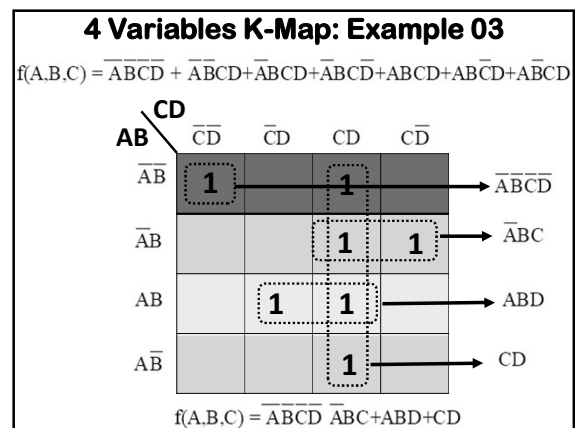
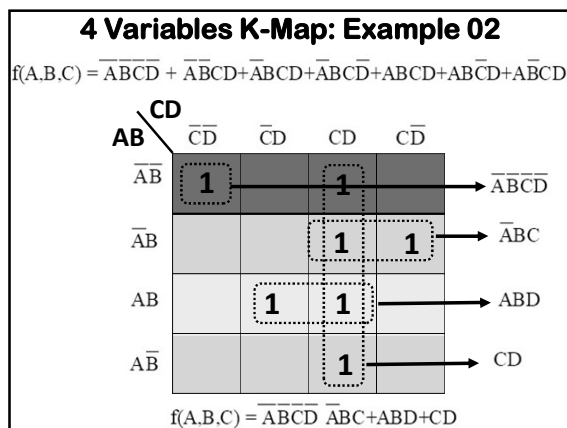
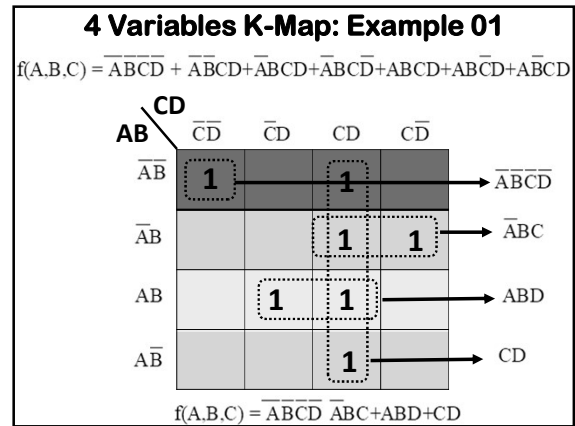
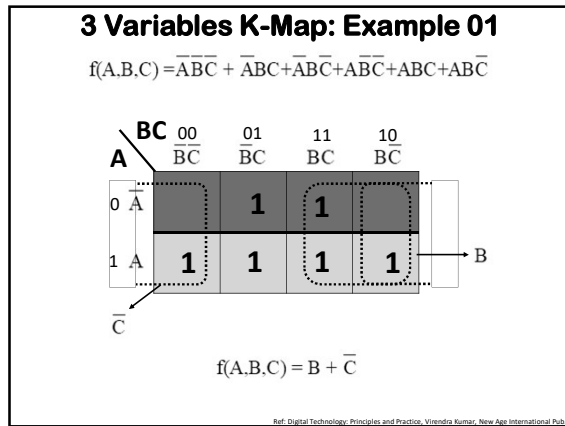
Ref: T15

2 Variables K-Map: Example 01

$$f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + A\overline{B}\overline{C} + A\overline{B}C + ABC + A\overline{B}C$$

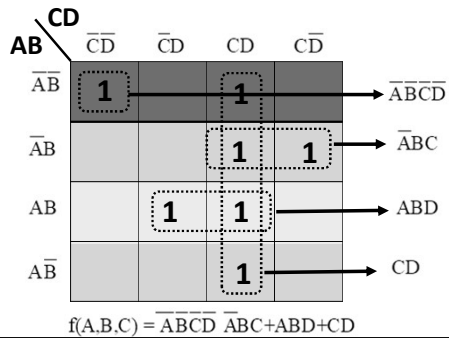


Ref: Digital Technology: Principles and Practice, Virendra Kumar, New Age International Pub.

