

TP : Introduction to design with ADS

Simulation of transmissions lines

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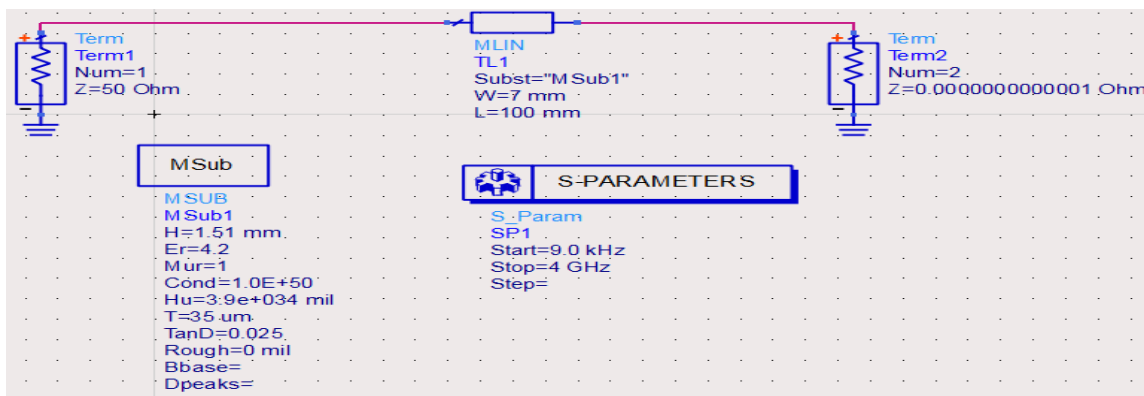
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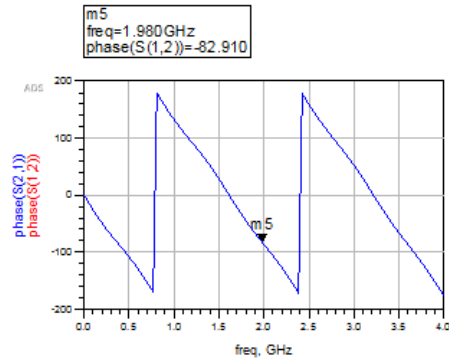
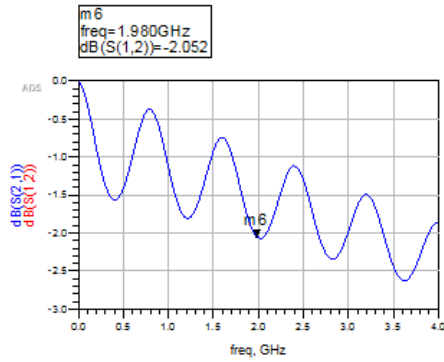
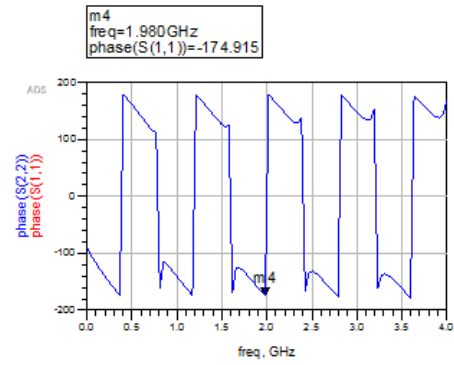
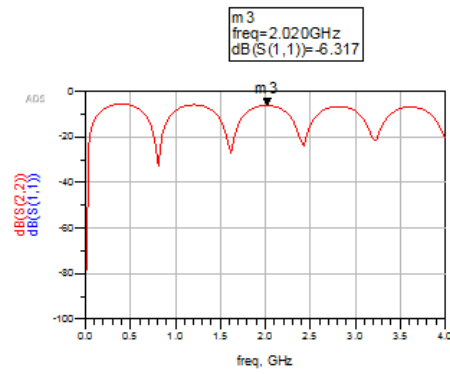
Introduction :

This laboratory session serves as an introductory lesson in electronic design, specifically regarding transmission lines. The aim is to comprehend microwave simulation techniques and interpret the outcomes. Our subsequent focus will be on analyzing frequency domain responses.

