

## Chapter C-09

(1) Rectangular window:-

$$w_R(n) = \begin{cases} 1 & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$$

(2) Hamming window:-

$$w_H(n) = \begin{cases} 0.54 - 0.46 \cos \frac{2\pi n}{N-1} & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$$

(3) Hanning window:-

$$w_{hn}(n) = \begin{cases} 0.5 - 0.5 \cos \frac{2\pi n}{N-1} & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$$

Filter: [frequency response and impulse response]

(1) Low-pass filters:- (frequency and impulse response)

$$H_d(\omega) = \begin{cases} e^{-j\alpha\omega}; & -w_c \leq \omega \leq w_c \\ 0; & \text{otherwise} \end{cases}$$

$$h_d(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(\omega) e^{j\omega n} d\omega$$

$$= \frac{\sin w_c(n-\alpha)}{\pi(n-\alpha)}$$

(2) High pass filters:

$$H_d(\omega) = \begin{cases} e^{-j\alpha\omega}; & -\pi \leq \omega \leq -w_c \\ e^{-j\alpha\omega}; & w_c \leq \omega \leq \pi \\ 0; & -w_c \leq \omega \leq w_c \end{cases}$$

$$h_d(n) = \frac{\sin(n-\alpha)\pi - \sin w_c(n-\alpha)}{\pi(n-\alpha)}$$

$$= \text{QD}$$

(3) Band-pass filter:-

$$H_d(\omega) = \begin{cases} e^{-j\omega}; & -w_2 \leq \omega \leq -w_{c1} \\ e^{-j\omega}; & w_{c1} \leq \omega \leq w_2 \\ e^{-j\omega}; & \text{otherwise} \end{cases}$$

$h_d(n) =$ 

$$\frac{\sin w_2(n-\alpha) - \sin w_{c1}(n-\alpha)}{\pi(n-\alpha)}$$

(4) Band-stop filter:-

$$H_d(\omega) = \begin{cases} e^{j\omega}; & -\pi \leq \omega \leq -w_{c2} \\ e^{-j\omega}; & -w_{c1} \leq \omega \leq w_{c2} \\ e^{-j\omega}; & w_{c2} \leq \omega \leq \pi \\ 0; & \text{otherwise} \end{cases}$$

$$h_d(n) = \frac{\sin w_{c1}(n-\alpha) + \sin \pi(n-\alpha) - \sin w_{c2}(n-\alpha)}{\pi(n-\alpha)}$$

the last value of  $\theta$  is  $\pi/2$ , so it's found  
that  $\omega$  is going to change its sign.

$\Rightarrow \omega = \sqrt{\frac{I}{\mu_0 S^2}} \cdot \sin(\theta) = (\frac{I}{\mu_0 S}) \cdot A$

Then we can see that  $\omega$  is changing  
sign, so  $\omega$  is oscillating.

- in unit of  $\text{rad s}^{-1}$  sufficient answer

$\Rightarrow \omega = (\frac{I}{\mu_0 S}) \cdot A$

so we can write  $\omega$  in terms of  $\theta$  by  
using  $\omega = \dot{\theta}$  we get  $\omega = (\frac{I}{\mu_0 S}) \cdot \dot{\theta}$

and now we can write  $\theta$  in terms of  $t$   
 $\theta = \theta_0 \cos(\omega t)$

so  $\omega = \frac{\theta_0}{t} \cdot \frac{1}{\cos(\theta_0)} = (\frac{\theta_0}{t}) \cdot \frac{1}{\cos(\theta_0)}$

and  $\omega = \frac{\theta_0}{t} \cdot \frac{1}{\cos(\theta_0)} = \frac{\theta_0}{t \cdot \cos(\theta_0)}$

Example: 9.7 (page-699):- A low pass filter is to be designed with the following desired

$$w_c = \frac{\pi}{4}$$

frequency response:-

$$H_d(e^{j\omega}) = \begin{cases} e^{-j2\omega} & -\frac{\pi}{4} \leq \omega \leq \frac{\pi}{4} \\ 0 & \frac{\pi}{4} \leq |\omega| \leq \pi \end{cases}$$

Determine the [filter co-efficient]  $h(n)$  if the window function is defined as:-

$$w(n) = \begin{cases} 1 & 0 \leq n \leq 4 \\ 0 & \text{otherwise} \end{cases} \quad \text{rectangular}$$

and determine the [frequency response]  $H(e^{j\omega})$  of the designed filter.

Sol<sup>n</sup> :-

The filter coefficients are given by

$$\begin{aligned} h_d(n) &= \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(\omega) e^{j\omega n} d\omega \\ &= \frac{1}{2\pi} \int_{-\pi/4}^{\pi/4} e^{-j2\omega} e^{j\omega n} d\omega. \end{aligned}$$

$$\begin{aligned}
 &= \frac{1}{2\pi} \int_{-\pi/4}^{\pi/4} e^{jw(n-2)} \cdot dw \\
 &= \frac{1}{2\pi} \int_{-\pi/4}^{\pi/4} \left[ \frac{e^{jw(n-2)}}{j(n-2)} \right] dw \\
 &= \frac{1}{\pi(n-2)} \left[ \frac{e^{j(n-2)\pi/4} - e^{-j(n-2)\pi/4}}{2j} \right]
 \end{aligned}$$

$$h_d(n) = \frac{1}{\pi(n-2)} \cdot \sin(n-2)\pi/4 = \frac{e^{j(n-2)\pi/4} - e^{-j(n-2)\pi/4}}{2j} = \sin(n-2)\pi/4$$

