

UNIVERSITY OF ILLINOIS AT URBANA-CHAMPAIGN  
Department of Electrical and Computer Engineering  
ECE 310 DIGITAL SIGNAL PROCESSING

## Homework 12

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Due: May 14, 2021

1. Consider the system shown below.



- (a) Let  $x[n] = \sin(n\frac{\pi}{2})$ . Determine and sketch  $y[n]$ .  
 (b) Let

$$X_d(\omega) = \frac{1}{1 - \frac{1}{2}e^{-j\omega}}$$

show that

$$Y_d(\omega) = \frac{1}{1 - \frac{1}{4}e^{-j\omega}}$$

**Solution**

- (a)  $x[n] = \sin(n\frac{\pi}{2}) = \{0, 1, 0, -1, 0, 1, 0, \dots\}$ .  
 After downsampling we are left with nothing:

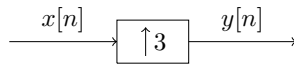
$$y[n] = \{0, 0, 0, 0, \dots\}$$

Notice down-sampling is not shift invariant,  $\{1, -1, 1, -1, \dots\}$  is not an accepted answer.

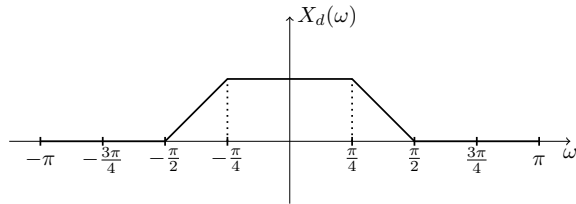
- (b) Recall downsampling in frequency domain:

$$\begin{aligned}
 Y_d(\omega) &= \frac{1}{D} \sum_{l=0}^{D-1} X_d\left(\frac{\omega - 2\pi l}{D}\right) \\
 &= \frac{1}{2} [X_d(\omega/2) + X_d(\omega/2 - \pi)] \\
 &= \frac{1}{2} \left( \frac{1}{1 - \frac{1}{2}e^{-j\omega/2}} + \frac{1}{1 - \frac{1}{2}e^{-j(\omega/2 - \pi)}} \right) \\
 &= \frac{1}{2} \left( \frac{1}{1 - \frac{1}{2}e^{-j\omega/2}} + \frac{1}{1 + \frac{1}{2}e^{-j(\omega/2)}} \right) \\
 &= \frac{1}{2} \frac{1 - \frac{1}{2}e^{-j\omega/2} + 1 + \frac{1}{2}e^{-j(\omega/2)}}{(1 - \frac{1}{2}e^{-j\omega/2})(1 + \frac{1}{2}e^{-j(\omega/2)})} \\
 &= \frac{1}{1 - \frac{1}{4}e^{-j\omega}}
 \end{aligned}$$

2. Consider the system shown below:

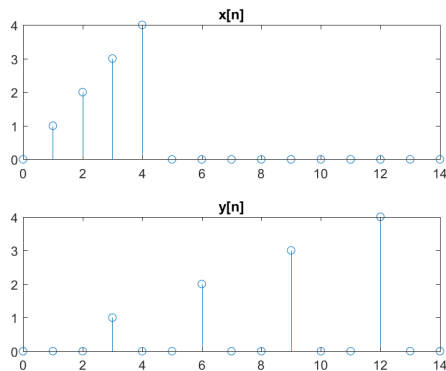


- (a) Let  $x[n] = n(u[n] - u[n - 5])$ . Determine and sketch  $y[n]$ .  
 (b) Let  $X_d(\omega)$  be as shown in the figure below. Determine and sketch  $Y_d(\omega)$ .



