

DIGITAL IMAGE PROCESSING



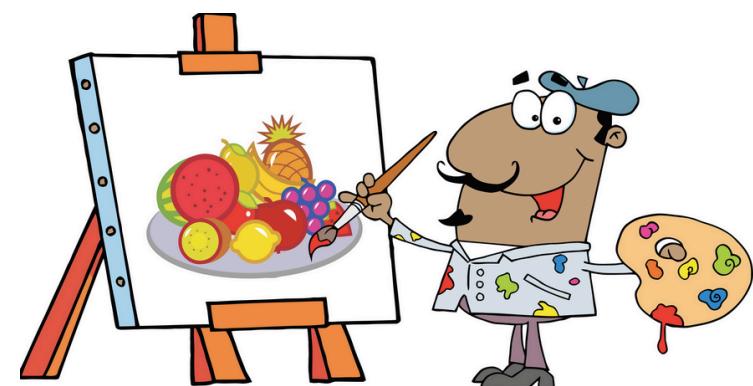
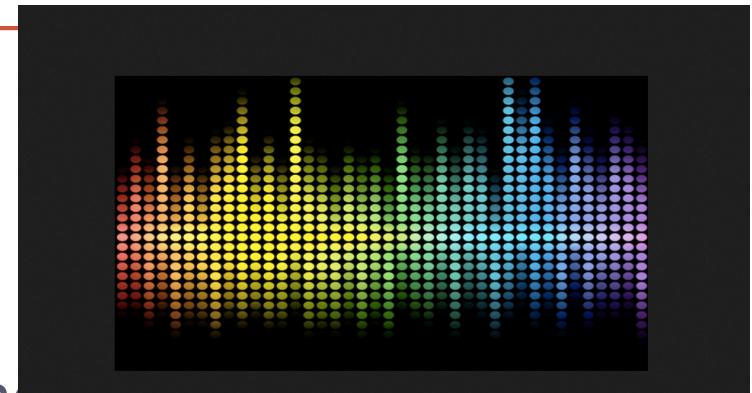
Lecture 4

Frequency domain

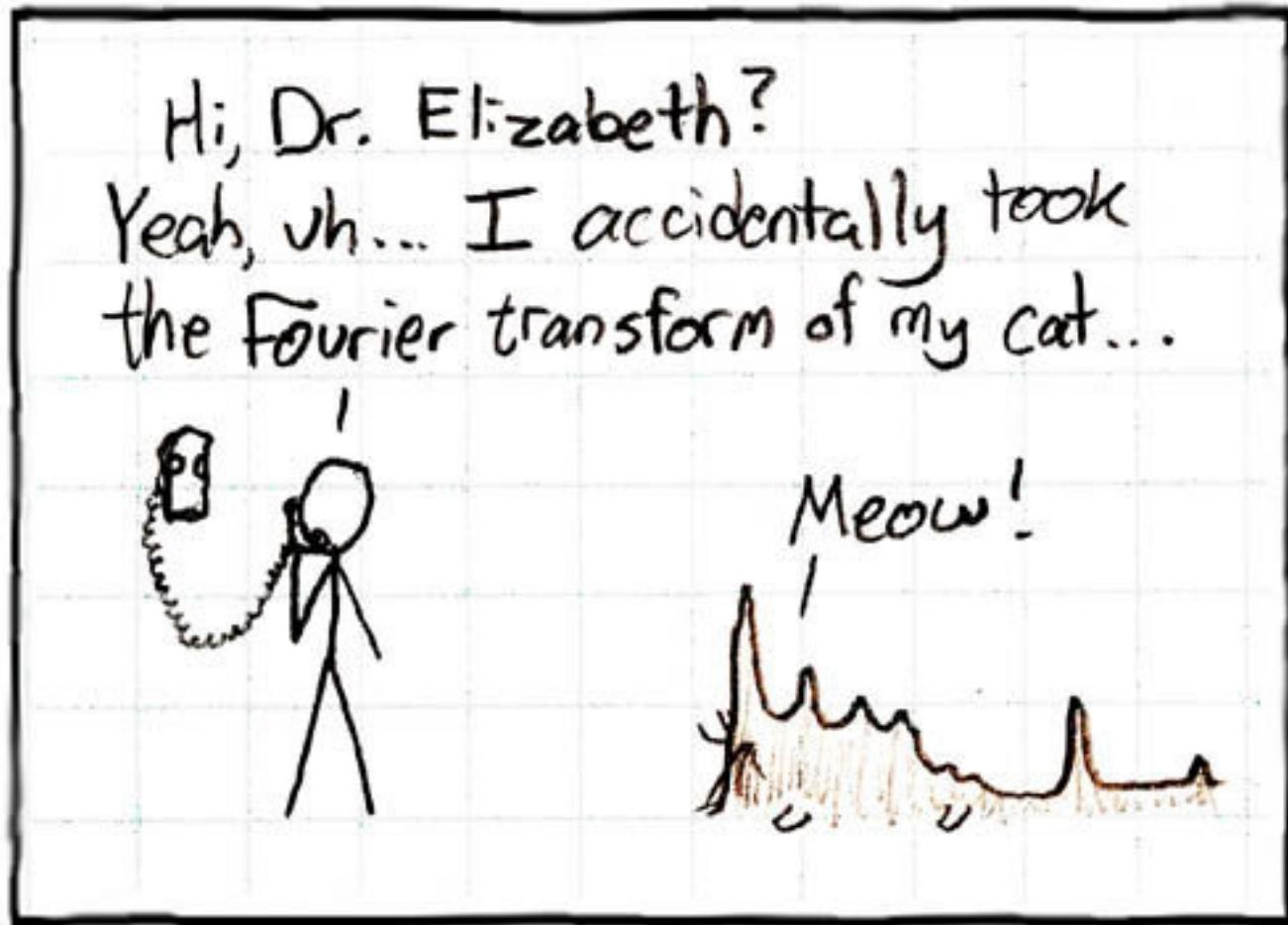
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Fourier Domain



2D Fourier transform and its applications

- Fourier transforms and spatial frequencies in 2D
 - Definition and meaning
- The Convolution Theorem
 - Applications to spatial filtering
- The Sampling Theorem and Aliasing

Much of this material is a straightforward generalization of the 1D Fourier analysis with which you are familiar.

A. Zisserman

Jean Baptiste Joseph Fourier (1768-1830)

A bold idea (1807):

Any univariate function can be rewritten as a weighted sum of sines and cosines of different frequencies.

...the manner in which the author arrives at these equations is not exempt of difficulties and...his analysis to integrate them still leaves something to be desired on the score of generality and even rigour.

- Don't believe it?
 - Neither did Lagrange, Laplace, Poisson and other big wigs
 - Not translated into English until 1878!
- But it's (mostly) true!
 - called Fourier Series
 - there are some subtle restrictions

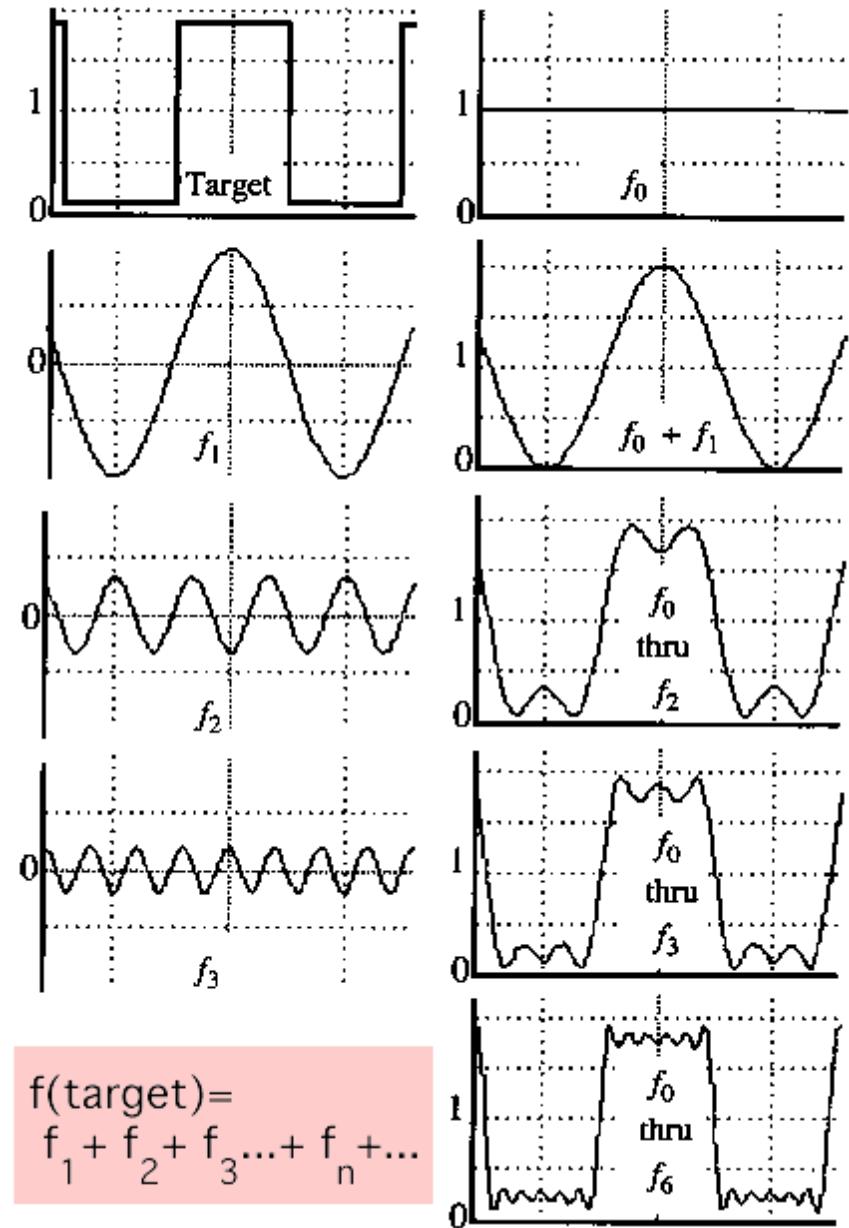


A sum of sines and cosines

Our building block:

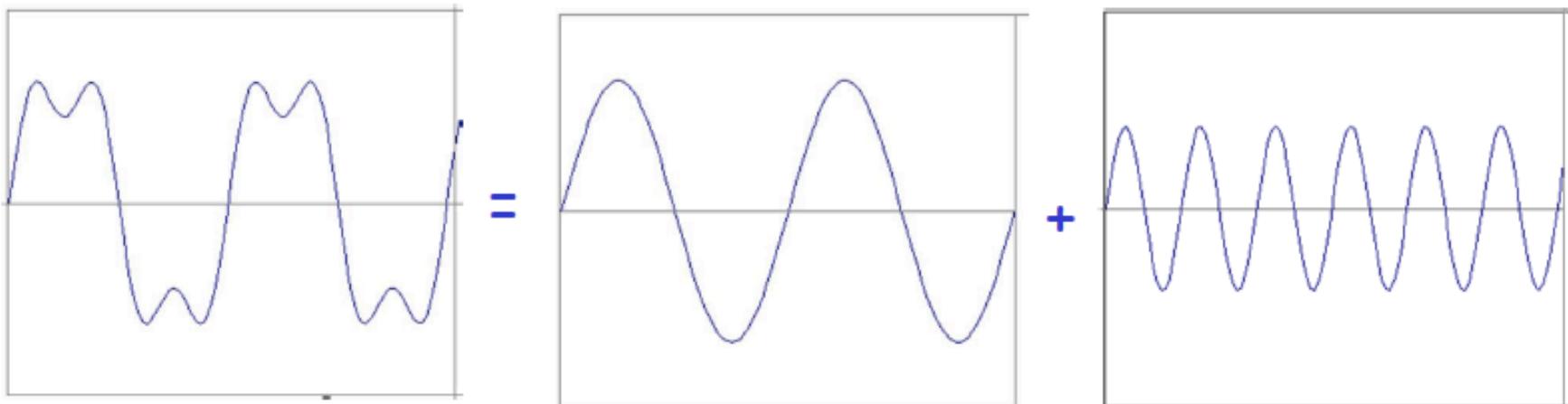
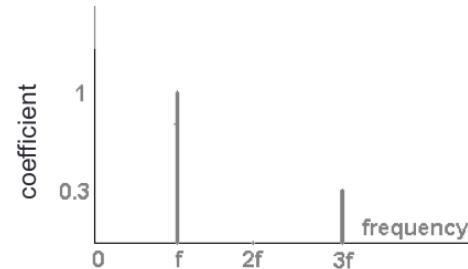
$$A \sin(\omega x) + B \cos(\omega x)$$

Add enough of them to get any signal $g(x)$ you want!



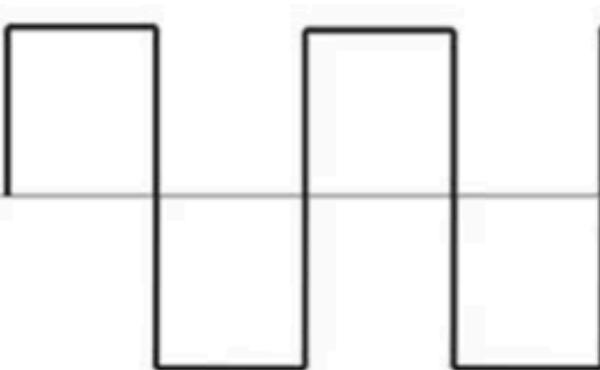
Reminder: 1D Fourier Series

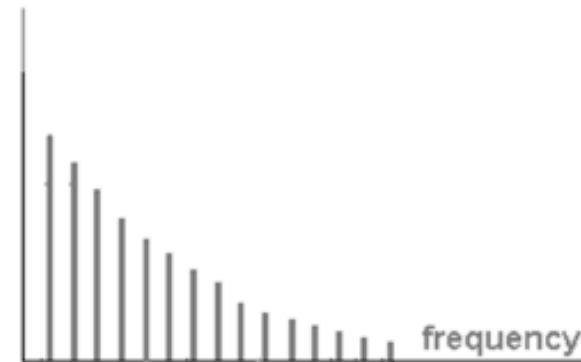
Example



$$f(x) = \sin x + \frac{1}{3} \sin 3x + \dots$$

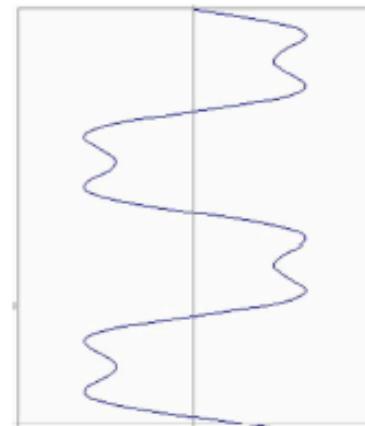
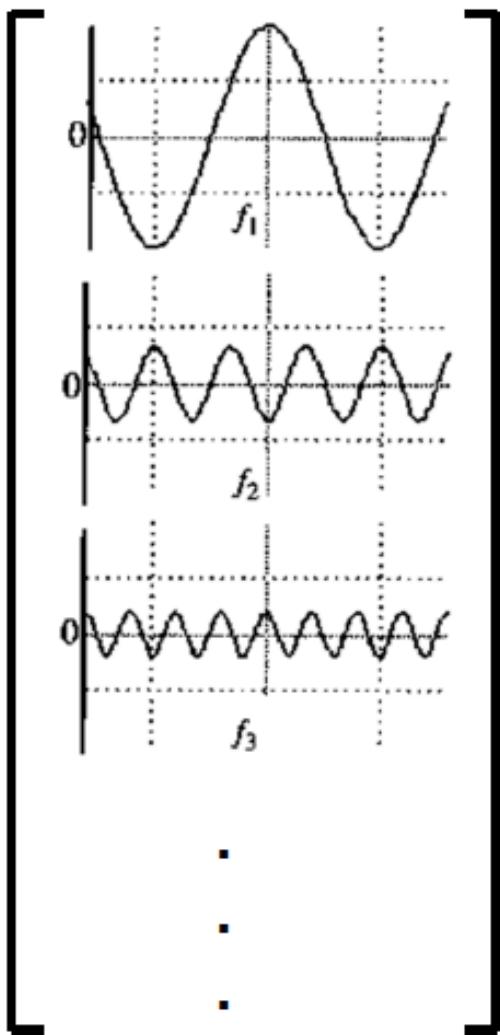
Fourier Series of a Square Wave


$$f(x) = \sum_{n=1,3,5,\dots} \frac{1}{n} \sin nx$$

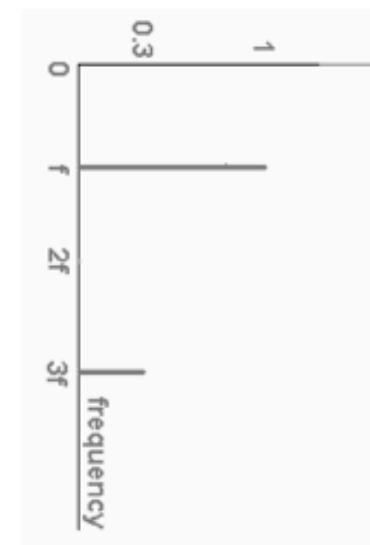


Fourier Series: Just a change of basis

$$\text{M } f(x) = F(\omega)$$



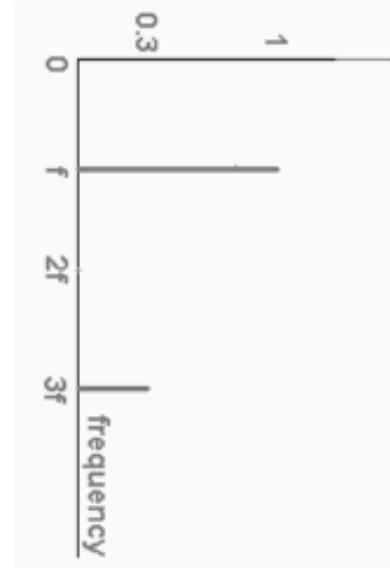
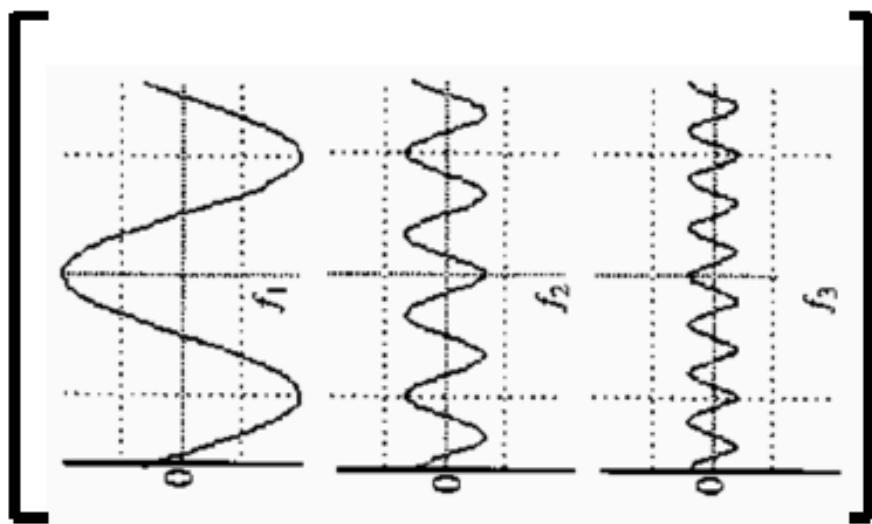
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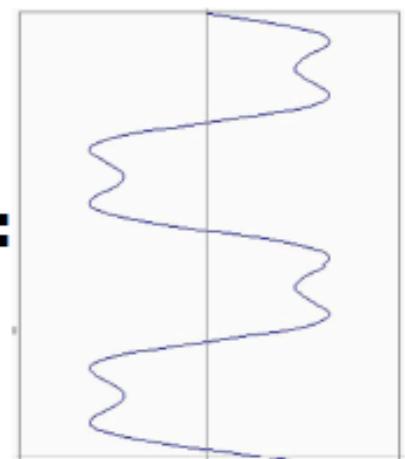
A. Zisserman

Inverse FT: Just a change of basis

$$\mathcal{M}^{-1} F(\omega) = f(x)$$



•



A. Zisserman

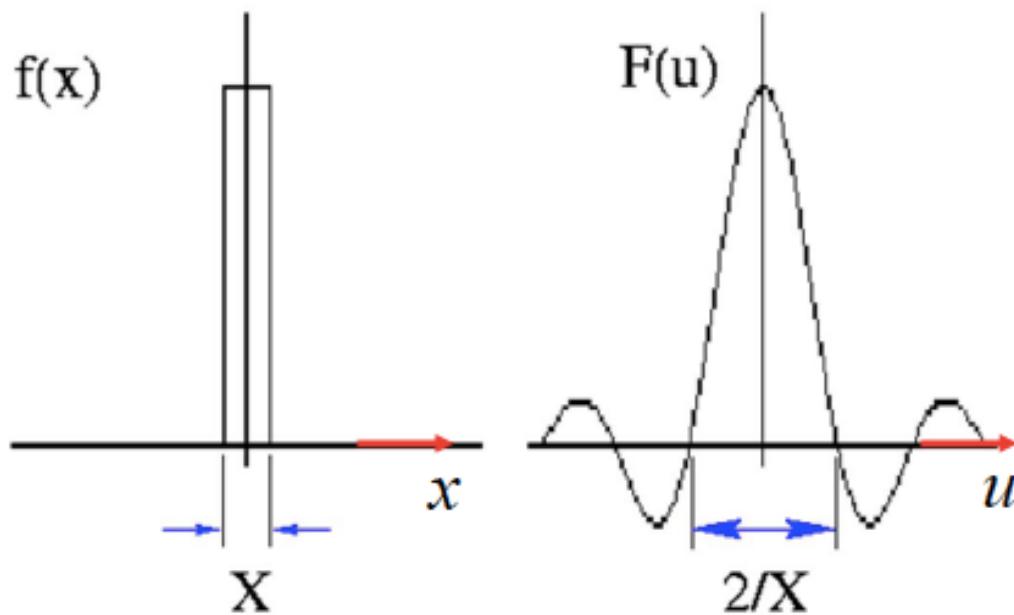
1D Fourier Transform

$$F(u) = \int_{-\infty}^{\infty} f(x) e^{-j2\pi ux} dx,$$

$$f(x) = \int_{-\infty}^{\infty} F(u) e^{j2\pi ux} du$$

1D Fourier Transform

$$F(u) = \int_{-\infty}^{\infty} f(x) e^{-j2\pi ux} dx,$$
$$f(x) = \int_{-\infty}^{\infty} F(u) e^{j2\pi ux} du$$

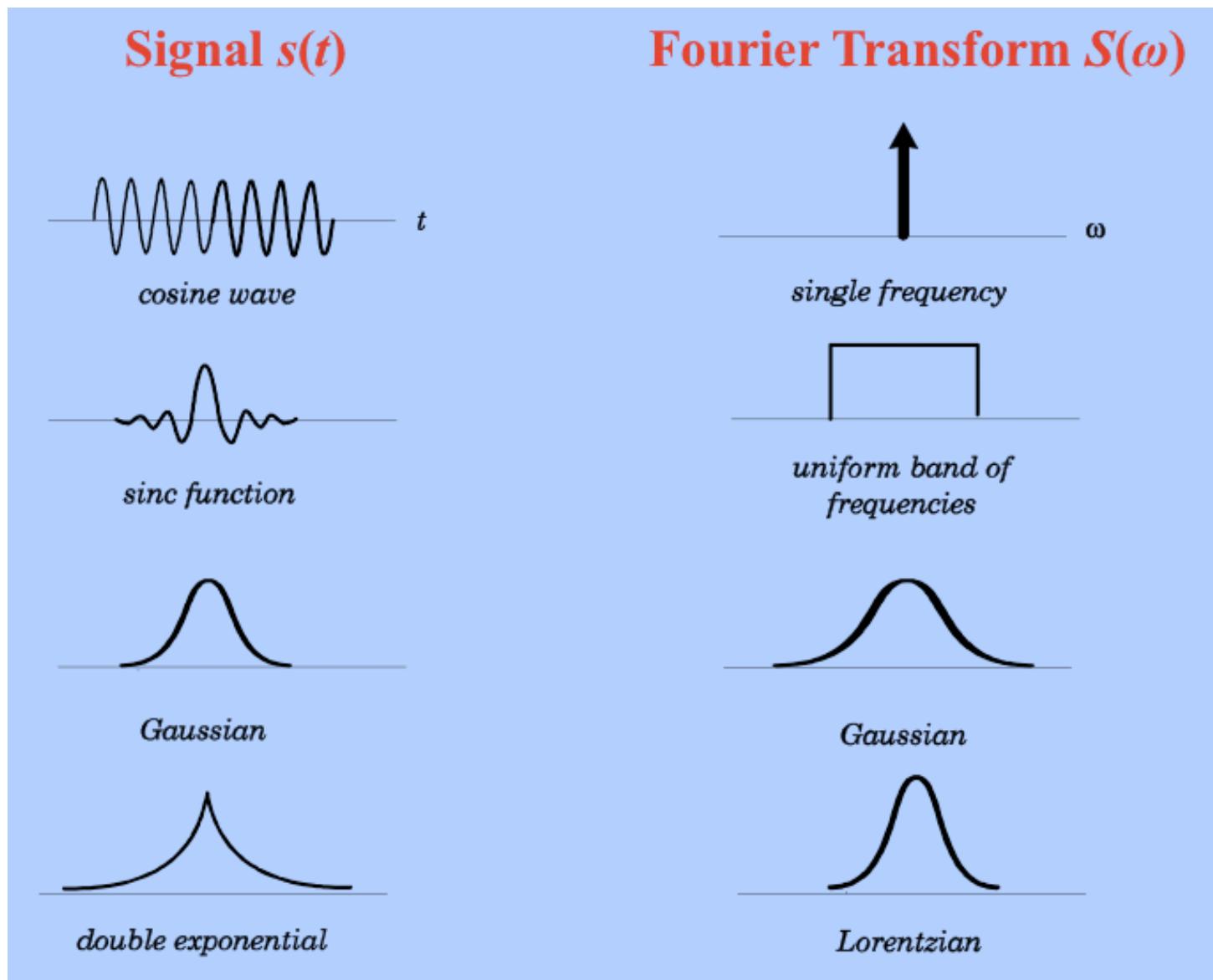


Example

$$f(x) = \begin{cases} 1, & |x| < \frac{X}{2}, \\ 0, & |x| \geq \frac{X}{2}. \end{cases}$$

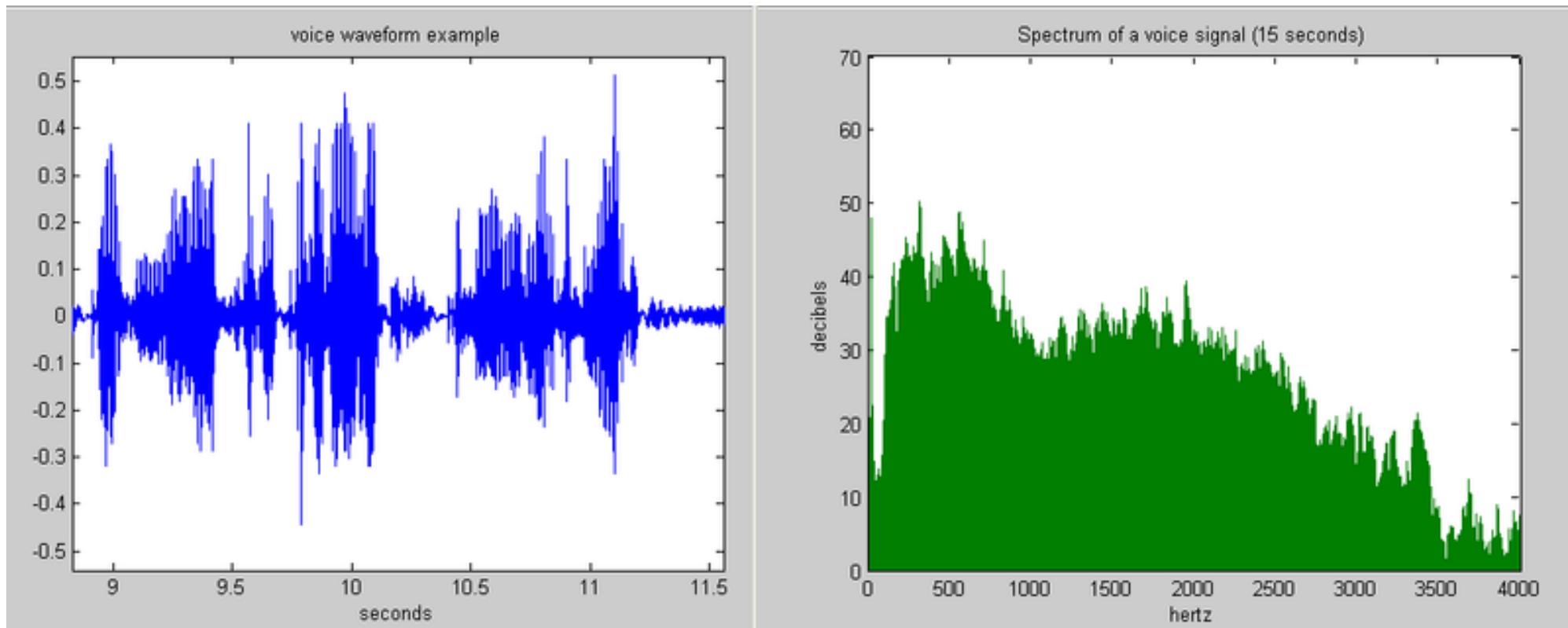
$$\begin{aligned} F(u) &= \int_{-\infty}^{\infty} f(x) e^{-j2\pi ux} dx \\ &= \int_{-X/2}^{X/2} e^{-j2\pi ux} dx \\ &= \frac{1}{-j2\pi u} [e^{-j2\pi uX/2} - e^{j2\pi uX/2}] \\ &= X \frac{\sin(\pi Xu)}{(\pi Xu)} = X \text{sinc}(\pi Xu). \end{aligned}$$