For  $n=2$  we apply L'Hospital's rule

$$h_d(2) = \lim_{n \rightarrow 2} \frac{1}{\pi} \cdot \frac{\sin(n-2)\pi/4}{(n-2)} = 0$$

$$= \frac{1}{\pi} \cdot \frac{\pi}{4} = \frac{1}{4} = \boxed{\sin(\lim_{\theta \rightarrow 0} \frac{\sin(\theta-1)}{(\theta-1)}) = 1}$$

$$\therefore h_d(0) = \frac{1}{\pi(-2)} \cdot \sin(-2)\frac{\pi}{4} = \frac{1}{-2\pi} (-\sin\frac{\pi}{2}) = \frac{1}{2\pi} (-1)$$

$$= \frac{1}{2\pi} (-1) = \frac{1}{2\pi}$$

$$h_d(1) = \frac{1}{\pi(1-2)} \cdot \sin((1-2)\frac{\pi}{4}) = \\ = \frac{1}{-\pi} (-\sin \frac{\pi}{4}) = \frac{1}{\pi} \cdot \frac{1}{\sqrt{2}} = \frac{1}{\sqrt{2}\pi}$$

$$h_d(3) = \frac{1}{\pi(3-2)} \cdot \sin((3-2)\frac{\pi}{4}) = \frac{1}{\pi} \sin \frac{\pi}{4} \\ = \frac{1}{\pi} \cdot \frac{1}{\sqrt{2}} = \frac{1}{\sqrt{2}\pi}$$

$$h_d(4) = \frac{1}{\pi(4-2)} \cdot \sin((4-2)\frac{\pi}{4}) = \frac{1}{2\pi} \cdot \sin 2\pi/2 = \frac{1}{2\pi}$$

The filter coefficient,  $[h(n) = h_d(n) \cdot w(n)]$

$$\therefore h(0) = \frac{1}{2\pi} = h(4); \quad \text{and } = (1)_b$$

$$\therefore h(1) = \frac{1}{\sqrt{2}\pi} = h_d(3);$$

$$\therefore h(2) = \frac{1}{\sqrt{2}}$$

$$\therefore h(3) = \frac{1}{\sqrt{2}\pi} = \frac{1}{\pi} \sin \frac{\pi}{4} : \frac{1}{\pi} = (0)_b$$

$$\therefore h(4) = (1) - \frac{1}{\sqrt{2}\pi} =$$

The frequency response  $H(\omega)$

$\therefore$  The ~~realizable digital filter~~ filter;

$$H(\omega) = \sum_{n=0}^{N-1} h(n) e^{-j\omega n}$$

$$= h(0) + h(1) e^{-j\omega} + h(2) e^{-2j\omega} + h(3) e^{-3j\omega} + h(4) e^{-4j\omega}$$

$$= e^{-2j\omega} [h(0) e^{2j\omega} + h(1) e^{j\omega} + h(2) + h(3) e^{-j\omega} + h(4) e^{-2j\omega}]$$

$$= e^{-2j\omega} \left[ \frac{1}{2\pi} \left\{ e^{2j\omega} + e^{-2j\omega} \right\} + \frac{1}{\sqrt{2}\pi} \left\{ e^{j\omega} + e^{-j\omega} \right\} + \frac{1}{4} \right]$$

$$= e^{-2j\omega} \left[ \frac{1}{4} + \frac{1}{2\pi} \cdot 2 \cos 2\omega + \frac{1}{\sqrt{2}\pi} \cdot 2 \cos \omega \right]$$

$$\therefore H(\omega) = e^{-2j\omega} \left[ \frac{1}{4} + \pi \cos 2\omega + \frac{\sqrt{2}}{\pi} \cos \omega \right]$$

$$\cos \theta = \frac{e^{j\theta} + e^{-j\theta}}{2}$$

$$= 2 \times \cos 2\omega$$

The realizable digital filter is :-

$$H(z) = \sum_{n=0}^{N-1} h(n) z^{-n}$$

$$= h(0) z^0 + h(1) z^{-1} + h(2) z^{-2} + h(3) z^{-3} + h(4) z^{-4}$$

$$= \frac{1}{2\pi} + \frac{1}{\sqrt{2}\pi} z^{-1} + \frac{1}{4} z^{-2} + \frac{1}{\sqrt{2}\pi} z^{-3} + \frac{1}{2\pi} z^{-4}$$

$$= z^{-2} \left[ \frac{1}{4} + \frac{1}{2\pi} (z^2 + z^{-2}) + \frac{1}{\sqrt{2}\pi} (z + z^{-1}) \right]$$

Ans:

$$\frac{1}{2\pi} + \frac{1}{\sqrt{2}\pi} z^{-1} + \frac{1}{4} z^{-2} + \frac{1}{\sqrt{2}\pi} (z + z^{-1})$$

Example 9.8

$$H_d(e^{j\omega}) = \begin{cases} 0; & -\pi/2 \leq \omega \leq \pi/2 \\ e^{-j2\omega}; & \pi/2 \leq |\omega| \leq \pi \end{cases}$$

$$w(n) = \begin{cases} 1; & 0 \leq n \leq 1 \\ 0; & \text{otherwise} \end{cases}$$

*rectangle*

Find filter coefficient  $h(n)$  and frequency response  $H(e^{j\omega})$ .

Sol:- Filter coefficients are:

$$h_d(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(\omega) \cdot e^{j\omega n} d\omega$$

$$= \frac{1}{2\pi} \int_{-\pi/2}^{\pi/2} e^{-j2\omega} \cdot e^{j\omega n} d\omega + \frac{1}{2\pi} \int_{\pi/2}^{\pi} e^{-j\omega^2} \cdot e^{j\omega n} d\omega$$

$$= \frac{1}{2\pi} \int_{-\pi}^{\pi/2} e^{j\omega(n-2)} d\omega + \frac{1}{2\pi} \int_{\pi/2}^{\pi} e^{j\omega(n-2)} d\omega.$$