Part 1 Simulation of an RF circuit using ADS built-in models. for 7mm line (B) at 2GHz

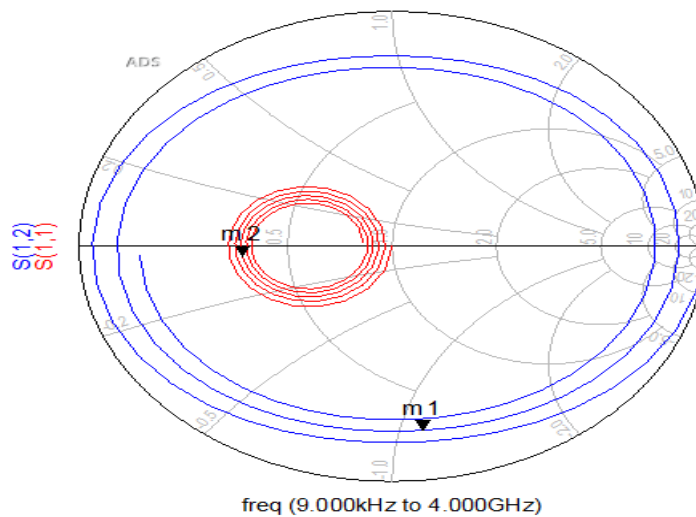
With $Z_c = 50 \text{ ohm}$





m1
freq=1.980GHz
S(1,2)=0.790 / -82.910
impedance = Z0 * (0.264 - j1.097)

m2
freq=1.980GHz
S(1,1)=0.481 / -174.915
impedance = Z0 * (0.351 - j0.039)



a- Comparaision between software simulation and real simulation values

	Real simulation in dB	ADS in dB
S11	-5.5	-6.3

S22	-5.5	-6.3
S21	-2.2	-2.052
S12	-2.2	-2.052

We notice that these results are quite close but just some difference due to real life phenomena (parasitic signal etc.)

$S_{11} = S_{22} = -5.5$ dB that means we reached the impedance matching almost no reflexion

$S_{12} = S_{21} = -2.2$ dB that means we have less losses in our transmission line

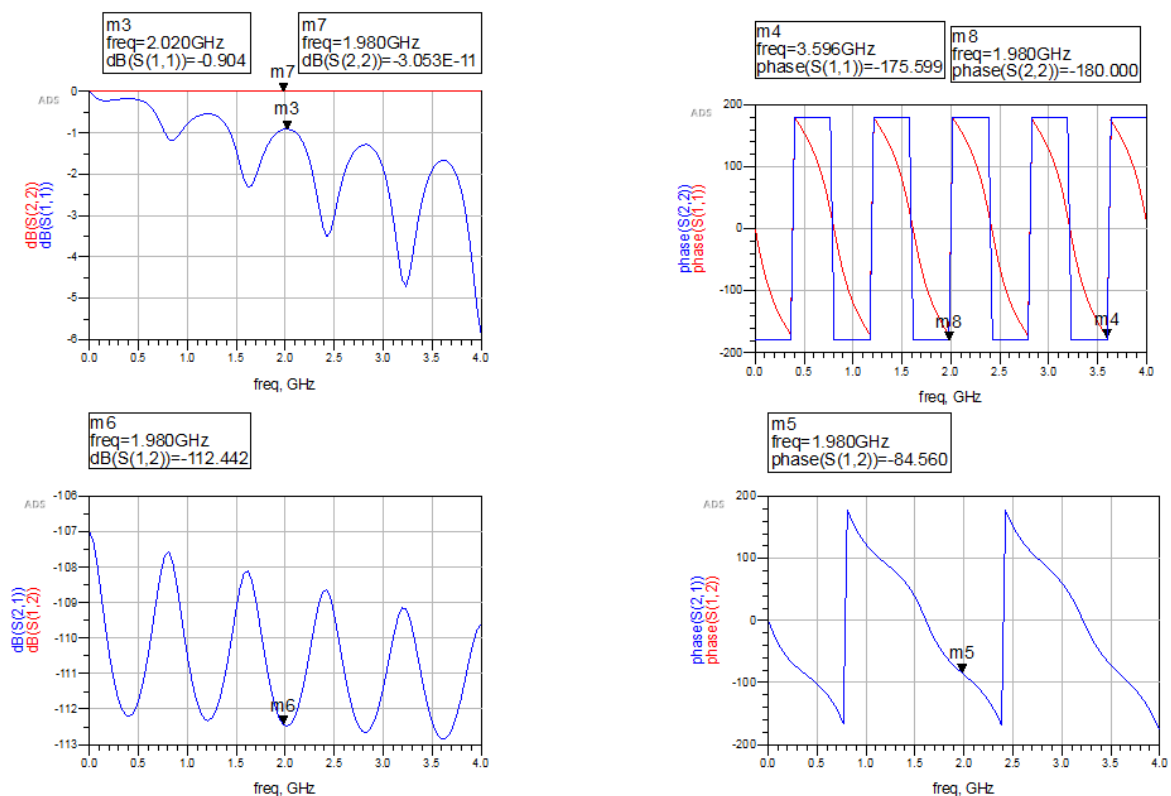
b- Paramaters evaluation in open cicuit case