Fourier Transform



Example: Music

- We think of music in terms of frequencies at different magnitudes

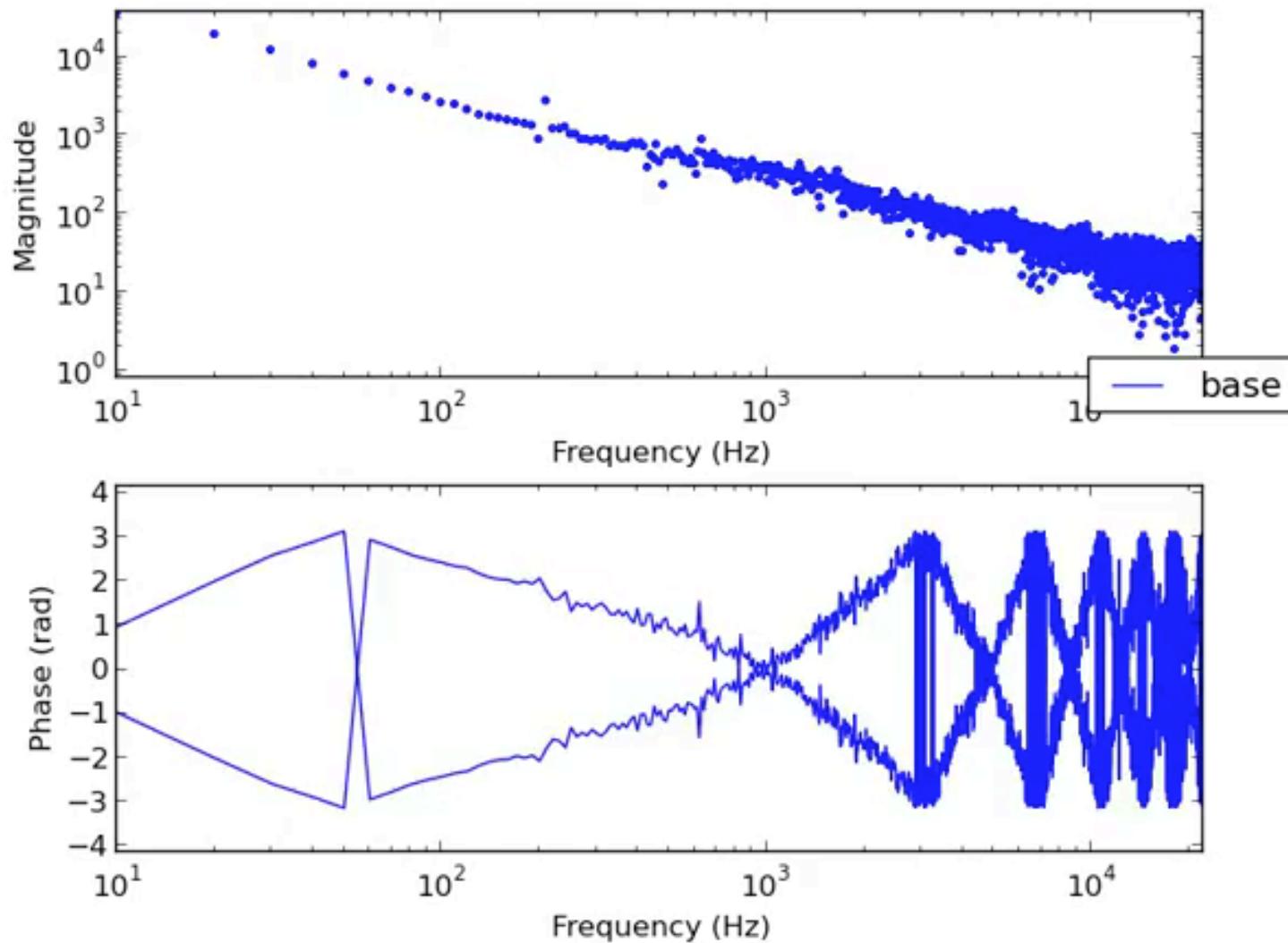


Fourier Analysis of a Piano



<https://www.youtube.com/watch?v=6SR81Wh2cqu>

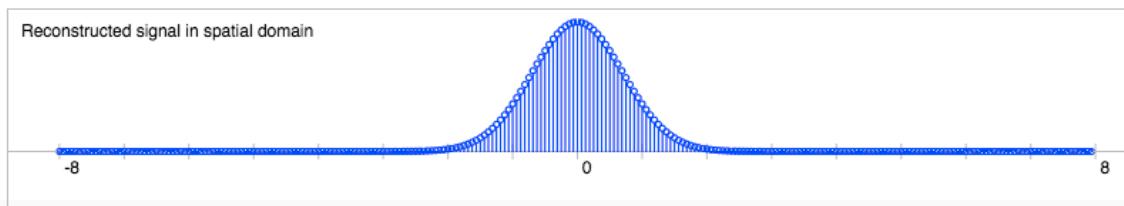
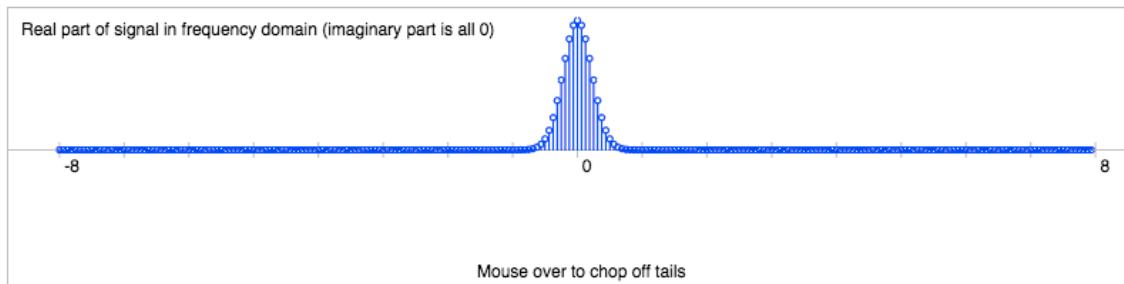
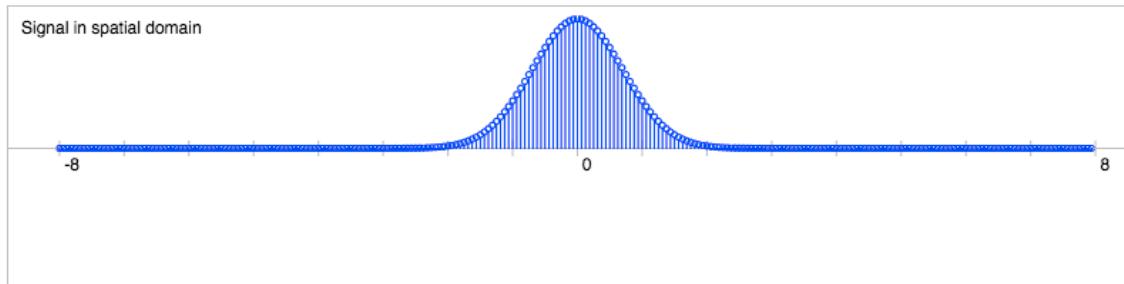
Fourier Analysis of a Piano



Discrete Fourier Transform Demo

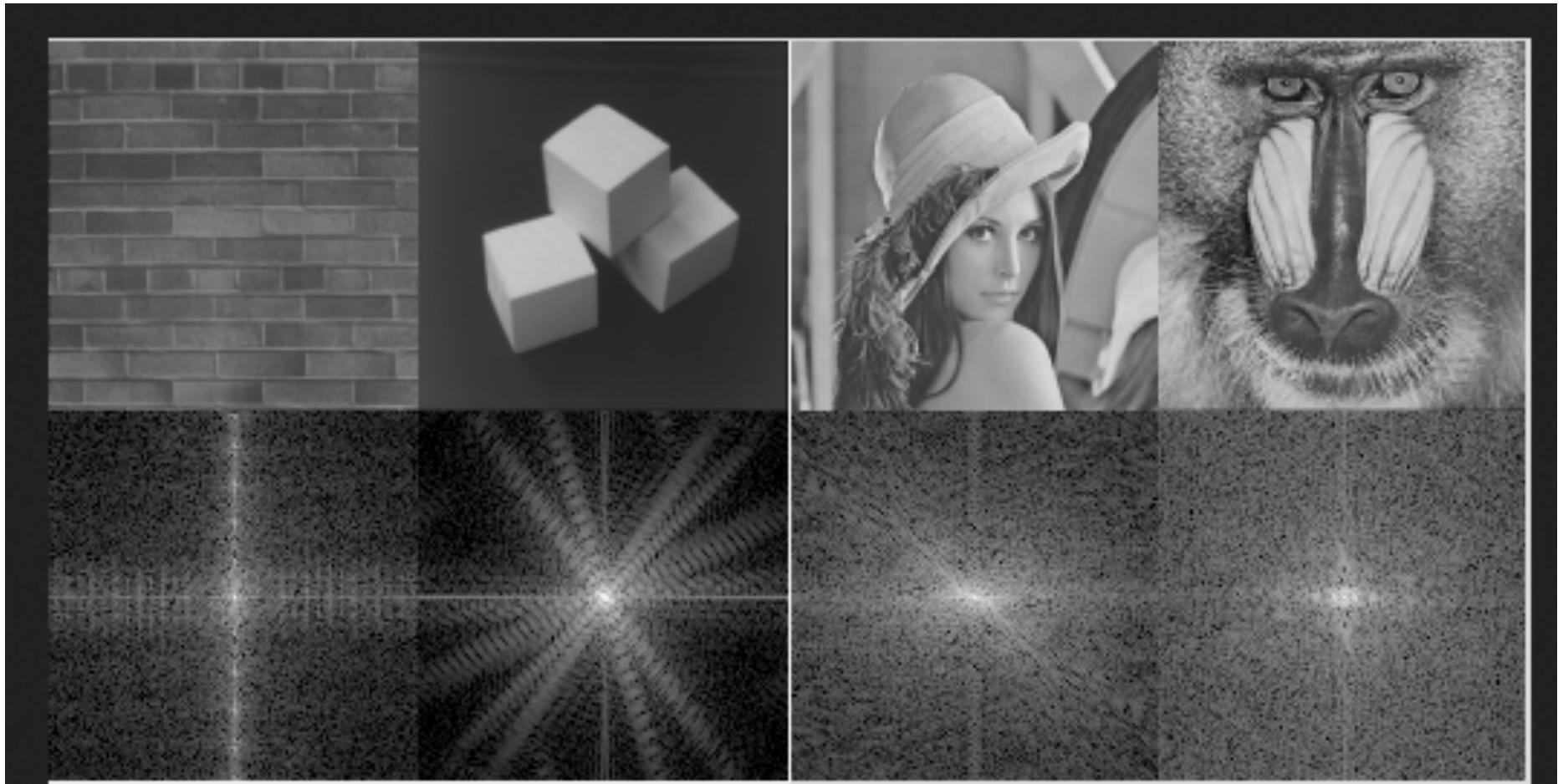
<http://madebyevan.com/dft/>

$$f(x) = \text{gaussian}(x)$$



Evan Wallace

2D Fourier Transform



2D Fourier Transform

Definition

$$F(u, v) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) e^{-j2\pi(ux+vy)} dx dy,$$

$$f(x, y) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F(u, v) e^{j2\pi(ux+vy)} du dv$$

where u and v are spatial frequencies.

Also will write FT pairs as $f(x, y) \Leftrightarrow F(u, v)$.

2D Fourier Transform

- $F(u, v)$ is complex in general,

$$F(u, v) = F_R(u, v) + jF_I(u, v)$$

- $|F(u, v)|$ is the **magnitude** spectrum
- $\arctan(F_I(u, v)/F_R(u, v))$ is the **phase** angle spectrum.
- Conjugacy: $f^*(x, y) \Leftrightarrow F(-u, -v)$
- Symmetry: $f(x, y)$ is **even** if $f(x, y) = f(-x, -y)$

Sinusoidal Waves

In 1D the Fourier transform is based on a decomposition into functions $e^{j2\pi ux} = \cos 2\pi ux + j \sin 2\pi ux$ which form an orthogonal basis. Similarly in 2D

$$e^{j2\pi(ux+vy)} = \cos 2\pi(ux + vy) + j \sin 2\pi(ux + vy)$$

The real and imaginary terms are sinusoids on the x, y plane. The maxima and minima of $\cos 2\pi(ux + vy)$ occur when

$$2\pi(ux + vy) = n\pi$$

Sinusoidal Waves

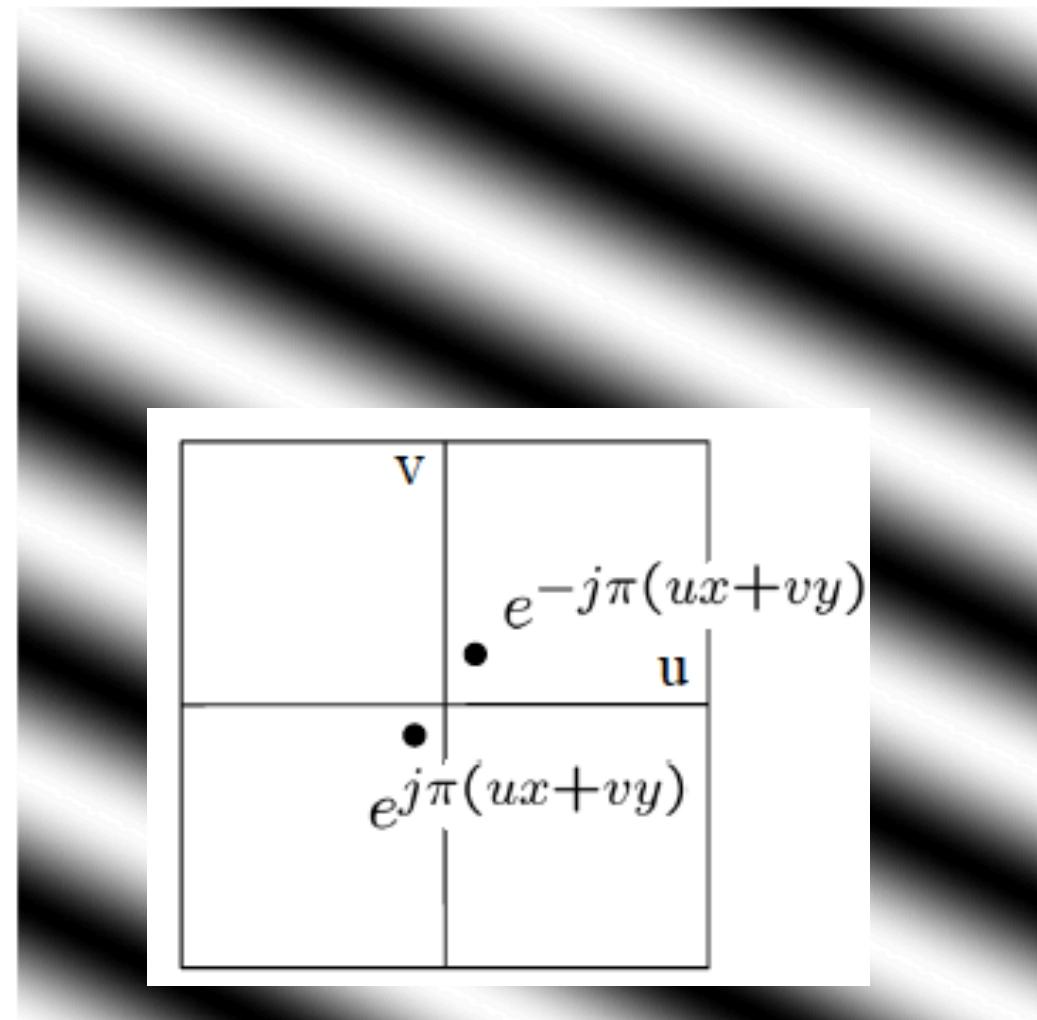
write $ux + vy$ using vector notation with $\mathbf{u} = (u, v)^\top$, $\mathbf{x} = (x, y)^\top$ then

$$2\pi(ux + vy) = 2\pi\mathbf{u} \cdot \mathbf{x} = n\pi$$

are sets of equally spaced parallel lines with normal \mathbf{u} and wavelength $1/\sqrt{u^2 + v^2}$.

Sinusoidal Waves

To get some sense of what basis elements look like, we plot a basis element --- or rather, its real part --- as a function of x, y for some fixed u, v . We get a function that is constant when $(ux+vy)$ is constant. The magnitude of the vector (u, v) gives a frequency, and its direction gives an orientation. The function is a sinusoid with this frequency along the direction, and constant perpendicular to the direction.



Sinusoidal Waves

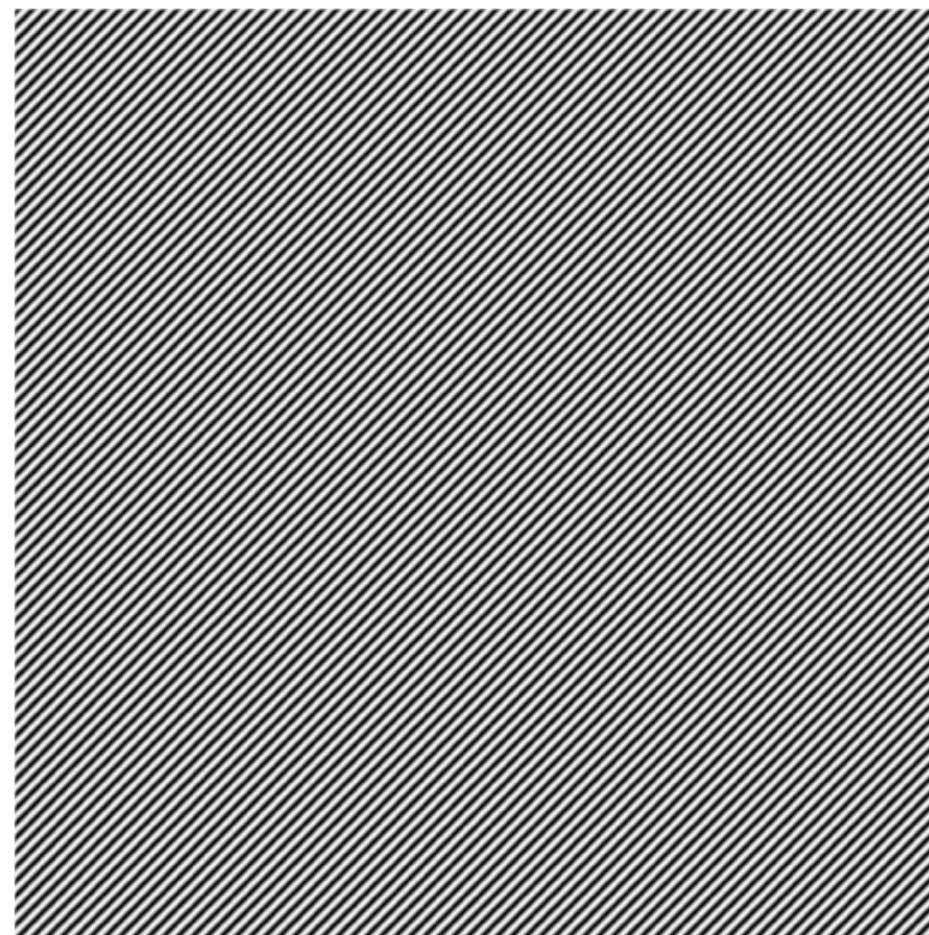
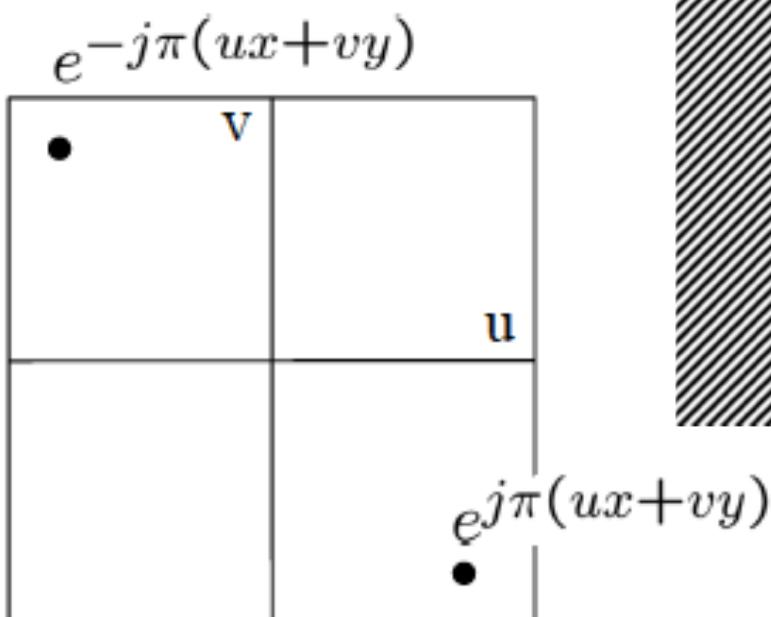
Here u and v are larger than in the previous slide.

$$\begin{array}{|c|c|} \hline v & \\ \hline e^{-j\pi(ux+vy)} & \\ \hline \bullet & u \\ \hline & \\ \hline & \bullet \\ \hline e^{j\pi(ux+vy)} & \\ \hline \end{array}$$



Sinusoidal Waves

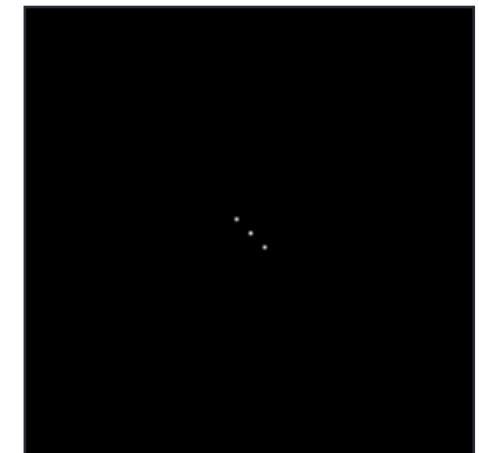
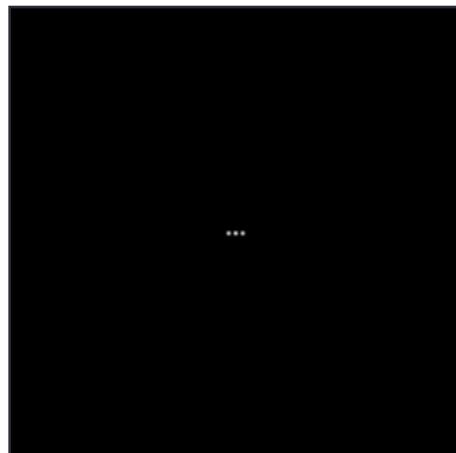
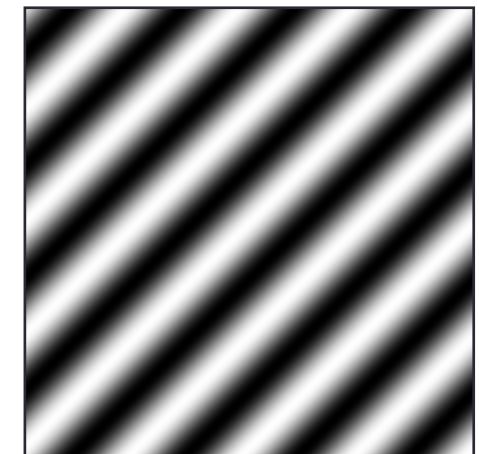
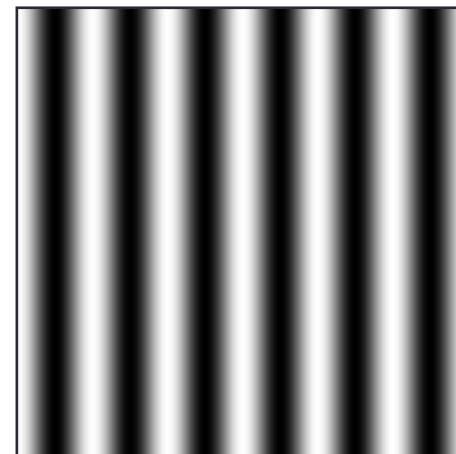
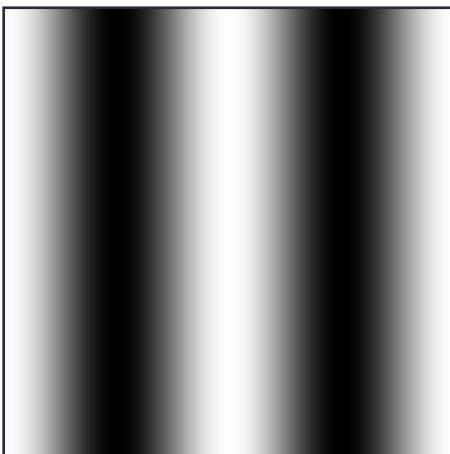
And larger still...



slide: B. Freeman

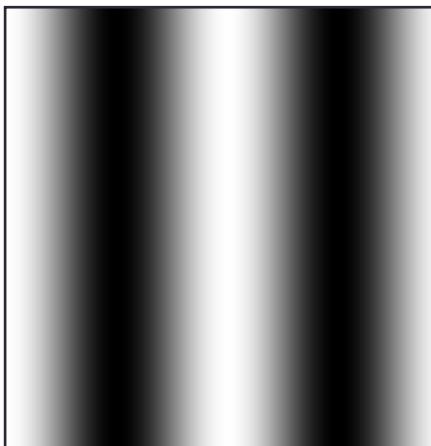
Fourier analysis in images

Intensity images

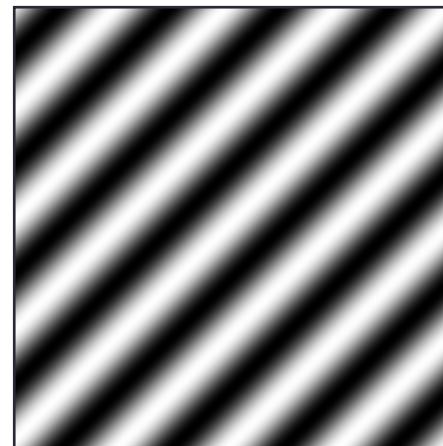


Fourier decomposition images

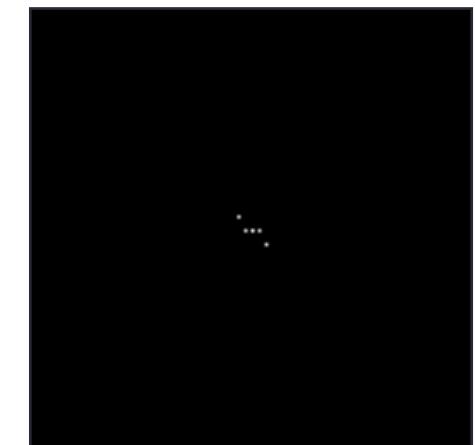
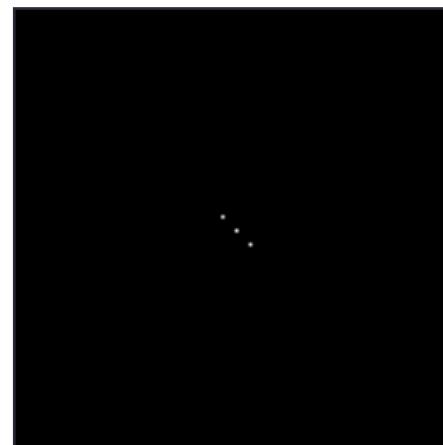
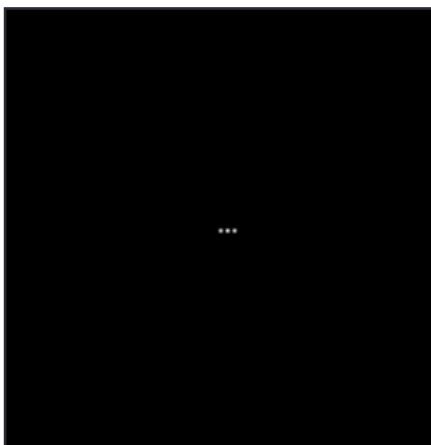
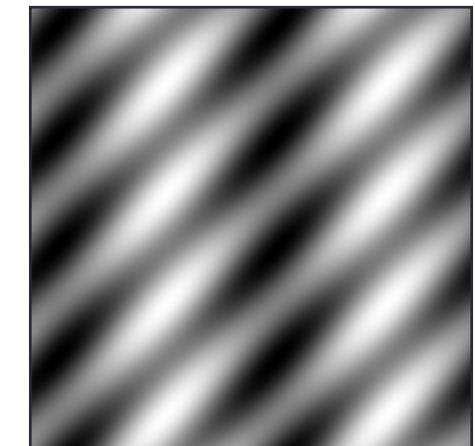
Signals can be composed



