# 第七章 带电粒子和电磁场的相互作用

- (1) 已知粒子运动求电磁场
- (2) 粒子激发电磁场对粒子的反作用
- (3) 粒子与外电磁场的相互作用



# § 1 运动带电粒子的势和辐射电磁场

### 1. 任意运动带电粒子的势

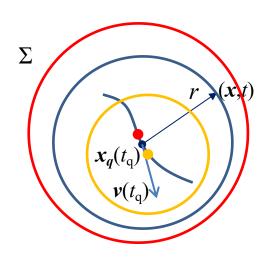


图1. 电荷运动轨迹 $x_q$ 和场点(x,t). 红圈、蓝圈、黄圈分别表示电荷在红点、蓝点、黄点激发的电磁场在t 时刻所到达的球面. 三个球面不可能有交点。

电荷在 $t_q$  激发的电磁场局限在以 $\mathbf{x}_q(t_q)$ 为圆心、以光速扩张的无穷薄球面上

$$r \equiv |\mathbf{x} - \mathbf{x}_q(t_q)| = c(t - t_q)$$

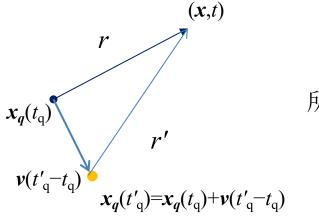


因为光速与发光电荷的运动无关,而电荷的速度小于光速,所以三个球面不可能有交点,即对给定的 $\mathbf{x}$ 和t只有唯一的 $t_q$ 满足下式

$$|\mathbf{x} - \mathbf{x}_q(t_q)| = c(t - t_q) \tag{1}$$

证明:设  $r = c(t - t_q)$ ,  $r' = |\mathbf{x} - \mathbf{x}_q(t_q')|$ ,  $t_q' > t_q$ .

 $\sum$ 



因为三角形两边和大于第三边,

$$r' + v(t'_q - t_q) > r = c(t - t_q)$$

所以

$$r' > c(t - t_q) - v(t'_q - t_q)$$
  
=  $c(t - t'_q) + (c - v)(t'_q - t_q)$   
>  $c(t - t'_q)$ 

所以, $x_q(t'_q)$ 发出的电磁波不可能在t 时刻达到x点。



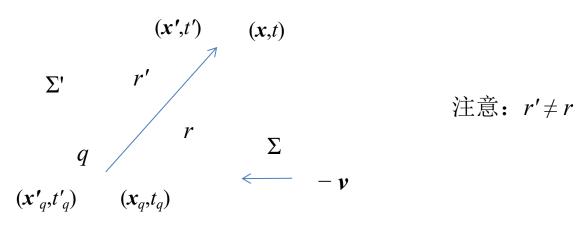
场方程

$$(\partial_{\nu}\partial^{\nu})A^{\mu} - \partial^{\mu}(\partial_{\nu}A^{\nu}) = -\mu_{0}j^{\mu}$$

带电物质仅通过四维电流密度与电磁势发生相互作用,所以带电物质激发的电磁势与电荷加速度无关。从而在电荷<mark>瞬时</mark>静止的惯性系Σ′中,电磁势等于静电荷激发的静电势(已取无穷远处为零的边界条件)

$$A_a'(r') = \left(0,0,0,\frac{i}{c}\frac{q}{4\pi\varepsilon_0 r'}\right) \tag{2}$$

q可以有加速度



电动力学 第七章



设在实验室惯性系 $\Sigma$ 的 $t_q$ 时刻电荷具有瞬时速度 $v(t_q)$ ,则 $\Sigma$ 相对 $\Sigma$ '以速度 $-v(t_q)$ 运动. 不妨设运动沿X轴方向. 由典型洛伦兹变换得 $\Sigma$ 系中的电磁势

$$\begin{pmatrix} A^{1} \\ A^{2} \\ A^{3} \\ \frac{i}{c} \varphi \end{pmatrix} = \begin{pmatrix} \gamma & 0 & 0 & -\frac{iv}{c} \gamma \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ \frac{iv}{c} \gamma & 0 & 0 & \gamma \end{pmatrix} \begin{pmatrix} 0 \\ 0 \\ 0 \\ \frac{i}{c} \frac{q}{4\pi\varepsilon_{0} r'} \end{pmatrix} = \gamma(v) \begin{pmatrix} \frac{qv}{4\pi\varepsilon_{0} c^{2} r'} \\ 0 \\ 0 \\ \frac{i}{c} \frac{q}{4\pi\varepsilon_{0} r'} \end{pmatrix}$$
(3)

其中 
$$v = v(t_q)$$
,和  $\gamma(v) = \frac{1}{\sqrt{1 - v^2(t_q)/c^2}}$ 

$$t' - t'_q = \frac{t - t_q - \frac{v}{c^2}(x^1 - x_q^1)}{\sqrt{1 - (v/c)^2}}$$



根据光速不变原理,在Σ′系源点与场点的距离

$$r' = c(t' - t'_q) = \frac{c(t - t_q) - \frac{1}{c}v(x^1 - x_q^1)}{\sqrt{1 - v^2/c^2}} = \frac{r - \frac{1}{c}v(x^1 - x_q^1)}{\sqrt{1 - v^2/c^2}}$$

通过空间转动,可以从电荷沿X轴运动的结果得到电荷任意运动的结果。

存在空间转动三维正交矩阵  $\hat{O}$  ,  $\hat{O}\hat{O}^T = \hat{O}^T\hat{O} = I$  , 使得

$$\begin{pmatrix} v^{1} \\ v^{2} \\ v^{3} \end{pmatrix} \xrightarrow{\hat{O}} \begin{pmatrix} v''^{1} \\ 0 \\ 0 \end{pmatrix} = \hat{O} \begin{pmatrix} v^{1} \\ v^{2} \\ v^{3} \end{pmatrix}, \qquad \begin{pmatrix} x^{1} - x_{q}^{1} \\ x^{2} - x_{q}^{2} \\ x^{3} - x_{q}^{3} \end{pmatrix} \xrightarrow{\hat{O}} \begin{pmatrix} x''^{1} - x_{q}''^{1} \\ 0 \\ 0 \end{pmatrix} = \hat{O} \begin{pmatrix} x^{1} - x_{q}^{1} \\ x^{2} - x_{q}^{2} \\ x^{3} - x_{q}^{3} \end{pmatrix}$$

从而

$$r' = \frac{r - \frac{1}{c} v''^{1} (x''^{1} - x_{q}^{"1})}{\sqrt{1 - v^{2} / c^{2}}} = \frac{r - \frac{1}{c} v^{T} \hat{O}^{T} \hat{O} r}{\sqrt{1 - v^{2} / c^{2}}} = \frac{r - \frac{1}{c} v^{T} r}{\sqrt{1 - v^{2} / c^{2}}} = \frac{r - \frac{1}{c} v^{T} r}{\sqrt{1 - v^{2} / c^{2}}} = \frac{r - \frac{1}{c} v \cdot r}{\sqrt{1 - v^{2} / c^{2}}}$$
(4)

其中  $\mathbf{r} = \mathbf{x} - \mathbf{x}_q$ 和 $\mathbf{r} = \sqrt{(\mathbf{r} \cdot \mathbf{r})} = c(t - t_q)$ . 右边表达式在空间坐标转动下不变,因此它对任意选取的笛卡尔坐标架普遍地成立.



同理可以通过空间转动,从电荷瞬时沿*X*轴运动的结果(3)式和(4)得到任意运动电荷辐射的电磁势。

#### 李纳-维谢尔势

$$A = \frac{q\mathbf{v}}{4\pi\varepsilon_0 c^2 (r - \frac{1}{c}\mathbf{v} \cdot \mathbf{r})}$$
 (5a)

$$\varphi = \frac{q}{4\pi\varepsilon_0(r - \frac{1}{c}\mathbf{v} \cdot \mathbf{r})} \tag{5b}$$

复杂性包含在v和r中,他们均依赖于 $t_q$ ,而后者是x和t 的隐函数。所以求E和B时涉及隐函数求导。



## 2. 连续分布带电体的势(推迟势)

考虑带电粒子集合 $\{q_n|n=1,2,3...\}$ , $q_n$ 的轨迹为 $\mathbf{x}_n(t')$ . 记 $\mathbf{r}_n(t')=\mathbf{x}-\mathbf{x}_n(t')$ .

$$r_{n} \frac{\partial r_{n}}{\partial t'} = -\frac{d\mathbf{x}_{n}}{dt'} \cdot \mathbf{r}_{n} = -\mathbf{v}_{n}(t') \cdot \mathbf{r}_{n}, \qquad (\text{Ar} \mathbf{x} \mathbf{n})$$

$$r_{n} - \frac{1}{c} \mathbf{v}_{n} \cdot \mathbf{r}_{n} = r_{n} + r_{n} \frac{\partial r_{n}}{\partial t'} = r_{n} \frac{\partial}{\partial t'} \left( t' - t + \frac{1}{c} r_{n} \right) \qquad (6)$$

