











SNOSD82A – JUNE 2018 – REVISED OCTOBER 2018

**TMP117** 

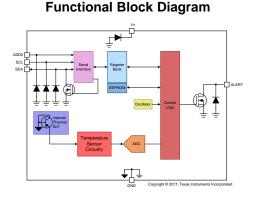
# TMP117x High-Accuracy, Low-Power, Digital Temperature Sensor With SMBus™- and I<sup>2</sup>C-Compatible Interface

#### 1 Features

- TMP117 High Accuracy Temperature Sensor
  - ±0.1°C (Maximum) From –20°C to +50°C
  - ±0.15°C (Maximum) From –40°C to +70°C
  - ±0.2°C (Maximum) From -40°C to +100°C
  - ±0.25°C (Maximum) From -55°C to +125°C
  - ±0.3°C (Maximum) From –55°C to +150°C
- Operating Temperature Range: -55°C to +150°C
- Low Power Consumption:
  - 3.5-μA, 1-Hz Conversion Cycle
  - 150-nA Shutdown Current
- Supply Range: 1.8 V to 5.5 V
- 16-Bit Resolution: 0.0078°C (1 LSB)
- Programmable Temperature Alert Limits
- Selectable Averaging
- Digital Offset for System Correction
- General-Purpose EEPROM: 48 Bits
- NIST Traceability
- SMBus<sup>™</sup>, I<sup>2</sup>C Interface Compatibility

### 2 Applications

- Medical Grade: Meets ASTM E1112 and ISO 80601-2-56
- Environmental Monitoring and Thermostats
- Wearables
- · Asset Tracking and Cold Chain
- Gas Meters and Heat Meters
- Test and Measurement
- RTDs Replacement: PT100, PT500, PT1000
- Cold-Junction Compensation of Thermocouples



### 3 Description

The TMP117 is a high-precision digital temperature sensor. It is designed to meet ASTM E1112 and ISO 80601 requirements for electronic patient thermometers. The TMP117 provides a 16-bit temperature result with a resolution of 0.0078°C and an accuracy of up to ±0.1°C across the temperature range of -20°C to 50°C with no calibration. The TMP117 has in interface that is I²C- and SMBus™-compatible, programmable alert functionality, and the device can support up to four devices on a single bus. Integrated EEPROM is included for device programming with an additional 48-bits memory available for general use.

The low power consumption of the TMP117 minimizes the impact of self-heating on measurement accuracy. The TMP117 operates from 1.8 V to 5.5 V and typically consumes 3.5  $\mu$ A.

For non-medical applications, the TMP117 can serve as a single chip digital alternative to a Platinum RTD. The TMP117 has an accuracy comparable to a Class AA RTD, while only using a fraction of the power of the power typically needed for a PT100 RTD. The TMP117 simplifies the design effort by removing many of the complexities of RTDs such as precision references, matched traces, complicated algorithms, and calibration.

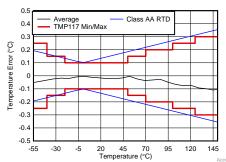
The TMP117 units are 100% tested on a production setup that is NIST traceable and verified with equipment that is calibrated to ISO/IEC 17025 accredited standards.

### **Device Information**(1)

PART NUMBER	PACKAGE	BODY SIZE (NOM)
TMP117	WSON (6)	2.00 mm × 2.00 mm

(1) For all available packages, see the package option addendum at the end of the data sheet.

### **Temperature Accuracy**





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### 4 Revision History

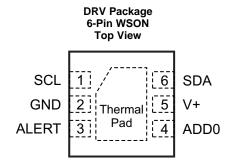
NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Original (June 2018) to Revision A			
•	Changed device status from Advanced Information to Production Data	1	
•	Changed shutdown current from: 250 nA to: 150 nA	1	

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## 5 Pin Configuration and Functions



### **Pin Functions**

PIN		TVDE	DESCRIPTION
NAME	WSON	TYPE	DESCRIPTION
ADD0	4	I	Address select. Connect to GND, V+, SDA, or SCL.
ALERT	3	0	Over temperature alert or data-ready signal. Open-drain output; requires a pullup resistor.
GND	2	_	Ground
SCL	1	I	Serial clock
SDA	6	I/O	Serial data input and open drain output; requires a pullup resistor.
V+	5	I	Supply voltage



### 6 Specifications

### 6.1 Absolute Maximum Ratings

Over free-air temperature range unless otherwise noted<sup>(1)</sup>

	· · · · · · · · · · · · · · · · · · ·			
		MIN	MAX	UNIT
Supply voltage	V+	-0.3	6	V
Voltage at	SCL, SDA, ALERT and ADD0	-0.3	6	V
Operating junction to	emperature, T <sub>J</sub>	-55	<b>-55</b> 150	
Storage temperature, T <sub>stg</sub>		-65	150	°C

<sup>(1)</sup> Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 6.2 ESD Ratings

			VALUE	UNIT
		Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±2000	
V <sub>(ESD)</sub>	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 (2)	±1000	V

<sup>(1)</sup> JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

### 6.3 Recommended Operating Conditions

		MIN	NOM	MAX	UNIT
V+	Supply voltage	1.8	3.3	5.5	V
V <sub>I/O</sub>	SCL, SDA, ALERT and ADD0	0		5.5	V
T <sub>A</sub>	Operating free-air temperature	-55		150	°C

### 6.4 Thermal Information

		TMP117	
	THERMAL METRIC <sup>(1)</sup>	DRV (WSON)	UNIT
		6 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	70.7	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	82.3	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	11.7	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	35.4	°C/W
ΨЈТ	Junction-to-top characterization parameter	2.2	°C/W
ΨЈВ	Junction-to-board characterization parameter	35.3	°C/W
$M_{T}$	Thermal Mass	5.1	mJ/°C

(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.

<sup>(2)</sup> JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



#### 6.5 Electrical Characteristics

Over free-air temperature range and V+ = 1.8 V to 5.5 V (unless otherwise noted); Typical specifications are at T<sub>A</sub> = 25°C and V+ = 3.3 V (unless otherwise noted)

	PARAMETE	R	TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT
EMPE	RATURE TO DIGI	TAL CONVER	TER					
			-20 °C to 50 °C		-0.1	±0.05	0.1	
			-40 °C to 70°C		-0.15	±0.05	0.15	
		TMP117	-40 °C to 100°C	8 averages	-0.2	±0.1	0.2	
			-55 °C to 125°C	1-Hz conversion cycle	-0.25	±0.1	0.25	
	Temperature		-55°C to 150°C	Thermal Pad unsoldered	-0.3	±0.1	0.3	°C
	accuracy	TMP117M	30°C to 45°C	(DRV Package)     I²C Input voltages: V <sub>IL</sub> ≤	-0.1	±0.05	0.1	٠.
		TIVIPTITIVI	0°C to 85°C	0.05 * V+, V <sub>IH</sub> ≥ 0.95 *	-0.2	±0.1	0.2	
			-40 °C to 100°C	V+	-0.2	±0.1	0.2	
		TMP117N	-55 °C to 125°C		-0.25	±0.1	0.25	
		-55°C to 150°C	-0.3	±0.1	0.3			
	DC power suppl	y sensitivity	One-shot mode, 8 Avera	ges		6		m°C/V
	Temperature res	solution (LSB)				7.8125		m°C
	Repeatability <sup>(1)</sup>		V+ = 3.3 V 8 averages 1-Hz conversion cycle			±1		LSB
	Long-term stabil	ity and drift	300 hours at 150°C <sup>(2)</sup>			±0.03		°C
	Temperature cychysteresis (3)	cling and	8 Averages			±2		LSB
	Conversion time	1	One-shot mode		13	15.5	17.5	ms
IGITA	L INPUT/OUTPUT	•					,	
	Input capacitano	e				4		рF
ін	Input logic high	level	SCL, SDA		0.7 * (V+)			V
IL	Input logic low le	evel	SCL, SDA				0.3 * (V+)	V
N	Input leakage cu	ırrent			-0.1		0.1	μΑ
OL.	SDA and ALER logic low level	Γ output	I <sub>OL</sub> = -3 mA		0		0.4	V

Repeatability is the ability to reproduce a reading when the measured temperature is applied consecutively, under the same conditions. Long term stability is determined using accelerated operational life testing at a junction temperature of 150°C.

Hysteresis is defined as the ability to reproduce a temperature reading as the temperature varies from room  $\rightarrow$  hot  $\rightarrow$ room $\rightarrow$ cold $\rightarrow$ room. The temperatures used for this test are -40°C, 25°C, and 150°C.