+



=

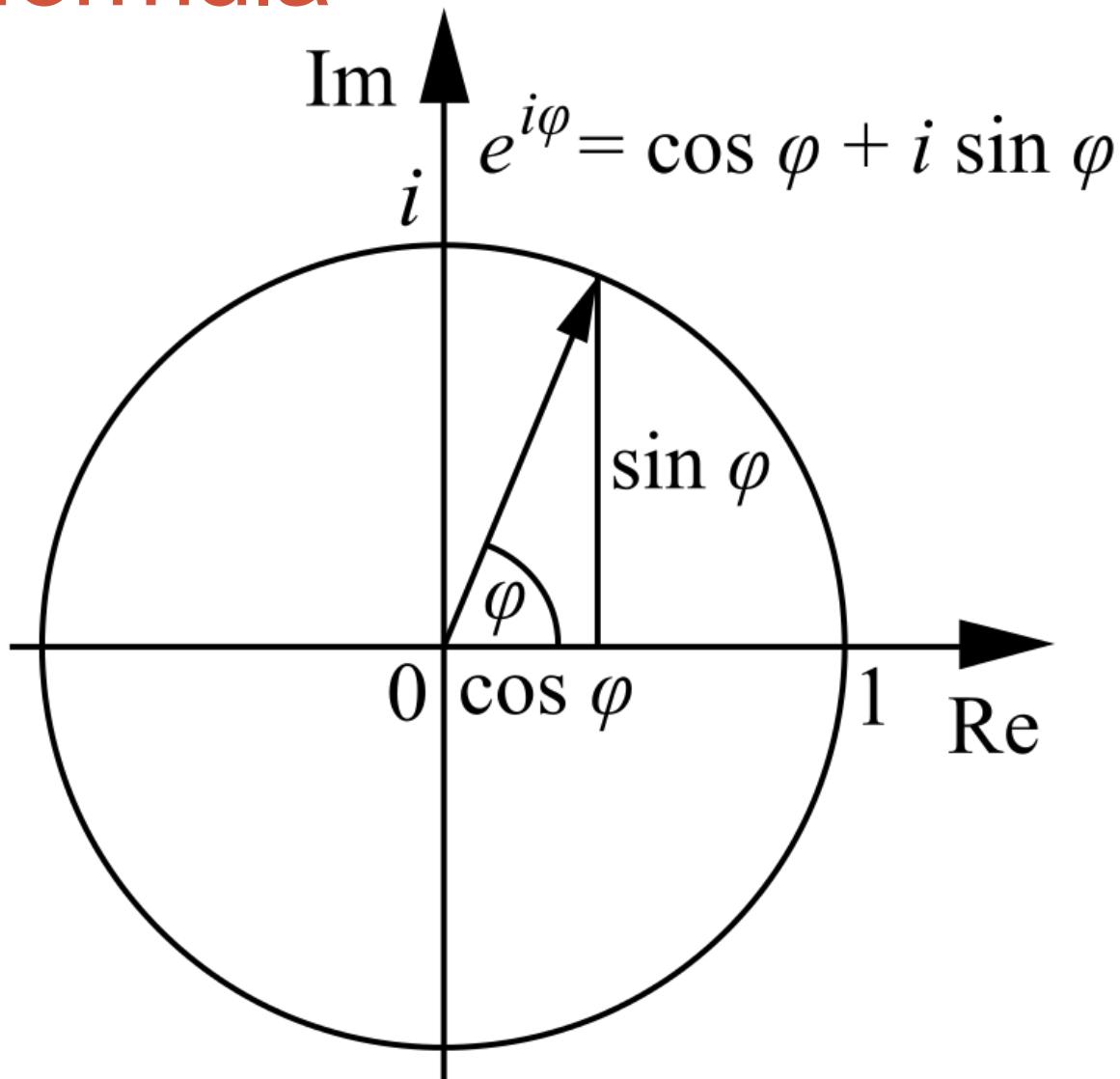


<http://sharp.bu.edu/~slehar/fourier/fourier.html#filtering>
More: <http://www.cs.unm.edu/~brayer/vision/fourier.html>

Fourier Transform

- Stores the amplitude and phase at each frequency:
 - For mathematical convenience, this is often notated in terms of real and complex numbers
 - Related by Euler's formula

Euler's formula



Fourier Transform

- Stores the amplitude and phase at each frequency:
 - For mathematical convenience, this is often notated in terms of real and complex numbers
 - Related by Euler's formula
 - Amplitude encodes how much signal there is at a particular frequency

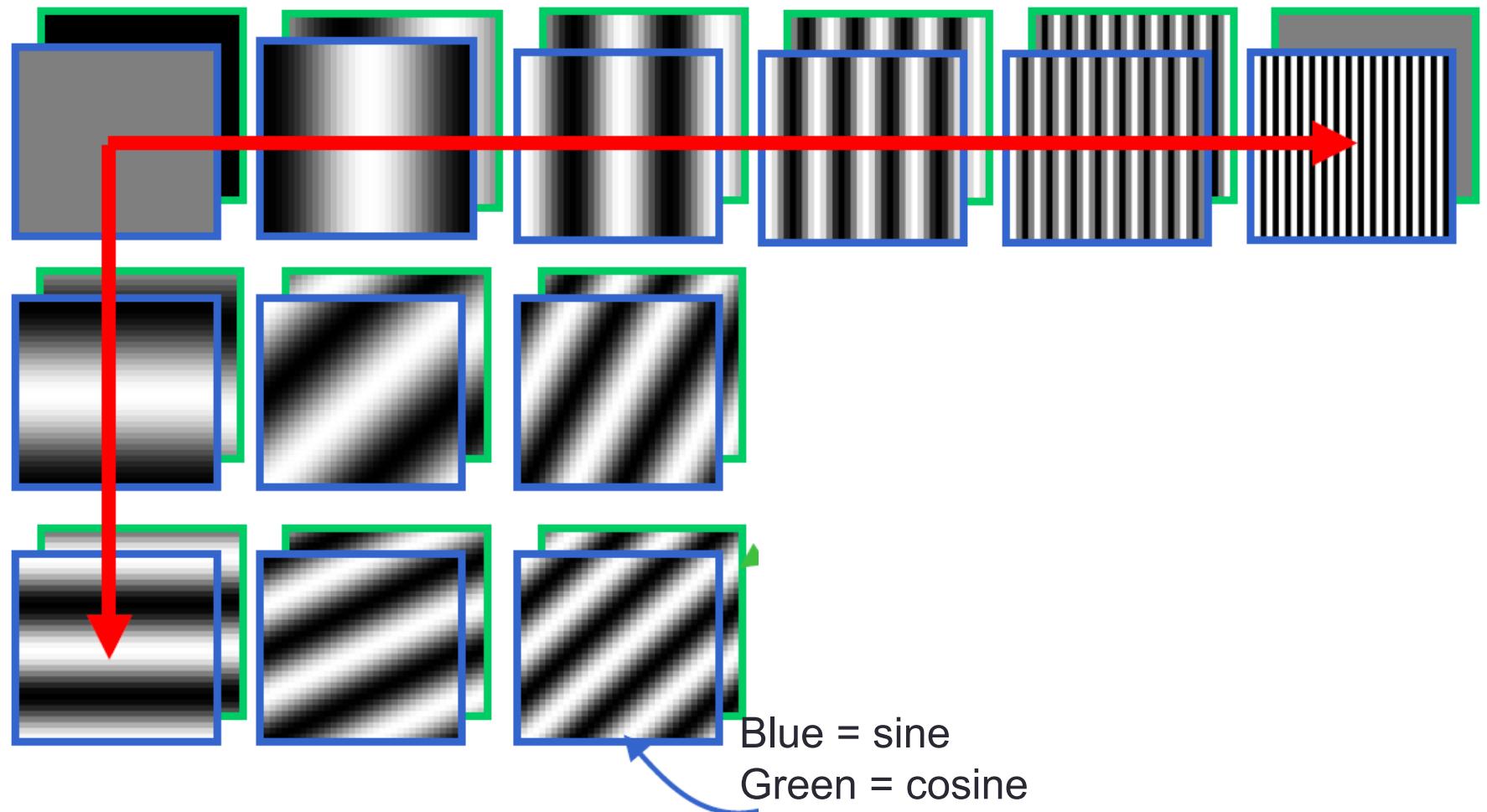
$$\text{Amplitude: } A = \pm \sqrt{\text{Re}(\omega)^2 + \text{Im}(\omega)^2}$$

- Phase encodes spatial information (indirectly)

$$\text{Phase: } \phi = \tan^{-1} \frac{\text{Im}(\omega)}{\text{Re}(\omega)}$$

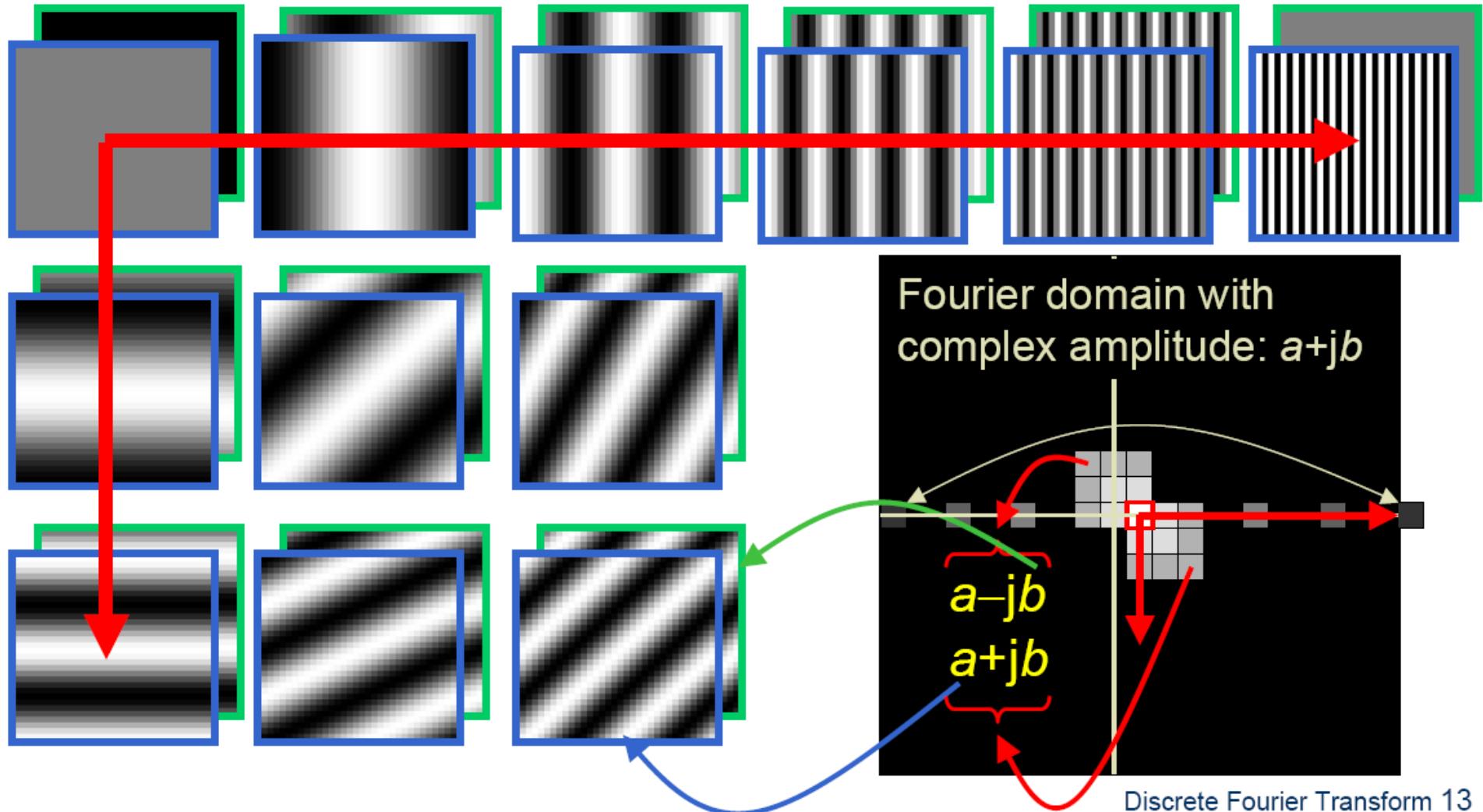
Fourier Bases

Teases away ‘fast vs. slow’ changes in the image.



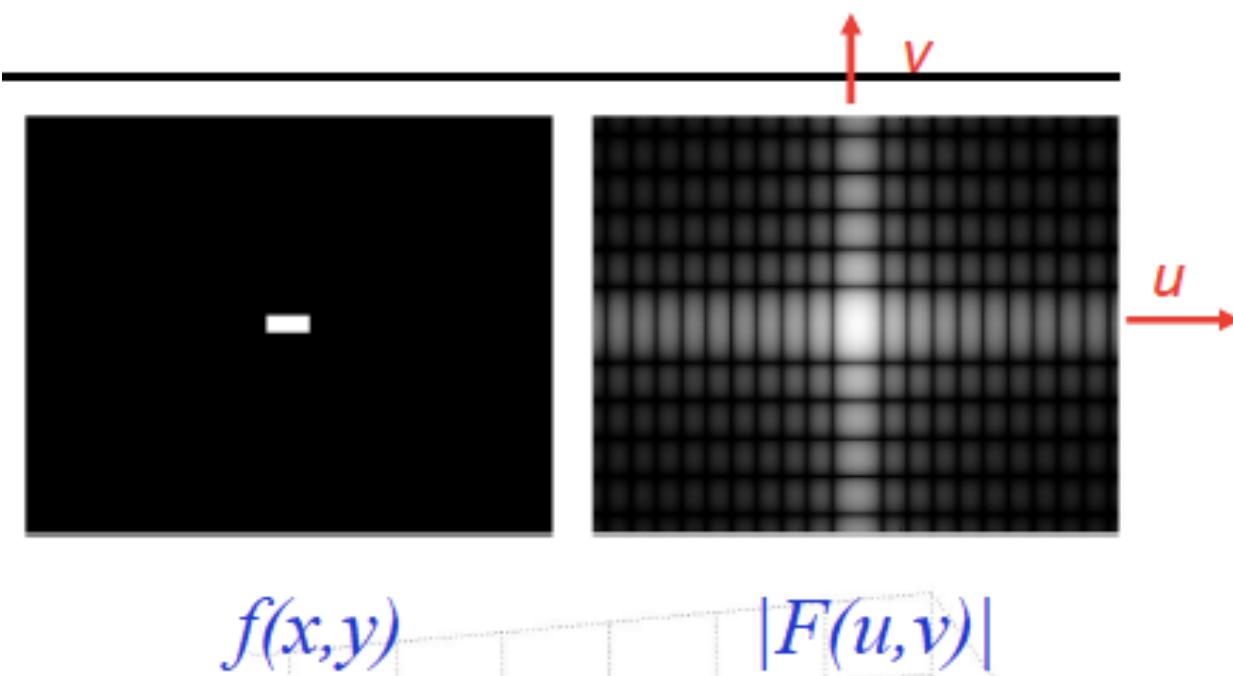
This change of basis is the Fourier Transform

Fourier Bases



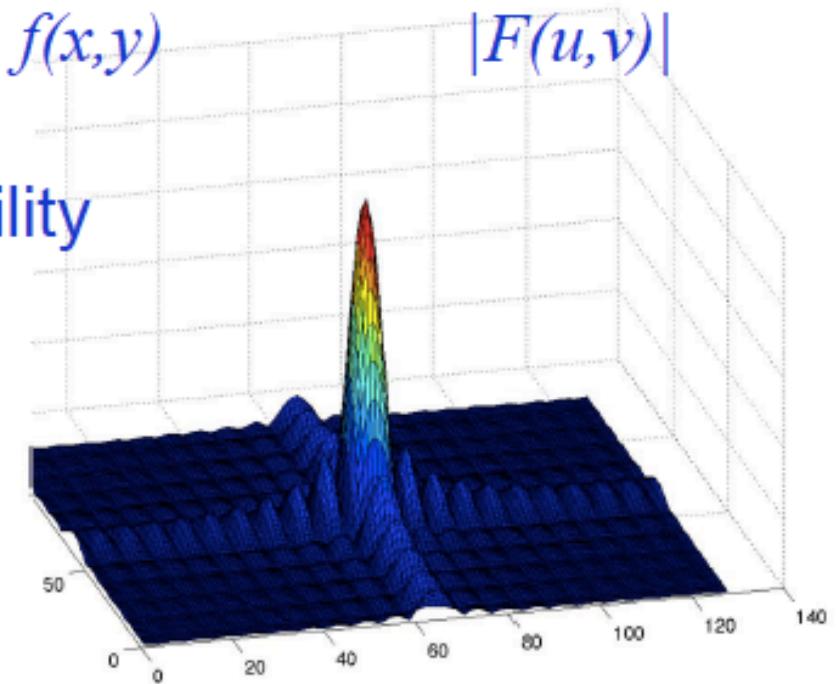
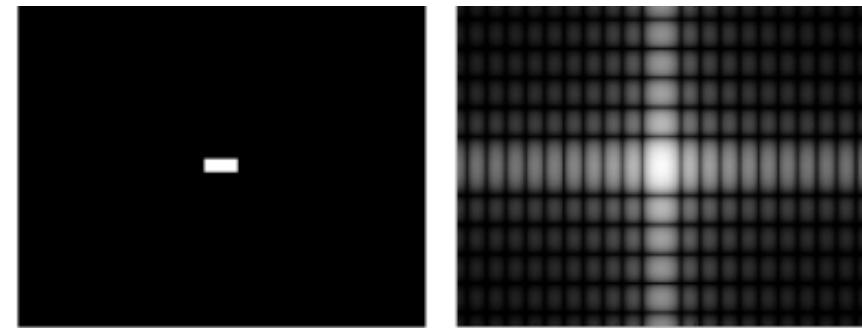
Important Fourier Transform Pairs

rectangle centred at origin
with sides of length X and Y



Important Fourier Transform Pairs

$$\begin{aligned} F(u, v) &= \int \int f(x, y) e^{-j2\pi(ux+vy)} dx dy, \\ &= \int_{-X/2}^{X/2} e^{-j2\pi ux} dx \int_{-Y/2}^{Y/2} e^{-j2\pi vy} dy, \quad \text{separability} \\ &= \left[\frac{e^{-j2\pi ux}}{-j2\pi u} \right]_{-X/2}^{X/2} \left[\frac{e^{-j2\pi vy}}{-j2\pi v} \right]_{-Y/2}^{Y/2}, \\ &= \frac{1}{-j2\pi u} [e^{-juX} - e^{juX}] \frac{1}{-j2\pi v} [e^{-jvY} - e^{jvY}], \\ &= XY \left[\frac{\sin(\pi Xu)}{\pi Xu} \right] \left[\frac{\sin(2\pi Yv)}{\pi Yv} \right] \\ &= XY \operatorname{sinc}(\pi Xu) \operatorname{sinc}(\pi Yv). \end{aligned}$$



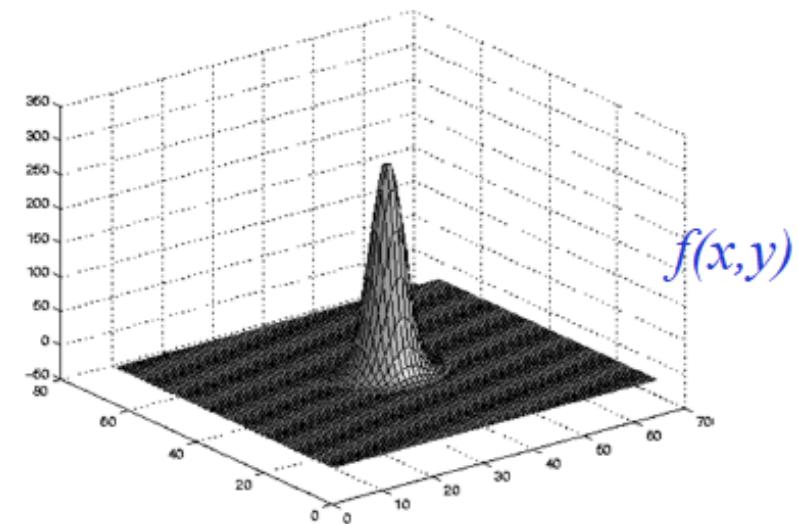
$$|F(u, v)|$$

Important Fourier Transform Pairs

Gaussian centred on origin

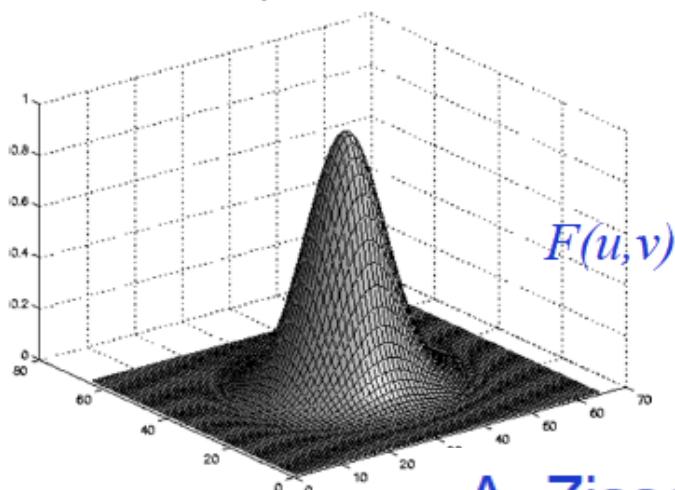
$$f(r) = \frac{1}{2\pi\sigma^2} e^{-r^2/2\sigma^2}$$

where $r^2 = x^2 + y^2$.



$$F(u, v) = F(\rho) = e^{-2\pi^2\rho^2\sigma^2}$$

where $\rho^2 = u^2 + v^2$.

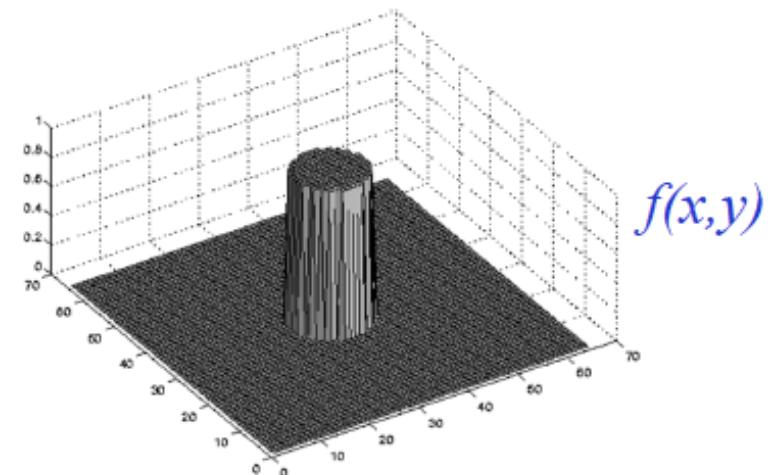


- FT of a Gaussian is a Gaussian
- Note inverse scale relation

Important Fourier Transform Pairs

Circular disk unit height and radius a centred on origin

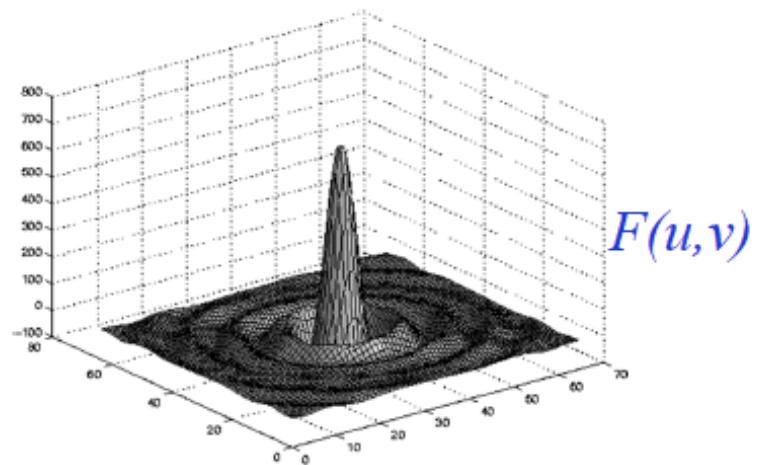
$$f(x, y) = \begin{cases} 1, & |r| < a, \\ 0, & |r| \geq a. \end{cases}$$



$$F(u, v) = F(\rho) = aJ_1(\pi a \rho)/\rho$$

where $J_1(x)$ is a Bessel function.

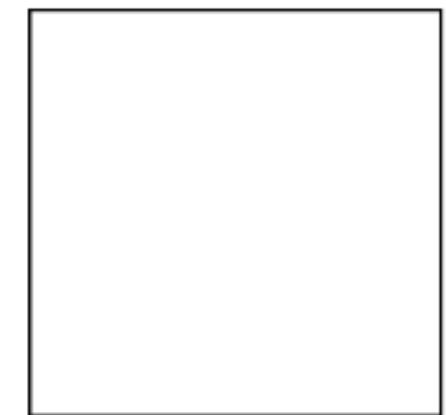
- rotational symmetry
- a ‘2D’ version of a sinc



Important Fourier Transform Pairs

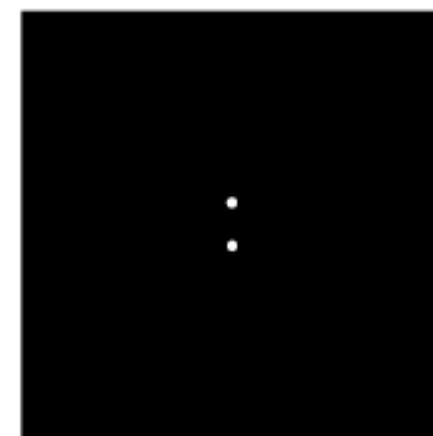
$$f(x, y) = \delta(x, y) = \delta(x)\delta(y)$$

$$\begin{aligned} F(u, v) &= \int \int \delta(x, y) e^{-j2\pi(ux+vy)} dx dy \\ &= 1 \end{aligned}$$



f(x,y)

F(u,v)



A. Zisserman

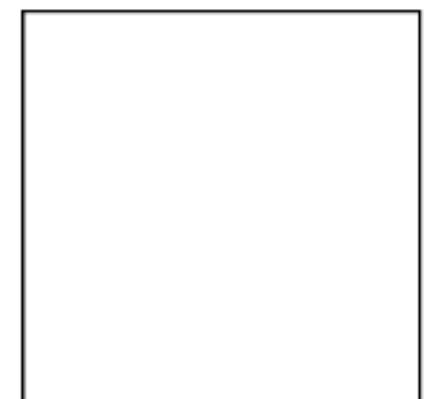
Important Fourier Transform Pairs

$$f(x, y) = \delta(x, y) = \delta(x)\delta(y)$$

$$\begin{aligned} F(u, v) &= \int \int \delta(x, y) e^{-j2\pi(ux+vy)} dx dy \\ &= 1 \end{aligned}$$

$$f(x, y) = \frac{1}{2} (\delta(x, y - a) + \delta(x, y + a))$$

$$\begin{aligned} F(u, v) &= \frac{1}{2} \int \int (\delta(x, y - a) + \delta(x, y + a)) e^{-j2\pi(ux+vy)} dx dy \\ &= \frac{1}{2} (e^{-j2\pi av} + e^{j2\pi av}) = \cos 2\pi av \end{aligned}$$



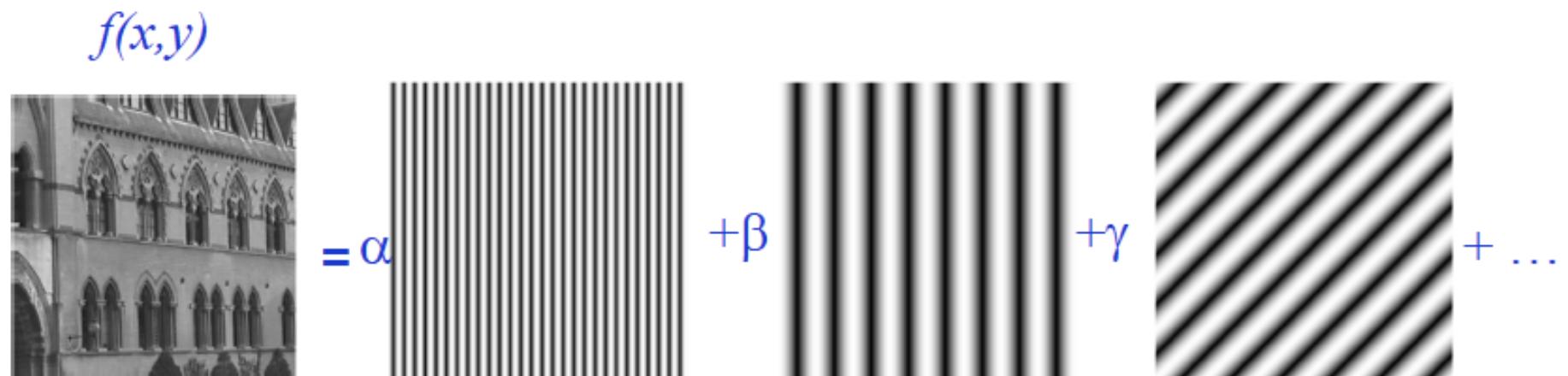
A. Zisserman

2D Fourier Decomposition

The spatial function $f(x, y)$

$$f(x, y) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F(u, v) e^{j2\pi(ux+vy)} du dv$$

is decomposed into a weighted sum of 2D orthogonal basis functions in a similar manner to decomposing a vector onto a basis using scalar products.