设  $q_n$  在  $t_n$  时刻发射的电磁波在t 时刻达到场点 x 即  $r_n(t_n) = c(t-t_n)$ . 可利用(6)式把 $q_n$ 激发的李纳-维谢尔矢势写成

$$A_n(\mathbf{x},t) = \frac{q_n \mathbf{v}_n(t_n)}{4\pi\varepsilon_0 c^2 (r_n - \frac{1}{c} \mathbf{v}_n(t_n) \cdot \mathbf{r}_n)} = \frac{q_n \mathbf{v}_n(t_n)}{4\pi\varepsilon_0 c^2 r_n \frac{\partial}{\partial t_n} (t_n - t + \frac{1}{c} r_n(t_n))}$$
(7)

为了方便对n求和,改写(7)式(目的是消去分母对n的依赖),

$$A_n(\mathbf{x},t) = \int dV' \int dt' \frac{\delta(t'-t_n)}{\frac{\partial}{\partial t_n} (t_n - t + \frac{1}{c} r_n(t_n))} \frac{\delta(\mathbf{x}' - \mathbf{x}_n(t')) q_n \mathbf{v}_n(t')}{4\pi \varepsilon_0 c^2 |\mathbf{x} - \mathbf{x}'|}$$
(8)



ਪੋਟੀ:  $f(t')=t'-t+r_n(t')/c$ 

因为  $r_n(t_n) = c(t-t_n)$ ,所以  $f(t_n) = 0$ ,即 $t_n$ 的是方程 f(t') = 0的根,而且前面已经分析过,这个根是唯一的。从而

$$\mathcal{S}(f(t')) = \mathcal{S}\left(\frac{\partial f(t')}{\partial t'}\Big|_{t'=t_n} (t'-t_n)\right) = \frac{1}{\frac{\partial f(t')}{\partial t'}\Big|_{t'=t_n}} \mathcal{S}(t'-t_n)$$
(9)

利用(9)式并记r = |x-x'|, (8)式写成

$$A_{n} = \int dV' \int dt' \delta(t' - t + \frac{1}{c} r_{n}(t')) \frac{\delta(\mathbf{x}' - \mathbf{x}_{n}(t')) q_{n} \mathbf{v}_{n}(t')}{4\pi \varepsilon_{0} c^{2} |\mathbf{x} - \mathbf{x}'|}$$

$$= \int dV' \int dt' \delta(t' - t + \frac{1}{c} |\mathbf{x} - \mathbf{x}'|) \frac{\delta(\mathbf{x}' - \mathbf{x}_{n}(t')) q_{n} \mathbf{v}_{n}(t')}{4\pi \varepsilon_{0} c^{2} |\mathbf{x} - \mathbf{x}'|}$$

$$= \int dV' \frac{\delta(\mathbf{x}' - \mathbf{x}_{n}(t - \frac{1}{c} r)) q_{n} \mathbf{v}_{n}(t - \frac{1}{c} r)}{4\pi \varepsilon_{0} c^{2} r}$$

$$(10)$$



回忆电流密度

$$J(x,t) = \sum_{n} q_n \delta(x - x_n(t)) v_n(t)$$
(11)

从而总矢势

$$A(\mathbf{x},t) = \sum_{n} A_{n} = \int dV' \frac{J(\mathbf{x}', t - \frac{1}{c}r)}{4\pi\varepsilon_{0}c^{2}r}$$
(12)

类似可得标势

$$\varphi(\mathbf{x},t) = \int dV' \frac{\rho(\mathbf{x}',t-\frac{1}{c}r)}{4\pi\varepsilon_0 r}$$
 (13)

公式(12)和(13)式正是推迟势公式,它们也可以在罗伦芝规范条件下直接从麦克斯韦方程方程组推导出来.推迟势反映了真空中光信号传播速度恒等于c.



## 3. 辐射电磁场

数学准备。记

$$s \equiv r - \frac{1}{c} \mathbf{v} \cdot \mathbf{r} \tag{14}$$

则(4)式给出

$$r' = \gamma(v)s \tag{15}$$

对给定的x和t,  $t_q$  唯一,故存在单值函数 $t_q = t_q(x,t)$ . 可证下面偏导数(<u>附录1</u>),

$$\frac{\partial t_q}{\partial t} = \frac{r}{s}, \qquad \frac{\partial r}{\partial t}\Big|_{x} = -\frac{r}{s}v, \qquad \frac{\partial r}{\partial t}\Big|_{x} = -\frac{r \cdot v}{s}, \qquad (16a)$$

$$\frac{\partial t_q}{\partial x^j} = -\frac{r^j}{cs}, \qquad \frac{\partial \mathbf{r}}{\partial x^j}\Big|_t = \hat{\mathbf{e}}_j + \frac{r^j}{cs}\mathbf{v}, \qquad \frac{\partial r}{\partial x^j}\Big|_t = \frac{r^j}{s}$$
(16b)



利用上述数学关系式, 计算李纳-维谢尔势对应的场强, 可得(附录2, 3)

$$\boldsymbol{E} = -\nabla \varphi - \frac{\partial \boldsymbol{A}}{\partial t} = \boldsymbol{E}_c + \boldsymbol{E}_{rad}, \qquad \boldsymbol{B} = \nabla \times \boldsymbol{A} = \boldsymbol{B}_c + \boldsymbol{B}_{rad}$$
(17)

非辐射场(最右表达式作了低速近似)

$$\boldsymbol{E}_{c} = \frac{q}{4\pi\varepsilon_{0}r^{2}} \frac{\hat{\boldsymbol{e}}_{r} - \frac{1}{c}\boldsymbol{v}}{(1 - \frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_{r})^{3}\gamma^{2}} \sim \frac{q}{4\pi\varepsilon_{0}r^{2}} \hat{\boldsymbol{e}}_{r}$$
(18)

$$\boldsymbol{B}_{c} = \frac{q}{4\pi\varepsilon_{0}c^{2}r^{2}} \frac{(1-\frac{v^{2}}{c^{2}})\boldsymbol{v} \times \hat{\boldsymbol{e}}_{r}}{(1-\frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_{r})^{3}} \sim \frac{q}{4\pi\varepsilon_{0}c^{2}r^{2}} \boldsymbol{v} \times \hat{\boldsymbol{e}}_{r}$$
(19)

辐射场(最右表达式作了低速近似)

$$\boldsymbol{E}_{rad} = \frac{q}{4\pi\varepsilon_0 c^2 r} \frac{\hat{\boldsymbol{e}}_r \times [(\hat{\boldsymbol{e}}_r - \frac{1}{c}\boldsymbol{v}) \times \dot{\boldsymbol{v}}]}{(1 - \frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r)^3} \sim \frac{q}{4\pi\varepsilon_0 c^2 r} \hat{\boldsymbol{e}}_r \times (\hat{\boldsymbol{e}}_r \times \dot{\boldsymbol{v}})$$
(20)

$$\boldsymbol{B}_{rad} = \frac{q}{4\pi\varepsilon_0 c^3 r} \frac{\dot{\boldsymbol{v}} \times \hat{\boldsymbol{e}}_r + \frac{1}{c} [(\dot{\boldsymbol{v}} \cdot \hat{\boldsymbol{e}}_r)(\boldsymbol{v} \times \hat{\boldsymbol{e}}_r) - (\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r)(\dot{\boldsymbol{v}} \times \hat{\boldsymbol{e}}_r)]}{(1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r)^3} = \frac{1}{c} \hat{\boldsymbol{e}}_r \times \boldsymbol{E}_{rad} \sim \frac{q}{4\pi\varepsilon_0 c^3 r} \dot{\boldsymbol{v}} \times \hat{\boldsymbol{e}}_r$$
(21)



## 辐射场的特点:

辐射场强以1/r衰减,而非辐射场强以1/r2衰减.