With $Z_c = 1M$ ohm

For an open circuit case we observe that the output reflexion coefficient S22 is at 0dB and its different from S11

when there's an open circuit at the output port, the reflection coefficient S22 would indeed be 0dB (zero) as all the signal is reflected back due to the open termination.

Regarding the transmission coefficient we observe that $S_{12} = S_{21} = -114$ dB we observe a strong attenuation of the transmitted signal at 2 GHz

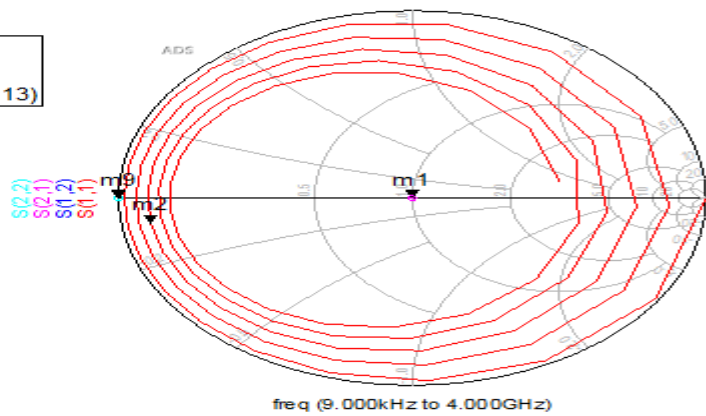


smith chart in open cicuit case

m9
freq=1.980GHz
S(2,2)=1.000 / -180.000
impedance = Z0 * (1.757E-12 - j1.947E-13)

m1
freq=1.980GHz
S(1,2)=2.387E-6 / -84.560
impedance = Z0 * (1.000 - j4.753E-6)

m2
freq=1.980GHz
S(1,1)=0.900 / -171.436
impedance = Z0 * (0.053 - j0.075)

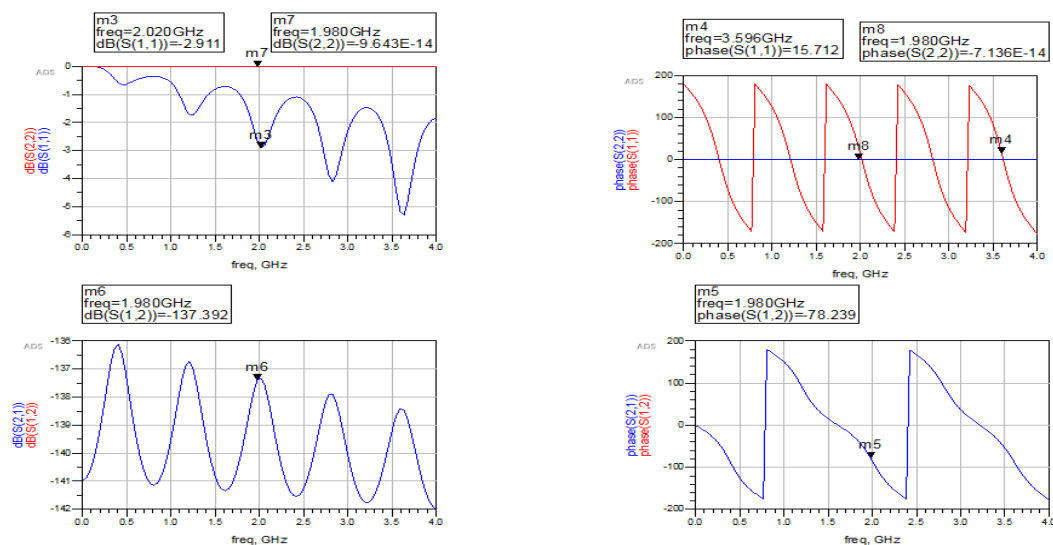


Regarding the smith chart as the m9 (S22) is at extrem right we can deduce that we are in total reflexion case on the output

b-2 Paramaters evaluation in short cicuit case ZL=0 ohm

For a short circuit case it's the same case than the previous one we have a total reflexion regarding the output S22 which is in fact correct by using theorical formula