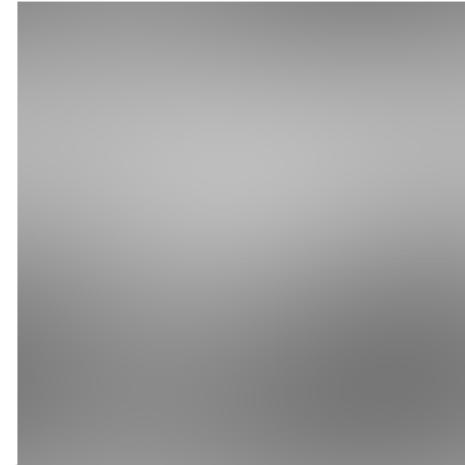
Basis reconstruction



Full image



First 1 basis fn



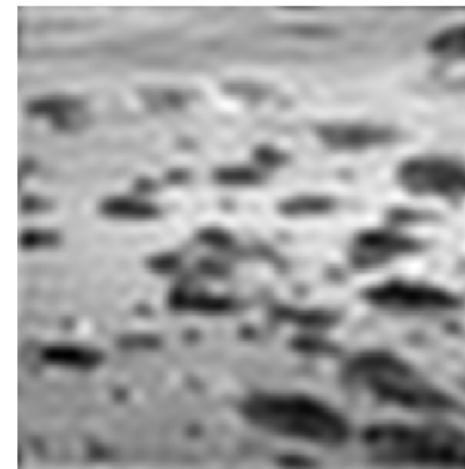
First 4 basis fns



First 9 basis fns

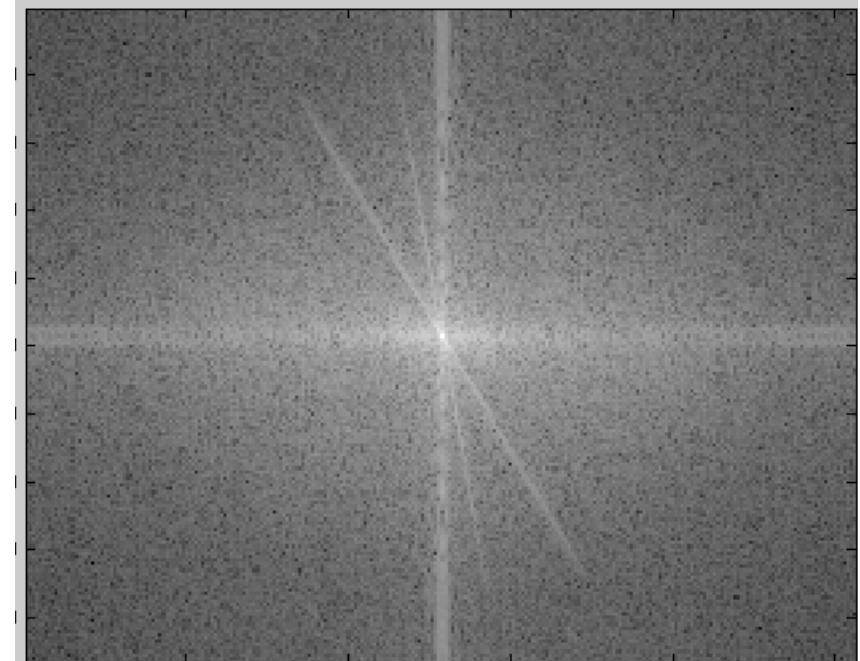


First 16 basis fns



First 400 basis fns

2D Fourier Transform of Real Images

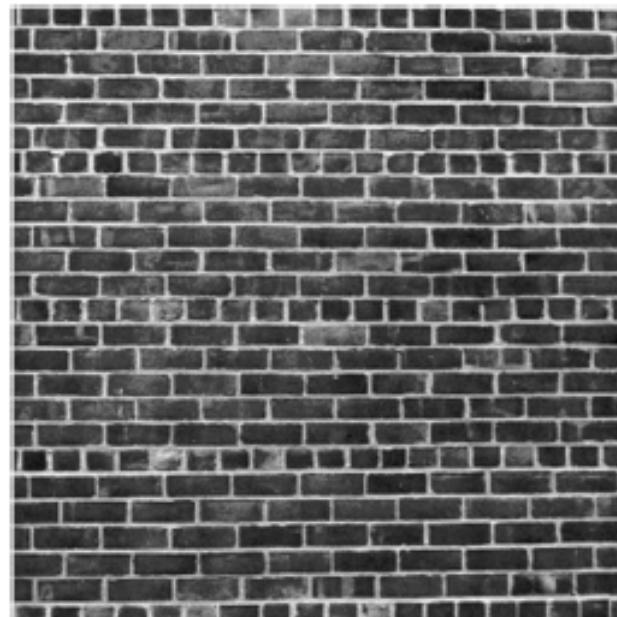


What does it mean to be at pixel x,y ?

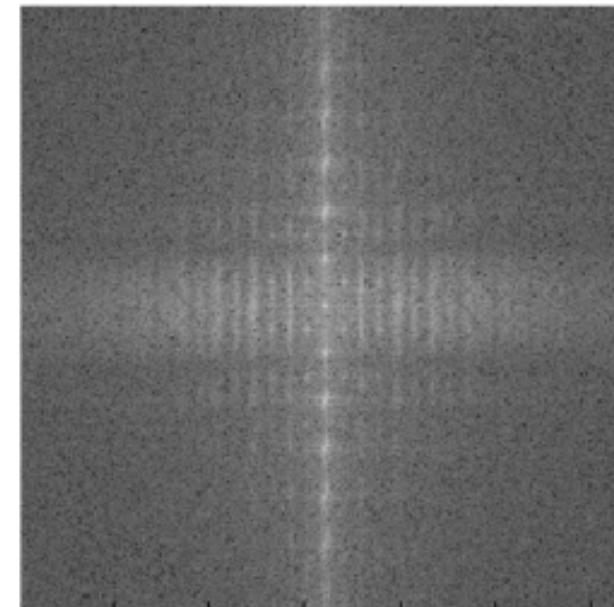
What does it mean to be more or less bright in the Fourier decomposition image?

2D Fourier Transform of Real Images

Image with periodic structure



$f(x,y)$

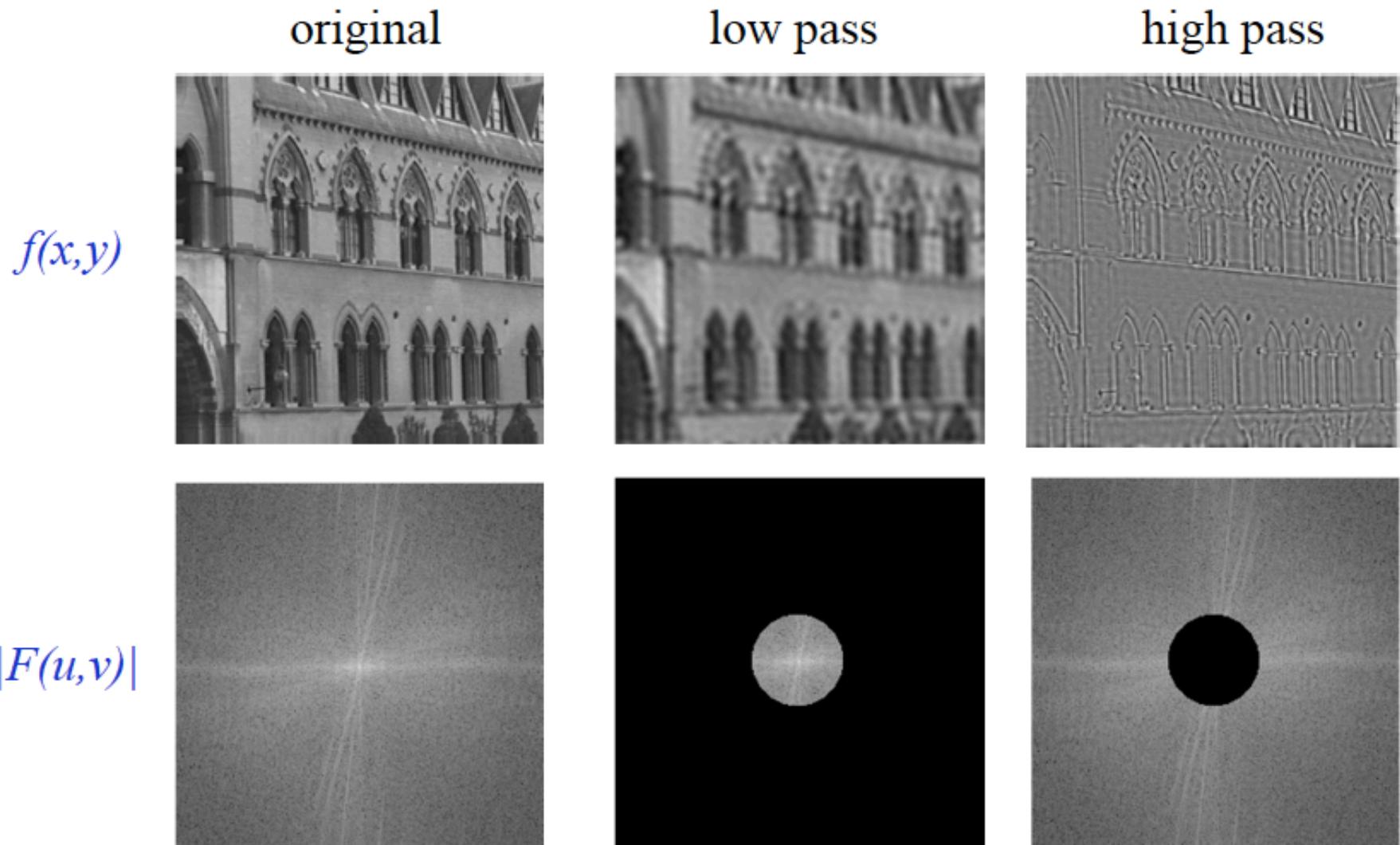


$|F(u,v)|$

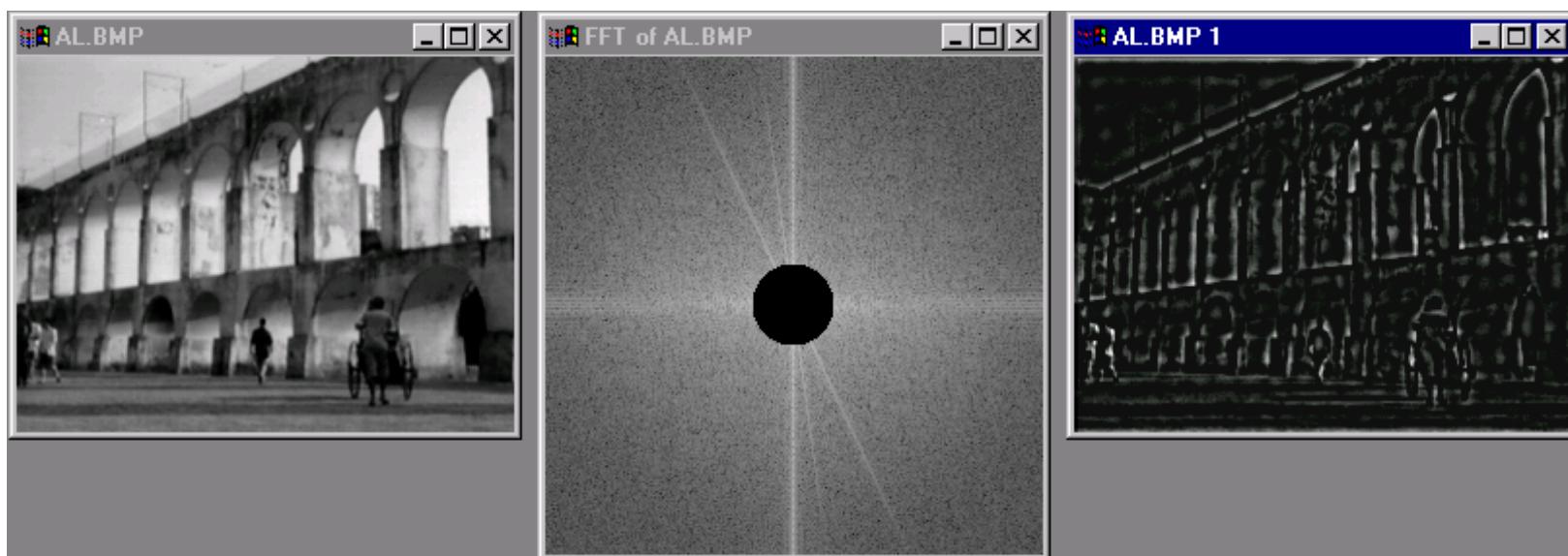
FT has peaks at spatial frequencies of repeated texture

A. Zisserman

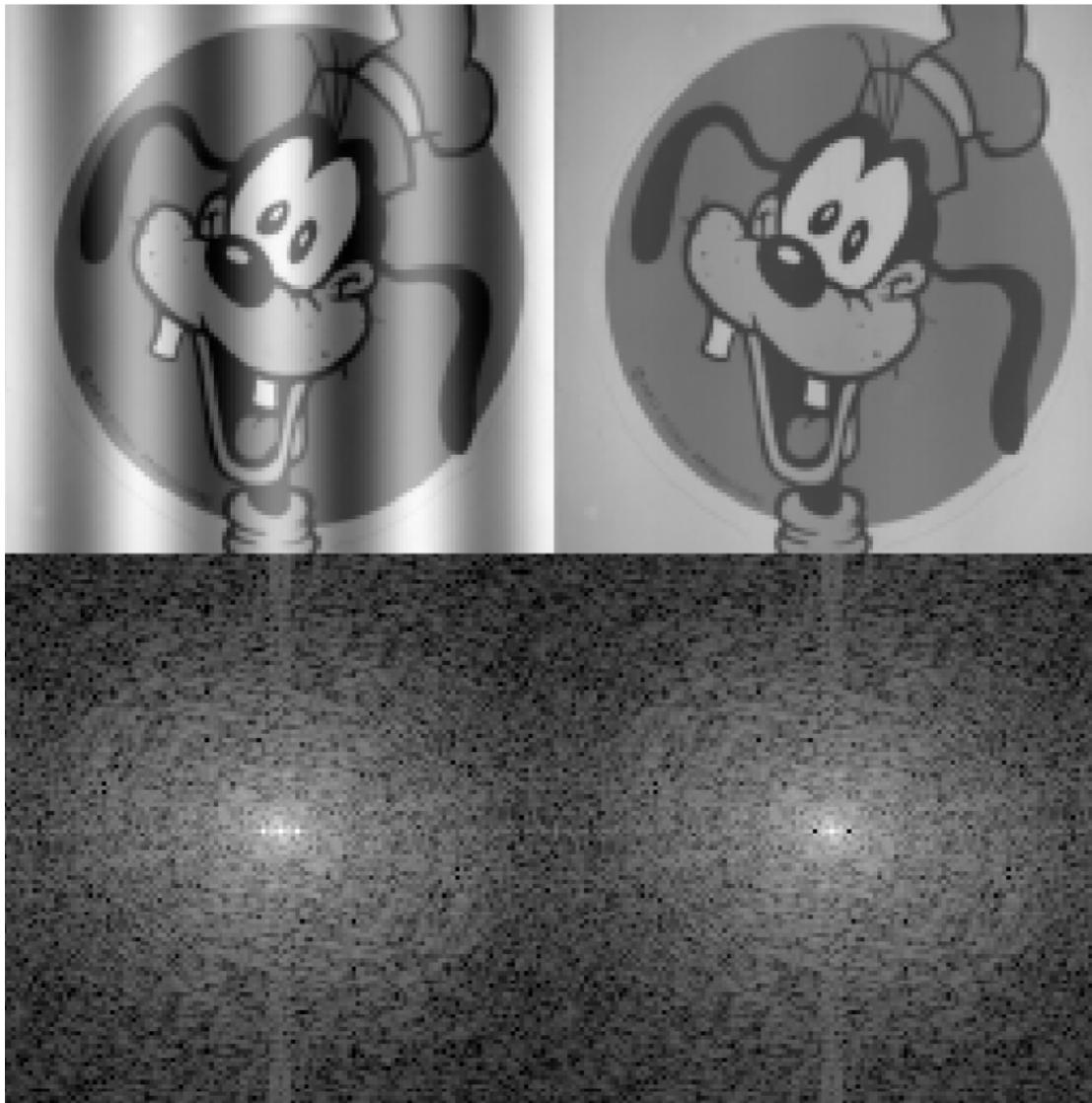
Low/High Pass Filters



Low and High Pass filtering



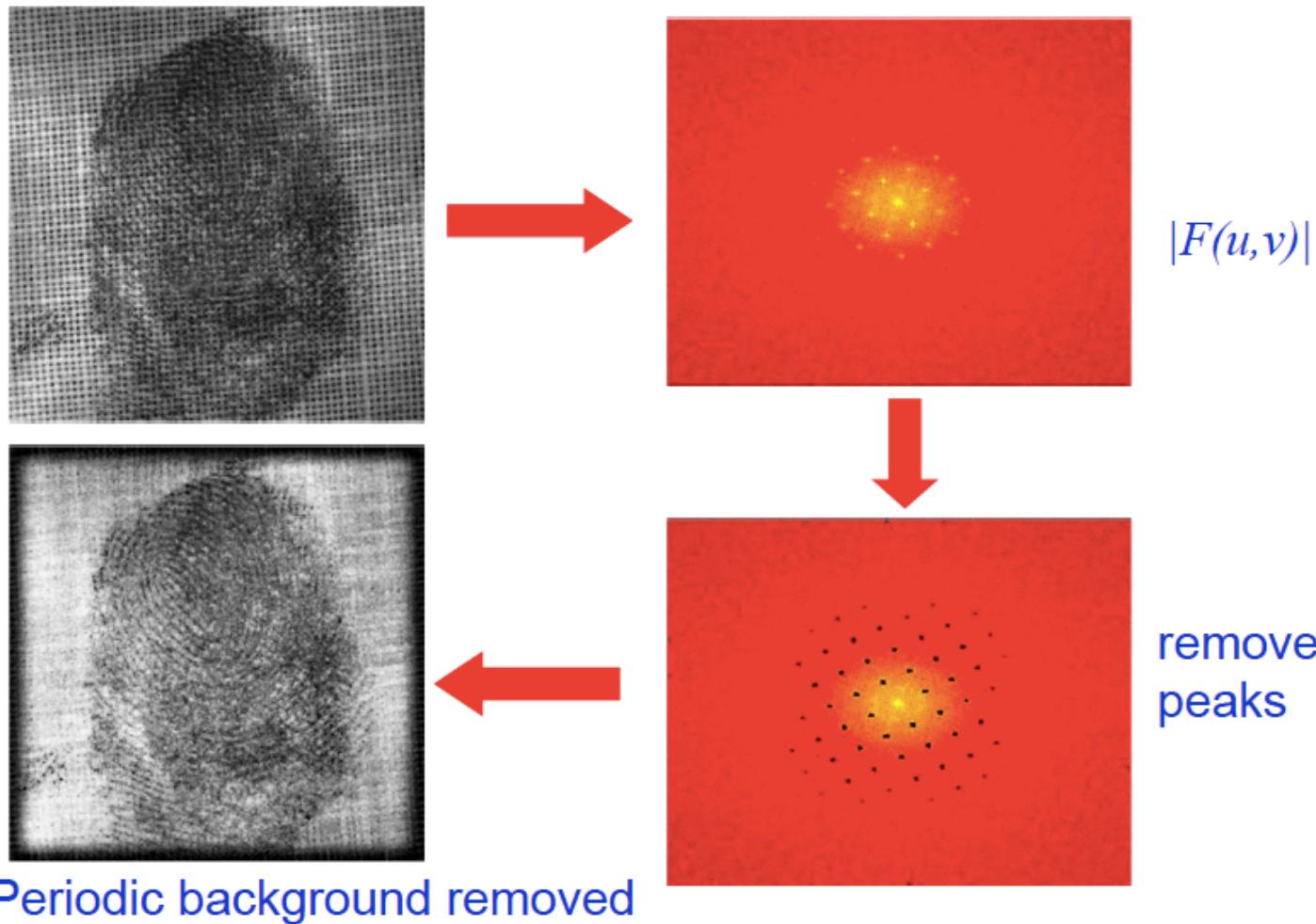
Removing frequency bands



Brayer

Removing frequency bands

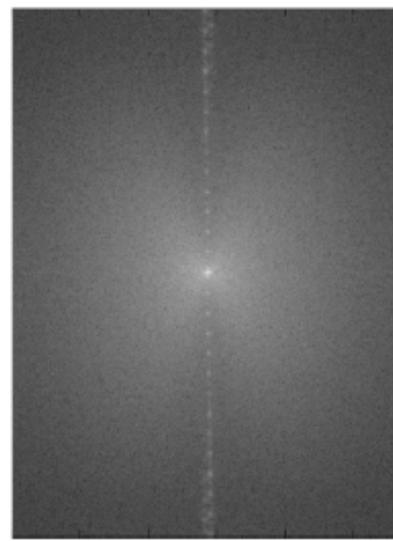
Example – Forensic application



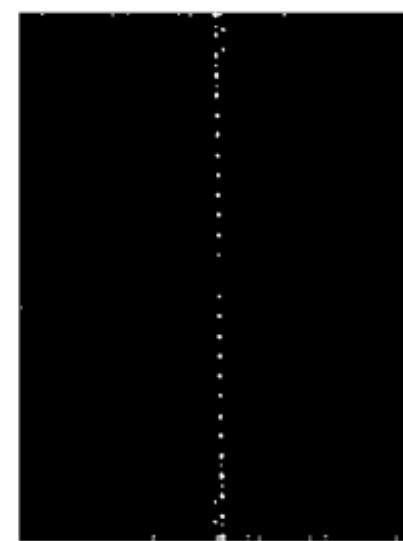
A. Zisserman

Removing frequency bands

Lunar orbital image (1966)