*using*

$\pm \pi/2$

$$= \frac{1}{2\pi} \left[ \frac{e^{jw(n-2)}}{j\pi(n-2)} \right]_{-\pi}^{-\pi/2} + \frac{1}{2\pi} \left[ \frac{e^{jw(n-2)}}{j(n-2)} \right]_{\pi/2}^{\pi}$$

$$= \frac{1}{2\pi j(n-2)} \left[ e^{-j(n-2)\pi/2} - e^{-j(n-2)\pi} + e^{j(n-2)\pi} - e^{j(n-2)\pi/2} \right]$$

$$= \frac{1}{\pi(n-2)} \left\{ \left[ \frac{e^{j(n-2)\pi} - e^{-j(n-2)\pi}}{2j} \right] - \left[ \frac{e^{j(n-2)\pi/2} - e^{-j(n-2)\pi/2}}{2j} \right] \right\}$$

$$h_d(n) = \frac{1}{\pi(n-2)} \left[ \sin(n-2)\pi - \sin(n-2)\pi/2 \right]$$

For  $n=2$  we apply L'Hospital's rule

$$h_2(2) = \lim_{n \rightarrow 2} \frac{1}{\pi} \left[ \frac{\sin((n-2)\pi)}{(n-2)} - \frac{\sin((n-2)\pi/2)}{(n-2)} \right]$$

$$= \frac{1}{\pi} [\pi - \pi/2] = \frac{1}{\pi} \cdot \frac{\pi}{2} = \frac{1}{2}$$

$$\therefore h_2(0) = \frac{1}{\pi(-2)} \left[ \sin(-2)\pi - \sin(-2)\pi/2 \right]$$

$$= \frac{1}{-2\pi} [-\sin 2\pi + \sin \pi/2]$$

$$= 0 \quad [\because \sin n\pi = \sin n\pi/2 = 0]$$

$$h_2(1) = \frac{1}{\pi(-1)} \left[ \sin(-1)\pi - \sin(-1)\pi/2 \right]$$

$$= \frac{1}{-\pi} [-\sin \pi + \sin \pi/2] = -\frac{1}{\pi}$$

$$h_2(3) = \frac{1}{\pi} \left[ \sin(\pi) - \sin \pi/2 \right] = -\frac{1}{\pi}$$

$$h_2(4) = \frac{1}{2\pi} \left[ \sin 2\pi - \sin 2\pi/2 \right] = 0$$

$$h(n) = h_1(n) \cdot w(n)$$

$$\therefore h(0) = 0 = h(1)$$

$$\therefore h(1) = -\frac{1}{\pi} = h(3)$$

$$\therefore h(2) = \frac{1}{2}$$

$\therefore$  Filter transfer function:

$$H(z) = \sum_{n=0}^{N-1} h(n) z^{-n}$$

$$= h(0) + h(1)z^1 + h(2)z^2 + h(3)z^3 + h(4)z^4$$

$$= 0 + -\frac{1}{\pi}z^1 + \frac{1}{2}z^2 - \frac{1}{\pi}z^3 + 0$$

$$= z^2 \left[ \frac{1}{2} - \frac{1}{\pi} \left\{ z + z^{-1} \right\} \right]$$

$$0 = \left[ \frac{\pi}{2} - \pi z - \pi z^{-1} \right] = \frac{1}{\pi z} = \text{Ans.}$$

$\therefore$  frequency response  $H(\omega)$

$$H(\omega) = \sum_{n=0}^4 h(n) e^{-jn\omega}$$

$$= h(0) + h(1)e^{-j\omega} + h(2)e^{-2j\omega} + h(3)e^{-3j\omega} + h(4)e^{-4j\omega}$$

$$= e^{-j2\omega} \left[ h(0) e^{2j\omega} + h(1) e^{j\omega} + h(2) e^{-j\omega} + h(3) e^{-2j\omega} + h(4) e^{-3j\omega} \right]$$

$$= e^{-j2\omega} \left[ \frac{1}{2} + 0 + \frac{1}{4} \cos \omega \right]$$

$$= e^{-2j\omega} \left[ \frac{1}{2} - \frac{1}{4} \cos \omega \right]$$

$$= \frac{1}{2} - \frac{1}{4} \cos \omega$$

Example: 9.9 - Using Hamming window;  $N=7$

$$H_d(e^{j\omega}) = \begin{cases} e^{-3j\omega}; & -\frac{\pi}{4} \leq \omega \leq \frac{\pi}{4} \\ 0; & \frac{\pi}{4} \leq |\omega| \leq \frac{\pi}{2} \end{cases}$$

Sol: Filter co-efficients are:

$$h_d(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(\omega) \cdot e^{j\omega n} d\omega$$

$$= \frac{1}{2\pi} \int_{-\frac{\pi}{4}}^{\frac{\pi}{4}} e^{-3j\omega} \cdot e^{j\omega n} d\omega$$
$$= \frac{1}{2\pi} \int_{-\frac{\pi}{4}}^{\frac{\pi}{4}} e^{j\omega(n-3)} d\omega$$

$$= \frac{1}{2\pi} \left[ \frac{e^{j\omega(n-3)}}{j(n-3)} \right]_{-\frac{\pi}{4}}^{\frac{\pi}{4}}$$

$\approx -0.05$

$$= \frac{1}{\pi(n-3)} \left[ \frac{e^{j(n-3)\frac{\pi}{4}} - e^{-j(n-3)\frac{\pi}{4}}}{2j} \right]$$

$$\therefore h_2(n) = \frac{1}{\pi(n-3)} \cdot \sin(n-3)\frac{\pi}{4}$$

For  $n=3$ , applying L'Hospital rule;

$$h_2(0^+)=\lim_{n \rightarrow 3} \frac{1}{\pi} \cdot \frac{\sin(n-3)\frac{\pi}{4}}{\pi(n-3)}$$

$$= \frac{1}{\pi} \cdot \frac{\pi}{4} = \frac{1}{4}$$

$$\therefore h_2(1) = \frac{1}{\pi(-3)} \cdot \sin(-3)\frac{\pi}{4}$$

$$= \frac{1}{-3\pi} \left( -\sin \frac{3\pi}{4} \right)$$

$$= \frac{1}{3\pi} \cdot \frac{\sqrt{2}}{2} \quad [ \text{Rad mod } 0 ]$$

$$= \frac{0.707}{3\pi}$$

$$h_d(1) = \frac{1}{\pi(-2)} \cdot \sin(-2) \frac{\pi}{4} =$$

$$= \frac{1}{2\pi} \sin \frac{\pi}{2} = \frac{1}{2\pi}$$

$$h_d(2) = \frac{1}{\pi(-1)} \sin(-1) \frac{\pi}{4} = \frac{1}{\pi} \sin \frac{\pi}{4} = \frac{1}{\pi} \cdot \frac{1}{\sqrt{2}}$$