与 | v | 成正比

 $E_{rad}$ 和  $B_{rad}$  互相正交并分别与 r 正交(横场。在自由空间的情形)

可直接验证

$$\boldsymbol{B}_{rad} = \frac{1}{c} \hat{\boldsymbol{e}}_r \times \boldsymbol{E}_{rad} \tag{22}$$

# § 2 高速运动带电粒子的辐射

# 1. 高速运动带电粒子的辐射功率和角分布

辐射场的能量密度

$$\mathbf{S} = \mathbf{E} \times \mathbf{H} = \frac{1}{\mu_0} \mathbf{E} \times \mathbf{B} = \frac{1}{\mu_0 c} \mathbf{E} \times (\hat{\mathbf{e}}_r \times \mathbf{E}) = \varepsilon_0 c E^2 \hat{\mathbf{e}}_r$$
 (23)

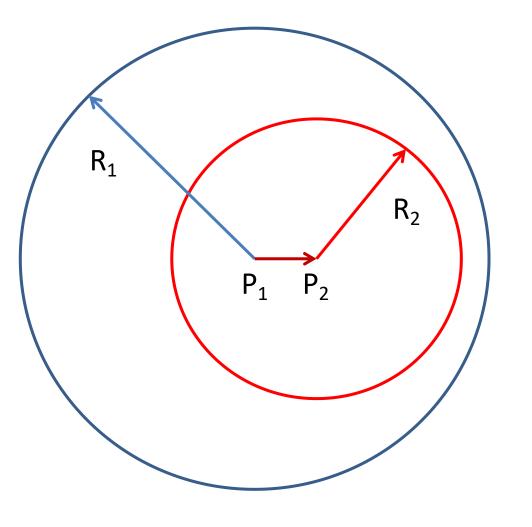
辐射场

$$\boldsymbol{E} = \frac{q}{4\pi\varepsilon_0 c^2 r} \frac{\hat{\boldsymbol{e}}_r \times [(\hat{\boldsymbol{e}}_r - \frac{1}{c}\boldsymbol{v}) \times \dot{\boldsymbol{v}}]}{(1 - \frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r)^3}$$
(24)

故t时刻x处(场点)

$$S = \frac{q^2}{16\pi^2 \varepsilon_0 c^3 r^2} \frac{\left|\hat{\boldsymbol{e}}_r \times \left[ (\hat{\boldsymbol{e}}_r - \frac{1}{c} \boldsymbol{v}) \times \dot{\boldsymbol{v}} \right] \right|^2}{\left(1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r \right)^6} \hat{\boldsymbol{e}}_r$$
 (25)

在源点附近S很复杂,所以一般不在源点附近求辐射功率.



P<sub>1</sub>: 电荷在t'时刻所在位置

P<sub>2</sub>: 电荷在t'+dt'时刻所在位置

在时刻t, $P_1$ 处发出的辐射达到半径为  $R_1$ 的球面

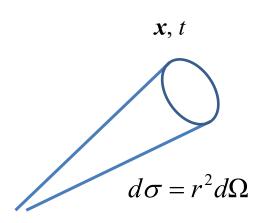
在时刻t, $P_2$ 处发出的辐射达到半径为 $R_2$ 的球面

$$R_1 = c(t - t') \qquad R_2 = c(t - t' - dt')$$

$$\overline{P_1 P_2} = v(t')dt'$$

电荷在dt'时间辐射的能量等于两球面 之间的辐射场能量

# 辐射功率角分布[方法一]



因为
$$x_q(t')$$
已知,从  $|x-x_q(t')|=c(t-t')$  可解出
$$t=t'+\frac{1}{c}|x-x_q(t')|\equiv t(x,t')$$

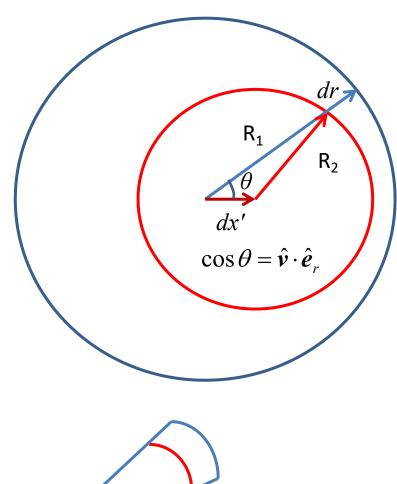
电荷在dt'时间产生的辐射经过x处面元 $d\sigma$ 的时间为

$$dt = \frac{\partial t}{\partial t'} \bigg|_{x} dt' = \left( \frac{\partial t'}{\partial t} \bigg|_{x} \right)^{-1} dt' = \left( 1 - \frac{1}{c} \mathbf{v} \cdot \hat{\mathbf{e}}_{r} \right) dt' \qquad (26)$$

记电荷辐射功率角分布为 $P(t',\Omega)$ ,即在dt'时间内辐射到 $d\Omega$ 立体角方向的能量为 $P(t',\Omega)dt'd\Omega$ (课本省略了变量 $\Omega$ ). 电荷在dt'辐射到 $d\Omega$ 的能量在 $t\sim t+dt$ 通过x处的 $d\sigma$  面元,

$$P(t',\Omega)dt'd\Omega = Sdtd\sigma$$

$$= Sr^{2} \left(1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_{r}\right) dt' d\Omega = \frac{q^{2}}{16\pi^{2} \varepsilon_{0} c^{3}} \frac{\left|\hat{\boldsymbol{e}}_{r} \times \left[\left(\hat{\boldsymbol{e}}_{r} - \frac{1}{c} \boldsymbol{v}\right) \times \dot{\boldsymbol{v}}\right]\right|^{2}}{\left(1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_{r}\right)^{5}} dt' d\Omega$$
(27)



体元

 $dV = r^2 d\Omega dr$ 

# 辐射功率角分布[方法二]

设dt'辐射到 $\theta$ 方向的场到达x处厚为dr的壳层.

$$R_1 = c(t - t') = r,$$

$$R_2 = c(t - t' - dt') = r - cdt'$$

$$R_2^2 = (R_1 - dr)^2 + dx'^2 - 2(R_1 - dr)dx'\cos\theta$$

从而(只需比较一阶无穷小量)

$$dr = (c - v\cos\theta)dt' = c(1 - \frac{1}{c}\mathbf{v}\cdot\hat{\mathbf{e}}_r)dt'$$

在 t 时刻 x 处的辐射能量密度  $w = \varepsilon_0 E^2$  立体角 $d\Omega$ 内dr厚体元dV的辐射能量即

$$P(t',\Omega)dt'd\Omega = wdV = \varepsilon_0 E^2(\mathbf{x},t)r^2 d\Omega dr$$
$$= \varepsilon_0 c E^2(\mathbf{x},t)r^2 (1 - \frac{1}{c}\mathbf{v} \cdot \hat{\mathbf{e}}_r)dt'd\Omega$$

它等于dt'辐射到 $d\Omega$ 的能量.

与(27)一致

### 辐射功率

记电荷辐射功率为P(t').

$$P(t') = \int P(t', \Omega) d\Omega \tag{28}$$

辐射功率角分布为

$$\frac{dP(t')}{d\Omega} = P(t', \Omega) = \frac{q^2}{16\pi^2 \varepsilon_0 c^3} \frac{\left|\hat{\boldsymbol{e}}_r \times \left[\left(\hat{\boldsymbol{e}}_r - \frac{1}{c}\boldsymbol{v}\right) \times \dot{\boldsymbol{v}}\right]\right|^2}{\left(1 - \frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r\right)^5}$$
(29)

故辐射功率为

$$P(t') = \frac{q^2}{16\pi^2 \varepsilon_0 c^3} \int \frac{\left|\hat{\boldsymbol{e}}_r \times \left[\left(\hat{\boldsymbol{e}}_r - \frac{1}{c}\boldsymbol{v}\right) \times \dot{\boldsymbol{v}}\right]\right|^2}{\left(1 - \frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r\right)^5} d\Omega$$
(30)

### 高速运动: v/c~1

$$1 - \frac{1}{c} \mathbf{v} \cdot \hat{\mathbf{e}}_r = 1 - \frac{v}{c} \cos \theta$$

当
$$\theta \sim 0$$
 时  $1 - \frac{1}{c} \mathbf{v} \cdot \hat{\mathbf{e}}_r \approx \frac{1}{2} (\gamma^{-2} + \theta^2)$ 

$$\frac{dP(t')}{d\Omega} = \frac{q^2}{16\pi^2 \varepsilon_0 c^3} \frac{\left|\hat{\boldsymbol{e}}_r \times \left[\left(\hat{\boldsymbol{e}}_r - \frac{1}{c}\boldsymbol{v}\right) \times \dot{\boldsymbol{v}}\right]\right|^2}{\left(1 - \frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r\right)^5}$$

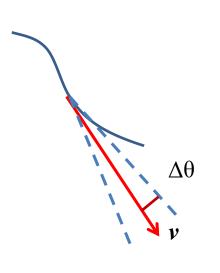
上式分母很小

$$\frac{1}{\left(1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r\right)^5} \approx \frac{32}{\left(\gamma^{-2} + \theta^2\right)^5}$$
 (31)

当 $\theta > \gamma^{-1}$ ,辐射功率迅速减少,从而射功率集中在粒子运动方向(此因子不涉及加速度方向)

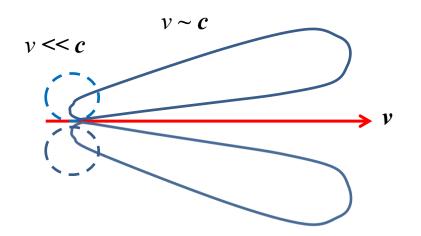
$$\Delta\theta \sim \frac{1}{\gamma} \tag{32}$$

电动力学第七章



# 2. 加速度平行速度情形

例如直线加速器



辐射功率角分布

$$\frac{dP(t')}{d\Omega} = \frac{q^2 \dot{\mathbf{v}}^2}{16\pi^2 \varepsilon_0 c^3} \frac{\sin^2 \theta}{\left(1 - \frac{v}{c} \cos \theta\right)^5}$$
(33)

辐射功率

$$P(t') = \frac{q^2 \dot{v}^2}{16\pi^2 \varepsilon_0 c^3} \int \frac{\sin^2 \theta}{\left(1 - \frac{v}{c} \cos \theta\right)^5} d\Omega$$
$$= \frac{q^2 \dot{v}^2}{6\pi \varepsilon_0 c^3} \gamma^6$$
(34)

电动力学 第七章

电荷受力(总力)

$$F(t') = \frac{d(\gamma m v)}{dt'}$$

$$= m \frac{\dot{v}(1 - (\frac{v}{c})^2) + \frac{1}{c^2} (v \cdot \dot{v}) v}{(1 - (\frac{v}{c})^2)^{3/2}} = \gamma^3 m \dot{v}$$
(35)

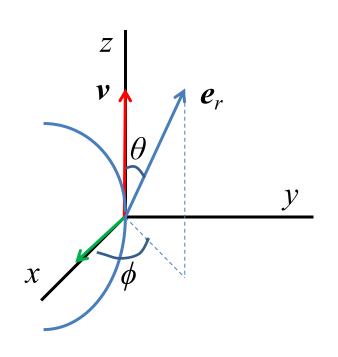
从而

$$P(t') = \frac{q^2}{6\pi\varepsilon_0 m^2 c^3} F^2 \tag{36}$$

直线加速运动电荷辐射功率只与其受力有关,而与粒子能量无关,由此可产生高能粒子.