$$|F(u,v)|$$

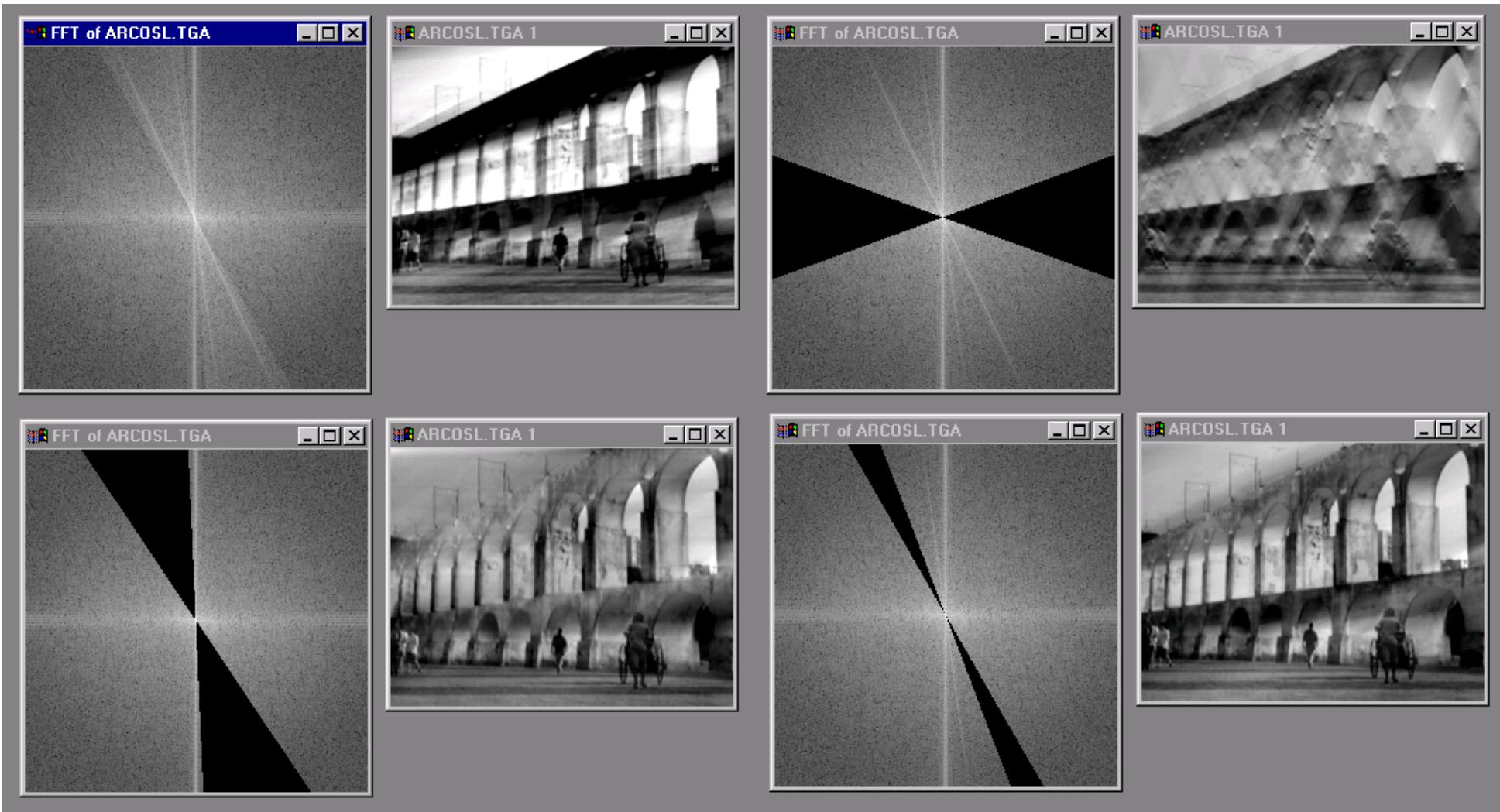


remove
peaks



join lines
removed

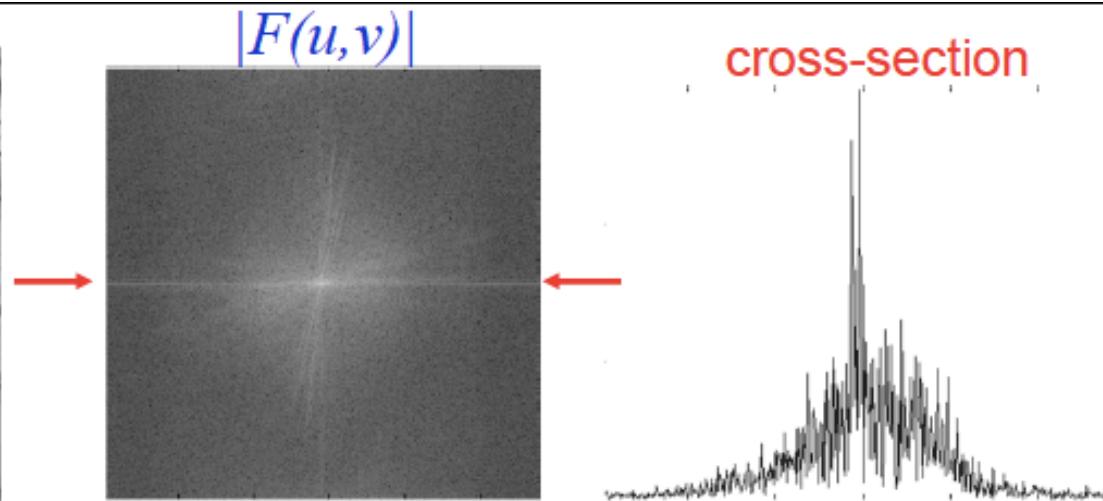
Editting frequencies



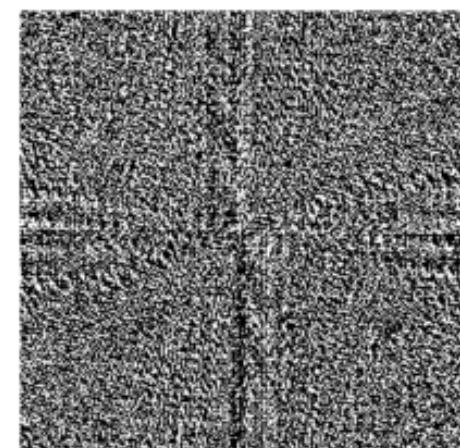
Magnitude vs. Phase



$f(x,y)$

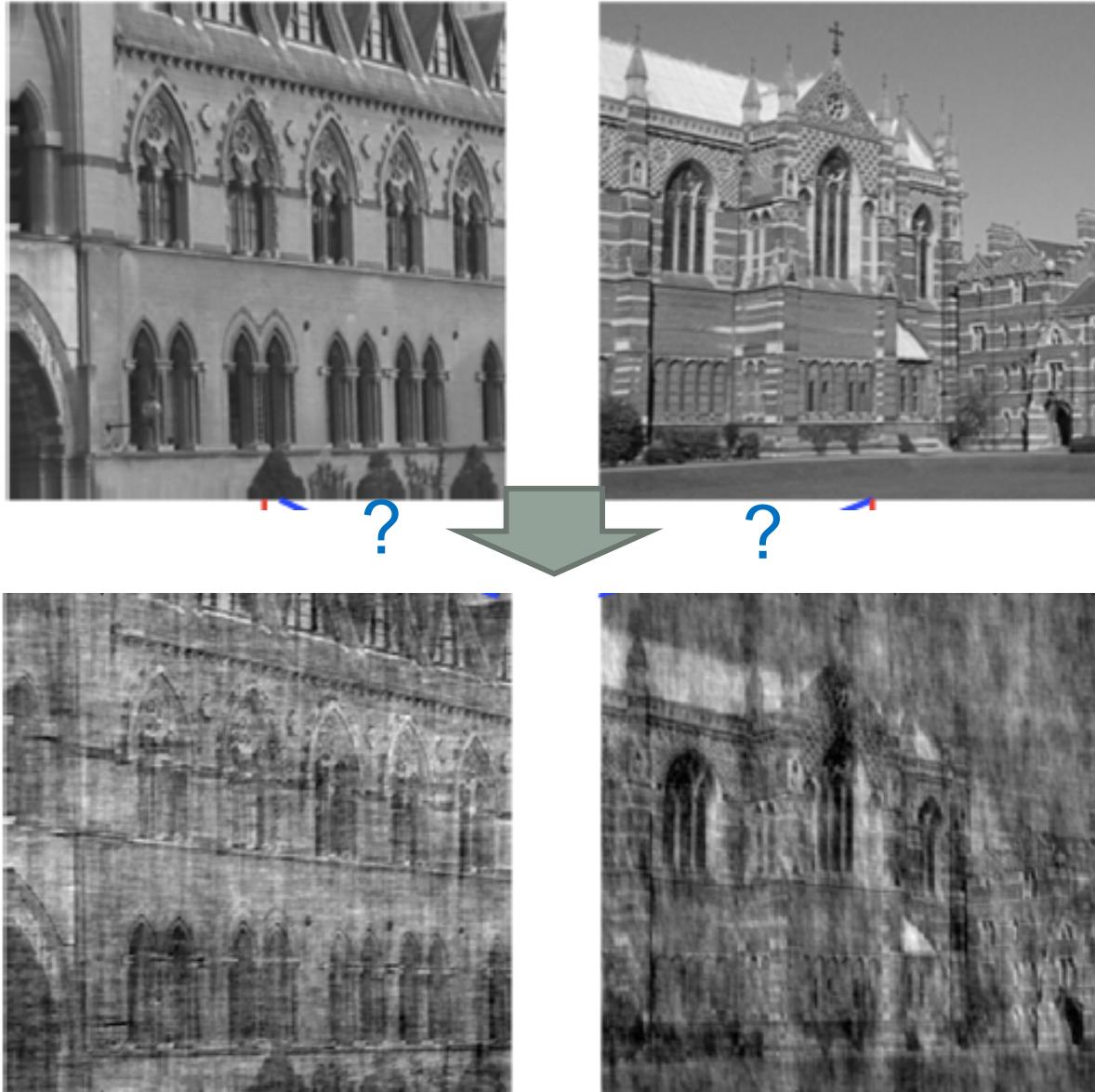


$|F(u,v)|$



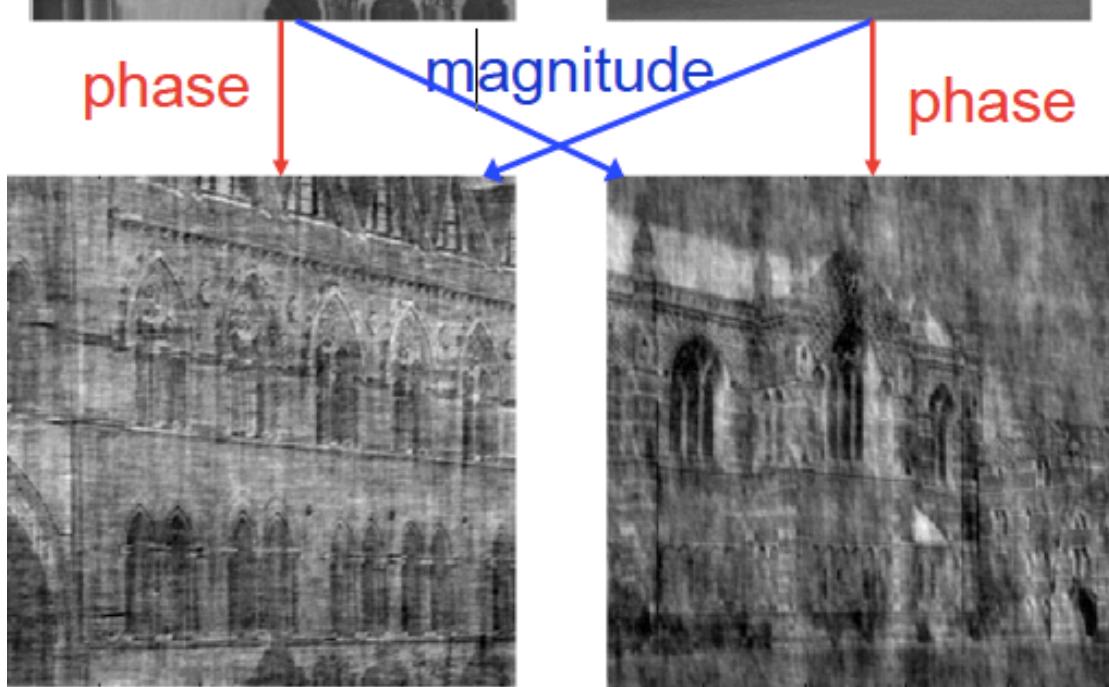
- $|f(u,v)|$ generally decreases with higher spatial frequencies
- phase appears less informative

The Importance of Phase



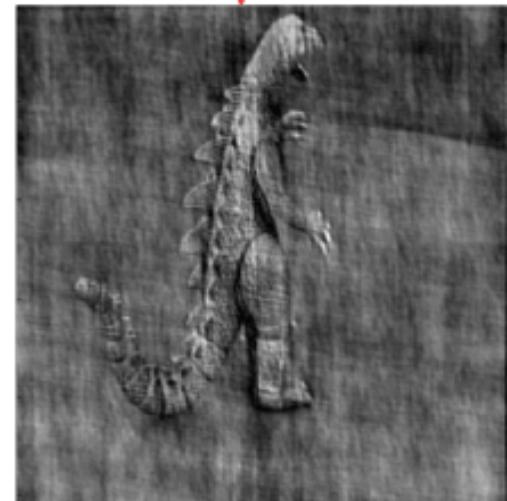
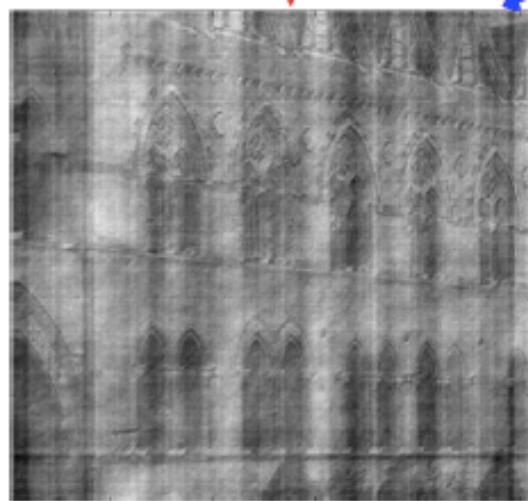
A. Zisserman

The Importance of Phase



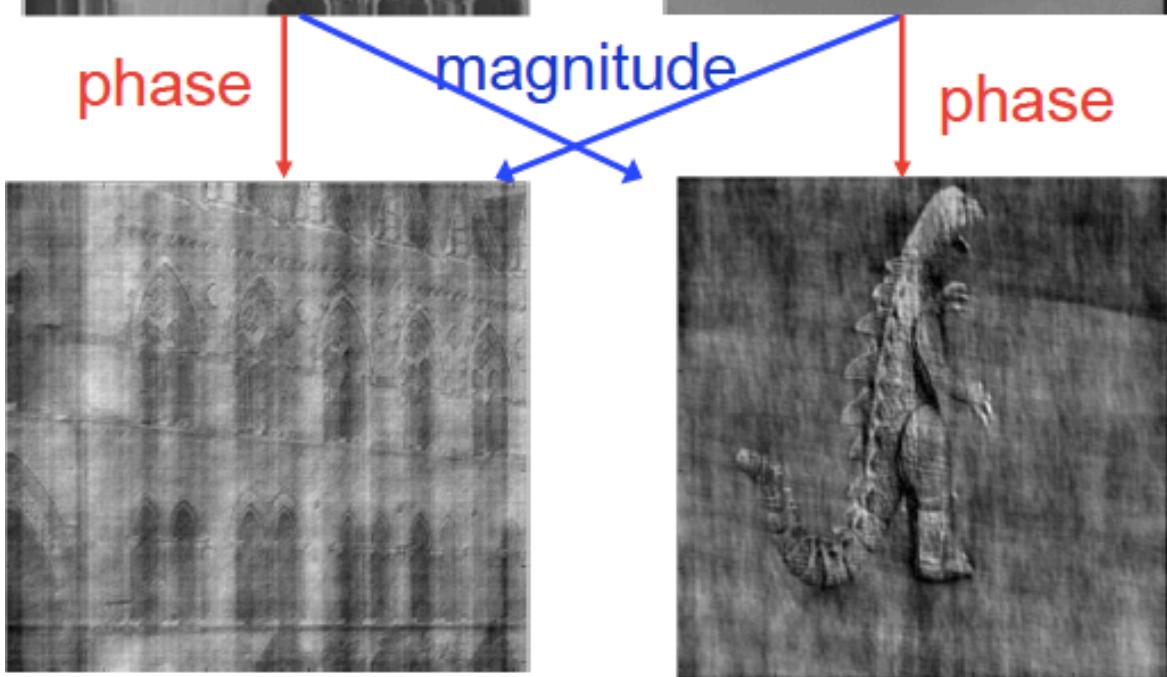
A. Zisserman

Phase and Magnitude- Another Example



A. Zisserman

The Importance of Phase



A. Zisserman

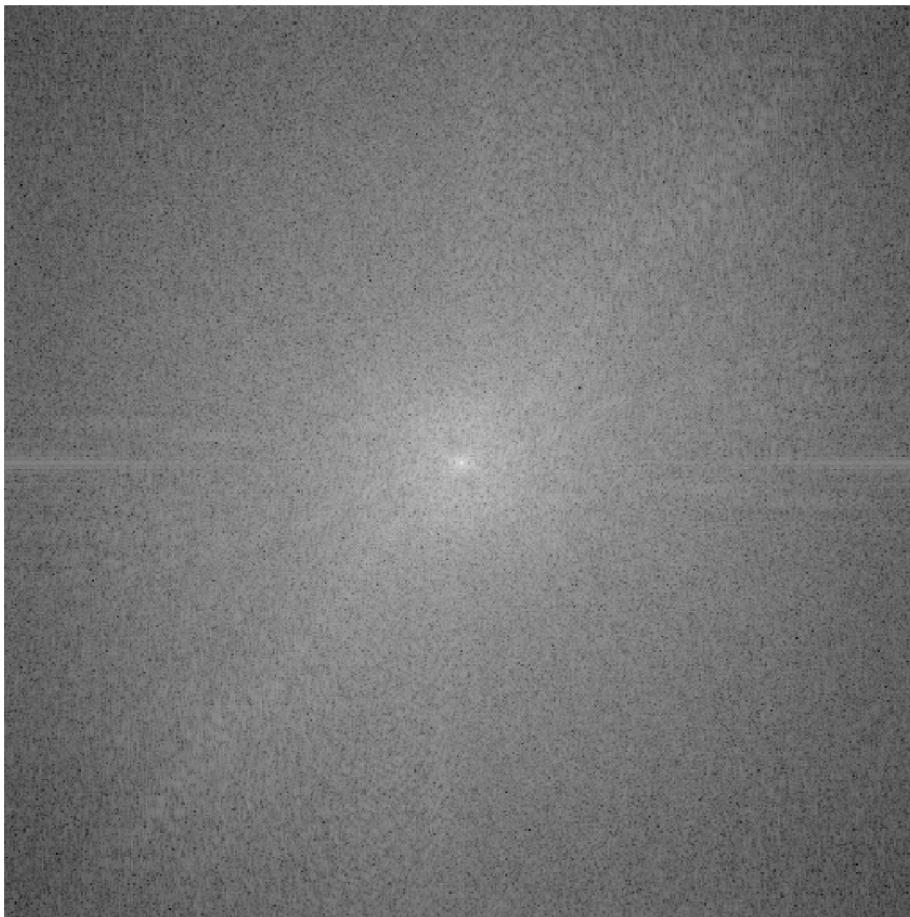
Phase and Magnitude- Yet Another Example



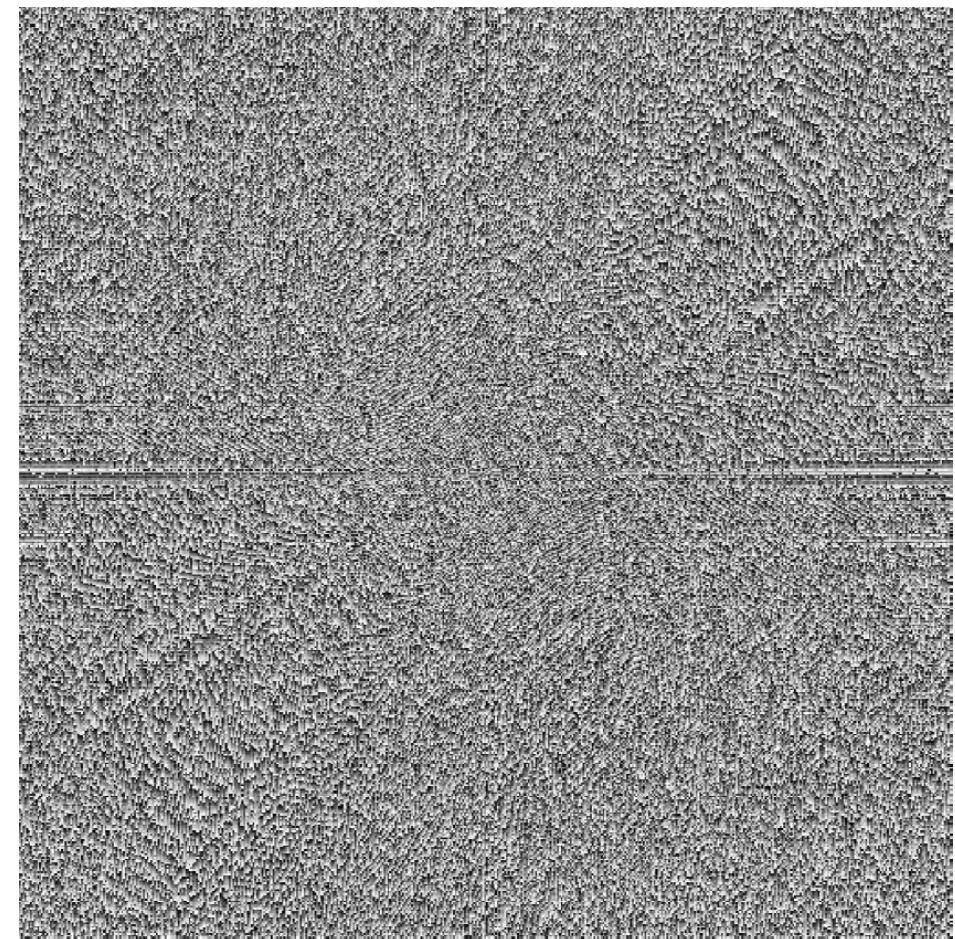
Efros

Phase and Magnitude- Yet Another Example

Amplitude



Phase



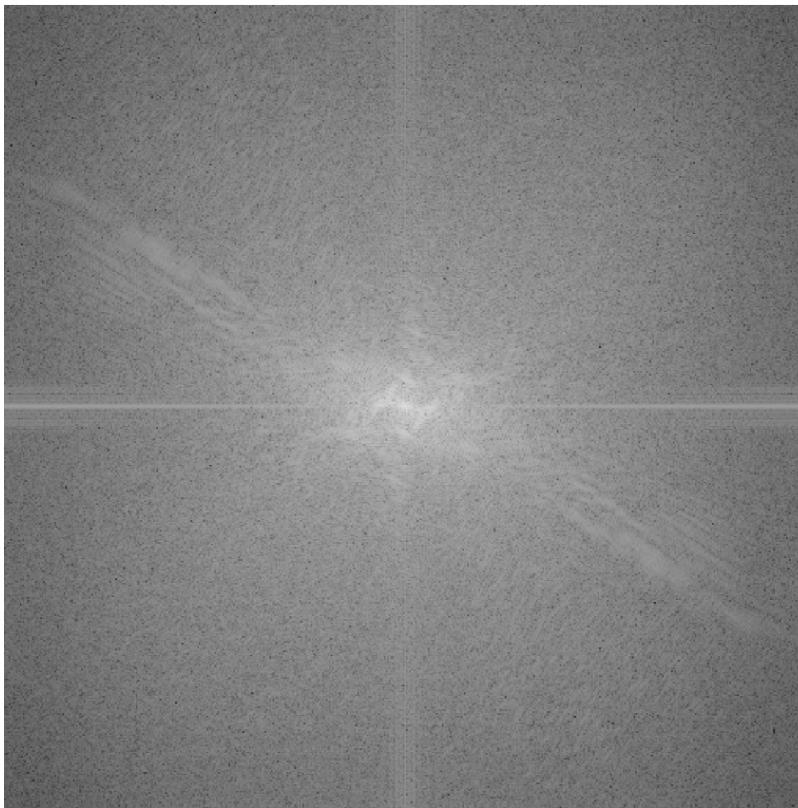
Efros

Phase and Magnitude- Yet Another Example

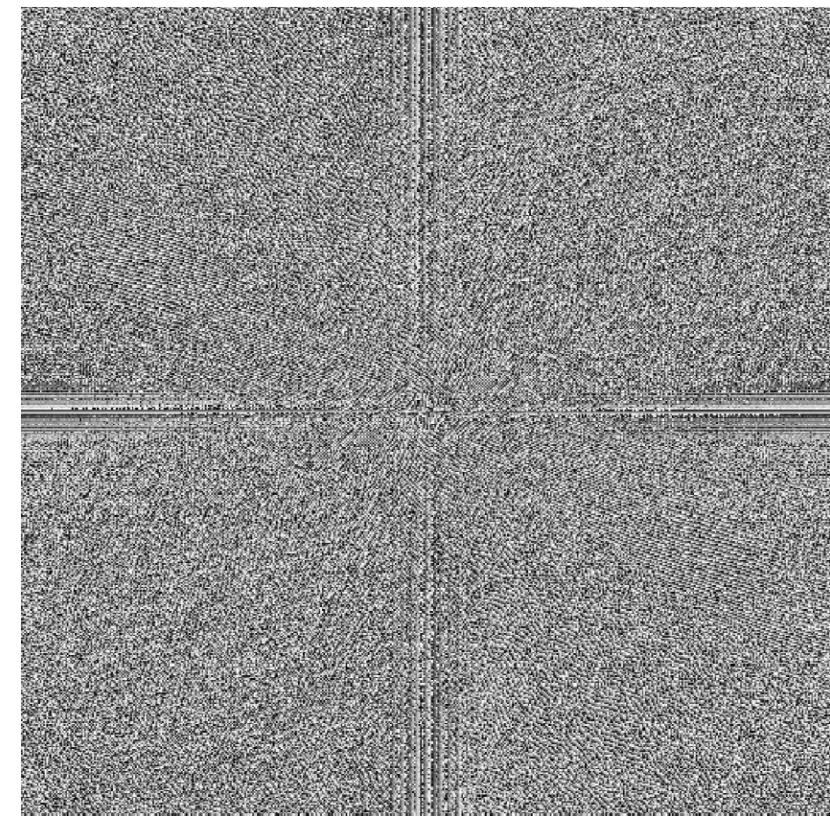


What about phase?

Amplitude



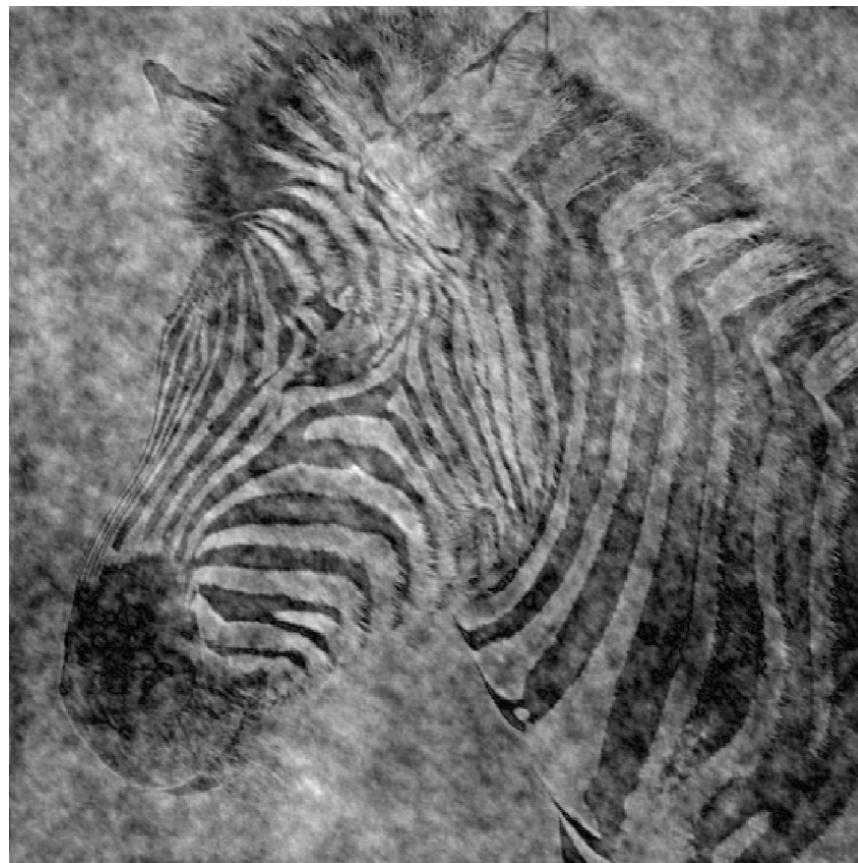
Phase



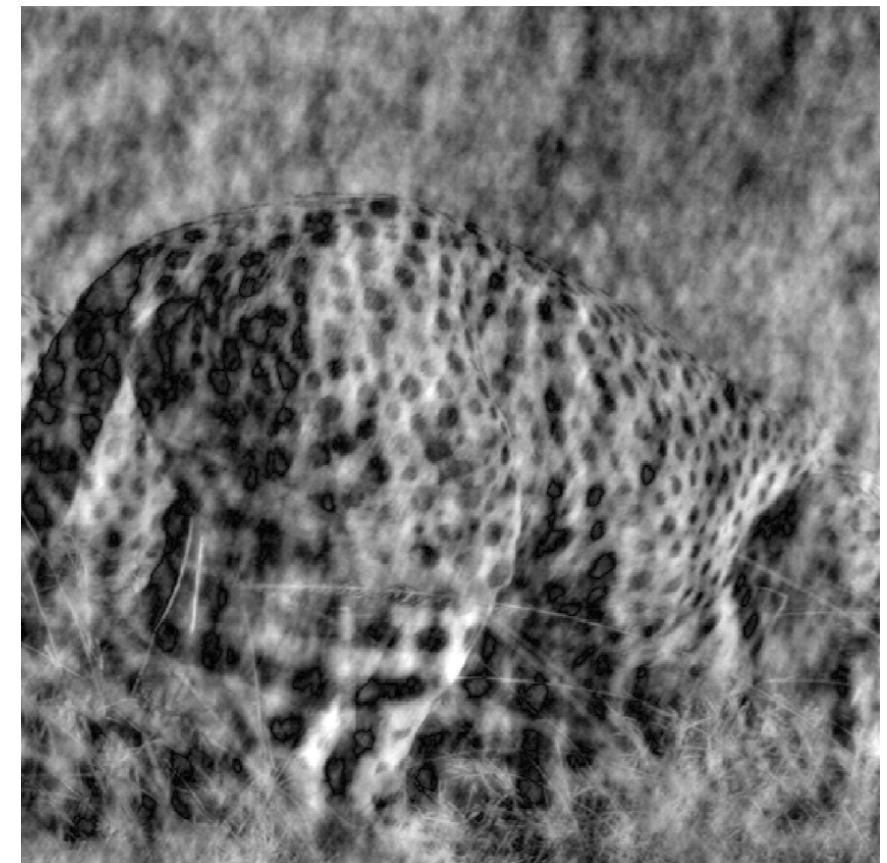
Efros

Cheebra

Zebra phase, cheetah amplitude



Cheetah phase, zebra amplitude



Phase and Frequency

- The frequency amplitude of natural images are quite similar
 - Heavy in low frequencies, falling off in high frequencies
 - Will *any* image be like that, or is it a property of the world we live in?
- Most information in the image is carried in the phase, not the amplitude
 - Not quite clear why

Properties of the Fourier Transform

As in the 1D case FTs have the following properties

- Linearity

$$\alpha f(x, y) + \beta g(x, y) \Leftrightarrow \alpha F(u, v) + \beta G(u, v).$$

- Similarity

$$f(ax, by) \Leftrightarrow \frac{1}{ab} F\left(\frac{u}{a}, \frac{v}{b}\right).$$

This applies, for example, when an image is scaled

- Shift

$$f(x - a, y - b) \Leftrightarrow e^{j2\pi(au+bv)} F(u, v).$$

This might apply, for example, if an object moved.

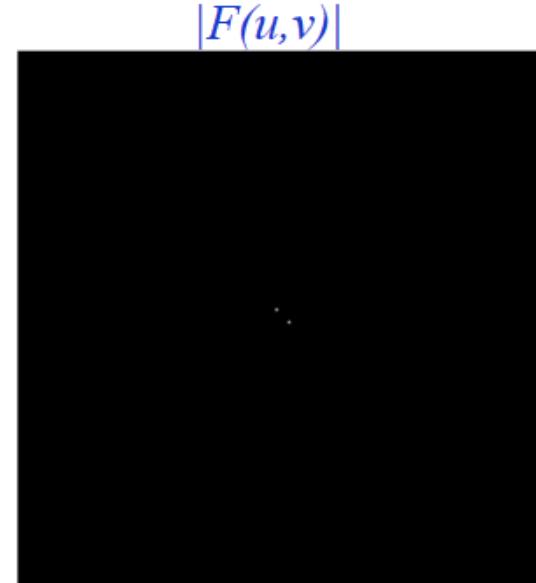
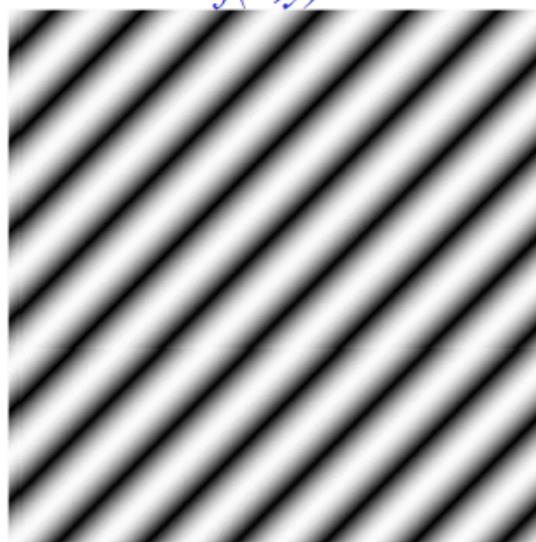
Properties of Fourier Transforms

In 2D can also rotate, shear etc

Under an affine transformation: $\mathbf{x} \rightarrow A\mathbf{x}$ $\mathbf{u} \rightarrow A^{-T}\mathbf{u}$

Example

How does $F(u,v)$ transform if $f(x,y)$ is rotated by 45 degrees?



If $A = R$ then $A^{-T} = R$.

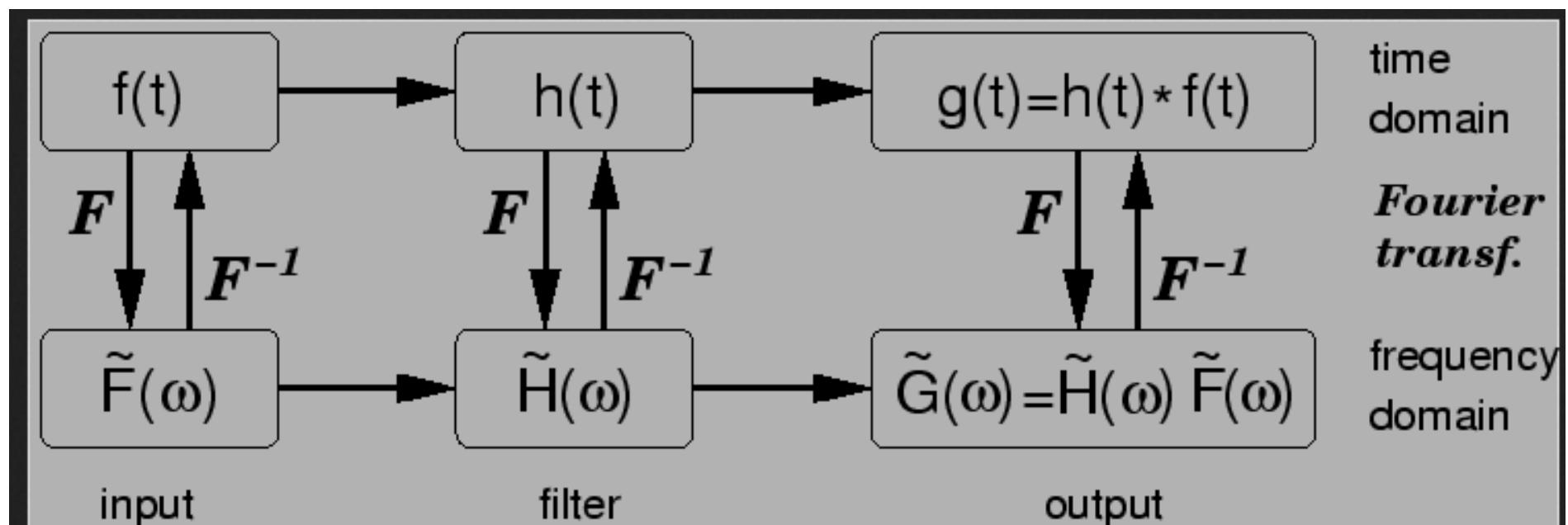
i.e. FT undergoes the same rotation.