### **Electrical Characteristics (continued)**

Over free-air temperature range and V+ = 1.8 V to 5.5 V (unless otherwise noted); Typical specifications are at  $T_A$  = 25°C and V+ = 3.3 V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
POWER	SUPPLY					
I <sub>Q_ACTIV</sub> E	Quiescent current during active conversion	Active Conversion, serial bus inactive		135	220	μΑ
		Duty cycle 1 Hz, averaging mode off, serial bus inactive. $T_A = 25  ^{\circ}\text{C}$		3.5	5	
I <sub>Q</sub>	Quiescent current	Duty cycle 1 Hz, 8 avgerages mode, serial bus inactive. $T_A = 25  ^{\circ}\text{C}$		16	22	μΑ
		Duty cycle 1 Hz, averaging mode off, serial bus active, SCL frequency = 400 kHz		15		
I <sub>SB</sub>	Standby current <sup>(4)</sup>	Serial bus inactive. SCL, SDA, and ADD0 = V+. T <sub>A</sub> = 25 °C		1.25	3.1	μΑ
	Shutdown current	Serial bus inactive, SCL, SDA, and ADD0 = V+. T <sub>A</sub> = 25 °C		0.15	0.5	μΑ
I <sub>SD</sub>	Shutdown current	Serial bus inactive, SCL, SDA and ADD0 = V+, T <sub>A</sub> = 150°C			5	μА
	Shutdown current	Serial bus active, SCL frequency = 400 kHz, ADD0 = V+		17		μΑ
I <sub>EE</sub>	EEPROM write quiescent current	ADC conversion off; serial bus inactive		240		μА
V <sub>POR</sub>	Power-on-reset threshold voltage	Supply rising		1.6		V
	Brownout detect	Supply falling		1.1		V
t <sub>RESET</sub>	Reset Time	Time required by device to reset		1.5		ms

<sup>(4)</sup> Quiescent current between conversions

### 6.6 Switching Characteristics

Over free-air tempearture range and V+ = 1.8 V to 5.5 V (unless otherwise noted); Typical specifications are at  $T_A$  = 25°C and V+ = 3.3 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
EEPROM					
Programming time			7		ms
Number of writes		1,000	50,000		Times
Data retention time		10	100		Years



### 6.7 Two-Wire Interface Timing

Over free-air tempearture range and V+ = 1.8 V to 5.5 V (unless otherwise noted)

		FAST-MODE	FAST-MODE	
		MIN	MAX	UNIT
f <sub>SCL</sub>	SCL operating frequency	1	400	KHz
t <sub>BUF</sub>	Bus free time between STOP and START conditions	1300		ns
t <sub>HD;STA</sub>	Hold time after repeated STRAT condition.  After this period, the first clock is generated (1)	600		ns
t <sub>SU;STA</sub>	Repeated START condition setup time	600		ns
t <sub>SU;STO</sub>	STOP condition setup time	600		ns
t <sub>HD;DAT</sub>	Data hold time	0		ns
t <sub>VD;DAT</sub>	Data valid time <sup>(2)</sup>		0.9	μs
t <sub>SU;DAT</sub>	Data setup time	100		ns
t <sub>LOW</sub>	SCL clock low period	1300		ns
t <sub>HIGH</sub>	SCL clock high period	600		ns
t <sub>F</sub> – SDA	Data fall time	20 × (V+ /5.5)	300	ns
t <sub>F</sub> , t <sub>R</sub> – SCL	Clock fall and rise time		300	ns
t <sub>R</sub>	Rise time for SCL ≤ 100 kHz		1000	ns
	Serial bus timeout (SDA bus released if there is no clock)	20	40	ms

- The maximum  $t_{\text{HD;DAT}}$  could be 0.9  $\mu s$  for Fast-Mode, and is less than the maximum  $t_{\text{VD;DAT}}$  by a transition time.  $t_{\text{VD;DATA}}$  = time for data signal from SCL "LOW" to SDA output ("HIGH" to "LOW", depending on which is worse).

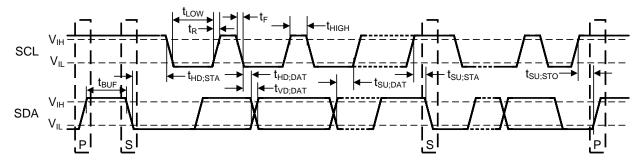
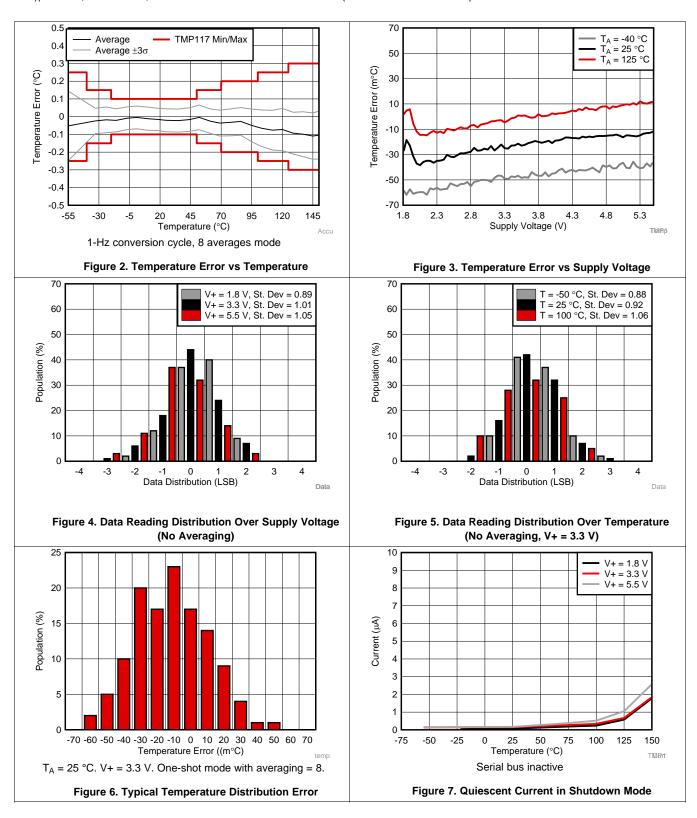


Figure 1. Two-Wire Timing Diagram



### 6.8 Typical Characteristics

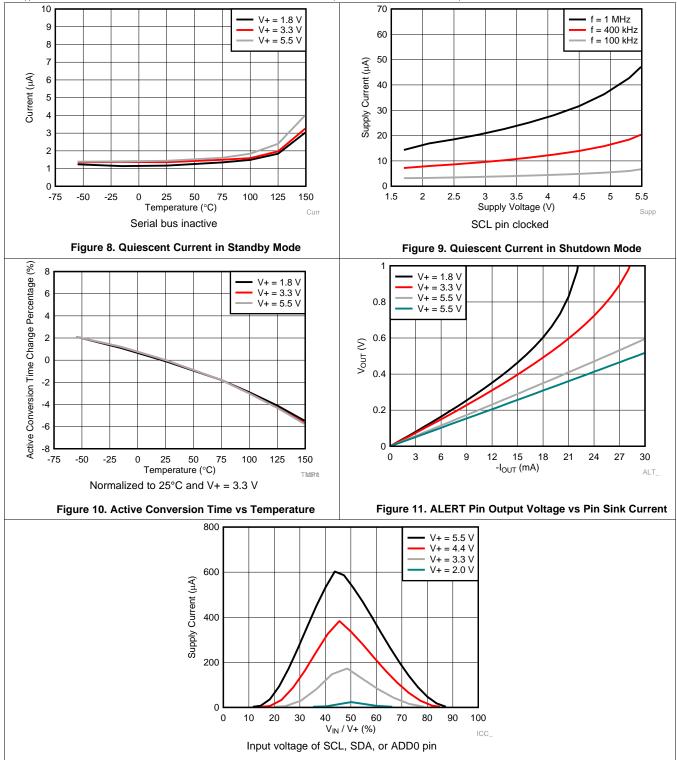
at T<sub>A</sub> = 25°C, V+ = 3.3 V, and measurement taken in oil bath (unless otherwise noted)





### **Typical Characteristics (continued)**

at  $T_A = 25$ °C, V+ = 3.3 V, and measurement taken in oil bath (unless otherwise noted)



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Figure 12. Supply Current vs Input Cell Voltage



### 7 Detailed Description

#### 7.1 Overview

The TMP117 is a digital output temperature sensor designed for thermal-management and thermal-protection applications. The TMP117 is two-wire, SMBus, and I<sup>2</sup>C interface-compatible. The device is specified over an ambient air operating temperature range of –55°C to +150°C. Figure 13 shows a block diagram of the TMP117.

### 7.2 Functional Block Diagrams

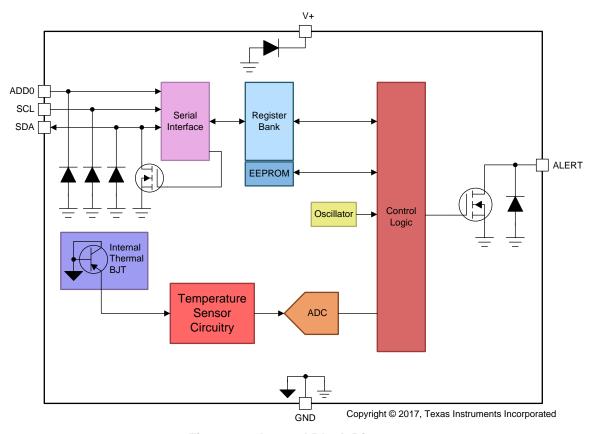


Figure 13. Internal Block Diagram

#### 7.3 Feature Description

#### 7.3.1 Power Up

After the supply voltage reaches within the operating range, the device requires 1.5 ms to power up before conversions can begin. The device can be programmed to start up in shutdown mode as well. See the *EEPROM Programming* section for more information. The temperature register reads –256°C before the first conversion.

#### 7.3.2 Averaging

Users can configure the device to report the average of multiple temperature conversions with the AVG[1:0] bits to reduce noise in the conversion results. When the TMP117 is configured to perform averaging with AVG set to 01, the device executes the configured number of conversions to eight. The device accumulates those conversion results and reports the average of all the collected results at the end of the process. As shown in the noise histograms of Figure 5 and Figure 6, the temperature result output has a repeatability of approximately ±3 LSBs when there is no averaging and ±1 LSB when the device is configured to perform eight averages.