# 3. 加速度垂直速度情形

例如回旋加速器的同步辐射



$$\mathbf{v} \cdot \hat{\mathbf{e}}_r = \mathbf{v} \cos \theta$$

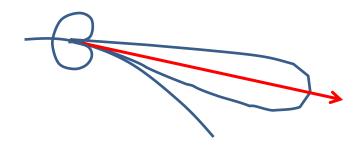
$$\dot{\mathbf{v}} \cdot \hat{\mathbf{e}}_r = \dot{\mathbf{v}} \sin \theta \cos \phi$$

$$\dot{\mathbf{v}} \cdot \mathbf{v} = 0$$

$$\hat{\mathbf{e}}_r \times \left[ (\hat{\mathbf{e}}_r - \frac{1}{c} \mathbf{v}) \times \dot{\mathbf{v}} \right]$$

$$\begin{aligned} & \left| \hat{\boldsymbol{e}}_r \times \left[ (\hat{\boldsymbol{e}}_r - \frac{1}{c} \boldsymbol{v}) \times \dot{\boldsymbol{v}} \right] \right|^2 \\ &= \dot{\boldsymbol{v}}^2 \left[ (1 - \frac{\boldsymbol{v}}{c} \cos \theta)^2 - (1 - \frac{\boldsymbol{v}^2}{c^2}) \sin^2 \theta \cos^2 \phi \right] \end{aligned}$$

 $= \dot{v}\sin\theta\cos\phi(\hat{\boldsymbol{e}}_r - \frac{1}{c}\boldsymbol{v}) - (1 - \frac{v}{c}\cos\theta)\dot{\boldsymbol{v}}$ 



辐射功率角分布

$$\frac{dP(t')}{d\Omega} = \frac{q^2 \dot{v}^2}{16\pi^2 \varepsilon_0 c^3} \frac{(1 - \frac{v}{c}\cos\theta)^2 - (1 - \frac{v^2}{c^2})\sin^2\theta\cos^2\phi}{(1 - \frac{v}{c}\cos\theta)^5}$$
(37)

辐射功率

$$P(t') = \frac{q^2}{16\pi^2 \varepsilon_0 c^3} \int \frac{dP(t')}{d\Omega} d\Omega = \frac{q^2 \dot{v}^2}{6\pi \varepsilon_0 c^3} \gamma^4$$
 (38)

电荷受力(总力)

$$F(t') = \frac{d(\gamma m v)}{dt'} = m \frac{\dot{v}(1 - (\frac{v}{c})^2) + \frac{1}{c^2} (v \cdot \dot{v}) v}{(1 - (\frac{v}{c})^2)^{3/2}} = \gamma m \dot{v}$$
(39)

辐射功率写成

$$P(t') = \frac{q^2}{6\pi\varepsilon_0 m^2 c^3} \gamma^2 F^2$$

$$= \frac{q^2}{6\pi\varepsilon_0 m^4 c^7} E_q^2 F^2,$$
(40)

电荷能量

$$E_q = \gamma mc^2. (41)$$

当*F*给定后,辐射功率与电荷能量平方成正比. 当辐射功率等于输入功率时,粒子能量不能继续增加. 但由此产生的"同步辐射"是很有用的.

# §3辐射的频谱分析

# 1. 频谱分析的一般公式

傅里叶变换

$$f(t) = \int_{-\infty}^{\infty} f_{\omega} e^{-i\omega t} d\omega \tag{42}$$

$$f_{\omega} = \frac{1}{2\pi} \int_{-\infty}^{\infty} f(t)e^{i\omega t}dt \tag{43}$$

若f(t)为实函数,

$$f_{\omega}^* = f_{-\omega} \tag{44}$$

$$\int_{-\infty}^{\infty} f^2(t)dt = 4\pi \int_{0}^{\infty} \left| f_{\omega} \right|^2 d\omega \tag{45}$$

### 例

(1) 电流密度

$$J(x,t) = \int_{-\infty}^{\infty} J_{\omega}(x)e^{-i\omega t}d\omega, \qquad J_{\omega}(x) = \frac{1}{2\pi} \int_{-\infty}^{\infty} J(x,t)e^{i\omega t}dt$$
(46)

(2) 推迟势

$$A(\mathbf{x},t) = \frac{\mu_0}{4\pi} \int \frac{\mathbf{J}(\mathbf{x}',t-\frac{r}{c})}{r} dV'$$

$$= \frac{\mu_0}{4\pi} \int \frac{dV'}{r} \int_{-\infty}^{\infty} \mathbf{J}_{\omega}(\mathbf{x}') e^{-i\omega t + i\omega\frac{r}{c}} d\omega$$

$$= \frac{\mu_0}{4\pi} \int_{-\infty}^{\infty} e^{-i\omega t} d\omega \left[ \int \frac{\mathbf{J}_{\omega}(\mathbf{x}')}{r} e^{i\omega\frac{r}{c}} dV' \right]$$
(47)

可见

$$A_{\omega}(\mathbf{x}) = \frac{\mu_0}{4\pi} \int \frac{\mathbf{J}_{\omega}(\mathbf{x}')}{r} e^{i\omega_c^r} dV' = \frac{\mu_0}{8\pi^2} \int e^{i\omega(t'+\frac{r}{c})} dt' \int \frac{\mathbf{J}(\mathbf{x}',t')}{r} dV'$$
(48)

类似可得

$$\varphi_{\omega}(\mathbf{x}) = \frac{1}{8\pi^{2} \varepsilon_{0}} \int e^{i\omega(t' + \frac{r}{c})} dt' \int \frac{\rho(\mathbf{x}', t')}{r} dV'$$
(49)

直接用电流密度计算势的频谱分量更方便.

#### 一般推迟势的频谱分量公式

$$A_{\omega}(\mathbf{x}) = \frac{\mu_0}{8\pi^2} \int e^{i\omega(t' + \frac{r}{c})} dt' \int \frac{\mathbf{J}(\mathbf{x}', t')}{r} dV', \tag{48}$$

$$\varphi_{\omega}(\mathbf{x}) = \frac{1}{8\pi^{2} \varepsilon_{0}} \int e^{i\omega(t' + \frac{r}{c})} dt' \int \frac{\rho(\mathbf{x}', t')}{r} dV'$$
(49)

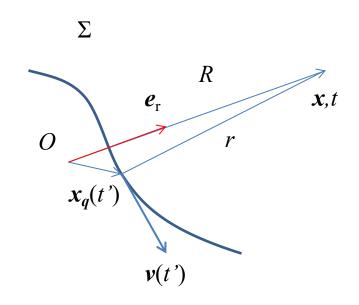
电流密度和电荷密度分别为

$$J(\mathbf{x}',t') = \rho(\mathbf{x}',t')\mathbf{v}(t'), \tag{50}$$

$$\rho(\mathbf{x}',t') = q\,\delta(\mathbf{x}' - \mathbf{x}_q(t')) \tag{51}$$

故

$$A_{\omega}(\mathbf{x}) = \frac{\mu_0}{8\pi^2} \int_{-\infty}^{\infty} \frac{q\mathbf{v}(t')}{r} e^{i\omega(t' + \frac{r}{c})} dt'$$
 (52)



远场近似(如同多极矩展开)

$$r \approx R - \hat{\boldsymbol{e}}_r \cdot \boldsymbol{x}_q(t') \tag{53}$$

$$A_{\omega}(\mathbf{x}) = \frac{\mu_0 q}{8\pi^2} \frac{e^{ikR}}{R} \int_{-\infty}^{\infty} \mathbf{v}(t') e^{i\omega(t' - \frac{1}{c}\hat{\mathbf{e}}_r \cdot \mathbf{x}_q)} dt'$$
(54)

其中 $k=\omega/c$ , R=|x|.