The transmission coefficient has again a strong attenuation -137 dB

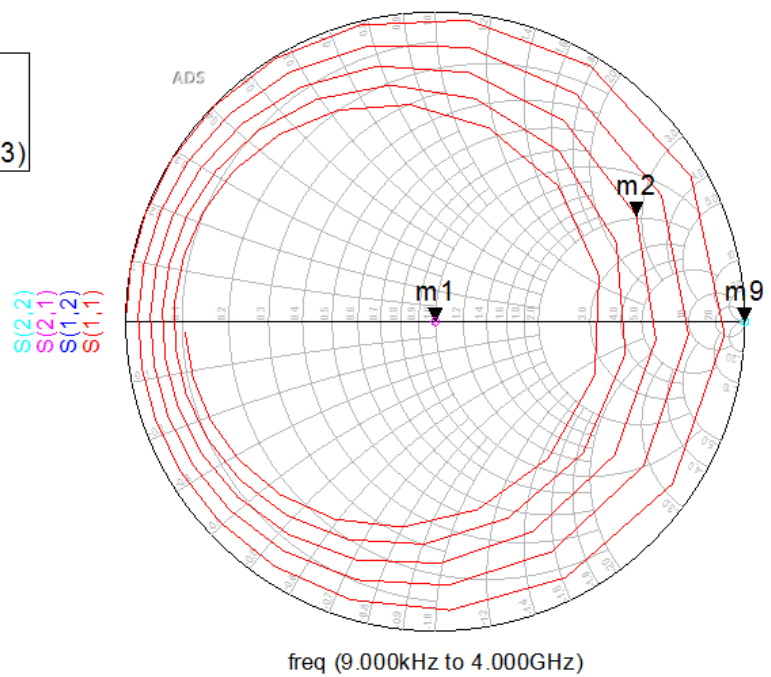


smith chart in short cicuit case

m9
 freq=1.980GHz
 $S(2,2)=1.000 / -7.136E-14$
 impedance = $Z_0 * (1.779E14 - j1.996E13)$

m1
 freq=1.980GHz
 $S(1,2)=1.350E-7 / -78.239$
 impedance = $Z_0 * (1.000 - j2.644E-7)$