$$= \frac{1}{\sqrt{2}\pi} = \frac{0.707}{\pi}$$

$$h_d(4) = \frac{1}{\pi} \sin \frac{\pi}{4} = \frac{1}{\pi} \cdot \frac{1}{\sqrt{2}} = \frac{1}{\sqrt{2}\pi} = \frac{0.707}{\pi}$$

$$h_d(5) = \frac{1}{2\pi} \sin \frac{\pi}{2} = \frac{1}{2\pi}$$

$$h_d(6) = \frac{1}{3\pi} \sin(3) \frac{\pi}{4} = \frac{0.707}{3\pi}$$

$$h_d(7) = \frac{1}{4\pi} \sin(4) \frac{\pi}{4} = \frac{1}{4} \sin \pi$$

$$\therefore h_2(0) = \frac{0.707}{3\pi} = h_2(6)$$

$$\therefore h_2(1) = \frac{1}{2\pi} = h_2(5)$$

$$\therefore h_2(2) = \frac{0.707}{\pi} = h_2(4)$$

$$\therefore h_2(3) = \frac{1}{\pi}$$

$$\therefore \underline{h_2(4)} =$$

For Hamming window we know

$$W(n) = \begin{cases} 0.54 - 0.46 \cos \frac{2\pi n}{N-1} & 0 \leq n \leq N-1 \\ 0; \text{ otherwise} & \end{cases}$$

$$\therefore W(0) = 0.54 - 0.46 \cos 0 = 0.08$$

$$\therefore W(1) = 0.54 - 0.46 \cos \frac{2\pi \cdot 1}{6} = 0.31$$

$$\therefore W(2) = 0.54 - 0.46 \cos \frac{2\pi \cdot 2}{6} = 0.77$$

$$\therefore W(3) = 0.54 - 0.46 \cos \frac{2\pi \cdot 3}{6} = 1$$

$$\therefore W(4) = 0.54 - 0.46 \cos \frac{2\pi \cdot 4}{6} = 0.77$$

$$\therefore W(5) = 0.54 - 0.46 \cos \frac{2\pi \cdot 5}{6} = 0.31$$

$$\therefore w(6) = 0.59 - .96 \cos \frac{2\pi \cdot 6}{6} = 0.08$$

The filter co-efficient;  $\frac{1}{\pi \Sigma} = \frac{1}{\pi \cdot 6} = (1)w$

$$h(n) = h_d(n) \cdot w(n)$$

$$\therefore h(0) = h_d(0) \cdot w(0) = \frac{0.707}{3\pi} \times 0.08 = 0.006$$

$$\therefore h(1) = h_d(1) \times w(1) = \frac{1}{2\pi} \times 0.31 = 0.049$$

$$\therefore h(2) = h_d(2) \times w(2) = \frac{0.707}{3\pi} \times 0.77 = 0.173$$

$$\therefore h(3) = \frac{1}{3} \times 1 = \frac{1}{3} = 0.25$$

$$\therefore h(4) = \frac{0.707}{\pi} \times 0.77 = 0.173$$

$$h(5) = \frac{1}{2\pi} \times 0.31 = 0.049$$

$$h(6) = \frac{0.707}{3\pi} \times 0.08 = 0.006$$

$$ff_0 = \frac{\pi \cdot 6}{6} \text{ or } 3\pi \cdot 0 - 1\pi \cdot 0 = (1)w$$

$$ff_0 = \frac{\pi \cdot 6}{6} \text{ or } 3\pi \cdot 0 - 1\pi \cdot 0 = (2)w$$

∴ Frequency response:

$$H(\omega) = \sum_{n=0}^6 h(n) e^{-j\omega n}$$

$$= h(0) e^{0} + h(1) e^{-j\omega} + h(2) e^{-2j\omega} + h(3) e^{-3j\omega} \\ + h(4) e^{-4j\omega} + h(5) e^{-5j\omega} + h(6) e^{-6j\omega}$$

$$= e^{-3j\omega} \left[ h(3) + \{ h(0) e^{3j\omega} + h(6) e^{-3j\omega} \} \right. \\ \left. + \{ h(4) e^{2j\omega} + h(5) e^{-2j\omega} \} + \{ h(2) e^{j\omega} + h(4) e^{-j\omega} \} \right]$$

$$= e^{-3j\omega} \left[ 0.25 + (2 \times 0.006) \cos 3\omega + \right. \\ \left. (2 \times 0.099) \cos 2\omega + (2 \times 0.173) \cos \omega \right]$$

$$= e^{-3j\omega} \left[ 0.25 + 0.012033\omega + 0.098 \cos 2\omega \right. \\ \left. + 0.346 \cos \omega \right]$$

Air

The transfer function:

$$H(z) = \sum_{n=0}^{N-1} h(n) z^{-n}$$

$$= h(0) + h(1) z^{-1} + h(2) z^{-2} + h(3) z^{-3} + h(4) z^{-4} + \\ h(5) z^{-5} + h(6) z^{-6}$$

$$= z^{-3} [h(3) + h(0)(z^3 + z^{-3}) + h(1)(z^2 + z^{-2}) \\ + h(2)(z + z^{-1})]$$

$$= z^{-3} [0.25 + 0.006(z^3 + z^{-3}) + 0.0749(z^2 + z^{-2}) \\ + 0.173(z + z^{-1})]$$

$$= 0.006 + 0.049z^{-1} + 0.173z^{-2} + 0.25z^{-3} +$$

$$+ 0.173z^{-4} + 0.049z^{-5} + 0.006z^{-6}$$

$$\boxed{0.006 + 0.049z^{-1} + 0.173z^{-2} + 0.25z^{-3} + 0.173z^{-4} + 0.049z^{-5} + 0.006z^{-6}}$$