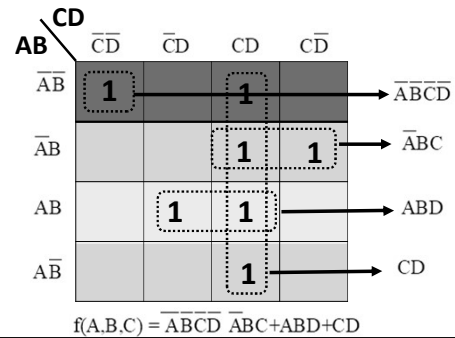


4 Variables K-Map: Example 06

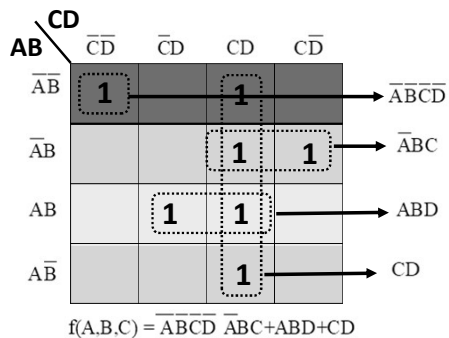
$$f(A,B,C) = \overline{A}\overline{B}\overline{C}\overline{D} + \overline{A}\overline{B}C\overline{D} + \overline{A}B\overline{C}\overline{D} + \overline{A}BC\overline{D} + \overline{A}B\overline{C}D + \overline{A}BCD + A\overline{B}\overline{C}\overline{D} + A\overline{B}C\overline{D}$$

**4 Variables K-Map: Example 07**

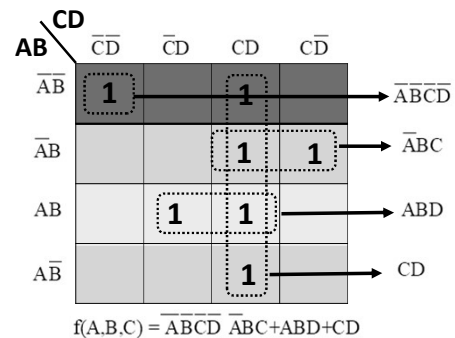
$$f(A,B,C) = \overline{A}\overline{B}\overline{C}\overline{D} + \overline{A}\overline{B}C\overline{D} + \overline{A}B\overline{C}\overline{D} + \overline{A}BC\overline{D} + \overline{A}B\overline{C}D + \overline{A}BCD + A\overline{B}\overline{C}\overline{D} + A\overline{B}C\overline{D}$$

**4 Variables K-Map: Example 08**

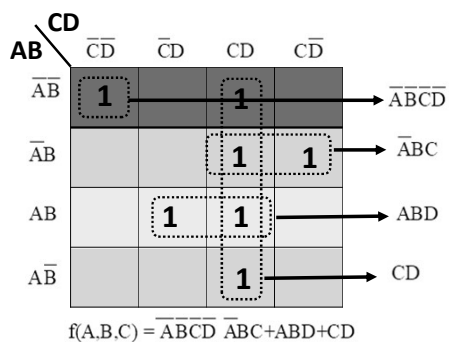
$$f(A,B,C) = \overline{A}\overline{B}\overline{C}\overline{D} + \overline{A}\overline{B}C\overline{D} + \overline{A}B\overline{C}\overline{D} + \overline{A}BC\overline{D} + \overline{A}B\overline{C}D + \overline{A}BCD + A\overline{B}\overline{C}\overline{D} + A\overline{B}C\overline{D}$$

**4 Variables K-Map: Example 09**

$$f(A,B,C) = \overline{A}\overline{B}\overline{C}\overline{D} + \overline{A}\overline{B}C\overline{D} + \overline{A}B\overline{C}\overline{D} + \overline{A}BC\overline{D} + \overline{A}B\overline{C}D + \overline{A}BCD + A\overline{B}\overline{C}\overline{D} + A\overline{B}C\overline{D}$$

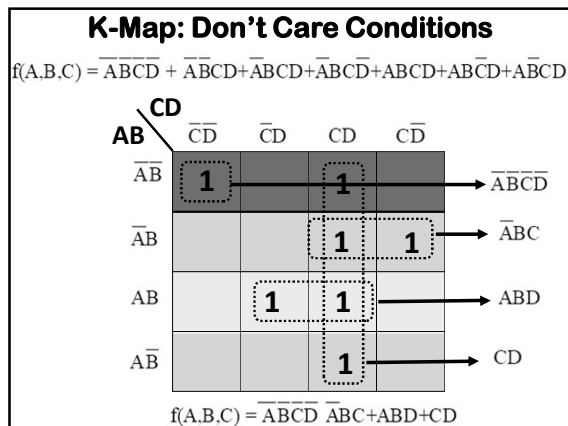
**4 Variables K-Map: Example 10**

$$f(A,B,C) = \overline{A}\overline{B}\overline{C}\overline{D} + \overline{A}\overline{B}C\overline{D} + \overline{A}B\overline{C}\overline{D} + \overline{A}BC\overline{D} + \overline{A}B\overline{C}D + \overline{A}BCD + A\overline{B}\overline{C}\overline{D} + A\overline{B}C\overline{D}$$