m2
 freq=1.980GHz
 $S(1,1)=0.734 / 27.812$
 impedance = $Z_0 * (1.919 + j2.851)$



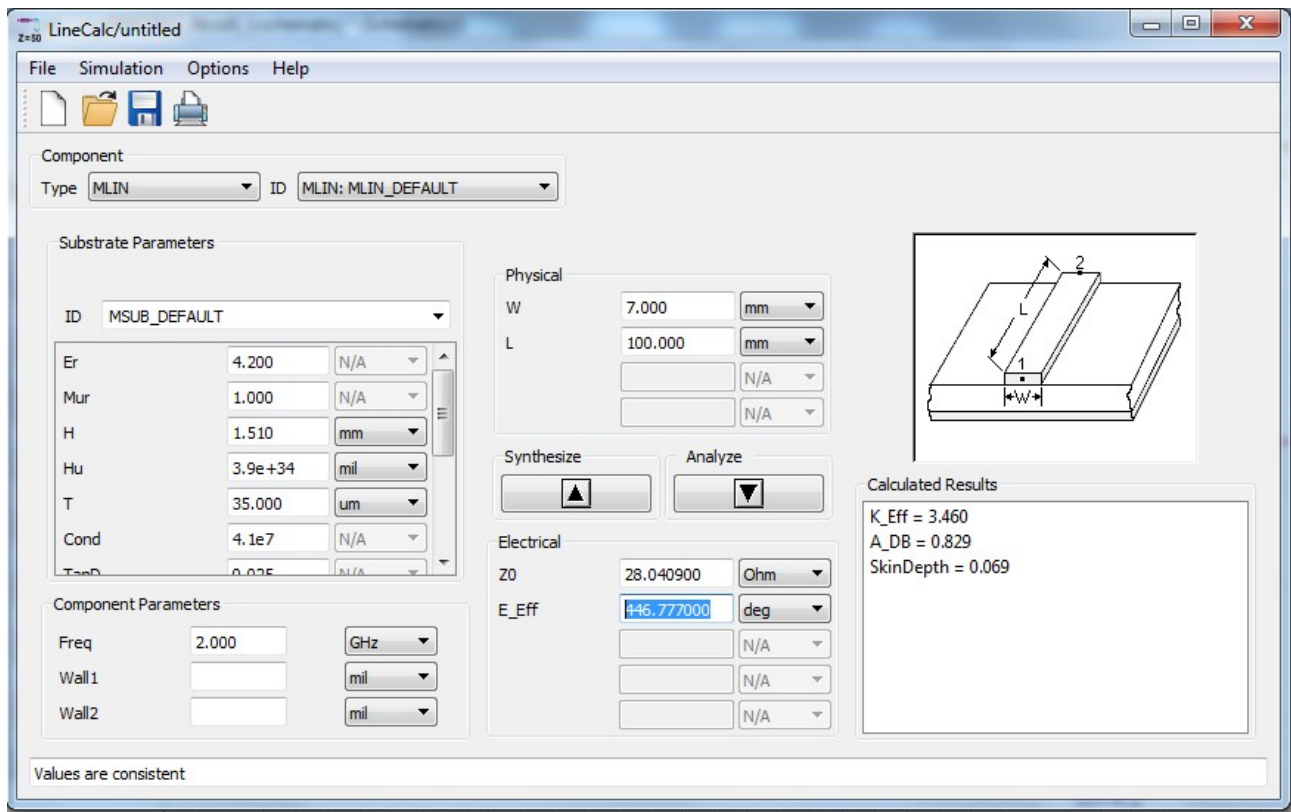
Regarding the smith is just confirmed the previous interpretation ; We observe that m1 (S12) is well centered on 1 that means the impedance there is 50 Ohm ; m9 (S22) is at 0 extrem left that means we have a total reflexion

c. Calculation of the impedance of the line

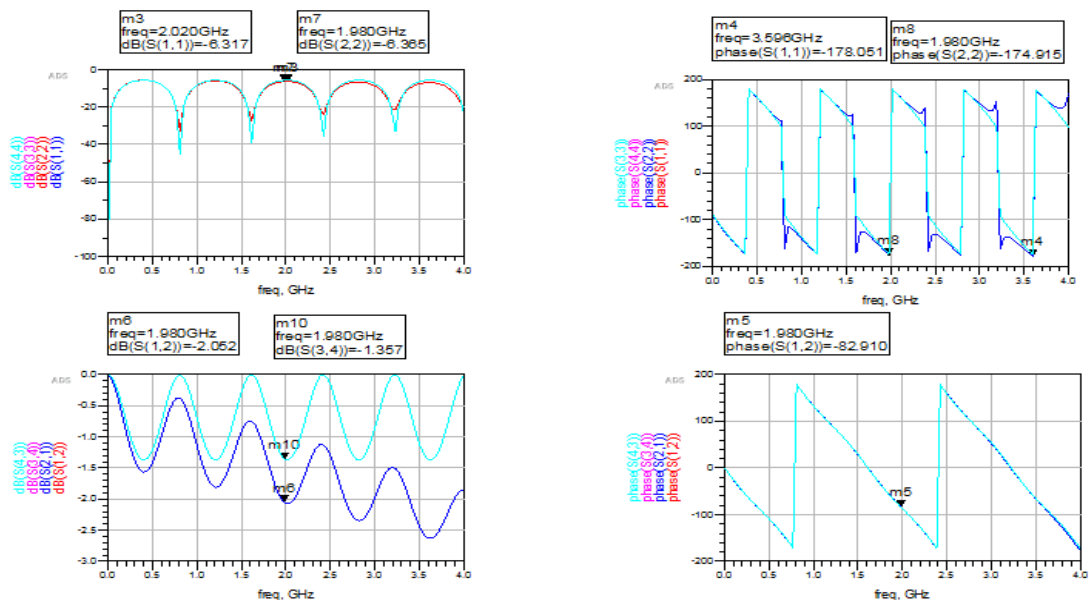
After well dedined the line properties we have calculated the Characteristic impedance of the line
 We find Z_c close to 50 ohm