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### **Feature Description (continued)**

Figure 14 shows the total conversion cycle time trade-off when using the averaging mode to achieve this improvement in noise performance. Averaging will increase the average active current consumption due to increasing the active conversion time in a conversion cycle. For example a single active conversion typically takes 15.5 ms, so if the device is configured to report an average of eight conversions, then the active conversion time is 124 ms (15.5 ms × 8). Use Equation 1 to factor in this increase in active conversion time to accurately calculate the average current consumption of the device. The average current consumption of the device can be decreased by increasing the amount of time the device spends in standby period as compared to active conversion. Under the factory EEPROM settings, the device is configured to report an average of eight conversions with a conversion cycle time of 1 second by default.

Averaging can be used in both the continuous conversion mode and the one-shot mode.

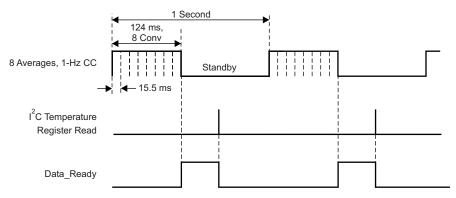


Figure 14. Averaging Timing Diagram

#### 7.3.3 Temperature Result and Limits

At the end of every conversion, the device updates the temperature register with the conversion result. The data in the result register is in two's complement format, has a data width of 16 bits and a resolution of 7.8125m°C. Table 1 shows multiple examples of possible binary data that can be read from the temperature result register and the corresponding hexadecimal and temperature equivalents.

The TMP117 also has alert status flags and alert pin functionality that use the temperature limits stored in the low limit register and high limit register. The same data format used for the temperature result register is used for data written to the high and low limit registers.

**TEMPERATURE REGISTER VALUE TEMPERATURE** (0.0078125°C RESOLUTION) (°C) **BINARY HEX** -2561000 0000 0000 0000 8000 -25 1111 0011 1000 0000 F380 -0.12501111 1111 1111 0000 FFF0 -0.00781251111 1111 1111 1111 **FFFF** 0000 0000 0000 0000 0000 0.0078125 0000 0000 0000 0001 0001 0.1250 0000 0000 0001 0000 0010 0000 0000 1000 0000 0800 25 0000 1100 1000 0000 0C80 100 0011 0010 0000 0000 3200 255.9921 0111 1111 1111 1111 7FFF

**Table 1. 16-Bit Temperature Data Format** 

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(1)



#### 7.4 Device Functional Modes

The TMP117 can be configured to operate in various conversion modes by using the MOD[1:0] bits. These modes provide flexibility to operate the device in the most power efficient way necessary for the intended application.

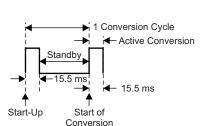
#### 7.4.1 Continuous Conversion Mode

When the MOD[1:0] bits are set to 00 or 10 in the configuration register, the device operates in continuous conversion mode. The device continuously performs temperature conversions in this mode, as shown in Figure 15, and updates the temperature result register at the end of every active conversion. The user can read the configuration register or the temperature result register to clear the Data\_Ready flag. Therefore, the Data\_Ready flag can be used to determine when the conversion completes so that an external controller can synchronize reading the result register with conversion result updates. The user can set the DR/nAlert\_EN bit in the configuration register to monitor the state of the Data\_Ready flag on the ALERT pin.

Every conversion cycle consists of an active conversion period followed by a standby period. The device typically consumes 135  $\mu$ A during active conversion and only 1.25  $\mu$ A during the low-power standby period. Figure 15 shows a current consumption profile of a conversion cycle while in continuous current mode. The duration of the active conversion period and standby period can be configured using the CONV[2:0] and AVG[1:0] bits in the configuration register, thereby allowing the average current consumption of the device to be optimized based on the application requirements. Changing the conversion cycle period also affects the temperature result update rate because the temperature result register is updated at the end of every active conversion.

Use Equation 1 to calculate the average current consumption of the device in continuous conversion mode.

(Active Current Consumption × Active Conversion Time) + (Standby Current Consumption × Standby Time)



Conversion Cycle Time

Figure 15. Conversion Cycle Timing Diagram

### 7.4.2 Shutdown Mode (SD)

When the MOD[1:0] bits are set to 01 in the configuration register, the device instantly aborts the currently running conversion and enters a low-power shutdown mode. In this mode, the device powers down all active circuitry and can be used in conjunction with the OS mode to perform temperature conversions. Engineers can use the TMP117 for battery-operated systems and other low-power consumption applications because the device typically only consumes 250 nA in SD mode.

Product Folder Links: TMP117



### **Device Functional Modes (continued)**

#### 7.4.3 One-Shot Mode (OS)

When MOD[1:0] bits are set to 11 in the configuration register, the TMP117 will run a temperature conversion referred to as a one-shot conversion. After the device completes a one-shot conversion, the device goes to the low-power shutdown mode. A one-shot conversion cycle, unlike the continuous conversion mode, only consists of the active conversion time and no standby period. Thus, the duration of a one-shot conversion is only affected by the AVG bit settings. The CONV bits do not affect the duration of a one-shot conversion. Figure 16 shows a timing diagram for this mode with an AVG setting of 00. At the end of a one-shot conversion, the Data\_Ready and ALERT flag in the configuration register is set. The Data\_Ready flag can be used to determine when the conversion completes. The user can perform an I<sup>2</sup>C read on the configuration register or temperature result register to clear the Data\_Ready flag. The user can also set the DR/nAlert\_EN bit in the configuration register to monitor the state of the Data\_Ready flag on the ALERT pin.

One-shot mode cannot be programmed to a default start-up mode. If the EEPROM is programmed to be in one-shot mode on start-up, it will default to shutdown mode instead.



Figure 16. One-Shot Timing Diagram With AVG[1:0] = 00

#### 7.4.4 Therm and Alert Modes

The built-in therm and alert functions of the TMP117 can alert the user if the temperature has crossed a certain temperature limit or if the device is within a certain temperature range. At the end of every conversion, including averaging, the TMP117 compares the converted temperature result to the values stored in the low limit register and high limit register. The device then either sets or clears the corresponding status flags in the configuration register, as described in this section.



### **Device Functional Modes (continued)**

#### 7.4.4.1 Alert Mode

When the T/nA bit in the configuration register is set to 0, the device is in alert mode. In this mode, the device compares the conversion result at the end of every conversion with the values in the low limit register and high limit register. If the temperature result exceeds the value in the high limit register, the HIGH\_Alert status flag in the configuration register is set. On the other hand, if the temperature result is lower than the value in the low limit register, the LOW\_Alert status flag in the configuration register is set. As shown in Figure 17, the user can run an I<sup>2</sup>C read from the configuration register to clear the status flags in alert mode.

When a user configures the device in alert mode, it affects the behavior of the ALERT pin. The device asserts the ALERT pin in this mode when either the HIGH\_Alert or the LOW\_Alert status flag is set, as shown in Figure 17. The user can either run an I<sup>2</sup>C read of the configuration register (which also clears the status flags) or run an SMBus alert response command (see the *SMBus Alert Function* section) to deassert the ALERT pin. The polarity of the ALERT pin can be changed by using the POL bit setting in the configuration register.

This mode effectively makes the device behave like a window limit detector. Thus this mode can be used in applications where detecting if the temperature goes outside of the specified range is necessary.

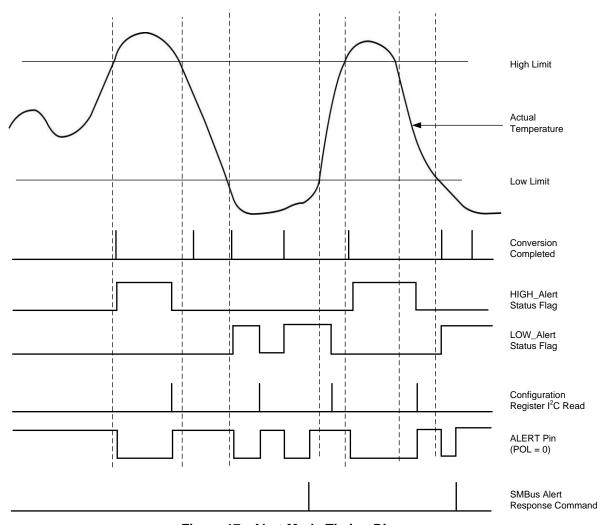


Figure 17. Alert Mode Timing Diagram

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### **Device Functional Modes (continued)**

#### 7.4.4.2 Therm Mode

When the T/nA bit in the configuration register is set to 1 the device is in therm mode. In this mode, the device compares the conversion result at the end of every conversion with the values in the low limit register and high limit register and sets the HIGH\_Alert status flag in the configuration register if the temperature exceeds the value in the high limit register. When set, the device clears the HIGH Alert status flag if the conversion result goes below the value in the low limit register. Thus, the difference between the high and low limits effectively acts like a hysteresis. In this mode, the LOW\_Alert status flag is disabled and always reads 0. Unlike the alert mode, I<sup>2</sup>C reads of the configuration register do not affect the status bits. The HIGH\_Alert status flag is only set or cleared at the end of conversions based on the value of the temperature result compared to the high and low

As in alert mode, configuring the device in therm mode also affects the behavior of the ALERT pin. In this mode, the device asserts the ALERT pin if the HIGH\_Alert status flag is set and deasserts the ALERT pin when the HIGH\_Alert status flag is cleared. In therm mode, the ALERT pin cannot be cleared by performing an I<sup>2</sup>C read of the configuration register or by performing an SMBus alert response command. As in alert mode, the polarity of the active state of the ALERT pin can be changed if the user adjusts the POL bit setting in the configuration register.