Properties of Fourier Transforms

- Fourier transform of a real signal is symmetric about the origin
- The energy of the signal is the same as the energy of its Fourier transform

See Szeliski Book (3.4)

The Convolution Theorem



<http://jclahr.com/science/psn/wielandt/node8.html>

Filtering Vs. Convolution in 1D

$$g(x) = \sum_i f(x + i)h(i)$$

filtering $f(x)$ with $h(x)$

$$f(x) \quad \boxed{100 \mid 200 \mid 100 \mid 200 \mid 90 \mid 80 \mid 80 \mid 100 \mid 100}$$

$$h(x) \quad \boxed{1/4 \mid 1/2 \mid 1/4}$$

molecule/template/kernel

$$g(x) \quad \boxed{\mid 150 \mid \quad \mid \quad \mid \quad \mid \quad \mid \quad \mid \quad \mid}$$

$$g(x) = \int f(u)h(x - u) du \quad \text{convolution of } f(x) \text{ and } h(x)$$

$$= \int f(x + u')h(-u') du' \quad \text{after change of variable } u' = u - x$$

$$= \sum f(x + i)h(-i)$$

Filtering Vs. Convolution in 1D

$$g(x) = \sum_i f(x + i)h(i)$$

filtering $f(x)$ with $h(x)$

$$\begin{aligned} g(x) &= \int f(u)h(x - u) du && \text{convolution of } f(x) \text{ and } h(x) \\ &= \int f(x + u')h(-u') du' && \text{after change of} \\ &= \sum f(x + i)h(-i) && \text{variable } u' = u - x \end{aligned}$$

Filtering Vs. Convolution in 1D

$$g(x) = \sum_i f(x + i)h(i)$$

filtering $f(x)$ with $h(x)$

$$f(x) \quad \boxed{100 \mid 200 \mid 100 \mid 200 \mid 90 \mid 80 \mid 80 \mid 100 \mid 100}$$

$$h(x) \quad \boxed{1/4 \mid 1/2 \mid 1/4}$$

molecule/template/kernel

$$g(x) \quad \boxed{\mid 150 \mid \quad \mid \quad \mid \quad \mid \quad \mid \quad \mid \quad \mid}$$

$$g(x) = \int f(u)h(x - u) du \quad \text{convolution of } f(x) \text{ and } h(x)$$

$$= \int f(x + u')h(-u') du' \quad \text{after change of variable } u' = u - x$$

$$= \sum f(x + i)h(-i)$$

Filtering Vs. Convolution in 1D

$$g(x) = \sum_i f(x + i)h(i)$$

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- note negative sign (which is a reflection in x) in convolution
- $h(x)$ is often symmetric (even/odd), and then (e.g. for even)

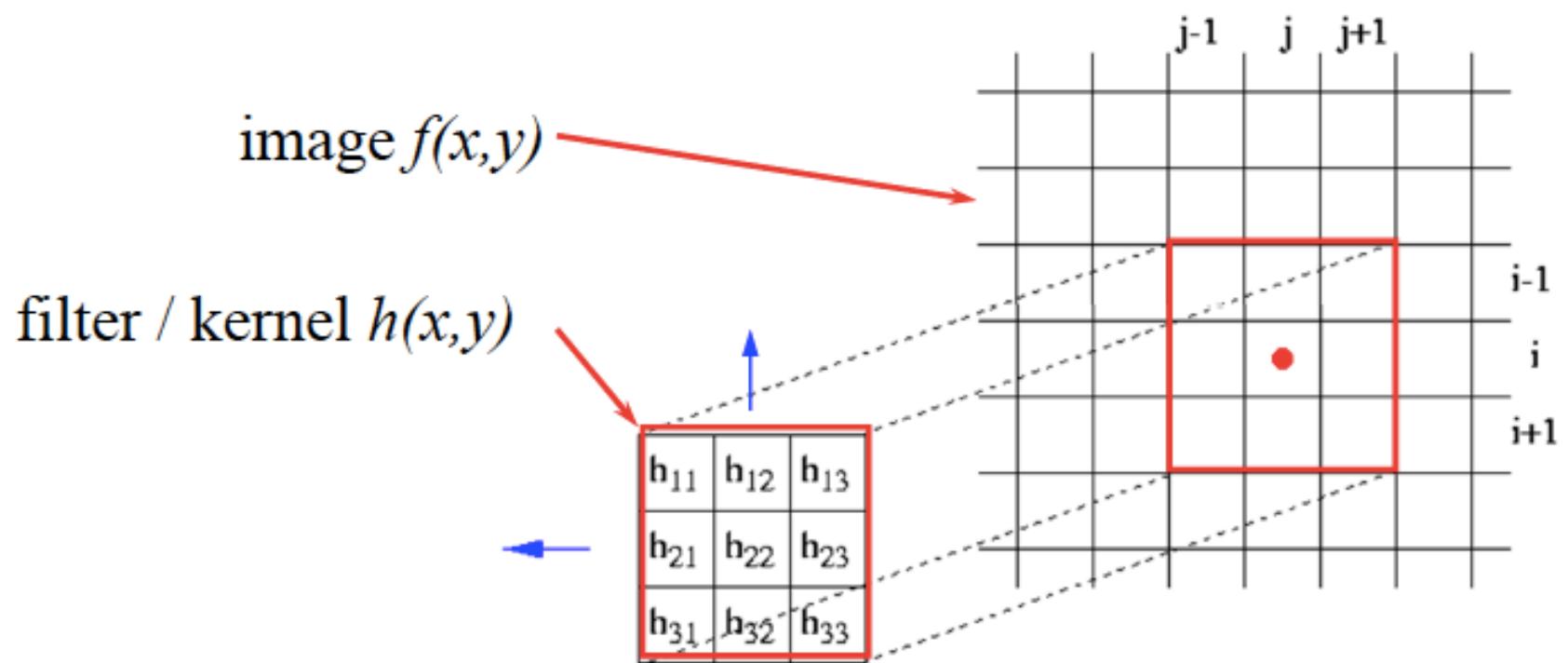
$$g(x) = \sum_i f(x + i)h(i)$$

Filtering Vs. Convolution in 2D

convolution

$$\begin{aligned}g(x, y) &= h(x, y) * f(x, y) = f(x, y) * h(x, y) \\&= \int \int f(u, v)h(x - u, y - v) du dv\end{aligned}$$

filtering



Filtering Vs. Convolution in 2D

convolution

$$\begin{aligned}g(x, y) &= h(x, y) * f(x, y) = f(x, y) * h(x, y) \\&= \int \int f(u, v)h(x - u, y - v) du dv\end{aligned}$$

filtering

$$g(x,y) = h_{11} f(i - 1, j - 1) + h_{12} f(i - 1, j) + h_{13} f(i - 1, j + 1) + \\h_{21} f(i, j - 1) + h_{22} f(i, j) + h_{23} f(i, j + 1) + \\h_{31} f(i + 1, j - 1) + h_{32} f(i + 1, j) + h_{33} f(i + 1, j + 1)$$

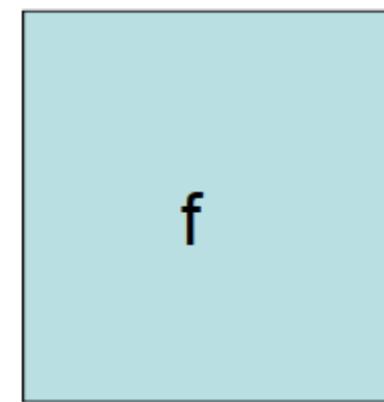
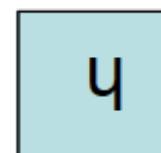
for convolution, reflect filter in x and y axes

Convolution

- Convolution:
 - Flip the filter in both dimensions (bottom to top, right to left)

$$g[i, j] = \sum_{u=-k}^k \sum_{v=-k}^k h[u, v] f[i - u, j - v]$$

convolution with h



Filtering vs. Convolution in 2D Matlab

2D filtering

- `g=filter2 (h, f) ;`

`f=image`
`h=filter`

$$g[m,n] = \sum_{k,l} h[k,l] f[m+k, n+l]$$

2D convolution

- `g=conv2 (h, f) ;`

$$g[m,n] = \sum_{k,l} h[k,l] f[m-k, n-l]$$

Convolution Theorem

$$f(x, y) * h(x, y) \Leftrightarrow F(u, v)H(u, v)$$

Space convolution = frequency multiplication

In words: the Fourier transform of the convolution of two functions is the product of their individual Fourier transforms

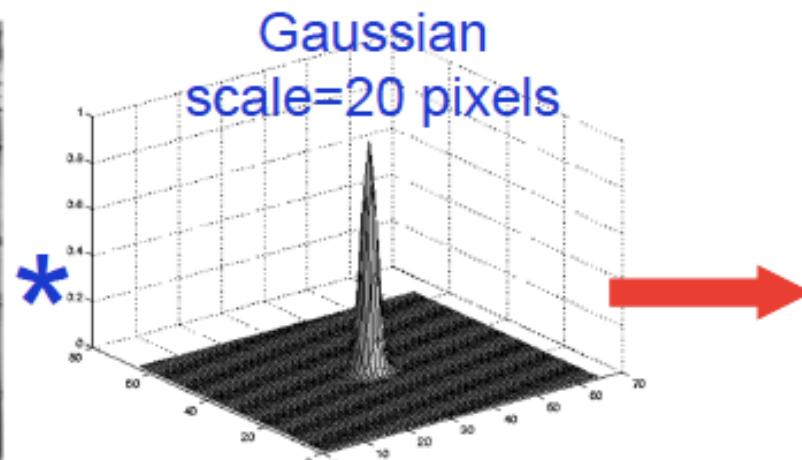
Why is this so important?

Because linear filtering operations can be carried out by simple multiplications in the Fourier domain

The Importance of Convolution Theorem

It establishes the link between operations in the frequency domain and the action of linear spatial filters

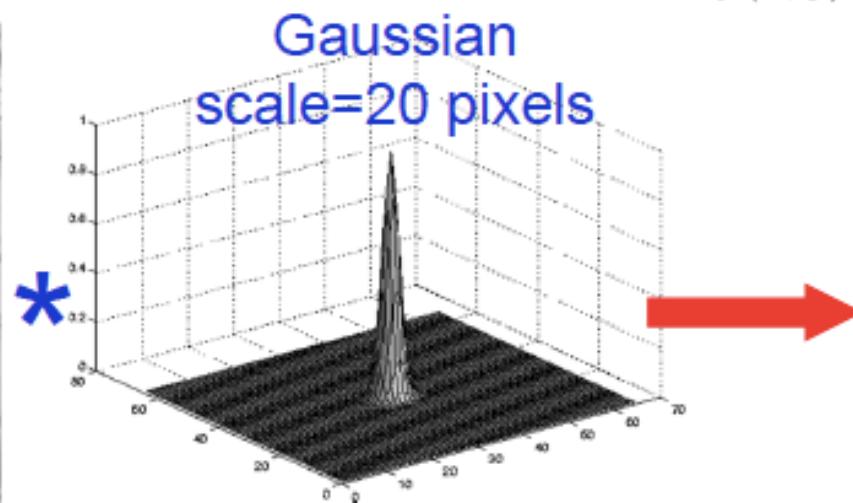
Example smooth an image with a Gaussian spatial filter



$$f(x, y) * g(x, y) \Leftrightarrow F(u, v)G(u, v)$$

The Importance of Convolution Theorem

Example smooth an image with a Gaussian spatial filter

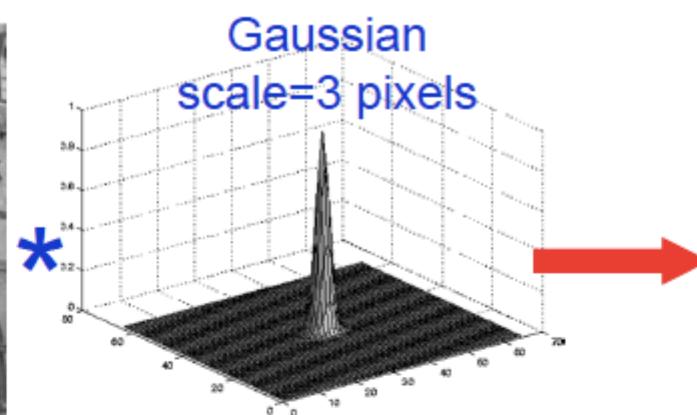
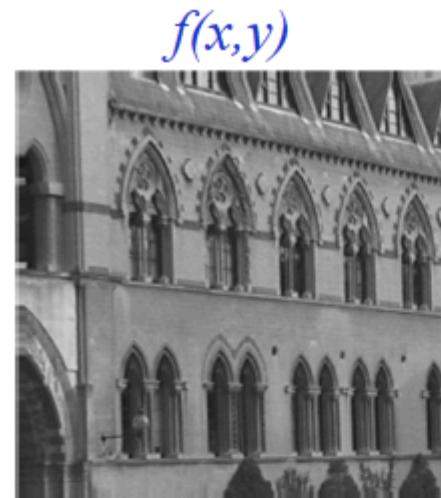


$$f(x, y) * g(x, y) \Leftrightarrow F(u, v)G(u, v)$$

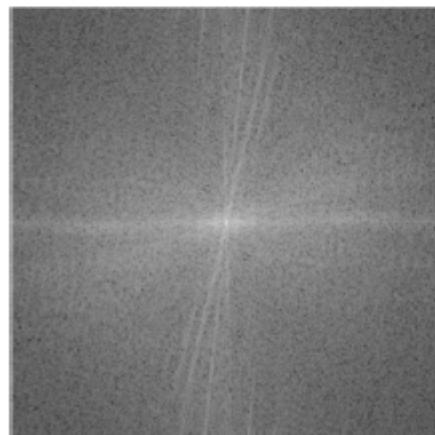


1. Compute FT of image and FT of Gaussian
2. Multiply FT's
3. Compute inverse FT of the result.

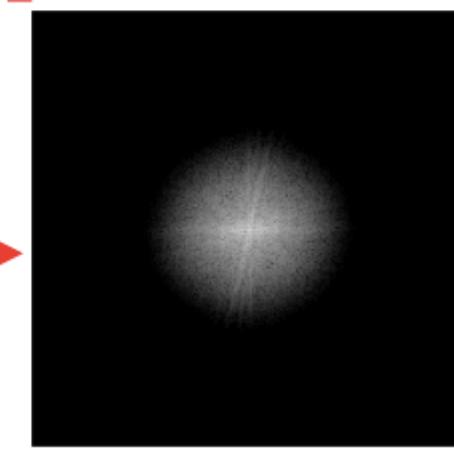
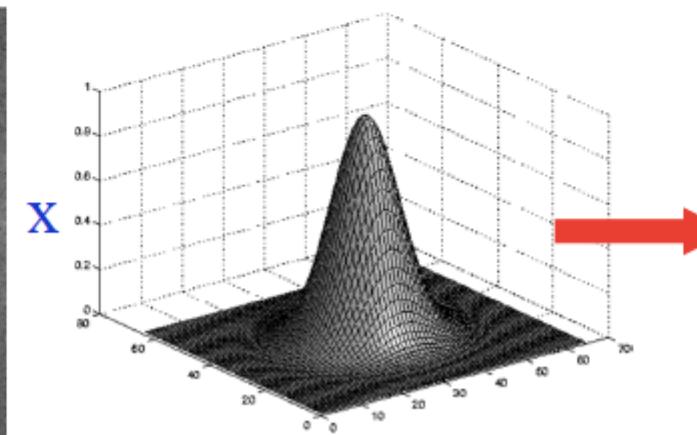
The Importance of Convolution Theorem



Fourier transform

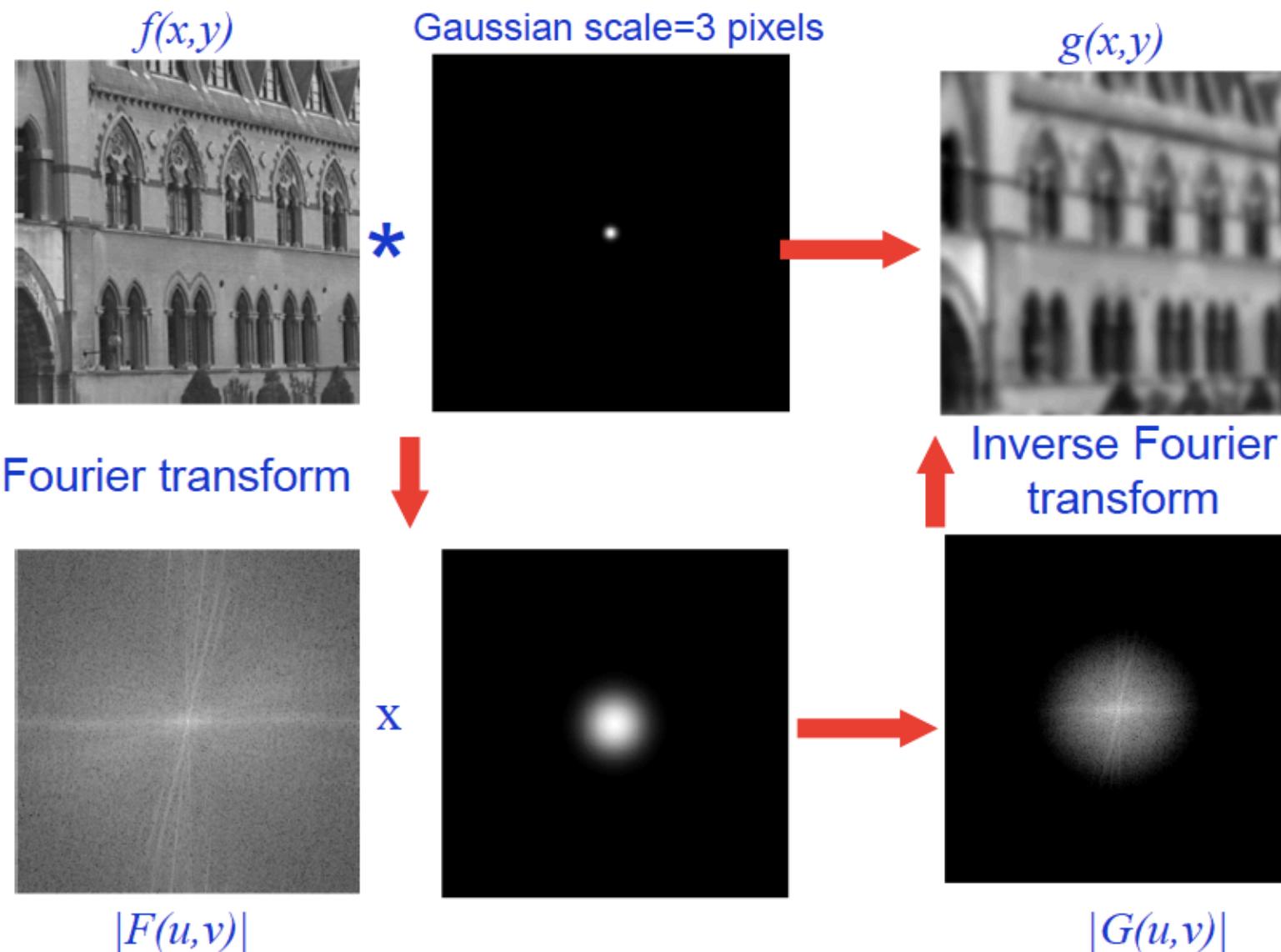


$$|F(u,v)|$$



$$|G(u,v)|$$

The Importance of Convolution Theorem



Filtering: Spatial Domain vs. Frequency Domain

There are two equivalent ways of carrying out linear spatial filtering operations:

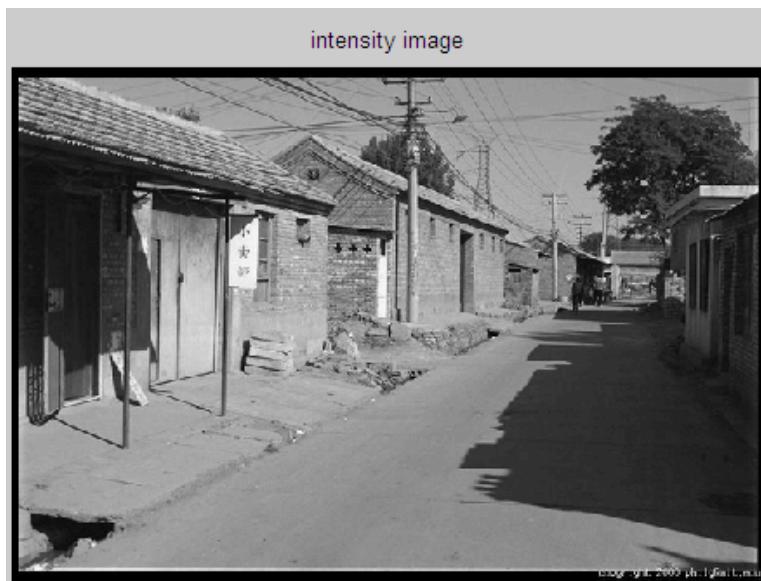
1. Spatial domain: convolution with a spatial operator
2. Frequency domain: multiply FT of signal and filter, and compute inverse FT of product

Why choose one over the other ?

- The filter may be simpler to specify or compute in one of the domains
- Computational cost

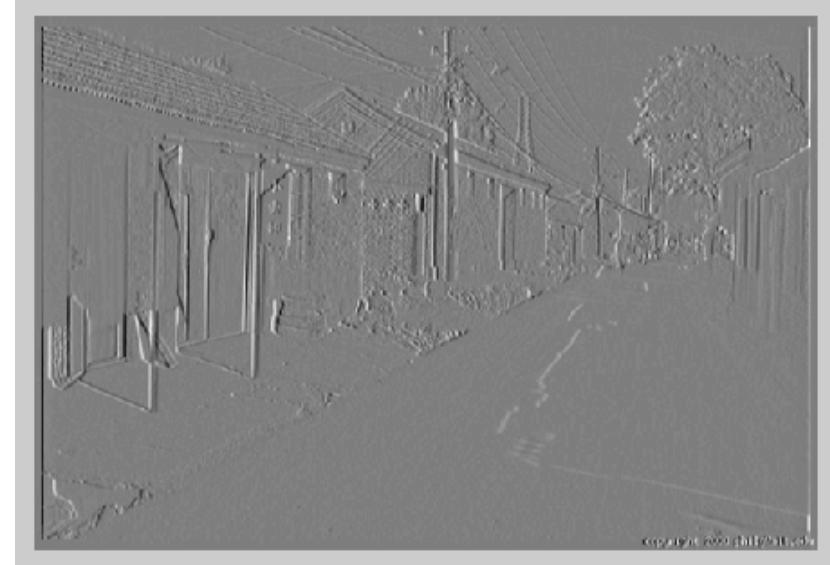
More on Filtering in spatial domain

1	0	-1
2	0	-2
1	0	-1



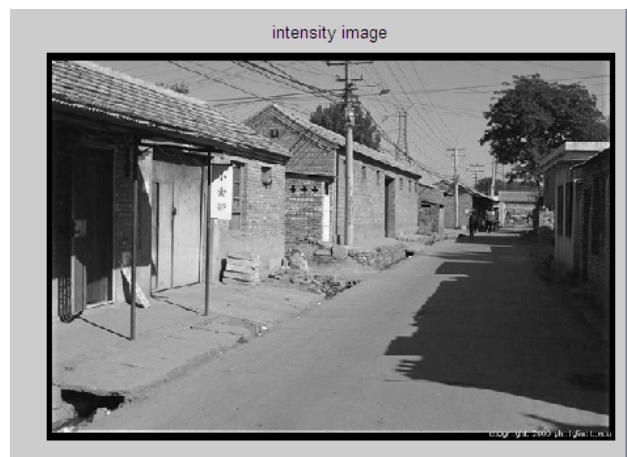
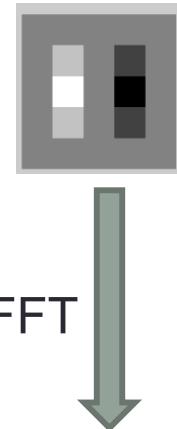
$$\ast =$$

A 3x3 kernel for edge detection, represented as a 3x3 grid of gray squares. The central square is white, and the squares in the three columns are arranged from shortest to tallest vertically, creating a gradient effect.

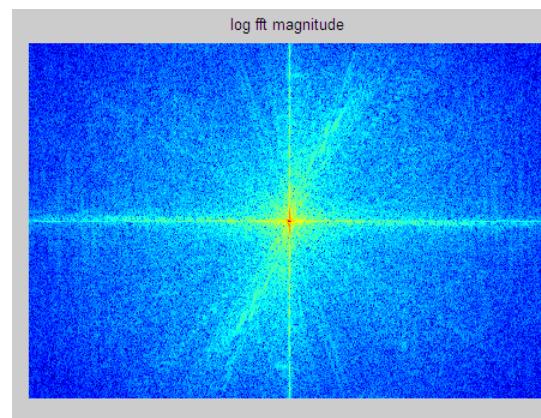


Hays

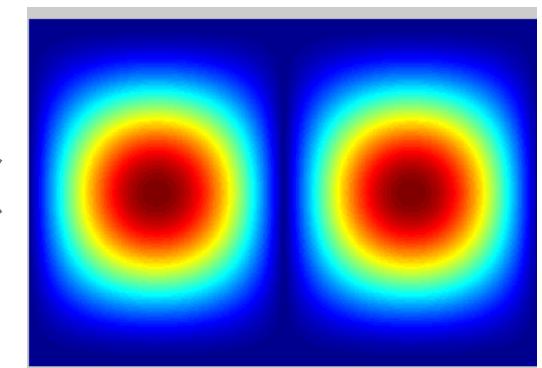
Filtering in frequency domain



FFT
→

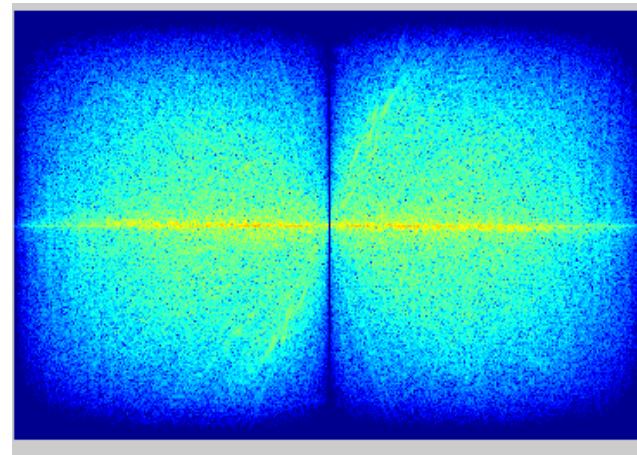
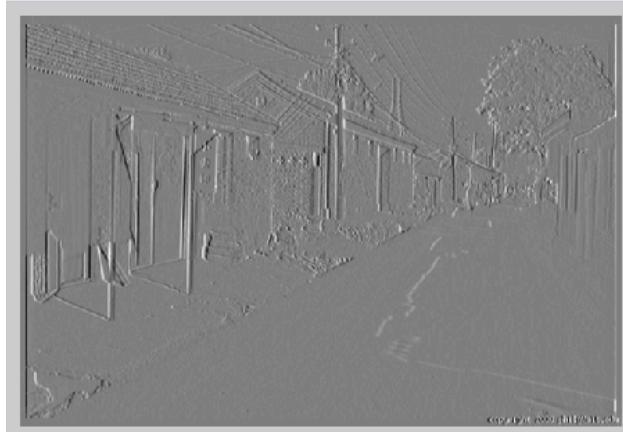


