

Lecture -16

Tuesday, 22 February 2022 8:12 AM

* Laplace transform :

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$

* Inverse Laplace transform using a combination of :

- Standard examples
- LT properties
- partial fractions

* we are interested in signals with rational form of $X(s)$

$$\text{i.e. } X(s) = \frac{A(s)}{B(s)} \quad \text{i.e. ratio of polynomials}$$

poles & zeros in the s -plane

poles decide the possible region of convergence (ROC)

ROC cannot contain any poles

* Properties of Laplace transform :

$$\text{given } x(t) \xleftrightarrow{LT} X(s) \quad \text{with ROC: } R_x$$

① Linearity

$$\text{i.e. } \alpha x_1(t) + \beta x_2(t)$$

② Time-shift

$$\text{i.e. } x(t-t_0)$$

③ Frequency-shift

$$\text{i.e. } X(s-s_0)$$

④ Time-reversal i.e. $x(-t)$

$$x(-t) \xleftrightarrow{LT} X(-s)$$

ROC: R_{-x}

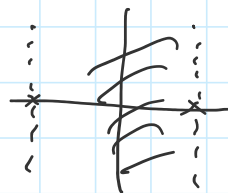
Ex. ROC: $\text{Re}(s) < \sigma$ i.e. R_x $s = \sigma + j\omega$

$$\sigma < \sigma$$

$$R_{-x} : s = -\sigma - j\omega$$

$$\Rightarrow -\sigma > -\sigma$$

$$R_{-x} : \text{Re}(s) > -\sigma$$



⑤ Time-scaling i.e. $x(at)$

$a \in \mathbb{R}$

$$a > 0 \Rightarrow x(at) \xleftrightarrow{LT} \frac{1}{a} X\left(\frac{s}{a}\right)$$

$$a < 0 \Rightarrow x(at) \xleftrightarrow{LT} -\frac{1}{a} X\left(\frac{s}{a}\right)$$

$$\Rightarrow x(at) \xleftrightarrow{LT} \frac{1}{|a|} X\left(\frac{s}{a}\right)$$

$a \in \mathbb{R}$

ROC: $a R_x$

⑥ Derivative in time i.e. $\frac{dx}{dt}$

$$\frac{dx}{dt} \xleftrightarrow{LT} s X(s)$$

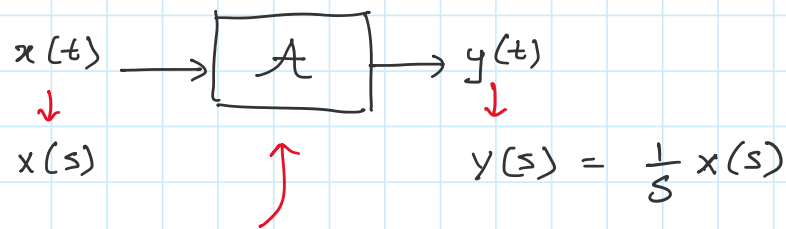
ROC: at least R_x

⑦ Derivative in frequency domain

$$-t x(t) \xleftrightarrow{LT} \frac{dX}{ds}$$

ROC: R_x

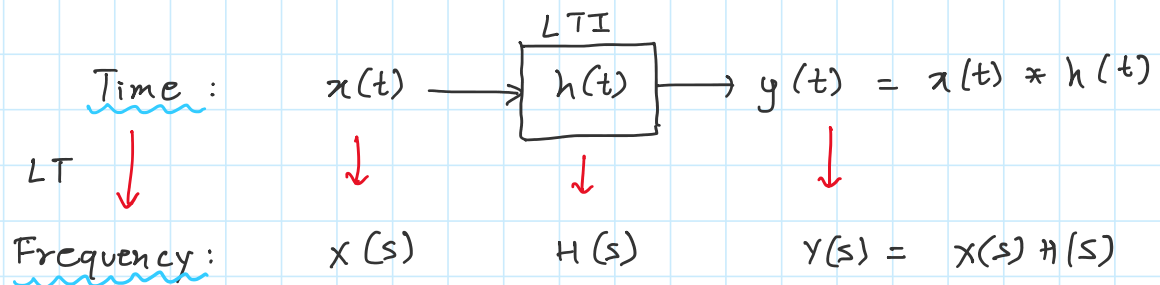
⑧ Integration i.e. $\int_{-\infty}^t x(\tau) d\tau = y(t)$