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曲 (54): 
$$A_{\omega}(\mathbf{x}) = \frac{\mu_0 q}{8\pi^2} \frac{e^{ikR}}{R} \int_{-\infty}^{\infty} \mathbf{v}(t') e^{i\omega(t' - \frac{1}{c}\hat{\mathbf{e}}_r \cdot \mathbf{x}_q)} dt' \quad 可得$$

磁场 (不用对分母求导)

$$\mathbf{B}_{\omega} = \nabla \times \mathbf{A}_{\omega} = ik\hat{\mathbf{e}}_{r} \times \mathbf{A}_{\omega}$$

$$= \frac{iq\omega}{8\pi^{2}\varepsilon_{0}c^{3}} \frac{e^{ikR}}{R} \int_{-\infty}^{\infty} \hat{\mathbf{e}}_{r} \times \mathbf{v}(t') e^{i\omega(t' - \frac{1}{c}\hat{\mathbf{e}}_{r} \cdot \mathbf{x}_{q})} dt'$$
(55)

电场

$$\boldsymbol{E}_{\omega} = -c\hat{\boldsymbol{e}}_{r} \times \boldsymbol{B}_{\omega}$$

$$= -\frac{iq\omega}{8\pi^{2}\varepsilon_{0}c^{2}} \frac{e^{ikR}}{R} \int_{-\infty}^{\infty} \hat{\boldsymbol{e}}_{r} \times (\hat{\boldsymbol{e}}_{r} \times \boldsymbol{v}) e^{i\omega(t' - \frac{1}{c}\hat{\boldsymbol{e}}_{r} \cdot \boldsymbol{x}_{q})} dt'$$
(56)

这两个式子和直接对辐射场E(t)和B(t)进行傅里叶变换的结果一致.

直接对辐射场E(t) 进行傅里叶变换:

$$\boldsymbol{E}_{\omega} = \frac{1}{2\pi} \int_{-\infty}^{\infty} \boldsymbol{E}_{rad} e^{i\omega t} dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} \boldsymbol{E}_{rad} e^{i\omega t} \partial_{t'} t dt'$$

曲 
$$\partial_{t'}t = 1 - \frac{1}{c}\mathbf{v} \cdot \hat{\mathbf{e}}_r$$
 和  $\mathbf{E}_{rad} = \frac{q}{4\pi\varepsilon_0 c^2 r} \frac{\hat{\mathbf{e}}_r \times [(\hat{\mathbf{e}}_r - \frac{1}{c}\mathbf{v}) \times \dot{\mathbf{v}}]}{(1 - \frac{1}{c}\mathbf{v} \cdot \hat{\mathbf{e}}_r)^3}$ 

$$\boldsymbol{E}_{\omega} = \frac{q}{8\pi^{2} \varepsilon_{0} c^{2}} \int \frac{\hat{\boldsymbol{e}}_{r} \times [(\hat{\boldsymbol{e}}_{r} - \frac{1}{c} \boldsymbol{v}) \times \dot{\boldsymbol{v}}]}{r(1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_{r})^{2}} e^{i\omega t} dt'$$

利用 
$$\frac{d}{dt'} \left[ \frac{\hat{\boldsymbol{e}}_r \times (\hat{\boldsymbol{e}}_r \times \boldsymbol{v})}{1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r} \right] = \frac{\hat{\boldsymbol{e}}_r \times [(\hat{\boldsymbol{e}}_r - \frac{1}{c} \boldsymbol{v}) \times \dot{\boldsymbol{v}}]}{(1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r)^2}$$
 得

$$\boldsymbol{E}_{\omega} = \frac{q}{8\pi^{2}\varepsilon_{0}c^{2}} \int \frac{e^{i\omega t}}{r} \frac{d}{dt'} \left[ \frac{\hat{\boldsymbol{e}}_{r} \times (\hat{\boldsymbol{e}}_{r} \times \boldsymbol{v})}{1 - \frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_{r}} \right] dt' = -\frac{q}{8\pi^{2}\varepsilon_{0}c^{2}} \int \frac{\hat{\boldsymbol{e}}_{r} \times (\hat{\boldsymbol{e}}_{r} \times \boldsymbol{v})}{1 - \frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_{r}} \frac{d}{dt'} \left[ \frac{e^{i\omega t}}{r} \right] dt'$$

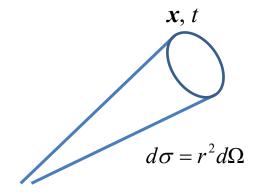
略去1/r<sup>2</sup>项,

$$\boldsymbol{E}_{\omega} = -\frac{iq\omega}{8\pi^{2}\varepsilon_{c}c^{2}} \int \frac{\hat{\boldsymbol{e}}_{r} \times (\hat{\boldsymbol{e}}_{r} \times \boldsymbol{v})}{r} e^{i\omega(t'-\frac{r}{c})} dt'$$

$$= \frac{1}{2} \int \frac{1}{2} \frac{1}{2} \int \frac{1}{2} \frac{1}{2} \int \frac{1}{2} \left( \frac{1}{2} \frac{1}{2} \right) dt'$$

电动力学第七章

# 辐射能量的角分布



$$\frac{dW}{d\Omega} = \int \frac{dP}{d\Omega} dt = \int_{-\infty}^{\infty} \mathbf{S} \cdot \hat{\boldsymbol{e}}_r R^2 dt$$
(57)

代入 
$$\mathbf{S} = \varepsilon_0 c E^2 \hat{\mathbf{e}}_r$$
,利用  $\int_{-\infty}^{\infty} f^2(t) dt = 4\pi \int_0^{\infty} |f_{\omega}|^2 d\omega$ 

$$\frac{dW}{d\Omega} = \varepsilon_0 c R^2 \int_{-\infty}^{\infty} E^2 dt = 4\pi \varepsilon_0 c R^2 \int_0^{\infty} \left| E_{\omega} \right|^2 d\omega \equiv \int_0^{\infty} \frac{dW_{\omega}}{d\Omega} d\omega \tag{58}$$

# 单位频率间隔辐射能量角分布 (角频率在ω)

$$\frac{dW_{\omega}}{d\Omega} = 4\pi\varepsilon_0 cR^2 |E_{\omega}|^2. \tag{59}$$

$$\left(\frac{dW_{\omega}}{d\Omega} = \frac{d^2W}{d\Omega d\omega}\right)$$

#### 对Ω积分得单位频率间隔辐射能量

$$W_{\omega} = 4\pi\varepsilon_0 cR^2 \oint |E_{\omega}|^2 d\Omega \tag{60}$$

先求 $E_{\omega}$ ,然后通过上面两式进行辐射能量的频谱分析.

电场傅里叶分量(复习)

$$\boldsymbol{E}_{\omega} = \frac{q}{8\pi^{2} \varepsilon_{0} c^{2}} \int \frac{\hat{\boldsymbol{e}}_{r} \times [(\hat{\boldsymbol{e}}_{r} - \frac{1}{c} \boldsymbol{v}) \times \dot{\boldsymbol{v}}]}{r(1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_{r})^{2}} e^{i\omega t} dt'$$

匀速运动, $\nu$ 为常矢量,故 $E_{\omega}$ =0.

远场近似  $r \approx R - \hat{\boldsymbol{e}}_R \cdot \boldsymbol{x}_q(t')$ ,

$$\boldsymbol{E}_{\omega} = -\frac{iq\omega}{8\pi^{2}\varepsilon_{0}c^{2}} \frac{e^{ikR}}{R} \hat{\boldsymbol{e}}_{R} \times \left[ \hat{\boldsymbol{e}}_{R} \times \int_{-\infty}^{\infty} v e^{i\omega(t' - \frac{1}{c}\hat{\boldsymbol{e}}_{R} \cdot \boldsymbol{x}_{q})} dt' \right]$$
(61)

周期直线运动  $x_q(t') = x_{q0} \cos(\omega_0 t')$ 

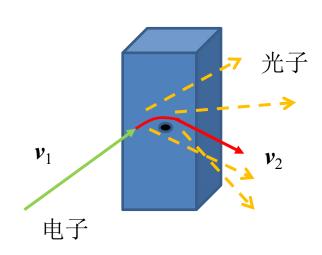
$$\boldsymbol{E}_{\omega} \approx -\frac{q\omega\omega_{0}}{16\pi^{2}\varepsilon_{0}c^{2}}\frac{e^{ikR}}{R}\hat{\boldsymbol{e}}_{R} \times \left[\hat{\boldsymbol{e}}_{R} \times \boldsymbol{x}_{q0} \int_{-\infty}^{\infty} e^{i(\omega-\omega_{0})t'-i\frac{\omega}{c}x_{q0R}\cos(\omega_{0}t')}dt'\right]$$
(62)

低速,  $\omega_0 x_{q0R} / c \ll 1$ 

$$\boldsymbol{E}_{\omega} \approx -\frac{q\,\omega\omega_{0}}{8\pi\varepsilon_{0}c^{2}}\frac{e^{i\boldsymbol{k}\boldsymbol{R}}}{R}\hat{\boldsymbol{e}}_{\boldsymbol{R}} \times \left(\hat{\boldsymbol{e}}_{\boldsymbol{R}} \times \boldsymbol{x}_{q0}\right)\delta(\omega - \omega_{0}) \tag{63}$$