Ex 9.10: Low-pass filter using rectangular window

taking 9 samples and cutoff frequency 1.2 rad/sec

Sol: For low pass filter.

$$H_d(\omega) = \begin{cases} e^{-j\omega w_c} & -w_c \leq \omega \leq w_c \\ 0 & \text{otherwise} \end{cases}$$

$$\begin{aligned} h_d(\omega) &= \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(\omega) e^{-j\omega n} d\omega \\ &= \frac{1}{2\pi} \int_{-w_c}^{w_c} e^{-j\omega w_c} e^{-j\omega n} d\omega \\ &= \frac{1}{2\pi} \int_{-w_c}^{w_c} e^{-j\omega(n-w_c)} d\omega \\ &= \frac{1}{2\pi} \left[ \frac{e^{-j\omega(n-w_c)}}{-j(n-w_c)} \right]_{-w_c}^{w_c} \end{aligned}$$

$$h_d(n) = \frac{1}{\pi(n-\omega)} \left[ \frac{e^{j\omega(n-\omega)} - e^{-j\omega(n-\omega)}}{2j} \right]$$

$$h_d(n) = \frac{1}{\pi(n-\omega)} \cos \omega(n-\omega)$$

$$h_d(n) = \frac{1}{\pi(n-\omega)} \cos \sin \omega(n-\omega) \quad \text{--- (1)}$$

For  $n=\infty$  using L'Hospital rule;

$$h_d(\infty) = \lim_{n \rightarrow \infty} \frac{\cos \sin \omega(n-\omega)}{\pi(n-\omega)}$$

$$h_d(\infty) = \frac{\omega}{\pi}$$

For rectangular window

$$w_R(n) = \begin{cases} 1 & \text{for } 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$$



$$w_c = 1.2 \text{ rad/sec} \text{ and } \alpha = \frac{\pi}{2} = \frac{\pi}{2} = \frac{\pi}{2} = \frac{\pi}{2}$$

$$\therefore h(0) = \frac{\sin(0-4) \times 1.2}{\pi(0-4)} = -0.0793$$

$$\therefore h(1) = \frac{\sin(1-4) \times 1.2}{\pi(1-4)} = -0.0470$$

$$\therefore h(2) = \frac{\sin(2-4) \times 1.2}{\pi(2-4)} = 0.1075$$

$$\therefore h(3) = \frac{\sin(3-4) \times 1.2}{\pi(3-4)} = 0.2967$$

$$\therefore h(4) = \frac{\sin(4-4) \times 1.2}{\pi(4-4)} = 0.382$$

$$\therefore h(5) = \frac{\sin(5-4) \times 1.2}{\pi(5-4)} = 0.2967$$

$$\therefore h(6) = \frac{\sin(6-4) \times 1.2}{\pi(6-4)} = 0.1075$$

$$\therefore h(7) = \frac{\sin(7-4) \times 1.2}{\pi(7-4)} = -0.047$$

$$\therefore h(8) = \frac{\sin(8-4) \times 1.2}{\pi(8-4)} = 0.0793$$

## Transfer function:

$$H(z) = \sum_{n=0}^{N-1} h(n) z^{-n}$$
$$= h(0) + h(1)z^{-1} + h(2)z^{-2} + h(3)z^{-3} + h(4)z^{-4}$$
$$+ h(5)z^{-5} + h(6)z^{-6} + h(7)z^{-7} + h(8)z^{-8}$$
$$= z^4 [h(4) + h(0) \{z^4 + z^{-4}\} + h(1) \{z^3 + z^{-3}\} +$$
$$+ h(2) \{z^2 + z^{-2}\} + h(3) \{z + z^{-1}\}]$$
$$= z^4 [0.382 + (-0.0793)(z^4 + z^{-4}) +$$
$$(-0.0970)(z^3 + z^{-3}) + 0.1075(z^2 + z^{-2}) +$$
$$0.2967(z + z^{-1})]$$

Ans

## Frequency response

$$H(\omega) = \sum_{n=0}^8 h(n) e^{-jn\omega}$$

$$= h(0) + h(1)e^{-j\omega} + h(2)e^{-2j\omega} + h(3)e^{-3j\omega}$$

$$+ h(4)e^{-4j\omega} + h(5)e^{-5j\omega} + h(7)e^{-7j\omega}$$

$$+ h(8)e^{-8j\omega} + h(6)e^{-6j\omega}$$

$$= e^{-4j\omega} \left[ h(4) + h(0) \left[ e^{4j\omega} + e^{-4j\omega} \right] + \right.$$

$$h(1) \left\{ e^{3j\omega} + e^{-3j\omega} \right\} + h(2) \left\{ e^{2j\omega} + e^{-2j\omega} \right\}$$

$$\left. + h(3) \left\{ e^{j\omega} + e^{-j\omega} \right\} \right]$$

$$= e^{-4j\omega} \left[ 0.382 + 2 \times (-0.0793) \cos 9\omega + \right.$$

$$2 \times (-0.0970) \cos 3\omega + 2 \times (0.1075) \cos 2\omega$$

$$\left. + 2(0.2967) \cos \omega \right]$$

Ex-9.11: high pass filter using Hamming Window  
cutoff frequency  $1.2 \text{ rad/sec}$  and  $N=9$