**Solution:**

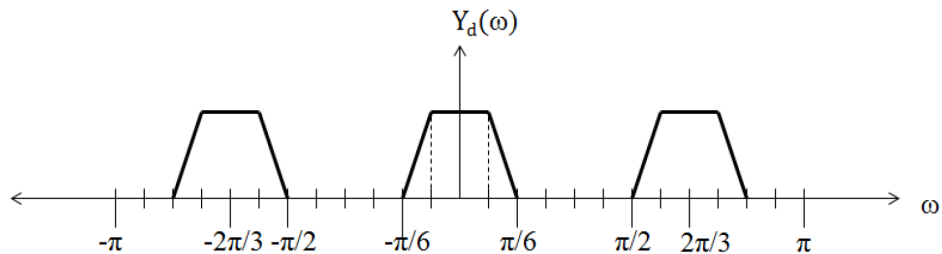
- (a) See figure for  $x[n]$  and  $y[n]$



- (b) Recall upsampling in the frequency domain:

$$Y_d(\omega) = X_d(I\omega)$$

We just have to shrink the whole axis by 3. Note that since  $X_d(\omega)$  is  $2\pi$  periodic, the copies at  $2\pi$ 's are also shrunk into the  $[-\pi, \pi]$  range of  $Y_d(\omega)$



3. Let

$$y[n] = \begin{cases} x[n/M], & n = \ell M, \text{ with integer } \ell \\ 0, & \text{otherwise} \end{cases}$$

Show that  $Y_d(\omega) = X_d(M\omega)$ .

**Solution:**

$$Y_d(\omega) = \sum_{n=-\infty}^{\infty} y[n]e^{-j\omega n} = \sum_{n=-\infty}^{\infty} x[n/M]e^{-j\omega n} = \sum_{\hat{n}=-\infty}^{\infty} x[\hat{n}]e^{-jM\hat{n}\omega} = X_d(M\omega)$$

where  $\hat{n} = n / M$

4. Let  $y[n] = x[Ln]$ . Show that

$$Y_d(\omega) = \frac{1}{L} \sum_{\ell=0}^{L-1} X_d\left(\frac{\omega - 2\ell\pi}{L}\right)$$

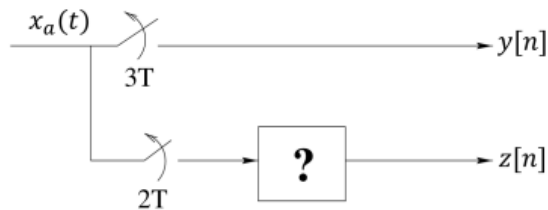
**Solution:**

$$Y_d(\omega) = \sum_{n=-\infty}^{\infty} y[n]e^{-j\omega n} = \sum_{n=-\infty}^{\infty} \tilde{x}[Ln]e^{-j\omega n} = \sum_{\hat{l}=-\infty}^{\infty} \tilde{x}[\hat{l}]e^{-j\frac{\omega}{L}\hat{l}} = \tilde{X}_d\left(\frac{\omega}{L}\right)$$

where

$$\begin{aligned} \tilde{x}[n] &= x[n]p[n] = x[n]\frac{1}{L}\sum_{l=0}^{L-1} e^{j\frac{2\pi}{L}ln} \\ \tilde{X}_d(\omega) &= \sum_{n=-\infty}^{\infty} \tilde{x}[n]e^{-j\omega n} = \frac{1}{L}\sum_{l=0}^{L-1}\sum_{n=-\infty}^{\infty} x[n]e^{-j(\omega - \frac{2\pi l}{L})n} = X_d\left(\omega - l\frac{2\pi}{L}\right) \\ \Rightarrow Y_d(\omega) &= \frac{1}{L}\sum_{l=0}^{L-1} X_d\left(\frac{\omega - 2\pi l}{L}\right) \end{aligned}$$

5. Consider the following system consisting of two synchronized ideal A/D convertors. Assume that the input analog signal  $x_a(t)$  is bandlimited to  $\Omega_0 = \pi/(3T)$ . Design a digital rate conversion subsystem marked with “?” using down-sampler(s), up-sampler(s), and digital filter(s) as necessary such that  $y[n] = z[n]$ . Draw a block diagram and determine all the essential parameters of the subsystem.



**Solution:**

$$\rightarrow \boxed{\uparrow 2} \rightarrow \boxed{LPF} \rightarrow \boxed{\downarrow 3} \rightarrow$$

where

$$T_1 = 1.5T_2 \Rightarrow \alpha = \frac{L}{M} = \frac{3}{2} \Rightarrow L = 3, M = 2$$

LPF