可见,辐射频率集中在粒子运动的频率附近。

# 2. 低速运动带电粒子在碰撞过程中的辐射频谱



轫致辐射,  $\Delta v = v_2 - v_1$ 

#### 物理过程特点:

$$\boldsymbol{E}_{\omega} = \frac{q}{8\pi^{2} \varepsilon_{0} c^{2}} \int \frac{\hat{\boldsymbol{e}}_{r} \times [(\hat{\boldsymbol{e}}_{r} - \frac{1}{c} \boldsymbol{v}) \times \dot{\boldsymbol{v}}]}{r(1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_{r})^{2}} e^{i\omega t} dt'$$

设 $v \ll c$ 。采用偶极近似

$$r \approx R - \boldsymbol{e}_r \cdot \boldsymbol{x}_q \approx R, \quad t \approx t' + \frac{1}{c}(R - \boldsymbol{e}_r \cdot \boldsymbol{x}_q) \approx t' + \frac{1}{c}R$$

$$\boldsymbol{E}_{\omega} = \frac{q}{8\pi^{2} \varepsilon_{0} c^{2}} \frac{e^{ikR}}{R} \int \hat{\boldsymbol{e}}_{r} \times (\hat{\boldsymbol{e}}_{r} \times \dot{\boldsymbol{v}}) e^{i\omega t'} dt'$$
 (64)

 $\dot{v}$  只在很短的时间  $\tau$  内非零,从而

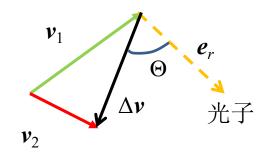
(1)若  $\omega \tau << 1$ ,因为 $e^{i\omega t'} \sim 1$ ,所以

$$\boldsymbol{E}_{\omega} \approx \frac{q}{8\pi^{2} \varepsilon_{0} c^{2}} \frac{e^{ikR}}{R} \hat{\boldsymbol{e}}_{r} \times (\hat{\boldsymbol{e}}_{r} \times \Delta \boldsymbol{v})$$
 (65)

(2)若  $\omega \tau >> 1$ ,因为 $e^{i\omega t}$  快速振荡,所以

$$\boldsymbol{E}_{\alpha} \approx 0 \tag{67}$$

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对(1): ωτ << 1 的情形,

$$|\mathbf{E}_{\omega}| \approx \frac{q}{8\pi^{2} \varepsilon_{0} c^{2}} \frac{1}{R} |\hat{\mathbf{e}}_{r} \times (\hat{\mathbf{e}}_{r} \times \Delta \mathbf{v})|$$

$$= \frac{q}{8\pi^{2} \varepsilon_{0} c^{2}} \frac{1}{R} |\Delta v \sin \Theta|$$
(68)

单位频率间隔辐射能量角分布为

$$\frac{dW_{\omega}}{d\Omega} = 4\pi\varepsilon_0 cR^2 |E_{\omega}|^2 = \frac{q^2}{16\pi^3 \varepsilon_0 c^3} |\Delta v|^2 \sin^2 \Theta$$
 (69)

单位频率间隔辐射能量

$$W_{\omega} = \int \frac{dW_{\omega}}{d\Omega} d\Omega = \frac{q^2}{6\pi^2 \varepsilon_0 c^3} |\Delta v|^2$$
 (70)

#### 与ω无关!

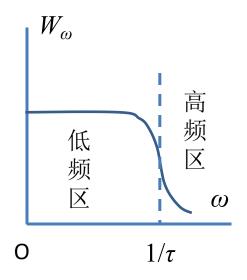
# 辐射能量按波长的分布

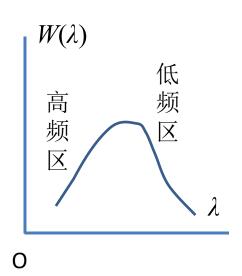
对(1):  $\omega \tau <<1$ ,有  $c\tau/\lambda <<1$ .

$$W(\lambda) = W_{\omega} \left| \frac{d\omega}{d\lambda} \right| = \frac{q^2}{3\pi\varepsilon_0} \left| \frac{\Delta v}{c} \right|^2 \frac{1}{\lambda^2}$$
 (71)

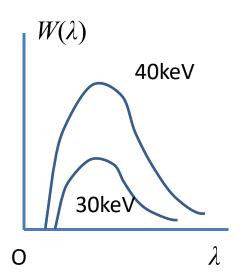
对(2):  $\omega \tau >> 1$ ,有  $c\tau/\lambda >> 1$ 

$$W(\lambda) = 0 \tag{72}$$





### 实验



短波区的截止波长是量子效应

$$E_e = \frac{1}{2}mv_1^2 = \hbar\omega_c \tag{73}$$

记光子数分布为 $N(\omega)$ 。对  $\omega\tau <<1$ ,

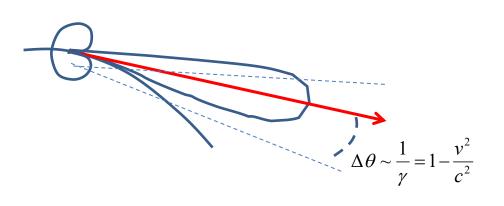
$$N(\omega)d\omega = \frac{W_{\omega}}{\hbar\omega}d\omega = \frac{2\alpha}{3\pi} \left(\frac{\Delta v}{c}\right)^2 \frac{d\omega}{\omega},\tag{74}$$

$$\alpha = \frac{e^2}{4\pi\varepsilon_0 \hbar c} = \frac{1}{137.0359895(61)} \tag{75}$$

## 3. 高速圆周运动带电粒子的辐射频谱

加速度沿垂直速度方向

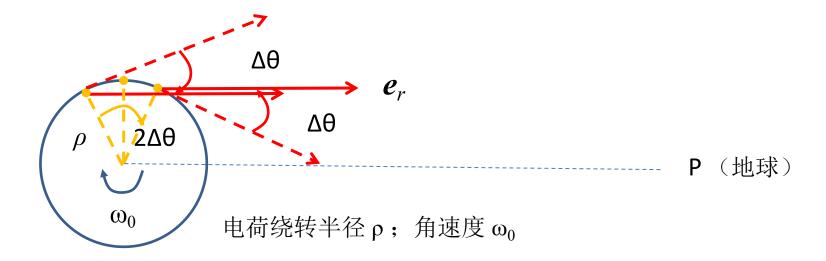
$$\frac{dP(t')}{d\Omega} = \frac{q^2 \dot{v}^2}{16\pi^2 \varepsilon_0 c^3} \frac{(1 - \frac{v}{c}\cos\theta)^2 - (1 - \frac{v^2}{c^2})\sin^2\theta\cos^2\phi}{(1 - \frac{v}{c}\cos\theta)^5}$$
(76)



$$\frac{1}{\left(1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r\right)^5} \approx \frac{32}{\left(\gamma^{-2} + \theta^2\right)^5} \tag{77}$$

高速圆周运动的带电粒子产生的辐射像一个旋转的探照灯一样

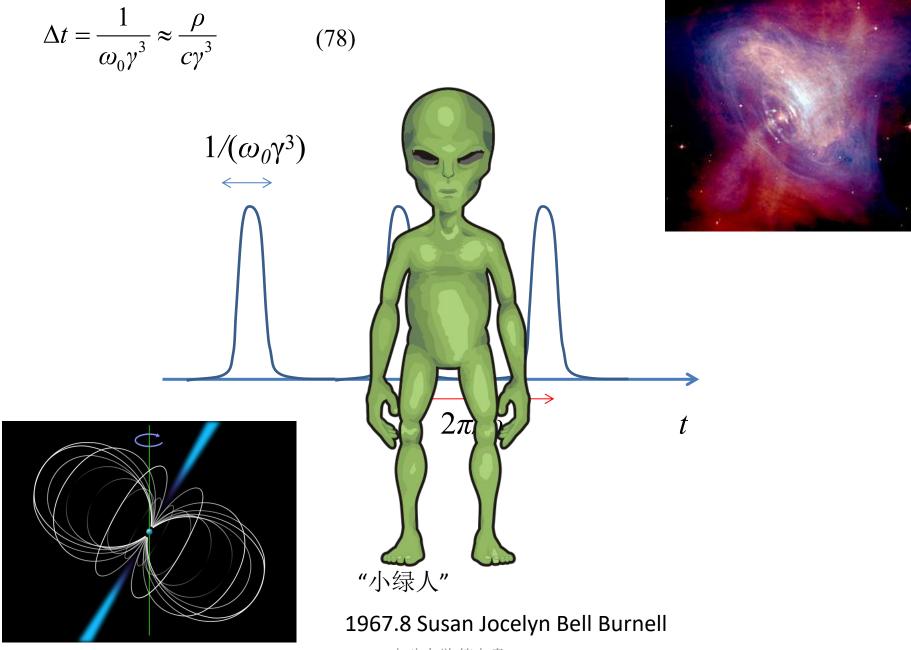
### 中子星



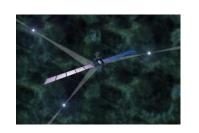
脉冲星转过角度 $\Delta\theta$ 的时间  $\Delta t'=2\Delta\theta/\omega_0$  收到脉冲星辐射的持续时间dt=(dt/dt') dt',  $\frac{dt}{dt'}=1-\frac{1}{c}\boldsymbol{v}\cdot\boldsymbol{e}_r$ 

$$\Delta t = \int_0^{\Delta t'} \frac{dt}{dt'} dt' = \int_0^{\Delta t'} \left( 1 - \frac{v}{c} \cos(\Delta \theta - \omega_0 t') \right) dt'$$

$$= \Delta t' - \frac{2v}{c\omega_0} \sin \frac{\omega_0 \Delta t'}{2} \approx \left( 1 - \frac{v}{c} \right) \Delta t' \approx \frac{\Delta t'}{2\gamma^2} = \frac{\Delta \theta}{\omega_0 \gamma^2} = \frac{1}{\omega_0 \gamma^3} \approx \frac{\rho}{c\gamma^3}$$