Sol: For high pass filter

$$H_d(w) = \begin{cases} e^{-jw_0}, & w_c \leq |w| \leq w_H \\ 0, & \text{otherwise} \end{cases}$$

$$\therefore h_d(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(w) e^{jwn} dw$$

$$= \frac{1}{2\pi} \int_{-\pi}^{-w_c} e^{-jw_n} e^{jwn} dw + \frac{1}{2\pi} \int_{w_c}^{\pi} e^{-jw_n} e^{jwn} dw$$

$$= \frac{1}{2\pi} \left[ \int_{-\pi}^{-w_c} e^{jw(n-\alpha)} dw \right] + \frac{1}{2\pi} \int_{w_c}^{\pi} e^{jw(n-\alpha)} dw$$

$$= \frac{1}{2\pi} \left[ \frac{e^{jw(n-\alpha)}}{j(n-\alpha)} \right]_{-\pi}^{-w_c} + \frac{1}{2\pi} \int_{w_c}^{\pi} \frac{e^{jw(n-\alpha)}}{j(n-\alpha)} dw$$

$$\begin{aligned}
 &= \frac{1}{\pi(n-\omega)} \left[ -\frac{e^{-j(n-\omega)w_c} - e^{-j(n-\omega)\pi}}{2j} \right] + \\
 &\quad \frac{1}{\pi(n-\omega)} \left[ \frac{e^{j(n-\omega)\pi} - e^{j(n-\omega)w_c}}{-2j} \right] \\
 &= \frac{1}{\pi(n-\omega)} \left[ \frac{e^{-j(n-\omega)\pi} - e^{-j(n-\omega)w_c}}{2j} - \frac{e^{j(n-\omega)\pi} - e^{j(n-\omega)w_c}}{2j} \right] \\
 &= \frac{1}{\pi(n-\omega)} \left[ \sin(n-\omega)\pi - \sin(n-\omega)w_c \right]
 \end{aligned}$$

If  $n=\infty$  then using L'Hopital's Law:

$$h_2(\omega) = \lim_{n \rightarrow \infty} \frac{\sin(n-\omega)\pi - \sin(n-\omega)w_c}{\pi(n-\omega)}$$

$$= \lim_{n \rightarrow \infty} \left[ \frac{\sin(n-\omega)\pi}{\pi(n-\omega)} - \frac{\sin(n-\omega)w_c}{\pi(n-\omega)} \right]$$

$$= \left[ -\frac{1}{\pi} - \frac{w_c}{\pi} \right] = 1 - \frac{w_c}{\pi}$$

$$\therefore h_2(0) = \lambda = \frac{N-1}{2} = \frac{9-1}{2} = 4; \quad w_c = 1.2$$

$$\therefore h_2(0) = \frac{\sin(0-4)\pi - \sin(0-4) \cdot 1.2}{\pi(0-4)} = 0.0792$$

$$\therefore h_2(1) = \frac{\sin(1-4)\pi - \sin(1-4) \cdot 1.2}{\pi(1-4)} = 0.0469$$

$$\therefore h_2(2) = \frac{\sin(2-4)\pi - \sin(2-4) \cdot 1.2}{\pi(2-4)} = -0.1075$$

$$\therefore h_2(3) = \frac{\sin(3-4)\pi - \sin(3-4) \cdot 1.2}{\pi(3-4)} = -0.2966$$

$$\therefore h_2(4) = 1 - \frac{1.2}{\pi} = 0.618$$

$$\therefore h_2(5) = \frac{\sin(5-4)\pi - \sin(5-4) \cdot 1.2}{\pi(5-4)} = -0.29$$

$$\therefore h_2(6) = \frac{\sin(6-4)\pi - \sin(6-4) \cdot 1.2}{\pi(6-4)} = -0.1075$$

$$\therefore h_1(7) = \frac{-\sin(7-4)\pi - \sin(7-4)1.2}{\pi(7-4)} = 0.0469$$

$$\therefore h_2(8) = \frac{-\sin(8-4)\pi - \sin(8-4)1.2}{\pi(8-4)} = 0.0792$$

For Hamming window:

$$w(n) = 0.54 - 0.46 \cos \frac{2\pi n}{N-1}$$

$$\therefore w(0) = 0.54 - 0.46 \cos \frac{2\pi \cdot 0}{8} = 0.08$$

$$\therefore w(1) = 0.54 - 0.46 \cos \frac{2\pi \cdot 1}{8} = 0.2147$$

$$\therefore w(2) = 0.54 - 0.46 \cos \frac{2\pi \cdot 2}{8} = 0.54$$

$$\therefore w(3) = 0.865$$

$$\therefore w(4) = 1$$

$$\therefore w(5) = 0.865$$

$$\therefore w(6) = 0.54$$

$$\therefore w(7) = 0.2147$$

$$\therefore w(8) = 0.08$$

$$\therefore h(0) = 0.0792 \times 0.18 = 0.0063$$

$$\therefore h(1) = 0.0969 \times 0.2147 = 0.0108$$

$$\therefore h(2) = -0.1075 \times 0.54 = -0.0588$$

$$\therefore h(3) = -0.2966 \times 0.865 = -0.2566$$

$$\therefore h(4) = 0.618 \times 1 = 0.618$$

$$\therefore h(5) = -0.2966 \times 0.865 = -0.2566$$

$$\therefore h(6) = -0.1075 \times 0.54 = -0.0588$$

$$\therefore h(7) = 0.0969 \times 0.2147 = 0.0108$$

$$\therefore h(8) = 0.0792 \times 0.88 = 0.0063$$

∴ Frequency response:

$$H(w) = \sum_{n=0}^8 h(n) e^{-jwn}$$

$$\begin{aligned} &= h(0) + h(1)e^{-jw} + h(2)(e^{-2jw} + h(3)e^{-3jw}) \\ &+ h(4)e^{-4jw} + h(5)e^{-5jw} + h(6)e^{-6jw} + h(7)e^{-7jw} \\ &+ h(8)e^{-8jw} \end{aligned}$$

$$= e^{-4jw} \left[ h(4) + 2h(3) \cos w + 2h(2) \cos 2w \right. \\ \left. + 2h(1) \cos 3w + 2h(0) \cos 4w \right]$$

$$= e^{-4jw} \left[ 0.618 - 0.5132 \cos w - 0.116 \cos 2w + \right. \\ \left. 0.0200033w + 0.012 \cos 4w \right]$$

$\therefore$  Transfer function:

$$H(z) = \sum_0^8 h(n) z^{-n}$$

$$= h(0) + h(1) z^{-1} + h(2) z^{-2} + h(3) z^{-3} +$$

$$h(4) z^{-4} + h(5) z^{-5} + h(6) z^{-6} + h(7) z^{-7} +$$

$$h(8) z^{-8}$$

$$= z^{-4} [h(4) + \dots ]$$

$z^{-4} \rightarrow 202.273$

An

# DSP

## Chap-19

9.12 :- Band Pass filter using Hanning window.