**K-Map: Don't Care Conditions**

- In some logic circuits certain input conditions never occur.
- Therefore, corresponding outputs never appear.
- In such cases outputs are not defined.
- Outputs either be HIGH or LOW.
- Such outputs conditions are represented as "Don't-Care" conditions in the Boolean Algebra.
- Output for the "Don't-Care" conditions in the Truth table and K-Map is represented as "x".
- In the minterms based BE the function with the "Don't-Care" conditions could be represented as follows:

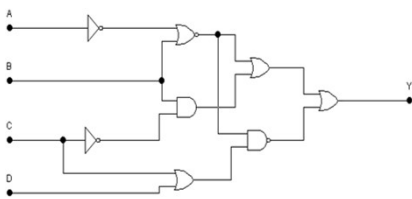
$$f(A,B,C) = \sum m(0,1,2,4,8,10) + d(5,7)$$
- The "Don't-Care" conditions are represented as d(5,7)



On a 4 Variables Karnaugh map

- One Cell requires 4 variables
- Two adjacent cells require 3 variables
- Four adjacent cells require 2 variables
- Eight adjacent cells require 1 variable
- Sixteen adjacent cells give a 1 or true

Simplify using Karnaugh map



First, we need to change the circuit to an SOP expression

Simplify using Karnaugh Map

$$Y = \overline{A} + \overline{B} + \overline{B}C + (\overline{A} + B)(C + D)$$

$$Y = \overline{A}\overline{B} + \overline{B}C + \overline{A}B(C + D)$$

$$Y = \overline{A}\overline{B} + \overline{B}C + \overline{A}B\overline{C} + \overline{A}BD$$

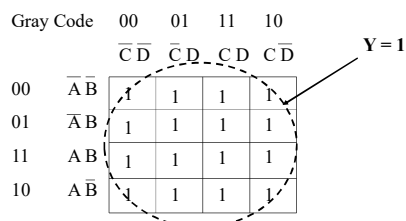
$$Y = \overline{A}\overline{B} + \overline{B}C + \overline{A}B\overline{C} + \overline{A}BD$$

$$Y = \overline{A}\overline{B} + \overline{B}C + (\overline{A} + B + \overline{C})(\overline{A} + B + D)$$

$$Y = \overline{A}\overline{B} + \overline{B}C + \overline{A} + \overline{A}B + \overline{A}D + B + BD + \overline{C} + \overline{C}D$$

← SOP expression

Simplify using Karnaugh map (cont'd)



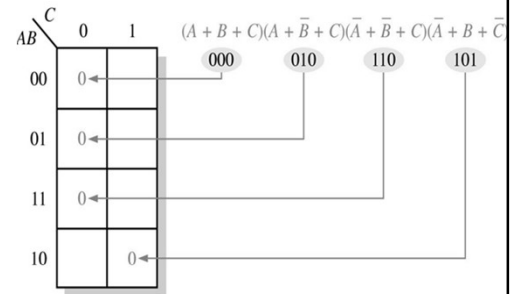
K-Map POS Minimization

3 Variables Karnaugh Map

Gray Code

	C	0	1
AB			
0 0		0	1
0 1		2	3
1 1		6	7
1 0		4	5

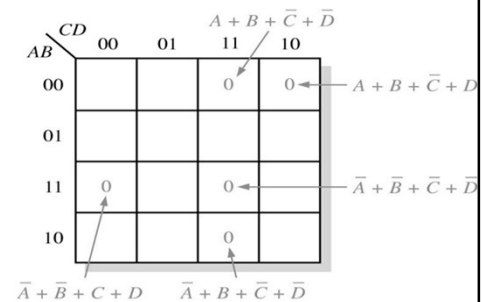
3 Variables Karnaugh Map (cont'd)