impulse response of Integrator system is unit step signal.
i.e. $h(t) = u(t)$

$$\text{ROC: } R_x \cap [R_c(s) > 0]$$

⑨ Convolution



$$Y(s) = \int_{-\infty}^{\infty} y(t) e^{-st} dt = \int_{-\infty}^{\infty} \left[\int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau \right] e^{-st} dt$$

$$= X(s) H(s)$$

ROC: at least $R_x \cap R_h$

* Differential equations:

linear constant coefficient differential equation * (LCCDE)

Ex.

$$\frac{dy}{dt} + ay(t) = x(t)$$

↓ LT

$$sY(s) + aY(s) = X(s)$$

... (system)



$$y(s)[a+s] = x(s)$$

$$y(s) = \frac{x(s)}{s+a}$$

$$\frac{y(s)}{x(s)} = \frac{1}{s+a}$$

convolution property

$= H(s)$ system function
transfer function

any LCCDE in Laplace domain can be expressed as ratio of polynomials.

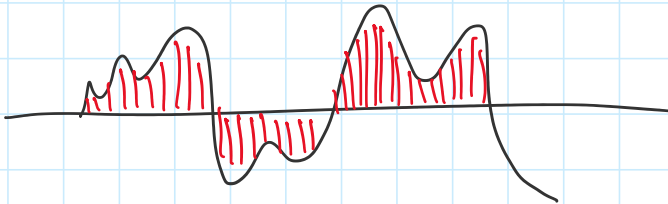
★ Discrete-time signals & systems (review for mini-project)

time is sampled

$$x(t) \longrightarrow \underline{x(nT_s)} = \underline{x[n]}$$

T_s - sampling interval

$$F_s = \frac{1}{T_s} \text{ - sampling frequency.}$$



$$x(t) = \sin(\omega_0 t)$$

$$x[n] = \sin(\omega_0 n T_s)$$

★ periodic DT signals.

N - period of DT signal
 $N T_s$ - period of CT

$$\hookrightarrow x[n] = x[n+N] \quad \forall n \quad \text{period} = N$$

Fourier series for DT periodic signals:

$$x[n] = \sum_{k=-\infty}^{\infty} c_k e^{j k \omega_0 n}$$

recall CT: $x(t) = \sum_{-\infty}^{\infty} d_k e^{jk\omega_0 t}$... synthesis

$$\omega_0 = \frac{2\pi}{T}$$

$$d_k = \frac{1}{T} \int_0^T x(t) e^{-jk\omega_0 t} dt \quad \dots \text{analysis}$$

For DT: only finite ^{basis} signals with period N

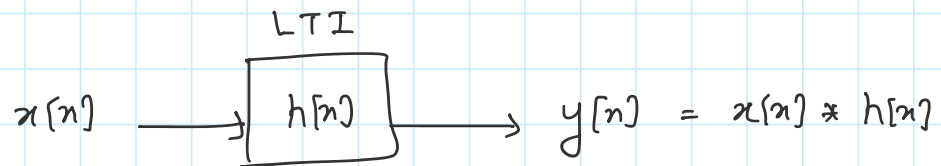
Basis: $\{e^{jk\omega_0 n}\}$, $\omega_0 = \frac{2\pi}{N}$ N is integer

$$k = 0, 1, 2, \dots, N-1$$

DT FS : synthesis $\rightarrow x[n] = \sum_{k=0}^{N-1} d_k e^{jk\omega_0 n}$ *

DT FS : analysis $\rightarrow d_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-jk\omega_0 n}$ *

discrete-time system (LTI)



Convolution sum: $y[n] = \sum_{m=-\infty}^{\infty} x[m] h[n-m]$