Frequency range 1 to 2 rad/sec and  $N=5$

Soln: For Band Pass filter:

$$H_d(\omega) = \begin{cases} e^{-j\omega\alpha}; & -w_{c_2} \leq \omega \leq w_{c_1} \\ e^{-j\omega\alpha}; & w_{c_1} \leq \omega \leq w_{c_2} \\ 0; & \text{otherwise} \end{cases}$$

$$\therefore h_d(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(\omega) e^{j\omega n} d\omega$$

$$= \frac{1}{2\pi} \int_{-w_{c_2}}^{-w_{c_1}} e^{-j\omega\alpha} \cdot e^{j\omega n} d\omega +$$

$$= \frac{1}{2\pi} \int_{w_{c_1}}^{w_{c_2}} e^{-j\omega\alpha} \cdot e^{j\omega n} d\omega$$

$$= \frac{1}{2\pi} \int_{-w_{c_2}}^{-w_{c_1}} e^{jw(n-\alpha)} dw + \frac{1}{2\pi} \int_{+w_{c_1}}^{w_{c_2}} e^{jw(n-\alpha)} dw$$

$$= \frac{1}{2\pi} \left[ \frac{e^{jw(n-\alpha)}}{j(n-\alpha)} \right]_{-w_{c_2}}^{-w_{c_1}} + \frac{1}{2\pi} \left[ \frac{e^{jw(n-\alpha)}}{j(n-\alpha)} \right]_{+w_{c_1}}^{w_{c_2}}$$

$$= \frac{1}{\pi(n-\alpha)} \cdot \left[ \frac{\frac{1}{e} - \frac{-j(n-\alpha)w_{c_1} - j(n-\alpha)w_{c_2}}{e}}{2j} \right] + \frac{j(n-\alpha)w_{c_1} - j(n-\alpha)w_{c_2}}{e}$$

$$= \frac{1}{\pi(n-\alpha)} \left[ \frac{\frac{1}{e} - \frac{-j(n-\alpha)w_{c_1} - j(n-\alpha)w_{c_2}}{e}}{2j} \right]$$

$$= \frac{1}{\pi(n-\alpha)} \left[ \sin(n-\alpha)w_{c_2} - \sin(n-\alpha)w_{c_1} \right]$$

$$\Rightarrow h_d(n) = \frac{\sin(n-\alpha) \cdot 2 - \sin(n-\alpha)}{\pi(n-\alpha)}$$

For  $n=2$ ;

$$h_d(2) = \lim_{n \rightarrow 2} \frac{\sin(n-\alpha) \cdot 2 - \sin(n-\alpha)}{\pi(n-\alpha)}$$

$$= \frac{2-1}{\pi} = \frac{1}{\pi}$$

$$\therefore \alpha = \frac{n-1}{2} = \frac{5-1}{2} = 2$$

$$\therefore h_d(0) = \frac{\sin(0-2) \cdot 2 - \sin(0-2)}{\pi(0-2)} = -0.2651$$

$$\therefore h_d(1) = 0.0215$$

$$\therefore h_d(2) = 0.3183$$

$$\therefore h_d(3) = 0.0215$$

$$\therefore h_d(4) = -0.2651$$

For hanning window;

$$w(n) = \begin{cases} 0.5 - 0.5 \cos \frac{\pi \cdot 2\pi n}{N-1}, & 0 \leq n \leq N-1 \\ 0, & \text{otherwise.} \end{cases}$$

$$\therefore w(0) = 0.5 - 0.5 \cos \frac{2\pi \cdot 0}{4} = 0$$

$$\therefore w(1) = 0.5$$

$$\therefore w(2) = 1$$

$$\therefore w(3) = 0.5$$

$$\therefore w(4) = 0$$

$$\therefore h(0) = h_d(0) \times w(0) = (-0.2651) \times (0) = 0$$

$$\therefore h(1) = \dots \quad h(2) = \dots \quad h(3) = \dots \quad h(4) = \dots$$

Using same method:

Frequency response:  $H(w) = \sum_{n=0}^N h(n) e^{-jwn}$

Transfer function  $H(z) = \sum_{n=0}^{N-1} h(n) z^{-n}$

9.13: Band-stop filter using rectangular frequency 1 to 2 and  $N=7$

Sol: For Band-stop filter,

$$H_d(\omega) = \begin{cases} e^{-j\omega\alpha} & -\pi \leq \omega \leq \omega_{c2} \\ e^{\beta} & -\omega_{c1} \leq \omega \leq \omega_{c1} \\ e^{-j\omega\alpha} & -\omega_{c2} \leq \omega \leq \pi \\ 0 & \text{otherwise} \end{cases}$$

$$\begin{aligned} h_d(n) &= \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(\omega) \cdot e^{j\omega n} d\omega \\ &= \frac{1}{2\pi} \left[ \int_{-\pi}^{\omega_{c2}} e^{-j\omega(n-\alpha)} d\omega + \int_{\omega_{c1}}^{\omega_{c1}} e^{\beta} d\omega + \int_{-\omega_{c2}}^{\pi} e^{-j\omega(n-\alpha)} d\omega \right] \\ &\quad + \frac{1}{2\pi} \int_{\omega_{c2}}^{\pi} e^{j\omega(n-\alpha)} d\omega. \end{aligned}$$

$$= \frac{1}{2\pi} \left[ \frac{e^{jw(n-\alpha)}}{j(n-\alpha)} \right]_{-\pi}^{+\pi} + \frac{1}{2\pi} \left[ \frac{e^{jw(n-\alpha)}}{j(n-\alpha)} \right]_{-\pi}^{+\pi} +$$

$$= \frac{1}{2\pi} \left[ \frac{e^{jw(n-\alpha)}}{j(n-\alpha)} \right]_{-\pi}^{\pi}$$