(the picture below was before calculation)

The simulations result are right



- Result comparison between the reflexion by defining Zc (S11 and S22) and the reflexion by using Zc calculate by the software (S33 and S44)



smith chart

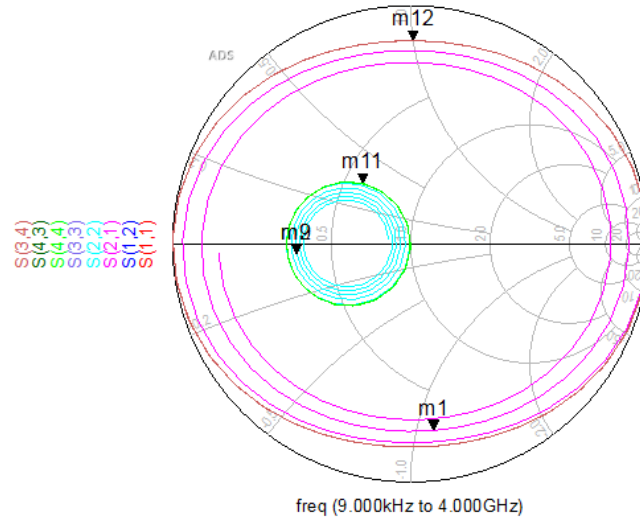
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m12
freq=1.212GHz
S(4,3)=0.853 / 89.338
impedance = $Z_0 * (0.159 + j0.999)$