×



||

Inverse FFT
←



Fast Fourier Transform in Matlab

- Filtering with fft (fft2 -> 2D)

```
im = double(imread('...'))/255;
im = rgb2gray(im); % "im" should be a gray-scale floating point image
[imh, imw] = size(im);

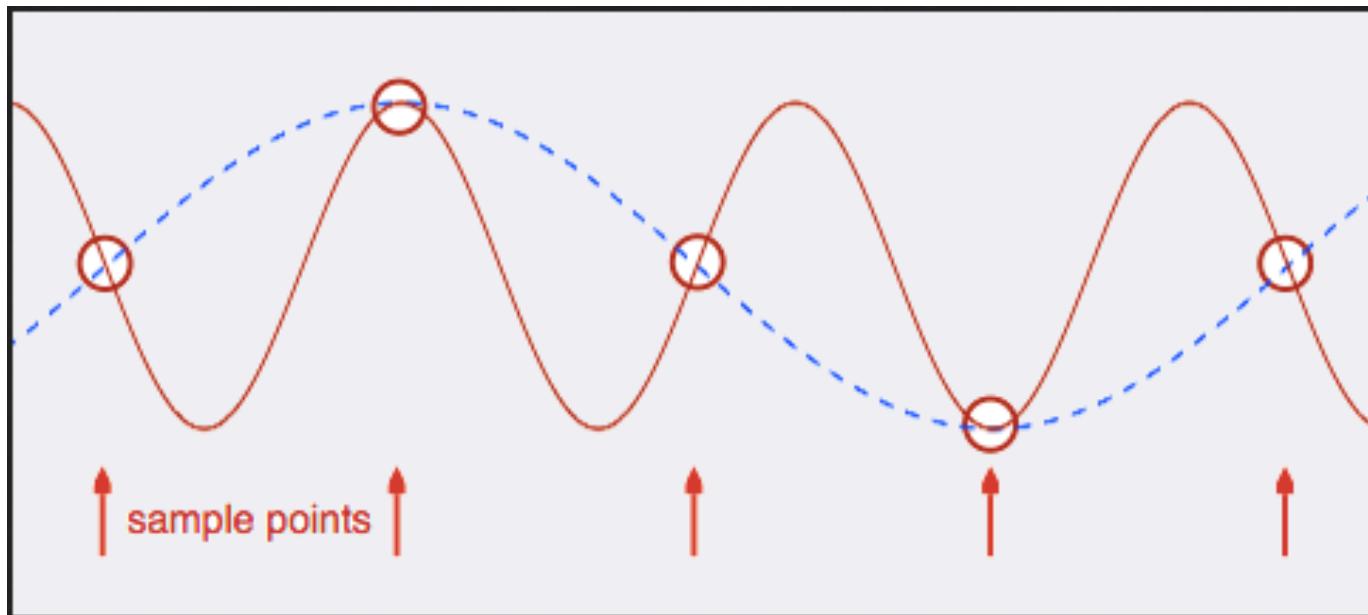
hs = 50; % filter half-size
fil = fspecial('gaussian', hs*2+1, 10);

fftsize = 1024; % should be order of 2 (for speed) and include padding
im_fft = fft2(im, fftsize, fftsize); % 1) fft im with padding
fil_fft = fft2(fil, fftsize, fftsize); % 2) fft fil, pad to same size as
image
im_fil_fft = im_fft .* fil_fft; % 3) multiply fft images
im_fil = ifft2(im_fil_fft); % 4) inverse fft2
im_fil = im_fil(1+hs:size(im,1)+hs, 1+hs:size(im, 2)+hs); % 5) remove padding
```

- Displaying with fft

```
figure(1), imagesc(log(abs(fftshift(im_fft))), axis image, colormap jet
```

Sampling Theorem

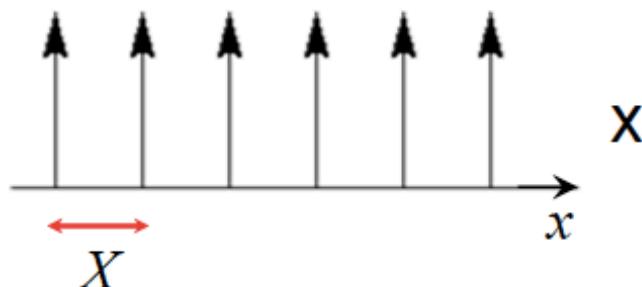


1D Sampling

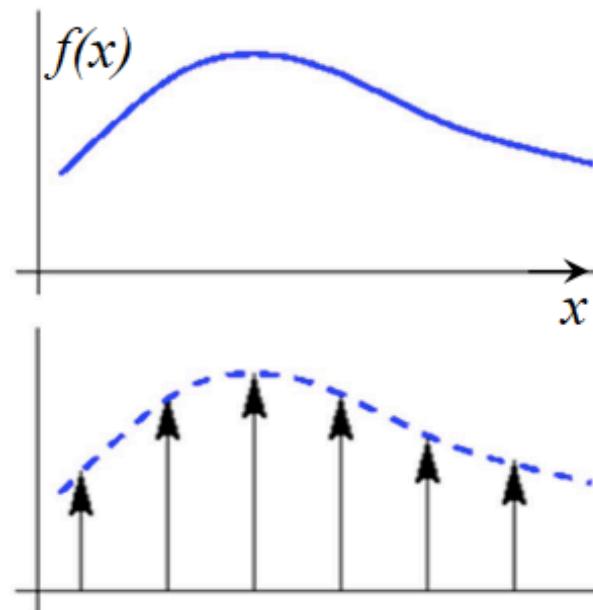
In 1D model the image as a set of point samples obtained by multiplying $f(x)$ by the **comb** function

$$\text{comb}(x) = \sum_{n=-\infty}^{\infty} \delta(x - nX)$$

an infinite set of delta functions spaced by X .

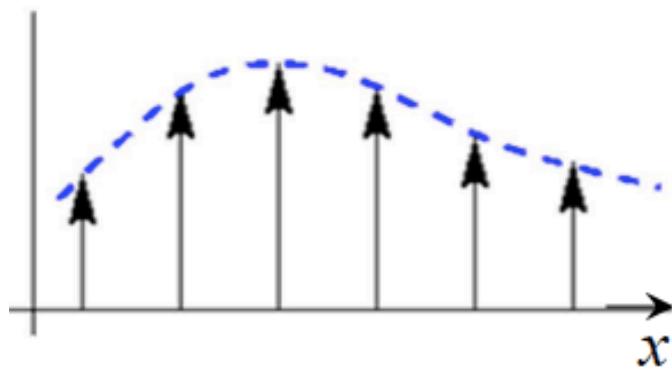


$$f_s(x) = \sum_{n=-\infty}^{\infty} \delta(x - nX)f(x)$$

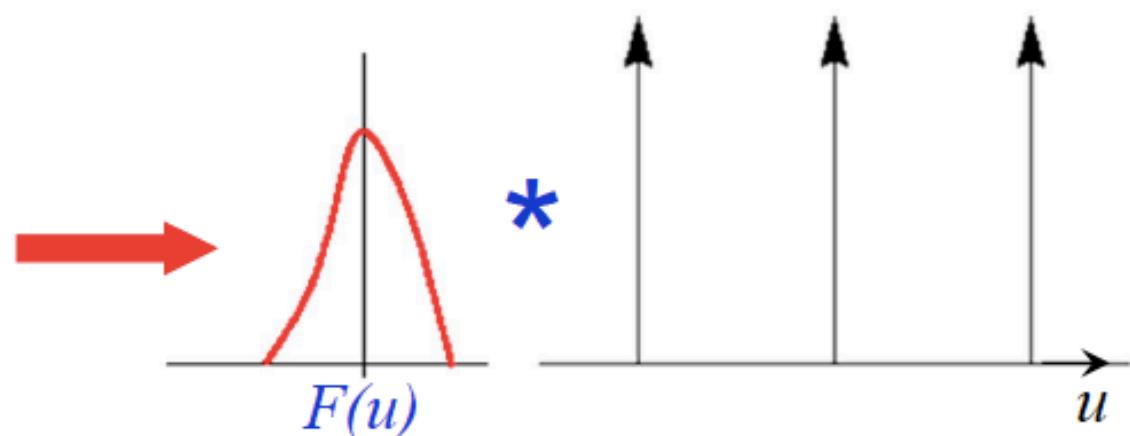


1D Sampling

spatial domain

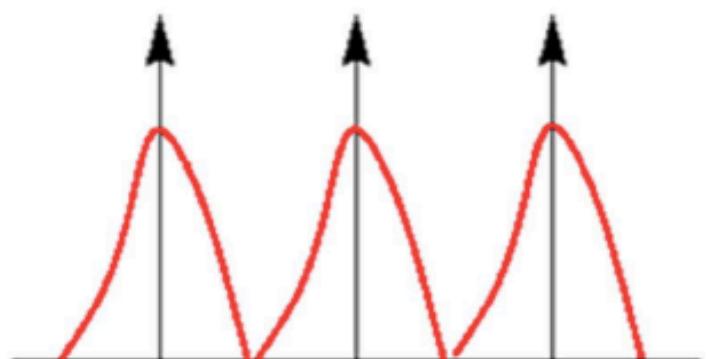


frequency domain



$$\begin{aligned}f_s(x) &= \sum_{n=-\infty}^{\infty} \delta(x - nX) f(x) \\&= \sum_{n=-\infty}^{\infty} f(nX) \delta(x - nX)\end{aligned}$$

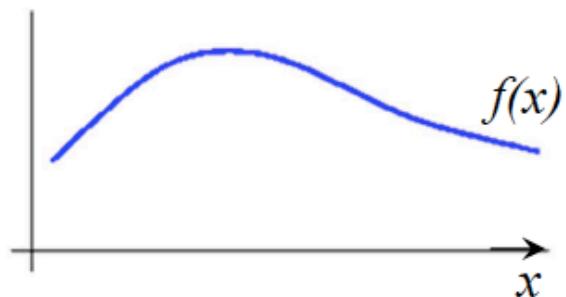
$$F_s(u) = \frac{1}{X} \sum_{n=-\infty}^{\infty} \delta(u - n/X) * F(u) = \frac{1}{X} \sum_{n=-\infty}^{\infty} F(u - n/X)$$



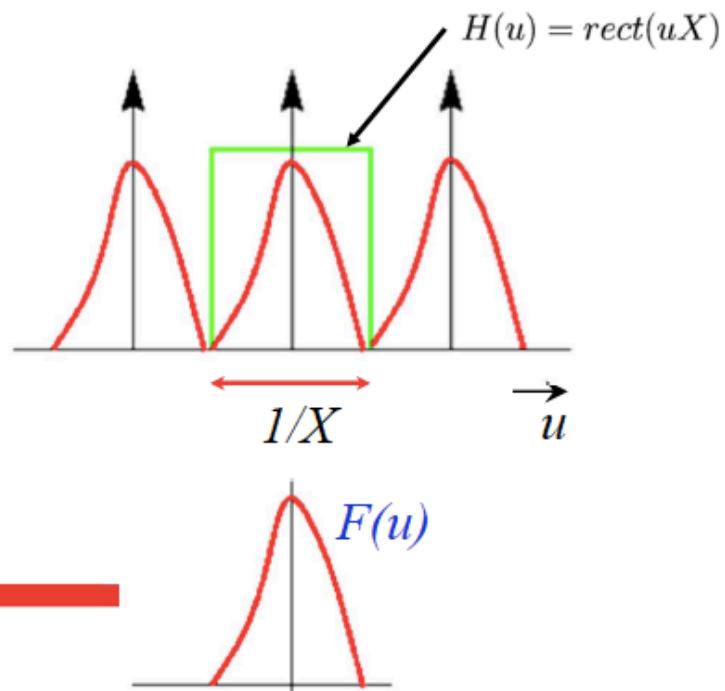
replicated copies of $F(u)$

1D Sampling

Apply a box filter



$$\begin{aligned}f(x) &= \sum_{n=-\infty}^{\infty} f(nX) \delta(x - nX) * \text{sinc} \frac{\pi x}{X} \\&= \sum_{n=-\infty}^{\infty} f(nX) \text{sinc} \frac{\pi}{X} (x - nX)\end{aligned}$$

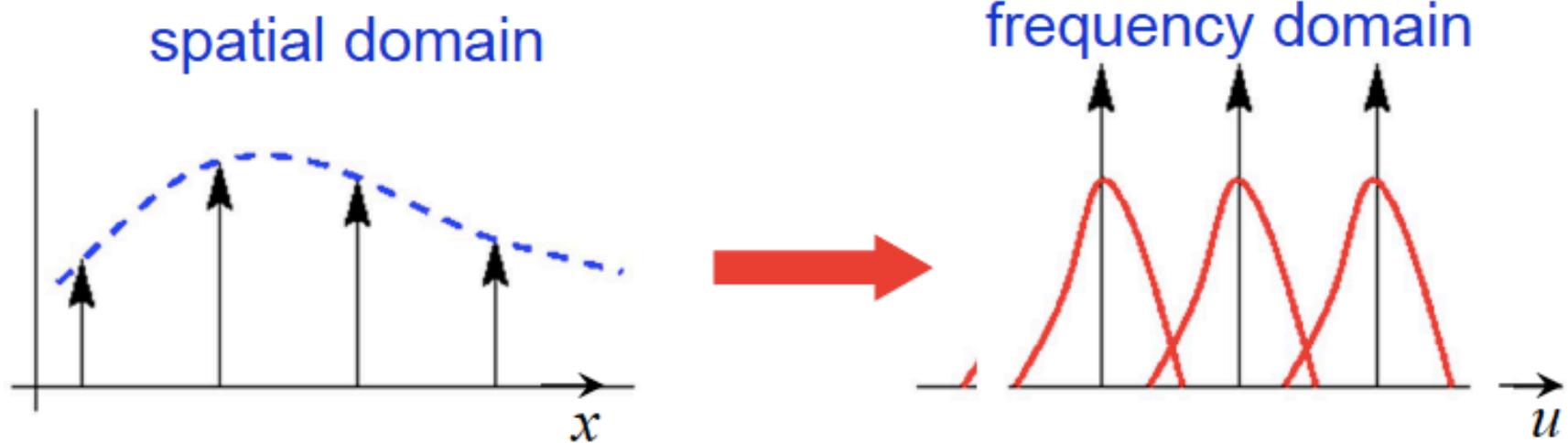


$$F(u) = F_s(u)H(u)$$

The original continuous function $f(x)$ is completely recovered from the samples provided the sampling frequency ($1/X$) exceeds twice the greatest frequency of the band-limited signal. (Nyquist sampling limit)

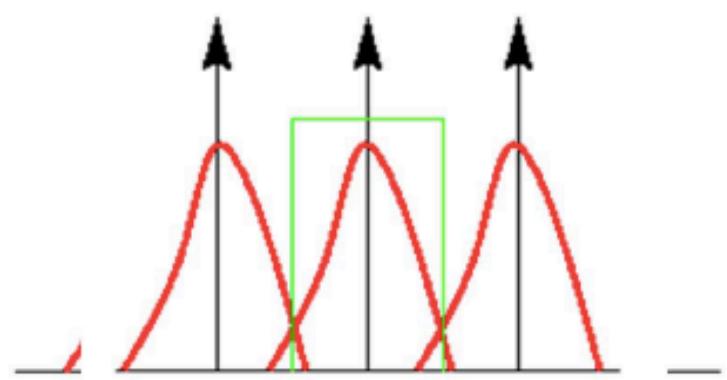
The Sampling Theorem and Aliasing

if sampling frequency is reduced ...



Frequencies above the Nyquist limit are 'folded back' corrupting the signal in the acceptable range.

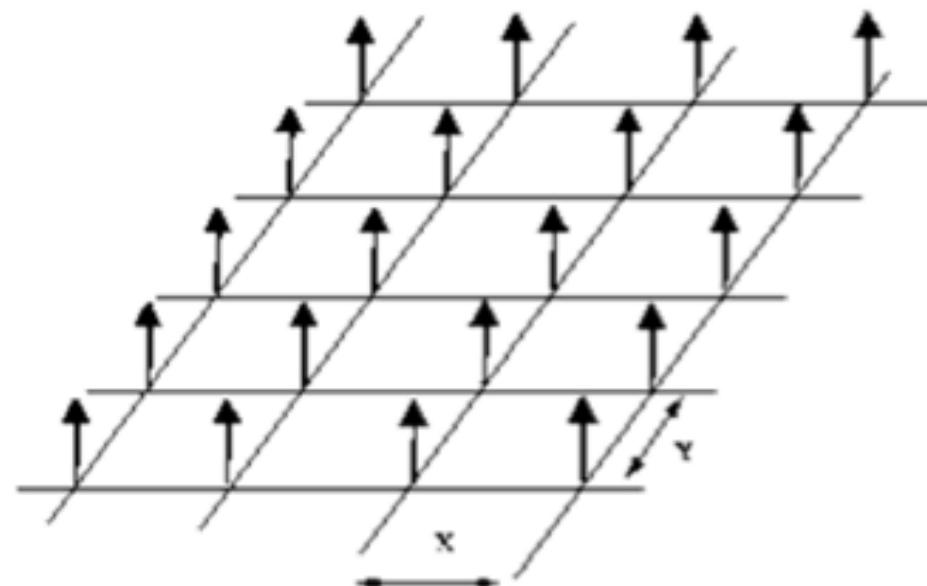
The information in these frequencies is not correctly reconstructed.



2D Sampling

In 2D the equivalent of a comb is a **bed-of-nails** function

$$\sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \delta(x - nX)\delta(y - mY)$$



2D Sampling

In 2D the equivalent of a comb is a **bed-of-nails** function

$$\sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \delta(x - nX) \delta(y - mY)$$

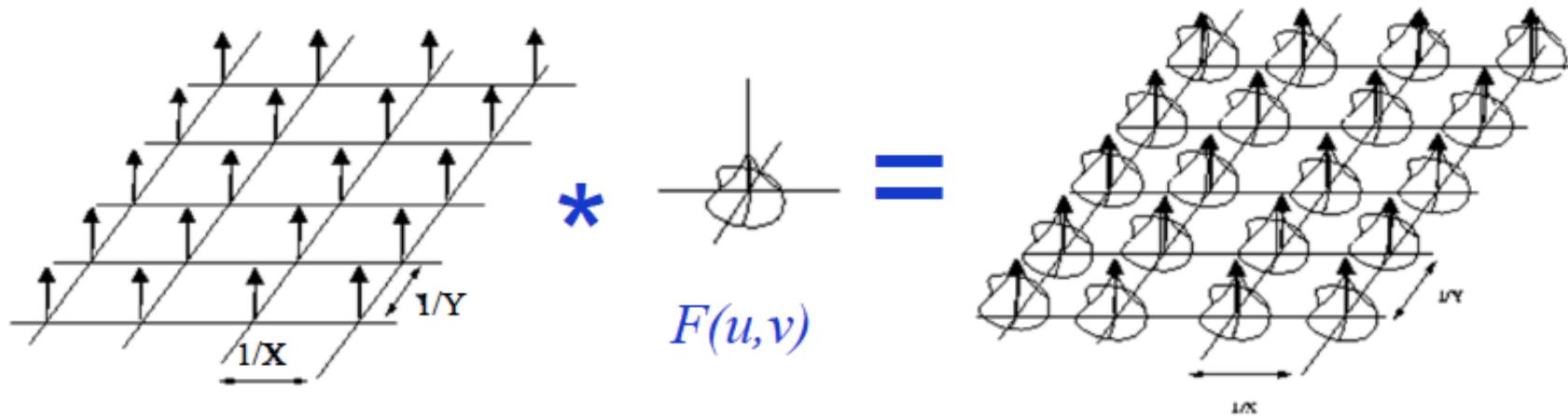
Fourier transform pairs

$$\sum_{n=-\infty}^{\infty} \delta(x - nX) \leftrightarrow \frac{1}{X} \sum_{n=-\infty}^{\infty} \delta(u - n/X)$$

$$\sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} \delta(x - nX) \delta(y - mY) \leftrightarrow \frac{1}{XY} \sum_{n=-\infty}^{\infty} \delta(u - n/X) \sum_{m=-\infty}^{\infty} \delta(v - m/Y)$$

Sampling Theorem in 2D

frequency domain



$$H(u, v) = \text{rect}(uX)\text{rect}(vY)$$

$$f(x, y) = \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} f(nX, mY) \text{sinc}\frac{\pi}{X}(x - nX) \text{sinc}\frac{\pi}{Y}(y - nY)$$

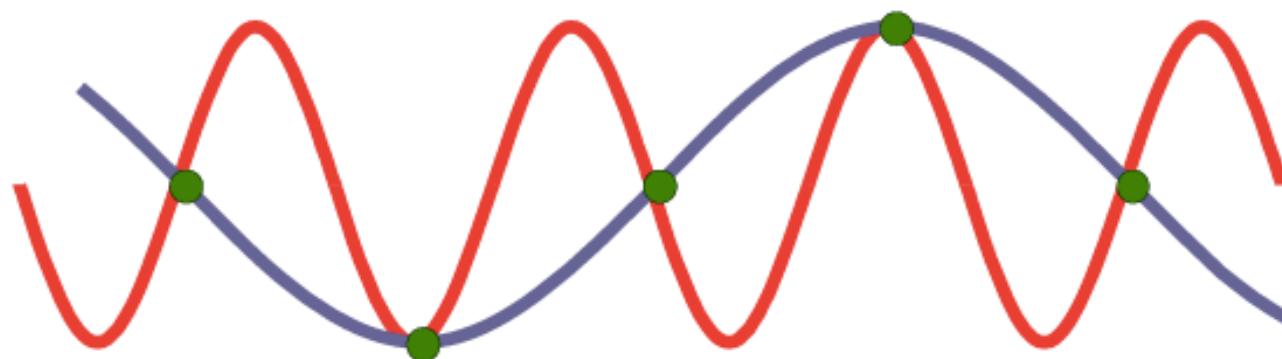
Sampling Theorem in 2D

If the Fourier transform of a function $f(x,y)$ is zero for all frequencies beyond u_b and v_b , i.e. if the Fourier transform is *band-limited*, then the continuous function $f(x,y)$ can be completely reconstructed from its samples as long as the sampling distances w and h along the x and y directions

are such that $w \leq \frac{1}{2u_b}$ and $h \leq \frac{1}{2v_b}$

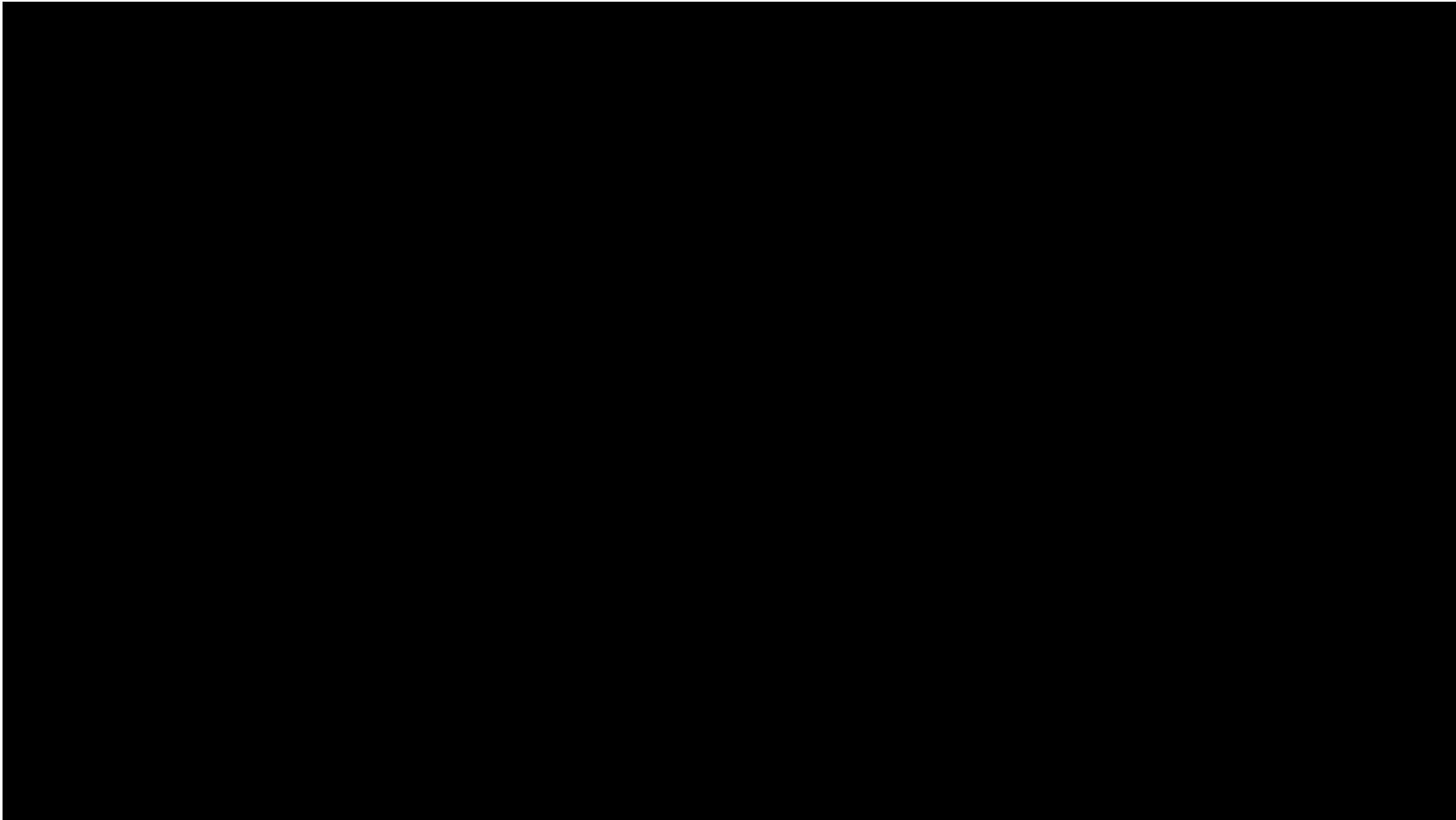
Aliasing: 1D Example

If the signal has frequencies above the Nyquist limit ...



Insufficient samples to distinguish the high and low frequency
aliasing: signals “travelling in disguise” as other frequencies

Aliasing in video

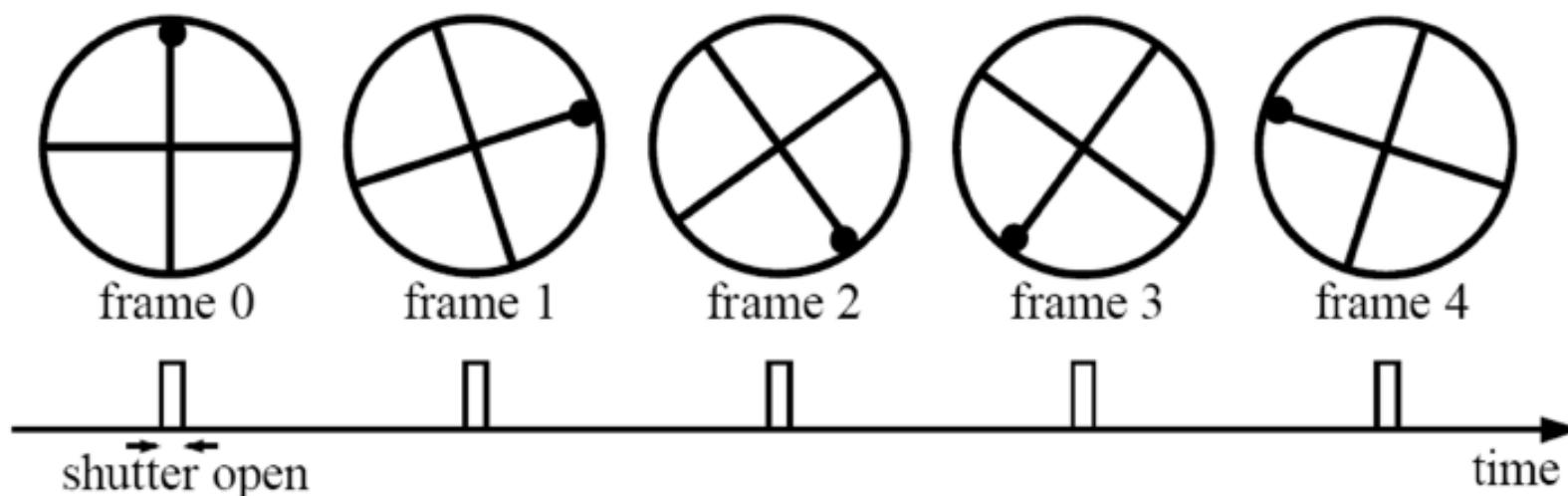


Aliasing in video

Imagine a spoked wheel moving to the right (rotating clockwise).

Mark wheel with dot so we can see what's happening.