Thus, this mode effectively makes the device behave like a high-limit threshold detector. This mode can be used in applications where detecting if the temperature has gone above a desired threshold is necessary. Figure 18 shows a timing diagram of this mode.

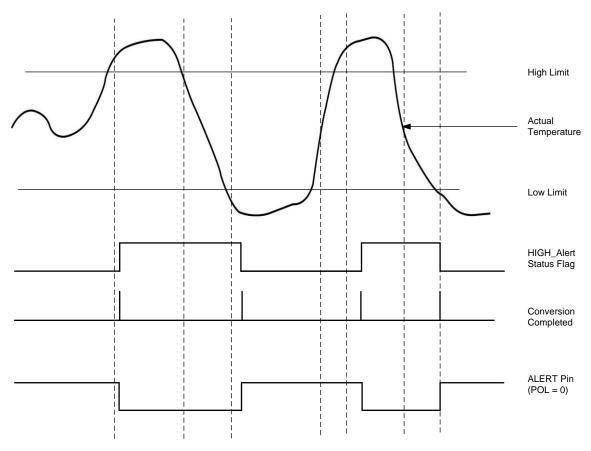


Figure 18. Therm Mode Timing Diagram



### 7.5 Programming

#### 7.5.1 EEPROM Programming

#### 7.5.1.1 EEPROM Overview

The device has a user-programmable EEPROM that can be used for two purposes:

- Storing power-on reset (POR) values of the high limit register, low limit register, conversion cycle time, averaging mode, conversion mode (continuous or shutdown mode), alert function mode (alert or therm mode), and alert polarity
- Storing four 16-bit locations for general-purpose use. See the EEPROM[4:1] registers for more information.

On reset, the device goes through a POR sequence that loads the values programmed in the EEPROM into the respective register map locations. This process takes approximately 1.5 ms. When the power-up sequence is complete, the device starts operating in accordance to the configuration parameters that are loaded from the EEPROM. Any I<sup>2</sup>C writes performed during this initial POR period to the limit registers or the configuration register are ignored. I<sup>2</sup>C read transactions can still be performed with the device during the power-up period. While the POR sequence is being executed, the EEPROM\_Busy status flag in the EEPROM unlock register is set.

During production, the EEPROM in the TMP117 is programmed with reset values as shown in Table 3. The *Programming the EEPROM* section describes how to change these values. A unique ID is also programmed in the general-purpose EEPROM locations during production. This unique ID is used to support NIST traceability. The TMP117 units are 100% tested on a production setup that is NIST traceable and verified with equipment that is calibrated to ISO/IEC 17025 accredited standards. Only reprogram the general-purpose EEPROM[4:1] locations if NIST traceability is not desired.

### 7.5.1.2 Programming the EEPROM

To prevent accidental programming, the EEPROM is locked by default. When locked, any I<sup>2</sup>C writes to the register map locations are performed only on the volatile registers and not on the EEPROM.

Figure 19 shows a flow chart describing the EEPROM programming sequence. To program the EEPROM, first unlock the EEPROM by setting the EUN bit in the EEPROM unlock register. After the EEPROM is unlocked, any subsequent I<sup>2</sup>C writes to the register map locations program a corresponding non-volatile memory location in the EEPROM. Programming a single location typically takes 7 ms to complete and consumes 230 μA. Do not perform any I<sup>2</sup>C writes until programming is complete. During programming, the EEPROM\_busy flag is set. Read this flag to monitor if the programming is complete. After programming the desired data, issue a general-call reset command to trigger a software reset. The programmed data from the EEPROM are then loaded to the corresponding register map locations as part of the reset sequence. This command also clears the EUN bit and automatically locks the EEPROM to prevent any further accidental programming. Avoid using the device to perform temperature conversions when the EEPROM is unlocked.

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### **Programming (continued)**

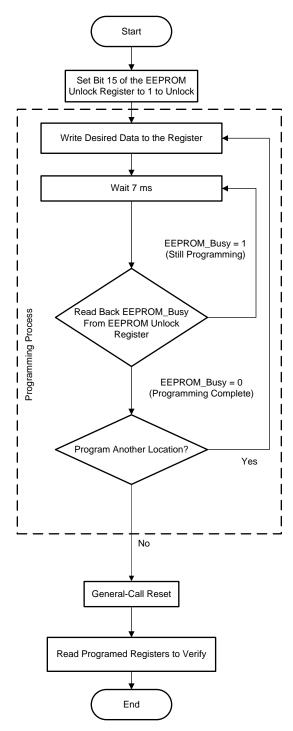


Figure 19. EEPROM Programming Sequence

Product Folder Links: TMP117



### **Programming (continued)**

#### 7.5.2 Pointer Register

Figure 20 shows the internal register structure of the TMP117. The 8-bit pointer register of the device is used to address a given data register. The reset value is 00.

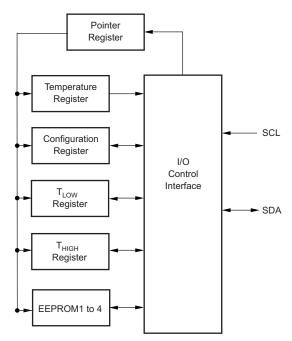


Figure 20. Internal Register Structures

#### 7.5.3 I<sup>2</sup>C and SMBus Interface

#### 7.5.3.1 Serial Interface

The TMP117 operates as a slave device only on the two-wire, SMBus and I<sup>2</sup>C interface-compatible bus. Connections to the bus are made through the open-drain I/O lines and the SDA and SCL pins. The SDA and SCL pins feature integrated spike-suppression filters and Schmitt triggers to minimize the effects of input spikes and bus noise. The device supports the transmission protocol for fast (1 kHz to 400 kHz) mode. Register bytes are sent with the most significant byte first, followed by the least significant byte.

#### 7.5.3.1.1 Bus Overview

The device that initiates the transfer is called a *master*, and the devices controlled by the master are *slaves*. The bus must be controlled by a master device that generates the serial clock (SCL), controls the bus access, and generates the START and STOP conditions.

To address a specific device, a START condition is initiated, indicated by pulling the data line (SDA) from a high-to low-logic level when the SCL pin is high. All slaves on the bus shift in the slave address byte on the rising edge of the clock, and the last bit indicates whether a read or write operation is intended. During the ninth clock pulse, the addressed slave generates an acknowledge and pulls the SDA pin low to respond to the master.

A data transfer is then initiated and sent over eight clock pulses followed by an acknowledge bit. During the data transfer, the SDA pin must remain stable when the SCL pin is high because any change in the SDA pin when the SCL pin is high is interpreted as a START or STOP signal.

When all data are transferred, the master generates a repeated START condition or a STOP condition.

#### 7.5.3.1.2 Serial Bus Address

To communicate with the TMP117, the master must first address slave devices through an address byte. The address byte has seven address bits and a read-write (R/W) bit that indicates the intent of executing a read or write operation.



### Programming (continued)

The TMP117 features an address pin to allow up to four devices to be addressed on a single bus. Table 2 describes the pin logic levels used to properly connect up to four devices. *x* represents the read-write (R/W) bit.

**Table 2. Address Pin and Slave Addresses** 

DEVICE TWO-WIRE ADDRESS	ADD0 PIN CONNECTION
1001000x	Ground
1001001x	V+
1001010x	SDA
1001011x	SCL

#### 7.5.3.1.3 Writing and Reading Operation

The user can write a register address to the pointer register to access a particular register on the TMP117. The value for the pointer register is the first byte transferred after the slave address byte with the R/W bit low. Every write operation to the TMP117 requires a value for the pointer register.

When reading from the TMP117, the last value stored in the pointer register by a write operation is used to determine which register is read during a read operation. To change the register pointer for a read operation, a new value must be written to the pointer register. The user can issue an address byte with the R/W bit low, followed by the pointer register byte to write a new value for the pointer register. No additional data is required. The master can then generate a START condition and send the slave address byte with the R/W bit high to initiate the read command. See Figure 22 for details of this sequence. If repeated reads from the same register are desired, it is not necessary to send the pointer register bytes continuously because the TMP117 retains the pointer register value until the value is changed by the next write operation.

Register bytes are sent with the most significant byte first, followed by the least significant byte.

#### 7.5.3.1.4 Slave Mode Operations

The TMP117 can operate as a slave receiver or slave transmitter. As a slave device, the TMP117 never drives the SCL line.

#### 7.5.3.1.4.1 Slave Receiver Mode

The first byte transmitted by the master is the slave address with the  $R/\overline{W}$  bit low. The TMP117 then acknowledges reception of a valid address. The next byte transmitted by the master is the pointer register. The TMP117 then acknowledges reception of the pointer register byte. The next byte(s) are written to the register addressed by the pointer register. The TMP117 acknowledges reception of each data byte. The master can terminate data transfer by generating a START or STOP condition.

#### 7.5.3.1.4.2 Slave Transmitter Mode

The first byte transmitted by the master is the slave address with the  $R/\overline{W}$  bit high. The slave acknowledges reception of a valid slave address. The next byte is transmitted by the slave and is the most significant byte of the register indicated by the pointer register. The master acknowledges reception of the data byte. The next byte transmitted by the slave is the least significant byte. The master acknowledges reception of the data byte. The master can terminate data transfer by generating a *not-acknowledge* on reception of any data byte or by generating a START or STOP condition.