m11
freq=3.071GHz
S(3,3)=0.324 / 128.367
impedance = $Z_0 * (0.594 + j0.337)$



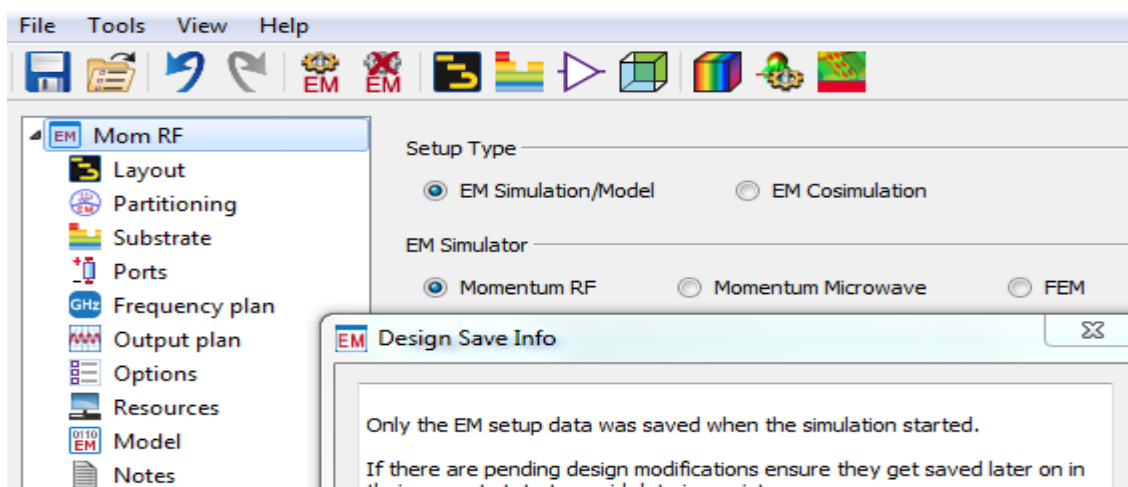
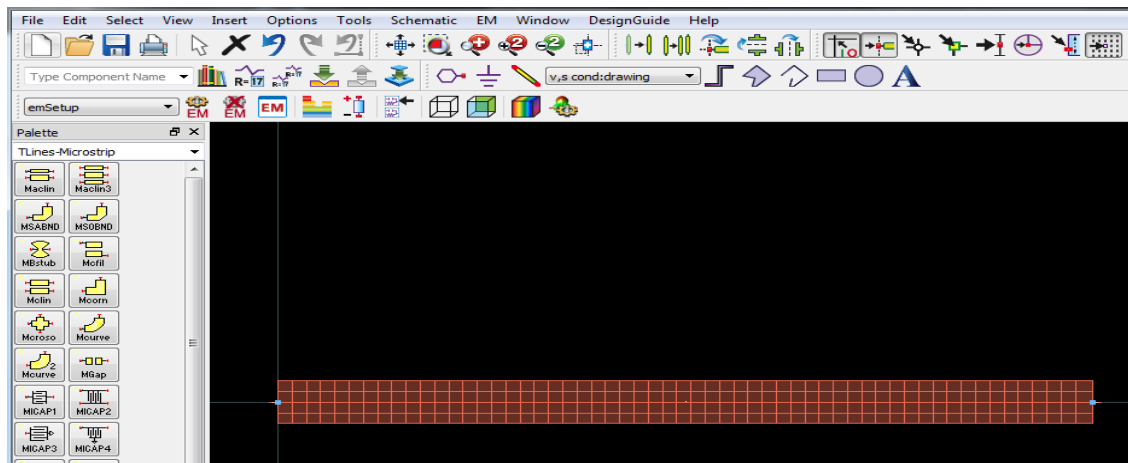
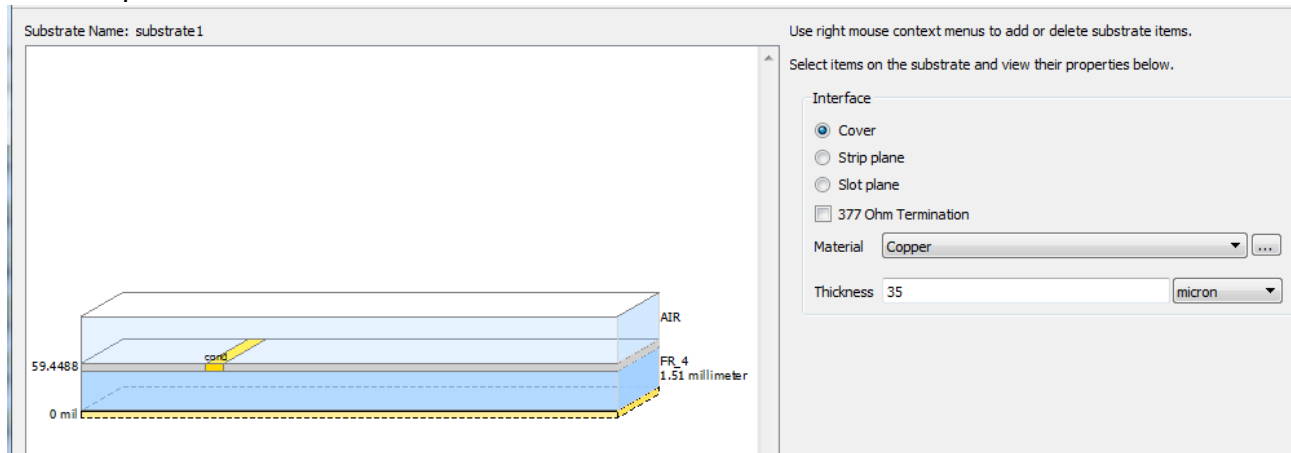
Regarding $S_{11} = S_{22}$ and $S_{33} = S_{44}$ we retrieve the same value for the both case with whatever the defined or the calculated Z_c that means the value is correct we could just define the properties of the line and the software would calculate the adapted Z_c by itself
Regarding the coefficients are equal in term of phases but magnitudes are different when we are getting close to higher frequencies.

The Smith Chart shows that $S_{43} = S_{21}$ has an inductive behavior while $S_{12} = S_{34}$ has a capacitor one And $S_{22} = S_{44}$ is real it is on the real axes