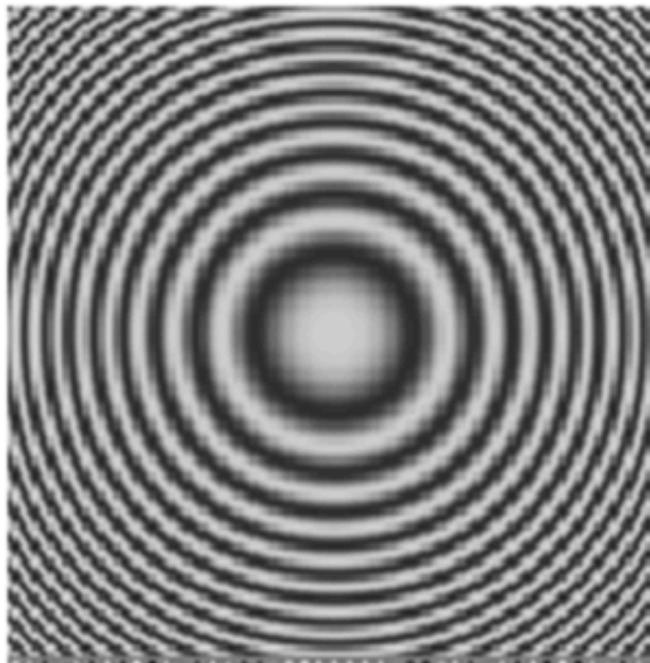
If camera shutter is only open for a fraction of a frame time (frame time = 1/30 sec. for video, 1/24 sec. for film):



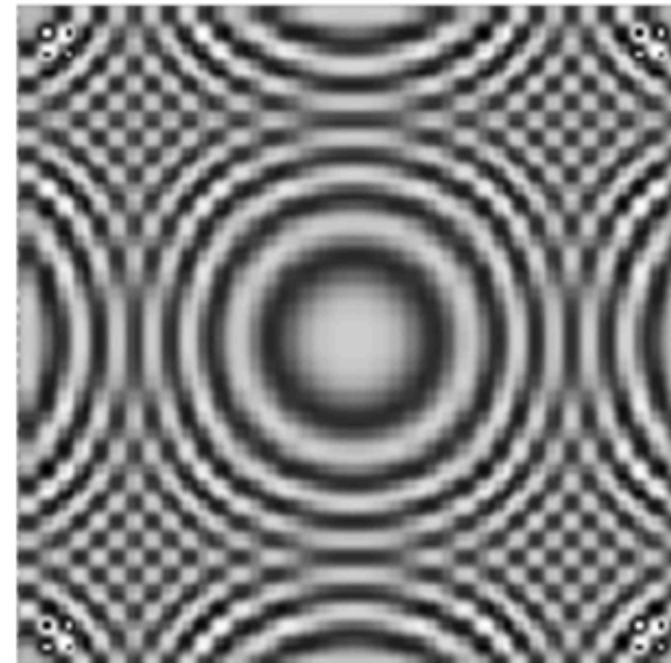
Without dot, wheel appears to be rotating slowly backwards!
(counterclockwise)

Aliasing in 2D: under-sampling reconstruction

original

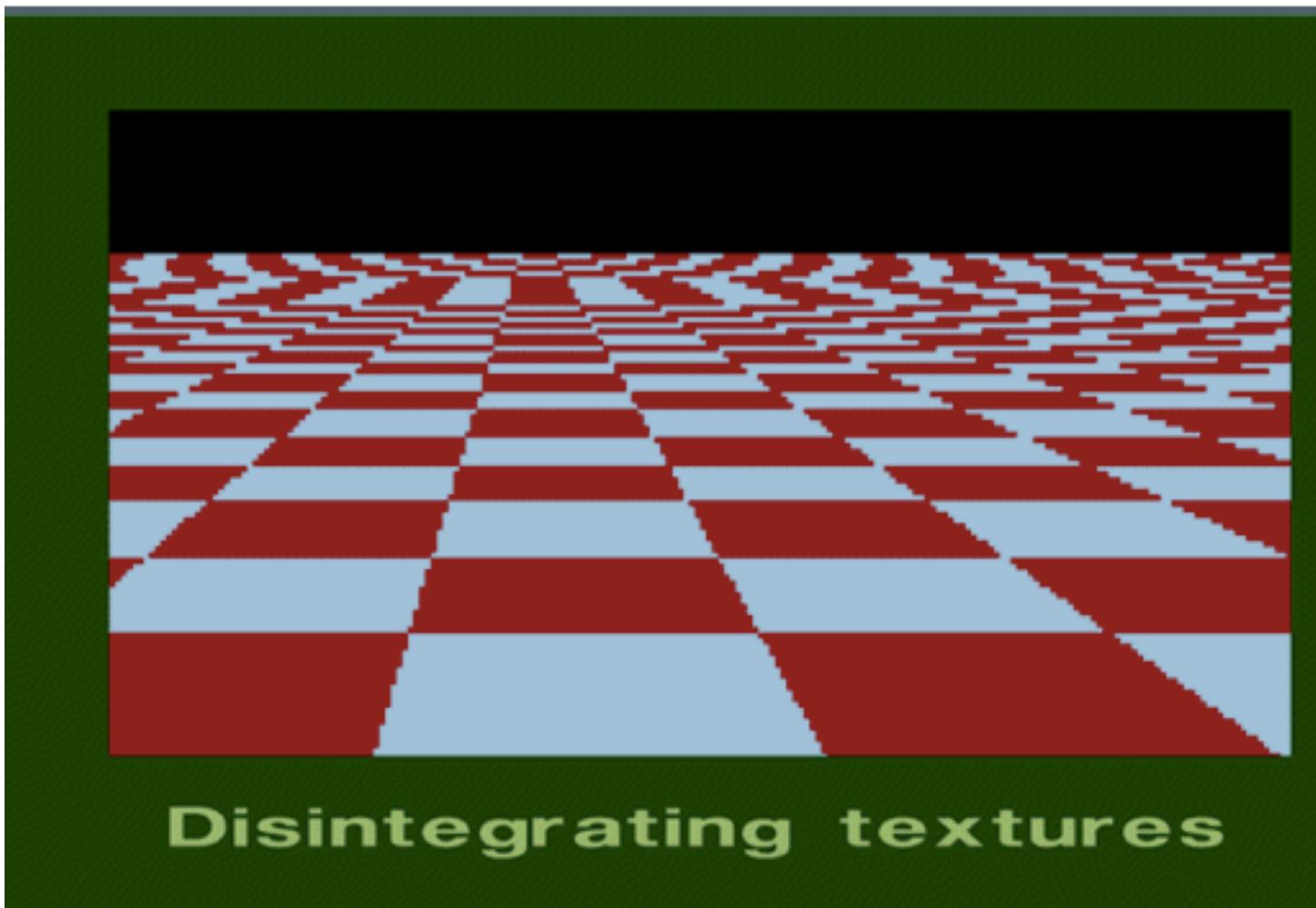


reconstruction



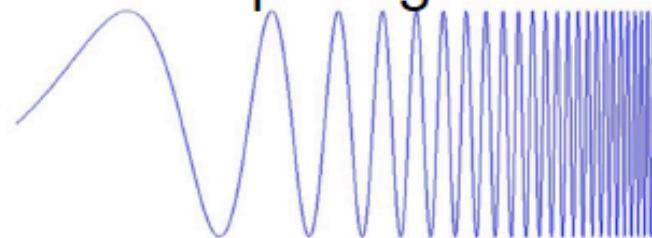
signal has frequencies
above Nyquist limit

Aliasing in Images

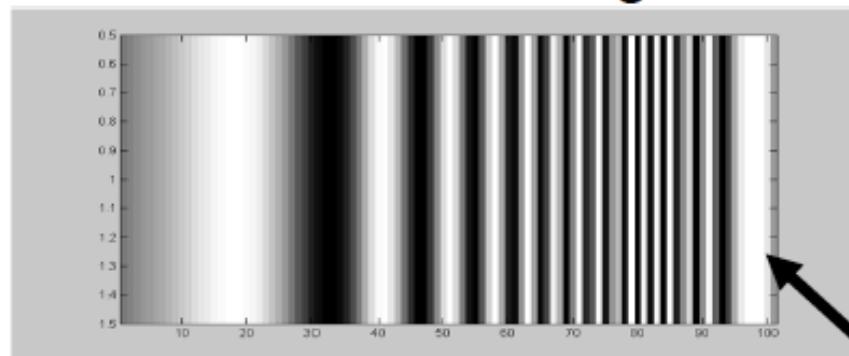


What's happening

Input signal:

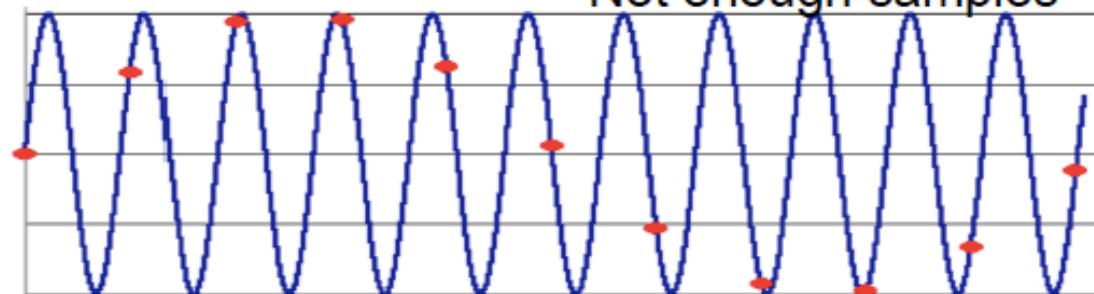


Plot as image:



```
x = 0:.05:5; imagesc(sin((2.^x).*x))
```

Aliasing
Not enough samples



Anti-aliasing

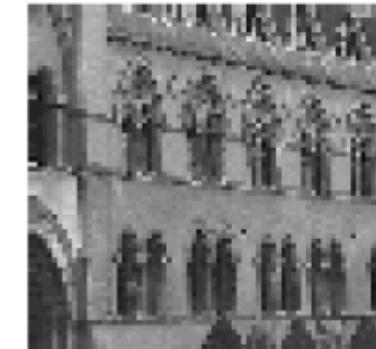
- Increase sampling frequency
 - e.g. in graphics rendering cast 4 rays per pixel
- Reduce maximum frequency to below Nyquist limit
 - e.g. low pass filter before sampling

Anti-aliasing

Example



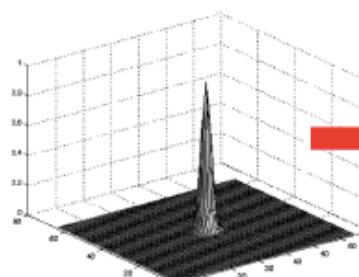
down sample by
factor of 4



4 x zoom



convolve with
Gaussian



down sample by
factor of 4

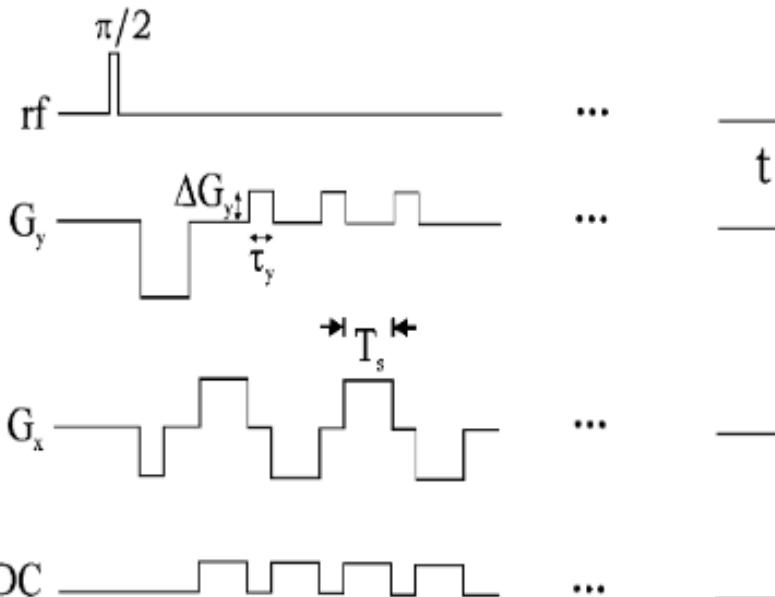


Aliasing in MRI

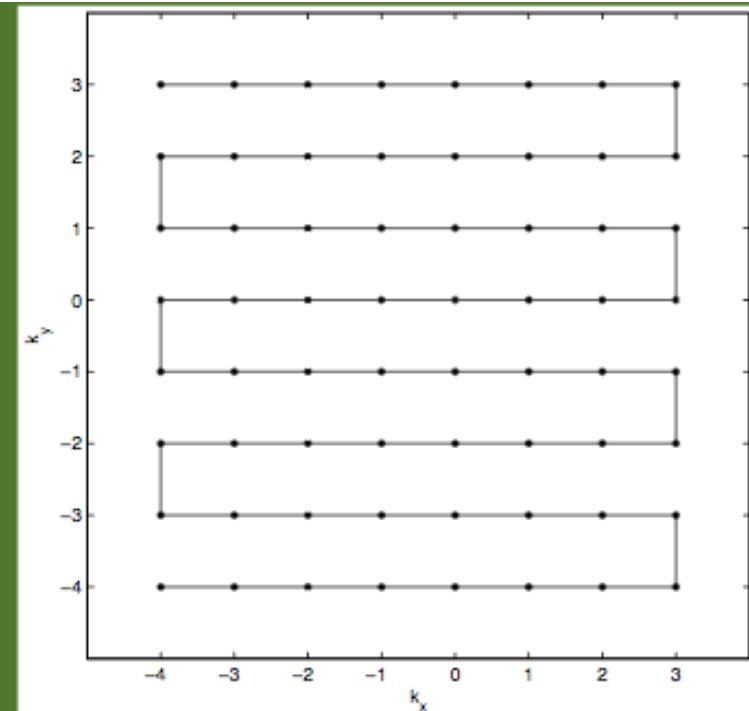


MRI

Aliasing in MRI

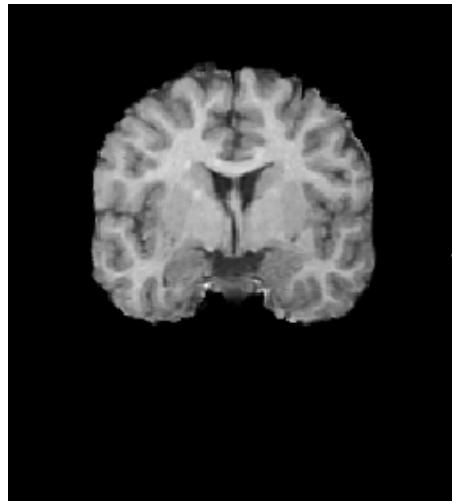


(a) Gradient Echo-EPI Pulse Sequence

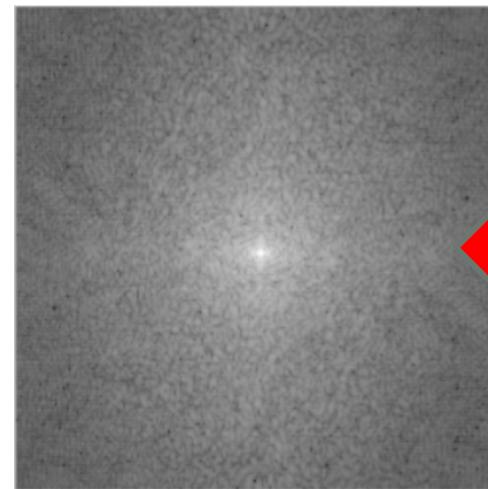


(b) k -Space Trajectory

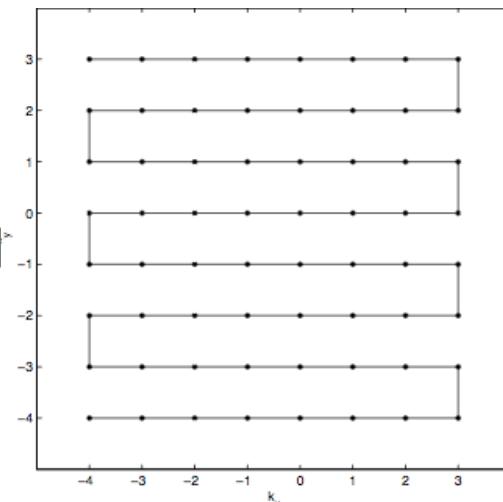
Aliasing in MRI



MRI



K-Space

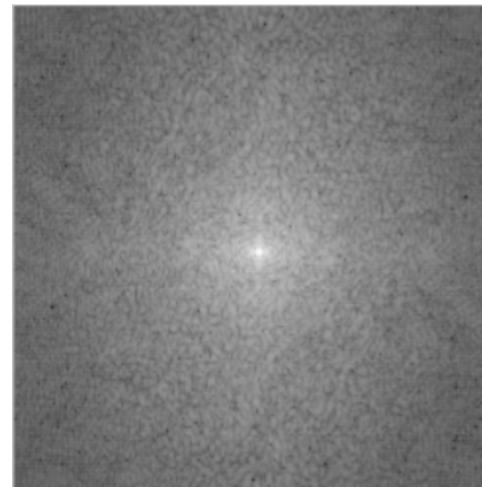


K-Space

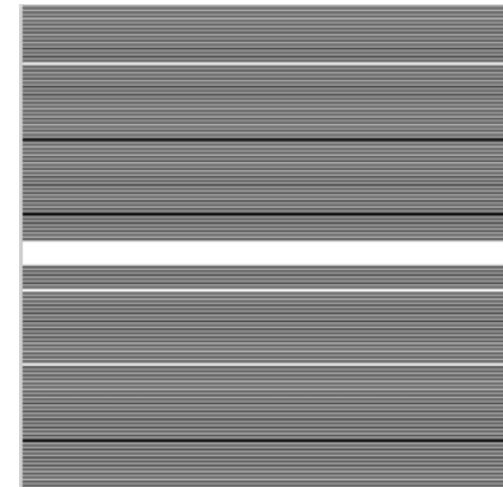
Aliasing in MRI



MRI

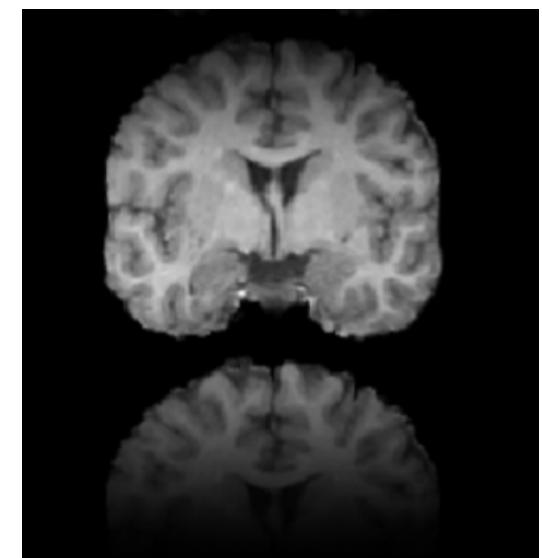


K-Space



Mask

Reconstructed
MRI



Ideas for final Projects

Solitaire Recognition

Chess recognition

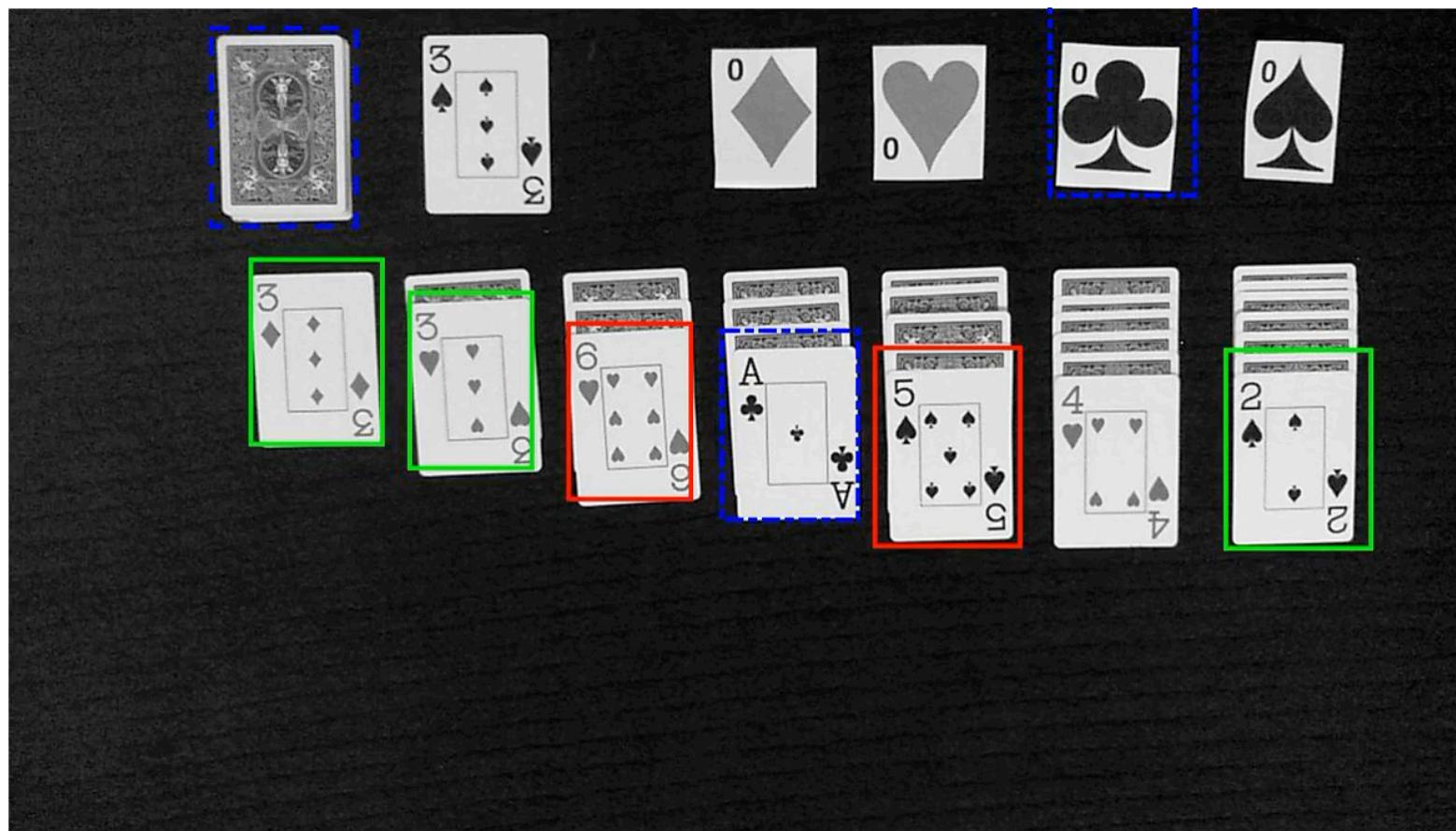
Real time Human Activity recognition

Face Recognition

Flying Object Detection

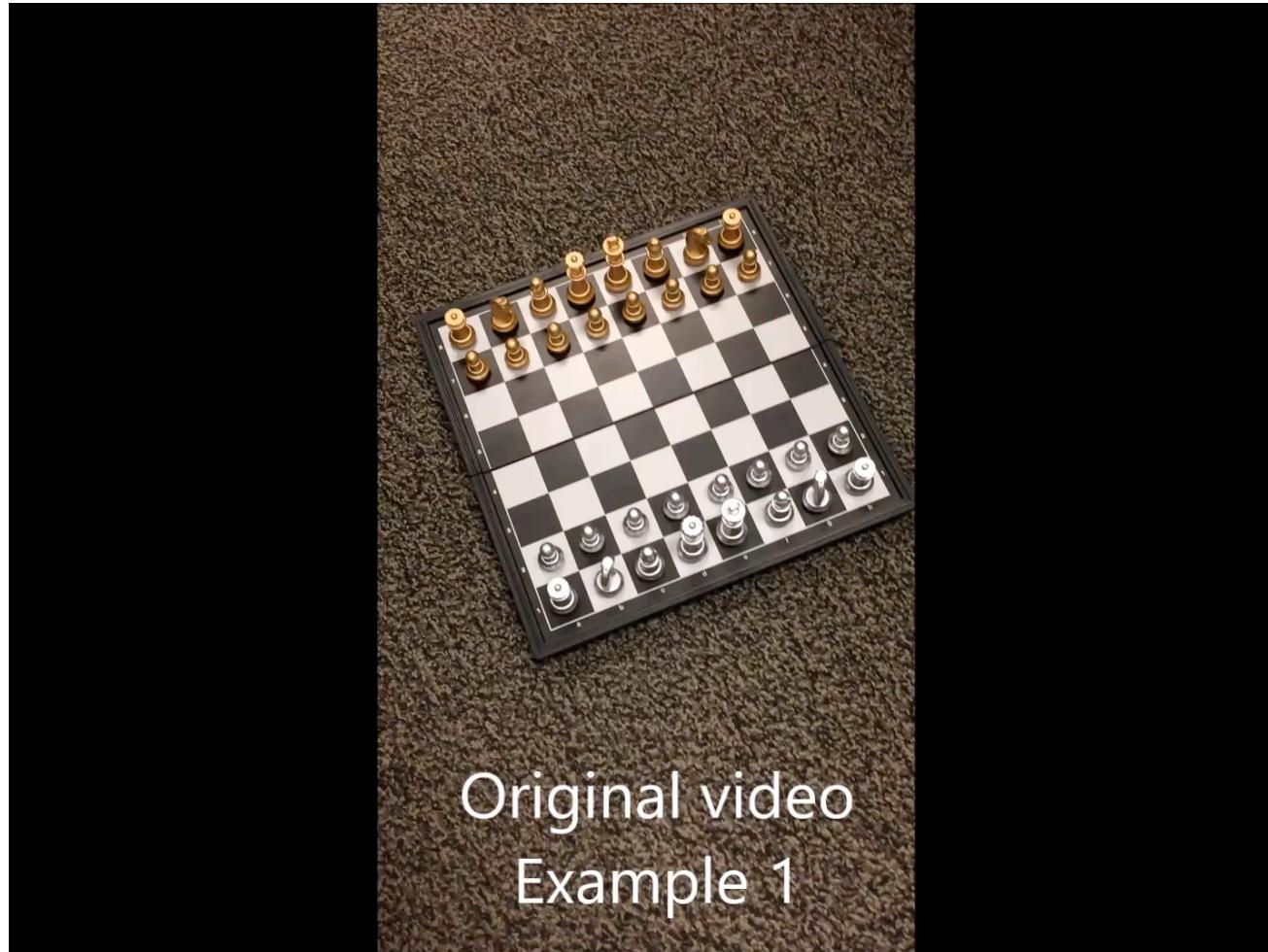
Ball detection (in soccer game)

Ideas for final Projects



EENG 512/CSCI 512 - Final Projects
Hoch, Garrett, Solitaire Recognition

Ideas for Final Projects



Original video
Example 1

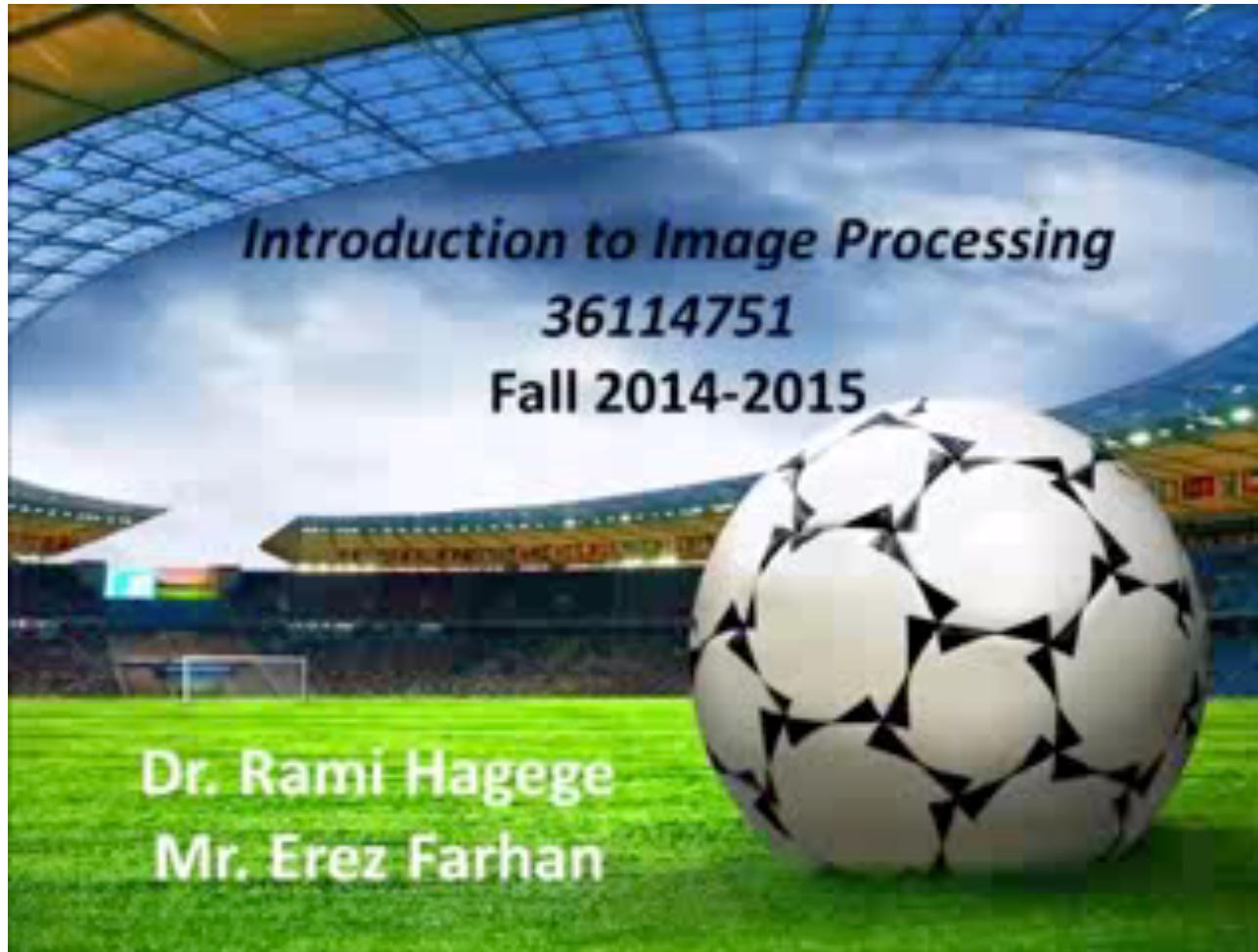
EENG 512/CSCI 512 - Final Projects
Xiao, Ke, *Chess Recognition*

Ideas for Final Projects



BGU – ECE 2013: Topaz, Ohad, Tsachi, Nadav

Ideas for Final Projects



BGU – ECE 2015:Doron, Boris, Alex

Ideas for Final Projects



BGU – ECE 2015:Nir, Tal & Shay – Interactive temple Run

Ideas for Final Projects



BGU – ECE 2013:Ariel, Tomer, Oren – Virtual Keyboard