#### 7.5.3.1.5 SMBus Alert Function

The TMP117 supports the SMBus alert function. When the ALERT pin is connected to an SMBus alert signal and a master senses that an alert condition is present, the master can send out an SMBus ALERT command (0001 1001) to the bus. If the ALERT pin is active, the device acknowledges the SMBus ALERT command and responds by returning the slave address on the SDA line. The eighth bit (LSB) of the slave address byte indicates if the alert condition is caused by the temperature exceeding  $T_{(HIGH)}$  or falling below  $T_{(LOW)}$ . The LSB is high if the temperature is greater than  $T_{(HIGH)}$ , or low if the temperature is less than  $T_{(LOW)}$ . See Figure 23 for details of this sequence.



If multiple devices on the bus respond to the SMBus ALERT command, arbitration during the slave address portion of the SMBus ALERT command determines which device clears the alert status of that device. The device with the lowest two-wire address wins the arbitration. If the TMP117 wins the arbitration, the TMP117 ALERT pin becomes inactive at the completion of the SMBus ALERT command. If the TMP117 loses the arbitration, the TMP117 ALERT pin remains active.

#### 7.5.3.1.6 General-Call Reset Function

The TMP117 responds to a two-wire, general-call address (0000 000) if the eighth bit is 0. The device acknowledges the general-call address and responds to commands in the second byte. If the second byte is 0000 0110, the TMP117 internal registers are reset to power-up values.

#### 7.5.3.1.7 Timeout Function

The TMP117 resets the serial interface if the SCL line is held low by the master or the SDA line is held low by the TMP117 for 35 ms (typical) between a START and STOP condition. The TMP117 releases the SDA line if the SCL pin is pulled low and waits for a START condition from the host controller. To avoid activating the timeout function, maintain a communication speed of at least 1 kHz for the SCL operating frequency.

### 7.5.3.1.8 Timing Diagrams

The TMP117 is two-wire, SMBus and  $I^2C$  interface-compatible. Figure 21 to Figure 24 show the various operations with the TMP117. Bus definitions are:

Bus Idle: Both SDA and SCL lines remain high.

**Start Data Transfer:** A change in the state of the SDA line from high to low when the SCL line is high defines a START condition. Each data transfer is initiated with a START condition.

**Stop Data Transfer:** A change in the state of the SDA line from low to high when the SCL line is high defines a STOP condition. Each data transfer is terminated with a repeated START or STOP condition.

**Data Transfer:** The number of data bytes transferred between a START and a STOP condition is not limited and is determined by the master device.

**Acknowledge:** Each receiving device, when addressed, is obliged to generate an acknowledge bit. A device that acknowledges must pull down the SDA line during the acknowledge clock pulse in such a way that the SDA line is stable low during the high period of the acknowledge clock pulse. The user must take setup and hold times into account. On a master receive, the termination of the data transfer can be signaled by the master generating a *not-acknowledge* (1) on the last byte transmitted by the slave.

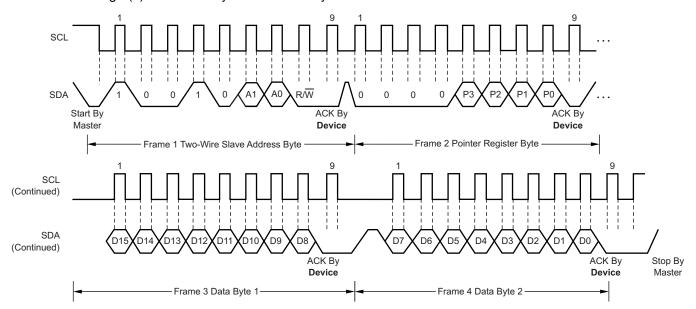


Figure 21. Write Word Command Timing Diagram

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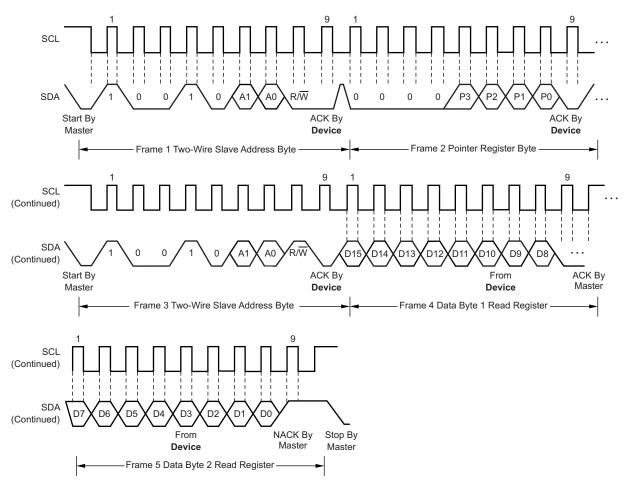


Figure 22. Read Word Command Timing Diagram

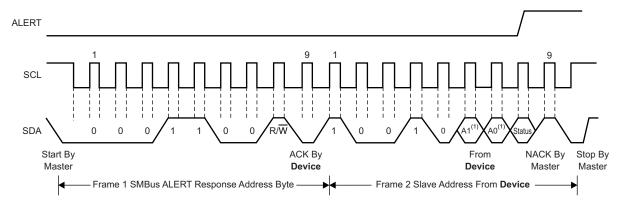


Figure 23. SMBus ALERT Timing Diagram

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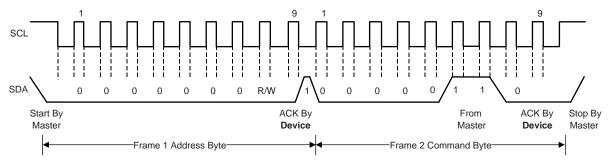


Figure 24. General-Call Reset Command Timing Diagram



### 7.6 Registers Map

### Table 3. Registers Map

ADDRESS	DEFAULT								REGISTI	ER DATA								
(HEX)	VALUE (HEX)	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	REGISTER NAME
00h	8000	T15	T14	T13	T12	T11	T10	T9	Т8	T7	T6	T5	T4	T3	T2	T1	T0	Temperature
01h	0220 <sup>(1)</sup>	HIGH_ Alert	LOW_ Alert	Data_ Ready	EEPROM _Busy	MOD1	MOD0	CONV2	CONV1	CONV0	AVG1	AVG0	T/nA	POL	DR/nAlert _EN	Soft_Rese t	_	Configuration
02h	6000 <sup>(1)</sup>	H15	H14	H13	H12	H11	H10	H9	H8	H7	H6	H5	H4	НЗ	H2	H1	H0	High Limit
03h	8000(1)	L15	L14	L13	L12	L11	L10	L9	L8	L7	L6	L5	L4	L3	L2	L1	L0	Low Limit
04h	0000	EUN	EEPROM _ Busy	_	_	_	_	_	_	_	_	_	_	_	_	_	_	EEPROM Unlock
05h	xxxx <sup>(1)</sup>	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	EEPROM1
06h	xxxx <sup>(1)</sup>	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	EEPROM2
07h	0000(1)	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	Temperature Offset
08h	xxxx <sup>(1)</sup>	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	EEPROM3
0Fh	x117	_	_	_	_	DID11	DID10	DID9	DID8	DID7	DID6	DID5	DID4	DID3	DID2	DID1	DID0	Device ID

<sup>(1)</sup> This value is stored in Electrically-Erasable, Programmable Read-Only Memory (EEPROM) during device manufacturing. The device reset value can be changed by writing the relevant code in the EEPROM cells (see the *EEPROM Overview* section).

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#### 7.6.1 Register Descriptions

Table 4. TMP117 Access Type Codes

ACCESS TYPE	CODE	DESCRIPTION
R	R	Read
Read-write	R/W	Read, write, or both
W	W	Write
-nx		Value after reset or the default value

### 7.6.1.1 Temperature Register (address = 00h) [default reset = 8000h]

This register is a 16-bit, read-only register that stores the output of the most recent conversion. One LSB equals 7.8125m°C. Data are represented in binary two's complement format. Following a reset, the temperature register reads –256°C until the first conversion, including averaging, is complete. See the *Power Up* section for more information.

Figure 25. Temperature Register

15	14	13	12	11	10	9	8
T15	T14	T13	T12	T11	T10	Т9	T8
R-1	R-0						
7	6	5	4	3	2	1	0
T7	T6	T5	T4	T3	T2	T1	T0
R-0							

**Table 5. Temperature Register Field Descriptions** 

BIT	FIELD	TYPE	RESET	DESCRIPTION
15:0	T[15:0]	R	8000h	16-bit, read-only register that stores the most recent temperature conversion results.



### 7.6.1.2 Configuration Register (address = 01h) [factory default reset = 0220h]

### Figure 26. Configuration Register

15	14	13	12	11	10	9	8
HIGH_Alert	LOW_Alert	Data_Ready	EEPROM_Busy	MOD1 (1)	MOD0 <sup>(2)</sup>	CONV2 <sup>(2)</sup>	CONV1 <sup>(2)</sup>
R-0	R-0	R-0	R-0	R/W-0	R/W-0	R/W-1	R/W-0
7	6	5	4	3	2	1	0
CONV0 <sup>(2)</sup>	AVG1 <sup>(2)</sup>	AVG0 <sup>(2)</sup>	T/nA <sup>(2)</sup>	POL <sup>(2)</sup>	DR/Alert <sup>(2)</sup>	Soft_Reset	_
R/W-0	R/W-0	R/W-1	R/W-0	R/W-0	R/W-0	R-0	R-0

<sup>(1)</sup> The MOD1 bit cannot be stored in EEPROM. The device can only be programmed to start up in shutdown mode or continuous conversion mode.