中国于2016年11月发射首颗脉冲星 导航试验卫星(XPNAV-1)



"中国天眼"500米口径球面射电望 远镜(FAST)





上海光源是一台高性能的中能第三代同步辐射光源,它的英文全名为Shanghai Synchrotron Radiation facility,简称SSRF。

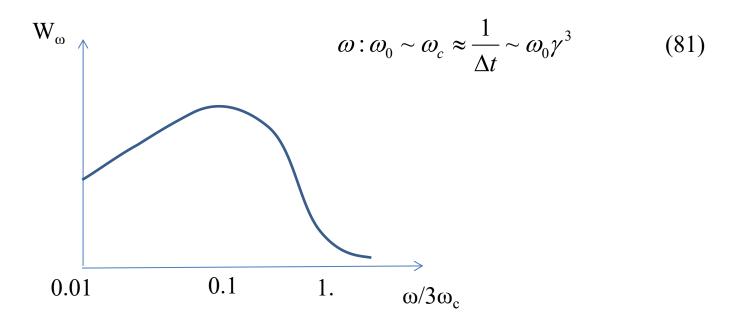
单位立体角内辐射能量

$$\frac{dW}{d\Omega} = \varepsilon_0 cR^2 \int_{-\infty}^{\infty} E^2 dt = 4\pi \varepsilon_0 cR^2 \int_0^{\infty} \left| E_{\omega} \right|^2 d\omega \equiv \int_0^{\infty} \frac{dW_{\omega}}{d\Omega} d\omega \tag{79}$$

单位频率间隔辐射能量

$$W_{\omega} = 4\pi\varepsilon_0 cR^2 \oint |\boldsymbol{E}_{\omega}|^2 d\Omega \tag{80}$$

同步辐射:  $E_e = 100$  MeV, 轨道半径  $\rho = 0.4$  m,  $\omega_0 \sim 8 \times 10^8$  s<sup>-1</sup>,  $\gamma \sim 200$ 



电动力学第七章

# 作业

$$\int_{-\infty}^{\infty} f^{2}(t)dt = 4\pi \int_{0}^{\infty} \left| f_{\omega} \right|^{2} d\omega$$

2. 从

$$\frac{dP(t')}{d\Omega} = \frac{q^2 \dot{v}^2}{16\pi^2 \varepsilon_0 c^3} \frac{(1 - \frac{v}{c} \cos \theta)^2 - (1 - \frac{v^2}{c^2}) \sin^2 \theta \cos^2 \phi}{(1 - \frac{v}{c} \cos \theta)^5}$$

推导*P(t')*.

3. p272第2题

### 附录1 推导偏导数关系



对给定的x和t,  $t_q$  唯一,因此存在函数关系 $t_q = t_q(x,t)$ . 对  $r = x - x_q$  求t的偏导数,

$$\frac{\partial \mathbf{r}}{\partial t}\Big|_{x} = -\frac{\partial \mathbf{x}_{q}(t_{q})}{\partial t}\Big|_{x} = -\frac{\partial t_{q}}{\partial t}\Big|_{x} \frac{d\mathbf{x}_{q}}{dt_{q}} = -\frac{\partial t_{q}}{\partial t}\Big|_{x} \mathbf{v} \tag{$\beta$ 1)}$$

对  $r = \sqrt{(\mathbf{r} \cdot \mathbf{r})}$  求t的偏导数,利用(附1),

$$\left. \frac{\partial r}{\partial t} \right|_{x} = \frac{\mathbf{r}}{r} \cdot \frac{\partial \mathbf{r}}{\partial t} \right|_{x} = -\frac{\mathbf{r} \cdot \mathbf{v}}{r} \frac{\partial t_{q}}{\partial t} \right|_{x} \tag{512}$$

此外,由 $r = c(t-t_q)$ ,

$$\left. \frac{\partial r}{\partial t} \right|_{x} = c \left( 1 - \frac{\partial t_q}{\partial t} \right|_{x} \right) \tag{$\beta$}$$

结合(附2)和(附3),

$$\frac{\partial t_q}{\partial t} = \frac{r}{r - \frac{1}{c} \mathbf{v} \cdot \mathbf{r}} = \frac{r}{s} \tag{5.4}$$

代回(附1)和(附2),

$$\frac{\partial \mathbf{r}}{\partial t}\bigg|_{\mathbf{r}} = -\frac{r}{s}\mathbf{v}, \qquad \frac{\partial r}{\partial t}\bigg|_{\mathbf{x}} = -\frac{\mathbf{r}\cdot\mathbf{v}}{s} \tag{5.5}$$



对  $r = x - x_q$  求xi的偏导数,考虑到 $x_q$ 是 $t_q$ 的函数,

$$\frac{\partial \mathbf{r}}{\partial x^{j}}\Big|_{t} = \hat{\mathbf{e}}_{j} - \frac{\partial \mathbf{x}_{q}}{\partial x^{j}}\Big|_{t} = \hat{\mathbf{e}}_{j} - \frac{\partial t_{q}}{\partial x^{j}}\Big|_{t} \dot{\mathbf{x}}_{q} = \hat{\mathbf{e}}_{j} - \frac{\partial t_{q}}{\partial x^{j}}\Big|_{t} \mathbf{v}$$
(\text{\text{\text{\text{\text{\text{\text{\$}}}}}}

对  $r = \sqrt{(r \cdot r)}$  求xi的偏导数,利用(附6),

$$\frac{\partial r}{\partial x^{j}}\Big|_{t} = \frac{\mathbf{r}}{r} \cdot \frac{\partial \mathbf{r}}{\partial x^{j}}\Big|_{t} = \frac{r^{j}}{r} - \frac{\partial t_{q}}{\partial x^{j}}\Big|_{t} \frac{\mathbf{r} \cdot \mathbf{v}}{r} \tag{5.17}$$

此外,由  $r = c(t-t_q)$ ,

$$\left. \frac{\partial r}{\partial x^{j}} \right|_{t} = -c \frac{\partial t_{q}}{\partial x^{j}}$$
 (\$\text{F\$\frac{1}{2}} \text{8})

结合(附7)和(附8),

$$\left. \frac{\partial t_q}{\partial x^j} \right|_{t} = -\frac{r^j}{c(r - \frac{1}{c} \mathbf{v} \cdot \mathbf{r})} = -\frac{r^j}{cs} \tag{5.9}$$

代回(附6)和(附8),

$$\frac{\partial \mathbf{r}}{\partial x^{j}} = \hat{\mathbf{e}}_{j} + \frac{r^{j}}{cs} \mathbf{v}, \qquad \frac{\partial r}{\partial x^{j}} = \frac{r^{j}}{s}$$
(1910)

#### 附录2 电场强度



$$\boldsymbol{E} = -\nabla \varphi - \frac{\partial \boldsymbol{A}}{\partial t} \tag{\mathref{M}} 11)$$

以李纳-维谢尔势代入

$$E = -\nabla \varphi - \frac{\partial A}{\partial t} = -\nabla \frac{q}{4\pi\varepsilon_0 (r - \frac{1}{c}\mathbf{v} \cdot \mathbf{r})} - \frac{\partial}{\partial t} \frac{q\mathbf{v}}{4\pi\varepsilon_0 c^2 (r - \frac{1}{c}\mathbf{v} \cdot \mathbf{r})}$$

$$= \frac{q}{4\pi\varepsilon_0 (r - \frac{1}{c}\mathbf{v} \cdot \mathbf{r})^2} \left[ \nabla r - \frac{1}{c}\nabla (\mathbf{v} \cdot \mathbf{r}) + \frac{\mathbf{v}}{c^2} \left( \partial_t r - \frac{1}{c}\partial_t (\mathbf{v} \cdot \mathbf{r}) \right) - (r - \frac{1}{c}\mathbf{v} \cdot \mathbf{r}) \frac{\partial_t \mathbf{v}}{c^2} \right]$$
(Fright 12)