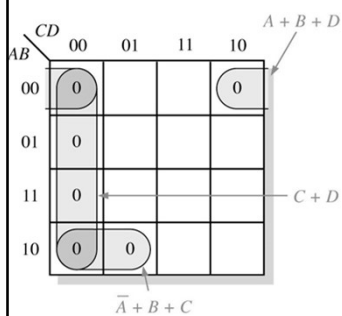
4 Variables Karnaugh Map

	CD	00	01	11	10
AB					
0 0		0	1	3	2
0 1		4	5	7	6
1 1		12	13	15	14
1 0		8	9	11	10

4 Variables Karnaugh Map (cont'd)



4 Variables Karnaugh Map (cont'd)



Karnaugh Map - Example

Mapping a Standard SOP expression

Example:

$$Y = \bar{A}\bar{B}\bar{C}\bar{D} + \bar{A}\bar{B}\bar{C}D + \bar{A}\bar{B}CD + \bar{A}BCD + \bar{A}\bar{B}C\bar{D} + \bar{A}\bar{B}C\bar{D}$$

Answer:

$$Y = \bar{B}\bar{D} + \bar{A}CD$$

Mapping a Standard POS expression

Example:

Using K-Map, convert the following standard POS expression into a minimum SOP expression

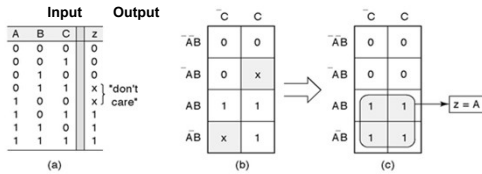
$$Y = A(\bar{B} + C)$$

Answer:

$$Y = \bar{A}\bar{B} + AC \quad \text{or standard SOP:} \quad Y = \bar{A}\bar{B}\bar{C} + \bar{A}\bar{B}C + ABC$$

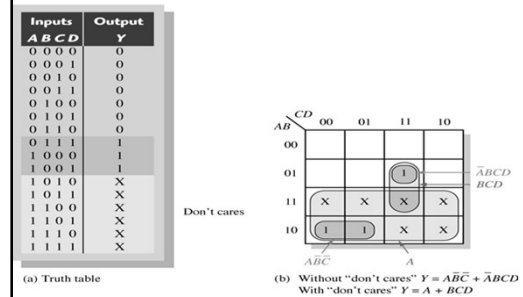
K-Map with "Don't Care" Conditions

Example :



3 variables with output "don't care (X)"

K-Map with "Don't Care" Conditions (cont'd)



4 variables with output "don't care (X)"

K-Map with "Don't Care" Conditions (cont'd)

• "Don't Care" Conditions

▪ Example:

Determine the minimal SOP using K-Map:

$$F(A, B, C, D) = \sum m(0, 2, 6, 8, 9, 10) \quad D(5, 12, 13, 14, 15)$$

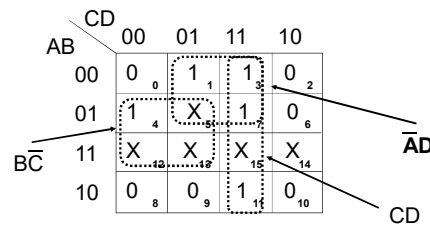
Answer:

$$F(A, B, C, D) = CD + \bar{B}\bar{C} + \bar{A}D$$

K-Map with "Don't Care" Conditions (cont'd)

Solution :

$$F(A, B, C, D) = \sum m(0, 2, 6, 8, 9, 10) \quad D(5, 12, 13, 14, 15)$$



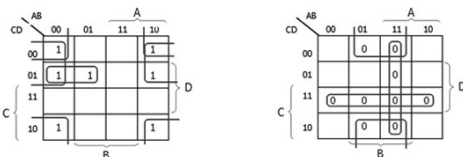
$$F(A, B, C, D) = CD + \bar{B}\bar{C} + \bar{A}D$$

K-map Product of Sums simplification

Example: Simplify the Boolean function $F_{ABCD} = \sum (0, 1, 2, 5, 8, 9, 10)$ in

(a) S-of-p

(b) P-of-s



Using the minterms (1s)
 $F_{ABCD} = \bar{B}\bar{D} + \bar{B}C + A\bar{C}D$

Using the maxterms (0s) and complementing F
 Grouping as if they were minterms, then using De Morgan's theorem to get F.
 $F'_{ABCD} = \bar{B}\bar{D} + \bar{C}D + AB$
 $F_{ABCD} = (B + D)(C + D)(A + B)$

Quine-McCluskey Method

- A **method** used for minimization of Boolean functions.
- Developed by Willard V. **Quine** and extended by Edward J. **McCluskey**.
- McCluskey developed the first algorithm for designing combinational circuits - the **Quine-McCluskey** logic minimization procedure as a doctoral student at **MIT**.