### **Table 6. Configuration Register Field Descriptions**

	Table 6. Configuration Register Field Descriptions											
BIT	FIELD	TYPE	RESET	DESCRIPTION								
15	HIGH_Alert	R	0	High Alert flag:  1: Set when the conversion result is higher than the high limit  0: Cleared on read of configuration register Therm mode:  1: Set when the conversion result is higher than the therm limit  0: Cleared when the conversion result is lower than the hysteresis								
14	LOW_Alert	R	0	Low Alert flag: 1: Set when the conversion result is lower than the low limit 0: Cleared when the configuration register is read Therm mode: Always set to 0								
13	Data_Ready	R	0	Data ready flag. This flag indicates that the conversion is complete and the temperature register can be read. Every time the temperature register or configuration register is read, this bit is cleared. This bit is set at the end of the conversion when the temperature register is updated. Data ready can be monitored on the ALERT pin by setting bit 2 of the configuration register.								
12	EEPROM_Busy	R	0	EEPROM busy flag. The value of the flag indicates that the EEPROM is busy during programming or power-up.								
11:10	MOD[1:0]	R/W	0	Set conversion mode. 00: Continuous conversion (CC) 01: Shutdown (SD) 10: Continuous conversion (CC), Same as 00 (reads back = 00) 11: One-shot conversion (OS)								
9:7	CONV[2:0]	R/W	100	Conversion cycle bit. See Table 7 for the standby time between conversions.								
6:5	AVG[1:0]	R/W	01	Conversion averaging modes. Determines the number of conversion results that are collected and averaged before updating the temperature register. The average is an accumulated average and not a running average.  00: No averaging 01: 8 Averaged conversions 10: 32 averaged conversions 11: 64 averaged conversions								
4	T/nA	R/W	0	Therm/alert mode select. 1: Therm mode 0: Alert mode								
3	POL	R/W	0	ALERT pin polarity bit. 1: Active high 0: Active low								
2	DR/Alert	R/W	0	ALERT pin select bit.  1: ALERT pin reflects the status of the data ready flag  0: ALERT pin reflects the status of the alert flags								
1	Soft_Reset	R/W	0	Software reset bit. When set to 1 it triggers software reset with a duration of 2 ms This bit will always read back 0								

<sup>(2)</sup> These bits can be stored in EEPROM. The factory setting for this register is 0220.



### **Table 6. Configuration Register Field Descriptions (continued)**

BIT	FIELD	TYPE	RESET	DESCRIPTION	
0	_	R	0	Not used	

### Table 7. Conversion Cycle Time in CC Mode

CONV[2:0]	AVG[1:0] = 00	AVG[1:0] = 01	AVG[1:0] = 10	AVG[1:0] = 11
000	15.5 ms	125 ms	500 ms	1 s
001	125 ms	125 ms	500 ms	1 s
010	250 ms	250 ms	500 ms	1 s
011	500 ms	500 ms	500 ms	1 s
100	1 s	1 s	1 s	1 s
101	4 s	4 s	4 s	4 s
110	8 s	8 s	8 s	8 s
111	16 s	16 s	16 s	16 s

If the time to complete the conversions needed for a given averaging setting is higher than the conversion setting cycle time, there will be no stand by time in the conversion cycle.

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#### 7.6.1.3 High Limit Register (address = 02h) [Factory default reset = 6000h]

This register is a 16-bit, read/write register that stores the high limit for comparison with the temperature result. One LSB equals 7.8125m°C. The range of the register is ±256°C. Negative numbers are represented in binary two's complement format. Following power-up or a general-call reset, the high-limit register is loaded with the stored value from the EEPROM. The factory default reset value is 6000h.

Figure 27. High Limit Register

15	14	13	12	11	10	9	8
H15	H14	H13	H12	H11	H10	H9	H8
R/W-0	R/W-1	R/W-1	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
7	6	5	4	3	2	1	0
H7	H6	H5	H4	H3	H2	H1	H0
R/W-0							

Table 8. High Limit Register Field Descriptions

BIT	FIELD	TYPE	RESET	DESCRIPTION
15:0	H[15:0]	R/W	6000h	16-bit, read/write register that stores the high limit for comparison with the temperature result.

#### 7.6.1.4 Low Limit Register (address = 03h) [Factory default reset = 8000h]

This register is configured as a 16-bit, read/write register that stores the low limit for comparison with the temperature result. One LSB equals 7.8125m°C. The range of the register is ±256°C. Negative numbers are represented in binary two's complement format. The data format is the same as the temperature register. Following power-up or reset, the low-limit register is loaded with the stored value from the EEPROM. The factory default reset value is 8000h.

Figure 28. Low Limit Register

15	14	13	12	11	10	9	8
L15	L14	L13	L12	L11	L10	L9	L8
R/W-1	R/W-0						
7	6	5	4	3	2	1	0
L7	L6	L5	L4	L3	L2	L1	LO
R/W-0							

**Table 9. Low Limit Register Field Descriptions** 

BIT	FIELD	TYPE	RESET	DESCRIPTION
15:0	L[15:0]	R/W	8000h	16-bit, read/write register that stores the low limit for comparison with the temperature result.



#### 7.6.1.5 EEPROM Unlock Register (address = 04h) [reset = 0000h]

#### Figure 29. EEPROM Unlock Register

15	14	13	12	11	10	9	8
EUN	EEPROM_Busy	_	_	_	_	_	_
R/W-0	R-0	R-0	R-0	R-0	R-0	R-0	R-0
7	6	5	4	3	2	1	0
_	_	_	_	_	_	_	_
R-0	R-0	R-0	R-0	R-0	R-0	R-0	R-0

Table 10. EEPROM Unlock Register Field Descriptions

BIT	FIELD	TYPE	RESET	DESCRIPTION
15	EUN	R/W	0	EEPROM unlock.  0: EEPROM is locked for programming: writes to all EEPROM addresses (such as configuration, limits, and EEPROM locations 1-4) are written to registers in digital logic and are not programmed in the EEPROM  1: EEPROM unlocked for programming: any writes to programmable registers program the respective location in the EEPROM
14	EEPROM_Busy	R	0	EEPROM busy. This flag is the mirror of the EEPROM busy flag (bit 12) in the configuration register.  0: Indicates that the EEPROM is ready, which means that the EEPROM has finished the last transaction and is ready to accept new commands  1: Indicates that the EEPROM is busy, which means that the EEPROM is currently completing a programming operation or performing power-up on reset load
13:0	_	R	0	Not used

#### 7.6.1.6 EEPROM1 Register (address = 05h) [reset = XXXXh]

The EEPROM1 register is a 16-bit register that be used as a scratch pad by the customer to store general-purpose data. This register has a corresponding EEPROM location. Writes to this address when the EEPROM is locked write data into the register and not to the EEPROM. Writes to this register when the EEPROM is unlocked causes the corresponding EEPROM location to be programmed. See the *Programming the EEPROM* section for more information. EEPROM[4:1] are preprogrammed during manufacturing with the unique ID that can be overwritten. To support NIST traceability do not delete or reprogram the EEPROM[1] register.

Figure 30. EEPROM1 Register

15	14	13	12	11	10	9	8
D15	D14	D13	D12	D11	D10	D9	D8
R/W-x							
7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
R/W-x							

Table 11. EEPROM1 Register Field Descriptions

BIT	FIELD	TYPE	RESET	DESCRIPTION
15:0	D[15:0]	R/W		This 16-bit register can be used as a scratch pad. To support NIST traceability do not delete or re-program this register.



### 7.6.1.7 *EEPROM2* Register (address = 06h) [reset = 0000h]

This register function the same as the EEPROM1 register.

#### Figure 31. EEPROM2 Register

15	14	13	12	11	10	9	8
D15	D14	D13	D12	D11	D10	D9	D8
R/W-x							
7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
R/W-x							

#### Table 12. EEPROM2 Register Field Descriptions

BIT	FIELD	TYPE	RESET	DESCRIPTION
15:0	D[15:0]	R/W	xxxxh	This 16-bit register can be used as a scratch pad.

#### 7.6.1.8 Temperature Offset Register (address = 07h) [reset = 0000h]

This 16-bit register is to be used as a user-defined temperature offset register during system calibration. The offset will be added to the temperature result after linearization. It has a same resolution of 7.8125m°C and same range of ±256°C as the temperature result register. The data format is the same as the temperature register. If the added result is out of boundary, then the temperature result will show as the maximum or minimum value.

Figure 32. Temperature Offset Register

15	14	13	12	11	10	9	8
D15	D14	D13	D12	D11	D10	D9	D8
R/W-0							
7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
R/W-0							

**Table 13. Temperature Offset Register Field Descriptions** 

BIT	FIELD	TYPE RESET DESCRIPTION		DESCRIPTION
15:0	D[15:0]	R/W	0	Temperature offset data from system calibration.



### 7.6.1.9 EEPROM3 Register (address = 08h) [reset = xxxxh]

This register function is the same as the EEPROM1 register. To support NIST traceability, do not delete or reprogram the EEPROM[1] register.

Figure 33. EEPROM3 Register

15	14	13	12	11	10	9	8
D15	D14	D13	D12	D11	D10	D9	D8
R/W-x							
7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
R/W-x							

### Table 14. EEPROM3 Register Field Descriptions

BIT	FIELD	TYPE	RESET	DESCRIPTION
15:0	D[15:0]	R/W		This 16-bit register is used as a scratch pad. To support NIST traceability, do not delete or re-program this register.

### 7.6.1.10 Device ID Register (address = 0Fh) [reset = 0117h]

This read-only register indicates the device ID.