应用(附4,5)式和(附9,10)式,

$$\nabla r = \frac{\hat{\boldsymbol{e}}_r}{1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r}, \qquad \nabla (\boldsymbol{v} \cdot \boldsymbol{r}) = \boldsymbol{v} + \frac{\frac{1}{c} \boldsymbol{v}^2 \hat{\boldsymbol{e}}_r}{1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r} - \frac{\frac{1}{c} (\hat{\boldsymbol{e}}_r \cdot \dot{\boldsymbol{v}}) \boldsymbol{r}}{1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r},$$

$$\partial_t r = -\frac{\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r}{1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r}, \qquad \partial_t \boldsymbol{v} = \frac{\dot{\boldsymbol{v}}}{1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r}$$

$$\partial_t (\boldsymbol{v} \cdot \boldsymbol{r}) = \frac{\boldsymbol{r} \cdot \dot{\boldsymbol{v}} - \boldsymbol{v}^2}{1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r}, \qquad \partial_t \boldsymbol{v} = \frac{\dot{\boldsymbol{v}}}{1 - \frac{1}{c} \boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r}$$

代回(附12)式,得到E为非辐射电场 $E_c$ 和辐射电场 $E_{rad}$ 之和

$$\boldsymbol{E}_{c} = \frac{q}{4\pi\varepsilon_{0}r^{2}} \frac{\hat{\boldsymbol{e}}_{r} - \frac{1}{c}\boldsymbol{v}}{(1 - \frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_{r})^{3}\gamma^{2}} \sim \frac{q}{4\pi\varepsilon_{0}r^{2}} \hat{\boldsymbol{e}}_{r}$$
 (\$\forall 14)

$$\boldsymbol{E}_{rad} = \frac{q}{4\pi\varepsilon_0 c^2 r} \frac{\hat{\boldsymbol{e}}_r \times [(\hat{\boldsymbol{e}}_r - \frac{1}{c}\boldsymbol{v}) \times \dot{\boldsymbol{v}}]}{(1 - \frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_r)^3} \sim \frac{q}{4\pi\varepsilon_0 c^2 r} \hat{\boldsymbol{e}}_r \times (\hat{\boldsymbol{e}}_r \times \dot{\boldsymbol{v}})$$
(\text{\text{15}})





$$\boldsymbol{B} = \nabla \times \boldsymbol{A}$$

(附16a)

在笛卡尔坐标中的分量形式(重复空间指标,j和k,隐含从1到3求和)

$$B^{i} = \varepsilon^{ijk} \partial_{i} A_{k} \tag{$M$16b}$$

以李纳-维谢尔势代入

$$B^{i} = \varepsilon^{ijk} \partial_{j} \frac{q v_{k}}{4\pi \varepsilon_{0} c^{2} s} = \frac{q \varepsilon^{ijk}}{4\pi \varepsilon_{0} c^{2} s} \partial_{j} v_{k} - \frac{q \varepsilon^{ijk} v_{k}}{4\pi \varepsilon_{0} c^{2} s^{2}} \partial_{j} s$$

$$= \frac{q \varepsilon^{ijk}}{4\pi \varepsilon_{0} c^{2} s} \dot{v}_{k} \frac{\partial t_{q}}{\partial x^{j}} - \frac{q \varepsilon^{ijk} v_{k}}{4\pi \varepsilon_{0} c^{2} s^{2}} \left[ \frac{\partial r}{\partial x^{j}} - \frac{r}{c} \cdot \frac{\partial v}{\partial x^{j}} - \frac{v}{c} \cdot \frac{\partial r}{\partial x^{j}} \right]$$

$$= \frac{q \varepsilon^{ijk}}{4\pi \varepsilon_{0} c^{2} s} \dot{v}_{k} \frac{\partial t_{q}}{\partial x^{j}} - \frac{q \varepsilon^{ijk} v_{k}}{4\pi \varepsilon_{0} c^{2} s^{2}} \left[ \frac{\partial r}{\partial x^{j}} - \frac{r}{c} \cdot \dot{v} \frac{\partial t_{q}}{\partial x^{j}} - \frac{v}{c} \cdot \frac{\partial r}{\partial x^{j}} \right]$$
()\text{\text{M17}}

其中  $s = r - \frac{1}{c} \mathbf{v} \cdot \mathbf{r}$ . 应用(附9、10)式,

$$\boldsymbol{B} = \frac{q\dot{\boldsymbol{v}} \times \boldsymbol{r}}{4\pi\varepsilon_0 c^3 s^2} - \frac{q}{4\pi\varepsilon_0 c^2 s^2} \left[ \frac{\boldsymbol{r} \times \boldsymbol{v}}{s} + \frac{(\dot{\boldsymbol{v}} \cdot \boldsymbol{r})(\boldsymbol{r} \times \boldsymbol{v})}{c^2 s} - \frac{v^2 (\boldsymbol{r} \times \boldsymbol{v})}{c^2 s} \right]$$

$$= \frac{q}{4\pi\varepsilon_0 c^2 r^2} \frac{(1 - \frac{v^2}{c^2}) \mathbf{v} \times \hat{\mathbf{e}}_r}{(1 - \frac{1}{c} \mathbf{v} \cdot \hat{\mathbf{e}}_r)^3} + \frac{q}{4\pi\varepsilon_0 c^3 r} \frac{\dot{\mathbf{v}} \times \hat{\mathbf{e}}_r + \frac{1}{c} [(\dot{\mathbf{v}} \cdot \hat{\mathbf{e}}_r)(\mathbf{v} \times \hat{\mathbf{e}}_r) - (\mathbf{v} \cdot \hat{\mathbf{e}}_r)(\dot{\mathbf{v}} \times \hat{\mathbf{e}}_r)]}{(1 - \frac{1}{c} \mathbf{v} \cdot \hat{\mathbf{e}}_r)^3}$$
 (\text{\text{\text{18}}})



B的第一项是非辐射项

$$\boldsymbol{B}_{c} = \frac{q}{4\pi\varepsilon_{0}c^{2}r^{2}} \frac{(1-\frac{v^{2}}{c^{2}})\boldsymbol{v} \times \hat{\boldsymbol{e}}_{r}}{(1-\frac{1}{c}\boldsymbol{v} \cdot \hat{\boldsymbol{e}}_{r})^{3}}$$
 (\$\mathref{19}\$)

第二项是辐射项

$$\mathbf{B}_{rad} = \frac{q}{4\pi\varepsilon_{0}c^{3}r} \frac{\dot{\mathbf{v}} \times \hat{\mathbf{e}}_{r} + \frac{1}{c}[(\dot{\mathbf{v}} \cdot \hat{\mathbf{e}}_{r})(\mathbf{v} \times \hat{\mathbf{e}}_{r}) - (\mathbf{v} \cdot \hat{\mathbf{e}}_{r})(\dot{\mathbf{v}} \times \hat{\mathbf{e}}_{r})]}{(1 - \frac{1}{c}\mathbf{v} \cdot \hat{\mathbf{e}}_{r})^{3}}$$

$$= \frac{q}{4\pi\varepsilon_{0}c^{3}r} \frac{\hat{\mathbf{e}}_{r} \times \hat{\mathbf{e}}_{r} \times (\hat{\mathbf{e}}_{r} \times \dot{\mathbf{v}}) + \frac{1}{c}[(\dot{\mathbf{v}} \cdot \hat{\mathbf{e}}_{r})\mathbf{v} - (\mathbf{v} \cdot \hat{\mathbf{e}}_{r})\dot{\mathbf{v}}] \times \hat{\mathbf{e}}_{r}}{(1 - \frac{1}{c}\mathbf{v} \cdot \hat{\mathbf{e}}_{r})^{3}}$$

$$= \frac{q}{4\pi\varepsilon_{0}c^{3}r} \frac{\hat{\mathbf{e}}_{r} \times \hat{\mathbf{e}}_{r} \times (\hat{\mathbf{e}}_{r} \times \dot{\mathbf{v}}) - \frac{1}{c}\hat{\mathbf{e}}_{r} \times [\hat{\mathbf{e}}_{r} \times (\mathbf{v} \times \dot{\mathbf{v}})]}{(1 - \frac{1}{c}\mathbf{v} \cdot \hat{\mathbf{e}}_{r})^{3}}$$

$$(\text{M}20)$$

与(附15)比较,

$$\boldsymbol{B}_{rad} = \frac{1}{c} \hat{\boldsymbol{e}}_r \times \boldsymbol{E}_{rad} \tag{5.1}$$

低速, v/c << 1,

$$\boldsymbol{B}_{c} \sim \frac{q}{4\pi\varepsilon_{0}c^{2}r^{2}}\boldsymbol{v} \times \hat{\boldsymbol{e}}_{r}, \qquad \boldsymbol{B}_{rad} \sim \frac{q}{4\pi\varepsilon_{0}c^{3}r}\dot{\boldsymbol{v}} \times \hat{\boldsymbol{e}}_{r} \qquad \qquad \frac{|B_{c}|}{|B_{rad}|} \sim \frac{\lambda}{r}$$
([\frac{\text{\text{122}}}{2}]