• Notable Publications:

- Quine, Willard Van Orman (October 1952). "The Problem of Simplifying Truth Functions". *The American Mathematical Monthly*. **59** (8): 521–531. doi:10.2307/2308219. JSTOR 2308219.
- Quine, Willard Van Orman (November 1955). "A Way to Simplify Truth Functions". *The American Mathematical Monthly*. **62** (9): 627–631. doi:10.2307/2307285. JSTOR 2307285.
- McCluskey, Jr., Edward J. (November 1956). "Minimization of Boolean Functions". *Bell System Technical Journal*. **35** (6): 1417–1444. doi:10.1002/j.1538-7305.1956.tb03835.x. Retrieved 2014-08-24.



Willard Van Orman Quine



Edward J. McCluskey

Quine-McCluskey Method: Important Terms

- Implicants: 1s $f(A,B,C) = \overline{A}\overline{B}C + \overline{A}B\overline{C} + ABC$
- Prime Implicants (PI)
 - Largest possible group of 1s
- Essential Prime Implicants (EPI)
 - PI that contains at least one Minterm which could not be further combined

AB \ CD	00 $\overline{C}\overline{D}$	01 $\overline{C}D$	11 CD	10 $C\overline{D}$
00 $\overline{A}\overline{B}$	1	1		
01 $\overline{A}B$		1	1	
11 AB	1	1	1	1
10 $A\overline{B}$			1	1

Quine-McCluskey Method

$f(A,B,C) = \sum m(0, 1, 3, 7, 8, 9, 11, 15)$

- Minterms

Minterms	Designations	Binary	A B C D
m_0	0	0 0 0 0	$\overline{A}\overline{B}\overline{C}\overline{D}$
m_1	1	0 0 0 1	$\overline{A}\overline{B}\overline{C}D$
m_3	3	0 0 1 1	$\overline{A}\overline{B}CD$
m_7	7	0 1 1 1	$\overline{A}BCD$
m_8	8	1 0 0 0	$A\overline{B}\overline{C}\overline{D}$
m_9	9	1 0 0 1	$A\overline{B}\overline{C}D$
m_{11}	11	1 0 1 1	$A\overline{B}CD$
m_{15}	15	1 1 1 1	$ABCD$

Quine-McCluskey Method

• Step 2: Table 2 [To make matched pairs between n^{th} and $(n+1)^{\text{th}}$]

Group of 1s	Matched Pairs	A B C D
0 (No of 1s is 0)	0	0 0 0 0
1 (No of 1s is 1)	1 8	0 0 0 1 1 0 0 0
2 (No of 1s is 2)	3 9	0 0 1 1 1 0 0 1
3 (No of 1s is 3)	7 11	0 1 1 1 1 0 1 1
4 (No of 1s is 4)	15	1 1 1 1

Quine-McCluskey Method

• Step 1: Table 1

Group	Minterms	A B C D	Matched Pairs	Variable
0	0	0 0 0 0	0, 1 0, 8	0 0 0 X X 0 0 0
1	1 8	0 0 0 1 1 0 0 0	1, 3 1, 9 8, 9	0 0 X 1 X 0 0 1 1 0 0 X
2	3 9	0 0 1 1 1 0 0 1	3, 7 3, 11 9, 11	0 X 1 1 X 0 1 1 1 0 X 1
3	7 11	0 1 1 1 1 0 1 1	7, 15 11, 15	X 1 1 1 1 X 1 1
4	15	1 1 1 1		

• Step 2:
• Table 2

Quine-McCluskey Method

Groups	Matched Minterms	A B C D	Matched Pairs	
0	0, 1 0, 8	0 0 0 X X 0 0 0	0,1; 8,9 0,8; 1,9	X 0 0 X X 0 0 X
1	1, 3 1, 9 8, 9	0 0 X 1 X 0 0 1 1 0 0 X	1,3; 9,11 1,9; 3,11	X 0 X 1 X 0 X 1
2	3, 7 3, 11 9, 11	0 X 1 1 X 0 1 1 1 0 X 1	3,7; 11,15 3,11; 7,15	X X 1 1 X X 1 1
3		X 1 1 1 1 X 1 1		
4				

• Step 3:
• Table 3

Quine-McCluskey Method

PI	Minterms Involved	Minterms							
		0	1	3	7	8	9	11	15
$\overline{B}\overline{C}$	0, 1, 8, 9	X	X			X	X		
$\overline{B}D$	1, 3, 9, 11		X	X			X	X	
CD	3, 7, 11, 15			X	X			X	X

$f = \overline{B}\overline{C} + CD$

Solve it with K-Map and Check