Figure 34. Device ID Register

15	14	13	12	11	10	9	8
Rev3	Rev2	Rev1	Rev0	DID11	DID10	DID9	DID8
R-x	R-x	R-x	R-x	R-0	R-0	R-0	R-1
7	6	5	4	3	2	1	0
DID7	DID6	DID5	DID4	DID3	DID2	DID1	DID0
R-0	R-0	R-0	R-1	R-0	R-1	R-1	R-1

**Table 15. Device ID Register Field Descriptions** 

BIT	FIELD	TYPE	RESET	DESCRIPTION
15:12	Rev[3:0]	R	0h	Indicates the revision number.
11:0	DID[11:0]	R	117h	Indicates the device ID.



### 8 Application and Implementation

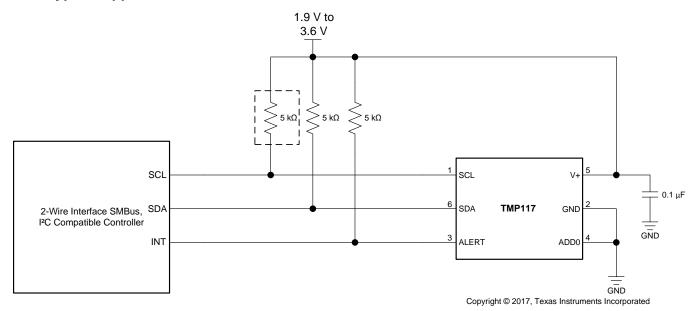
#### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 8.1 Application Information

The TMP117 is used to measure the temperature of the board location where the device is mounted. The programmable address options allow up to four locations on the board to be monitored on a single serial bus. For more information, refer to the related *TMP116 Ambient Air Temperature Measurement* (SNOA966), *Replacing Resistance Temperature Detectors with the TMP116 Temp Sensor* (SNOA969), and *Temperature Sensors: PCB Guidelines for Surface Mount Devices* (SNOA967) application reports on ti.com.

### 8.2 Typical Application



NOTE: The SDA and ALERT pins require pullup resistors.

Figure 35. Typical Connections

#### 8.2.1 Design Requirements

The TMP117 operates only as a slave device and communicates with the host through the  $I^2C$ -compatible serial interface. SCL is the input pin, SDA is a bidirectional pin, and ALERT is the output. The TMP117 requires a pullup resistor on the SDA, and ALERT pins. The recommended value for the pullup resistors is 5 k $\Omega$ . In some applications, the pullup resistor can be lower or higher than 5 k $\Omega$ . A 0.1- $\mu$ F bypass capacitor is recommended to be connected between V+ and GND. An SCL pullup resistor is required if the system microprocessor SCL pin is open-drain. Use a ceramic capacitor type with a temperature rating that matches the operating range of the application, and place the capacitor as close as possible to the V+ pin of the TMP117. The ADD0 pin can be connected directly to GND, V+, SDA and SCL for address selection of four possible unique slave ID addresses. Table 1 explains the addressing scheme. The ALERT output pin can be connected to a microcontroller interrupt that triggers an event that occurred when the temperature limit exceeds the programmable value in registers 02h and 03h. The ALERT pin can be left floating or connected to ground when not in use.



### **Typical Application (continued)**

#### 8.2.2 Detailed Design Procedure

#### 8.2.2.1 Noise and Averaging

The device temperature sampling distribution (with averaging disabled) covers an area of approximately six neighboring codes. The noise area of the six codes remains the same at full supply and full temperature range with a standard deviation of approximately 1 LSB. The device provides an averaging tool for 1, 8, 32, or 64 conversions. As shown in Figure 6, the 8-sample averaging reduces the internal noise distribution to a theoretical minimum of 2 LSB. This averaging means that if the system temperature slowly changes and the supply voltage is stable, then the 8-sample averaging can be enough to neutralize the device noise and provide stable temperature readings. However, if the system environment is noisy (such as when measuring air flow temperatures, power supply fluctuations, intensive communication on a serial bus, and so forth), then higher averaging numbers are recommended to be used.

#### 8.2.2.2 Self-Heating Effect (SHE)

During ADC conversion, some power is dissipated that heats the device despite the small power consumption of the TMP117. Consider the self-heating effect (SHE) for certain precise measurements. Figure 36 shows the device SHE in still air at 25°C after the supply is switched on. The device package is soldered to the 11-mm × 20-mm × 1.1-mm size coupon board. The board is placed horizontally, with the device on top. The TMP117 is in continuous conversion mode with 64 sampling averaging and zero conversion cycle time. There is no digital bus activity aside from reading temperature data one time each second. As shown in Figure 36, the SHE stabilization time in still air is greater when the device dissipates more power.

The SHE drift is strongly proportional to the device dissipated power. The SHE drift is also proportional to the device temperature because the consumption current with the same supply voltage increases with temperature. Figure 37 shows the SHE drifts versus temperature and dissipated power at 25°C for the same coupon board and the same conditions described previously.

To estimate the SHE for similar size boards, calculate the device consumption power for  $25^{\circ}$ C and use the corresponding power line shown in Figure 37. For example, in CC mode without DC at a 3.3-V supply at  $25^{\circ}$ C, the device dissipates 410  $\mu$ Wt. So self-heating in still air is approximately  $40m^{\circ}$ C for the described condition and rises to  $52m^{\circ}$ C at  $150^{\circ}$ C.



### **Typical Application (continued)**

The following methods can reduce the SHE:

- System calibration removes not only the self-heating error and power-supply rejection ratio (PSRR) effect but
  also compensates the temperature shift caused by the thermal resistance between the device and the
  measured object.
- If practical, use the device one-shot mode. If continuous conversion is needed, use the conversion cycle
  mode with significant standby time. For example, in most cases an 8-sample averaging (125 ms) with a 1second conversion cycle provides enough time for the device to cool down to the environment temperature
  and removes the SHE.
- Use the minimal acceptable power supply voltage.
- Use a printed-circuit board (PCB) layout that provides minimal thermal resistance to the device.
- Avoid using small-value pullup resistors on the SDA and ALERT pins. Instead, use pullup resistors larger than 2 kΩ.
- Ensure that the SCL and SDA signal levels remain below 10% or above 90% of the device supply voltage.
- Avoid heavy bypass traffic on the data line. Communication to other devices on the same data line increases
  the supply current even if the device is in SD mode.
- · Use the highest available communication speed.

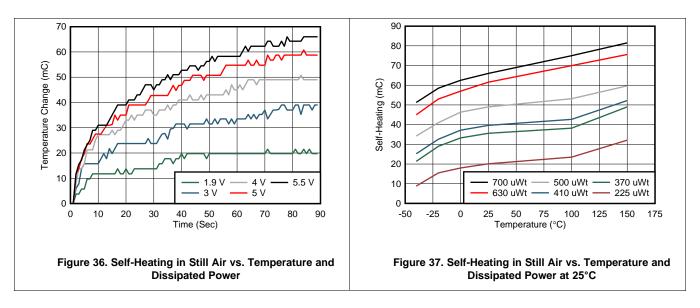
#### 8.2.2.3 Synchronized Temperature Measurements

When four temperature measurements are needed in four different places simultaneously, triggering a reset is recommended. In this method, four devices are programed with control registers set to CC mode with a conversion cycle time of 16 s. All four devices are connected to same two-wire bus with four different bus addresses. The bus general-call reset command is issued by the master. This command triggers all devices to reset (which takes approximately 1.5 ms) and triggers a simultaneous temperature sampling according to configuration registers setting. The master has 16 seconds to read data from the devices.



### **Typical Application (continued)**

#### 8.2.3 Application Curves



### 9 Power Supply Recommendations

The TMP117 operates on a power-supply range from 1.8 V to 5.5 V. The device is trimmed for operation at a 3.3-V supply, but can measure temperature accurately in the full supply range. A power-supply bypass capacitor is required, which must be placed as close to the supply and ground pins of the device as possible. A recommended value for this supply bypass capacitor is 100 nF. Applications with noisy or high-impedance power supplies may require additional decoupling capacitors to reject power-supply noise.

The TMP117 is a very low-power device that generates low noise on the supply bus. The user can apply an RC filter to the V+ pin of the device to further reduce noise that the TMP117 might propagate to other components.  $R_F$  in Figure 38 must be less than 0.5 k $\Omega$  and  $C_F$  must be at least 100 nF. The package thermal pad is not connected to the device ground and should be left unsoldered for best measurement accuracy. If the thermal pad is soldered, it must be left floating or grounded.

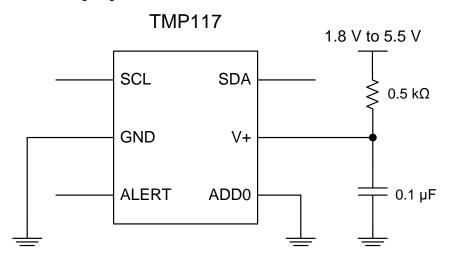


Figure 38. Noise-Reduction Techniques

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### 10 Layout

### 10.1 Layout Guidelines

#### NOTE

To achieve a high-precision temperature reading for a rigid PCB, do not solder down the thermal pad. For a flexible PCB, the user can solder the thermal pad to increase board level reliability.

For more information on board layout, refer to the related *Precise Temperature Measurements With TMP116* (SNOA986) and *Wearable Temperature Sensing Layout Considerations Optimized for Thermal Response* (SNIA021) application reports on ti.com.

Place the power-supply bypass capacitor as close as possible to the supply and ground pins. The recommended value of this bypass capacitor is 0.1  $\mu$ F. In some cases, the pullup resistor can be the heat source, therefore, maintain some distance between the resistor and the device. Mount the TMP117 on the PCB pad to provide the minimum thermal resistance to the measured object surface or to the surrounding air. The recommended PCB layout minimizes the device self-heating effect, reduces the time delay as temperature changes, and minimizes the temperature offset between the device and the object.