Part 2 – Momentum simulation

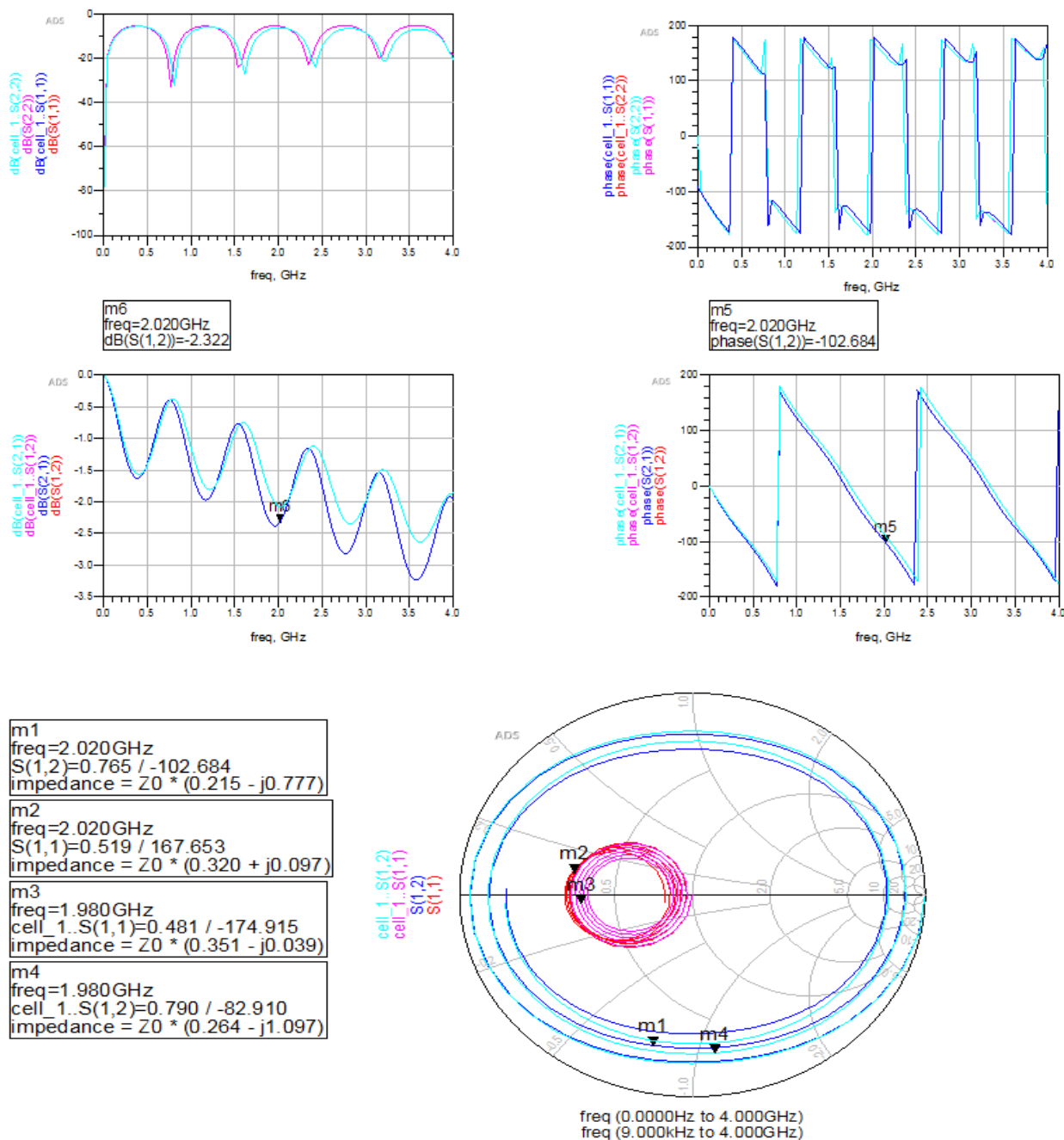
1 - Definition of the line properties on the Layout

Line microstrip from length equal à 100 mm Engraved on one substrate dielectric FR4 ($\epsilon_r=4.2$; $\text{tg}\delta=0.025$) 1.51mm thick. Metallization thickness (copper) east equal to 35 μm . The line has a width equal to 7 mm.



2-Simulation results

Comparison between the Momentum S-parameters and the previous result (cell 1 at defined $Z_c=50\text{ Ohm}$)



In fact the both simulations results are quite similar this show we have the same reflexion coefficient with a very low reflexion ; and a reflexion coefficient with less losses -2.2 dB

It is just another way of simulation

the smith chart show us S parameter modulus and the normalized impedance

m4 and m1 show a capacitive impedance behavior (S12 in the both case)

Conclusion :

This laboratory work demonstrates various methods of microwave simulation. We explored the utilization of the Smith chart, S-parameters, and Momentum simulation. The importance of adaptation in information transmission was highlighted. ADS provides diverse simulation options to analyze our line before execution.