- 1. Soldering the TMP117 thermal pad to the PCB minimized the thermal resistance to the PCB, reduces the response time as temperature changes and minimizes the temperature offset between the device and measured object. Simultaneously the soldering of the thermal pad will, however, introduce mechanical stress that can be a source of additional measurement error. For cases when system calibration is not planned, TI recommends not soldering the thermal pad to the PCB. Due to the small thermal mass of the device, not soldering the thermal pad will have a minimal impact on the described characteristics.
- 2. Minimize thermal resistance from the PCB to the environment:
  - Design the pin-soldering pads to be as large as possible, especially if these pads are corner pads.
  - Use a PCB with thicker copper layers, when possible.
  - If the PCB has unused internal layers, extend these layers under the TMP117.
  - Cover the top and bottom unused board space with the copper layer.
- 3. If the device is used to measure air temperature:
  - Miniaturize the board to reduce thermal mass. Smaller thermal mass results in faster thermal response.
  - Place two copper planes to the top and bottom of the exposed pad.
  - Remove the top solder mask.

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- Cover any exposed copper with solder paste to prevent oxidation
- Thermal isolation is required to avoid thermal coupling from heat source components through the PCB.
- Avoid running the copper plane underneath the temperature sensor.
- Maximize the air gap between the sensor and the surrounding copper areas (anti-etch), especially when close to the heat source.
- Create a PCB cutout between sensor and other circuits. Leave a narrow channel away from heat source components as a routing bridge into the island.
- Avoid running traces on top if the heat source is top side. Instead, route all signals on the bottom side.
- Place the board vertically to improve air flow and to reduce dust collection.



### 10.2 Layout Example

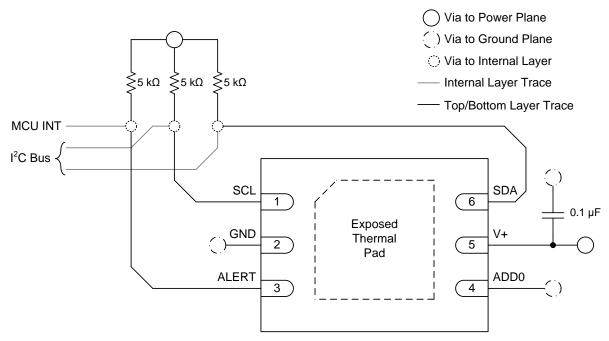


Figure 39. Layout Recommendation

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### 11 Device and Documentation Support

### 11.1 Documentation Support

#### 11.1.1 Related Documentation

For related documentation see the following:

- TMPx75 Temperature Sensor With I<sup>2</sup>C and SMBus Interface in Industry Standard LM75 Form Factor and Pinout (SBOS288)
- TMP275 ±0.5°C Temperature Sensor With I2C and SMBus Interface in Industry Standard LM75 Form Factor and Pinout (SBOS363)
- TMP116 Ambient Air Temperature Measurement (SNOA966)
- Replacing Resistance Temperature Detectors with the TMP116 Temp Sensor (SNOA969)
- Temperature Sensors: PCB Guidelines for Surface Mount Devices (SNOA967)
- Precise Temperature Measurements With TMP116 (SNOA986)
- Wearable Temperature Sensing Layout Considerations Optimized for Thermal Response (SNIA021)

### 11.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 11.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community T's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

#### 11.4 Trademarks

E2E is a trademark of Texas Instruments.

SMBus is a trademark of Intel Corporation.

All other trademarks are the property of their respective owners.

#### 11.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 11.6 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

### 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



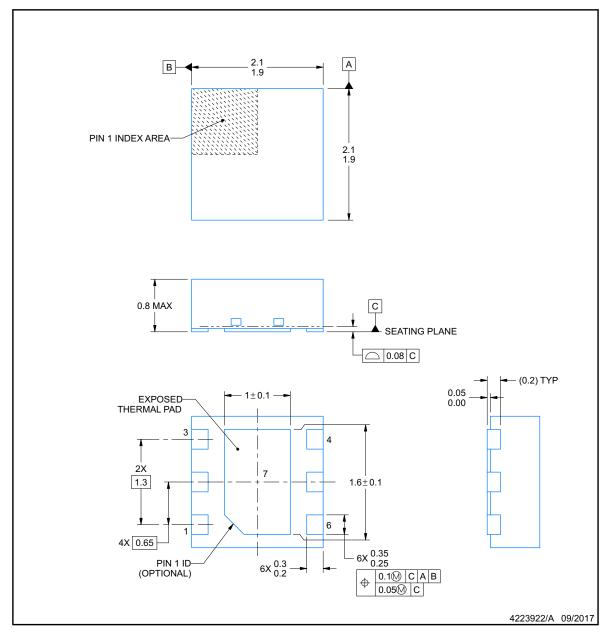
### **DRV0006B**



### **PACKAGE OUTLINE**

### WSON - 0.8 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.



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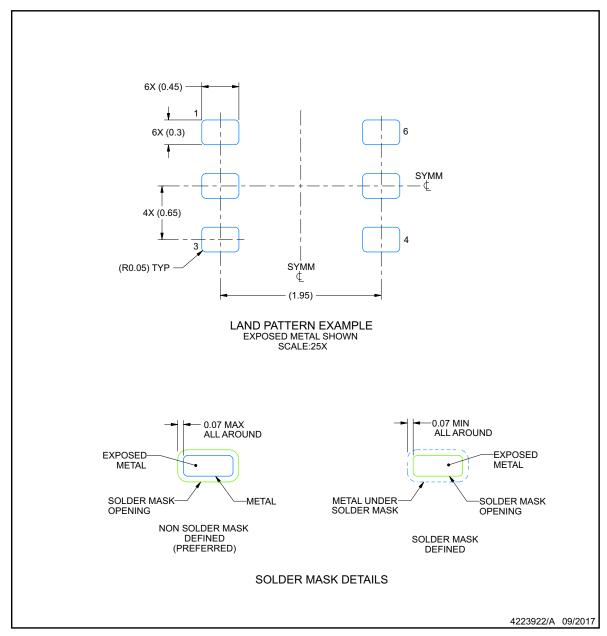


### **EXAMPLE BOARD LAYOUT**

### **DRV0006B**

### WSON - 0.8 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



NOTES: (continued)

3. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).



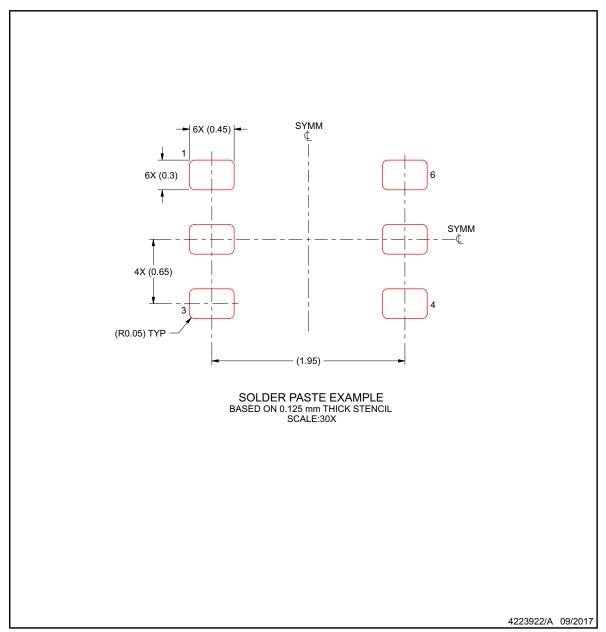


### **EXAMPLE STENCIL DESIGN**

### **DRV0006B**

### WSON - 0.8 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



NOTES: (continued)

4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.







26-Oct-2018

#### **PACKAGING INFORMATION**

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TMP117AIDRVR	ACTIVE	WSON	DRV	6	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 150	T117	Samples
TMP117AIDRVT	ACTIVE	WSON	DRV	6	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 150	T117	Samples
TMP117MAIDRVR	ACTIVE	WSON	DRV	6	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	0 to 85	117M	Samples
TMP117MAIDRVT	ACTIVE	WSON	DRV	6	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	0 to 85	117M	Samples
TMP117NAIDRVR	ACTIVE	WSON	DRV	6	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 150	117N	Samples
TMP117NAIDRVT	ACTIVE	WSON	DRV	6	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 150	117N	Samples

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.



### **PACKAGE OPTION ADDENDUM**

26-Oct-2018

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE MATERIALS INFORMATION

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### TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TMP117AIDRVR	WSON	DRV	6	3000	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
TMP117AIDRVT	WSON	DRV	6	250	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
TMP117MAIDRVR	WSON	DRV	6	3000	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
TMP117MAIDRVT	WSON	DRV	6	250	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
TMP117NAIDRVR	WSON	DRV	6	3000	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
TMP117NAIDRVT	WSON	DRV	6	250	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2

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\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TMP117AIDRVR	WSON	DRV	6	3000	210.0	185.0	35.0
TMP117AIDRVT	WSON	DRV	6	250	210.0	185.0	35.0
TMP117MAIDRVR	WSON	DRV	6	3000	210.0	185.0	35.0
TMP117MAIDRVT	WSON	DRV	6	250	210.0	185.0	35.0
TMP117NAIDRVR	WSON	DRV	6	3000	210.0	185.0	35.0
TMP117NAIDRVT	WSON	DRV	6	250	210.0	185.0	35.0

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