$$= e^{-j(n-\alpha)w_{C_2}} - e^{-j(n-\alpha)\pi} - e^{j(n-\alpha)w_C} - e^{-j(n-\alpha)w_{C_2}}$$

$$= \frac{1}{\pi(n-\alpha)} \left[ \frac{e^{j(n-\alpha)\pi} - e^{j(n-\alpha)w_C} + e^{j(n-\alpha)w_{C_2}} - e^{-j(n-\alpha)\pi}}{2j} \right]$$

$$= \frac{1}{\pi(n-\alpha)} \left[ \frac{e^{j(n-\alpha)\pi} - e^{-j(n-\alpha)\pi}}{2j} + \frac{e^{j(n-\alpha)w_C} - e^{-j(n-\alpha)w_C}}{2j} \right]$$

$$\sin(n-\alpha)\pi + \sin(n-\alpha)w_C = \sin(n-\alpha)w_{C_2}$$

$$= \frac{\sin(n-\alpha)\pi + \sin(n-\alpha)w_C}{\pi(n-\alpha)}$$

$$\therefore \Delta = \frac{N-1}{2} = \frac{6}{2} = 3$$

$$w_{C_2} = 2; \quad w_C = 1$$

For  $\omega \rightarrow n = \infty$

$$h_d(\omega) = \lim_{n \rightarrow \infty} \frac{\sin(n-\omega)\pi + \sin(n-\omega) - \sin(n-\omega) \cdot 2}{\pi(n-\omega)}$$
$$= \frac{\pi + 1 - 2}{\pi} = \frac{\pi - 1}{\pi} = 0.6816$$

$$\therefore h_d(0) = \frac{0.0462}{0.0996} \quad \therefore h_d(1) = -0.0216$$

$$\therefore h_d(1) = 0.2651 \quad \therefore h_d(5) = 0.2651$$

$$\therefore h_d(2) = -0.0216 \quad \therefore h_d(6) = 0.0996$$

$$\therefore h_d(3) = 0.6816$$

$$\therefore h(n) = w(n) h_d(n);$$

$$\therefore h(n) = h_d(n);$$

$$\therefore \text{Frequency response } H(\omega) = \sum_{n=0}^6 h(n) e^{-j\omega n}$$

$$\therefore \text{Transfer function } H(z) = \sum_{n=0}^6 h(n) z^{-n}$$

$$z = e^{j\omega} \quad \bar{z} = e^{-j\omega}$$

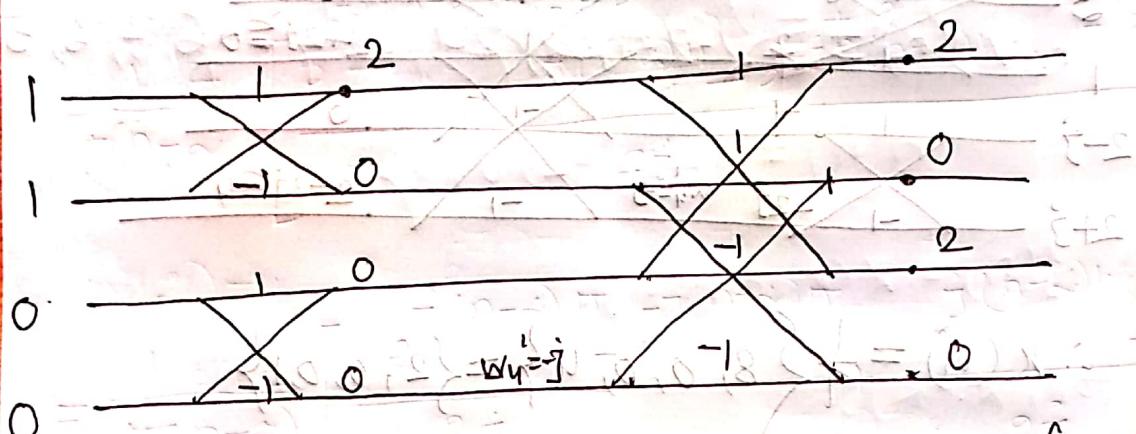
Ans

## Question

2017

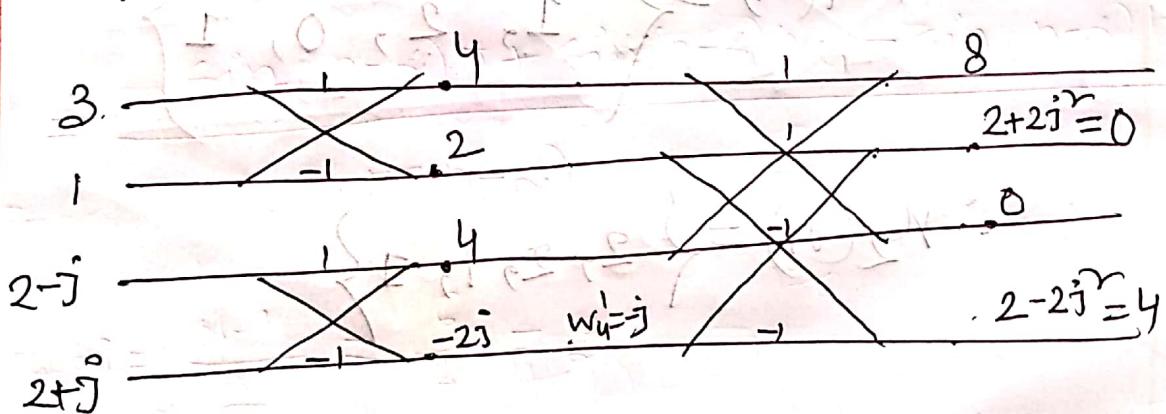
3(c): Find IDFT using DIT FFT

$$X(k) = \{1, 0, 1, 0\} \Rightarrow X^*(k) = \{1, 0, 1, 0\}$$



$$\therefore X(n) = \frac{1}{4} \{ 2, 0, 2, 0 \} = \left\{ \frac{1}{2}, 0, \frac{1}{2}, 0 \right\}$$

$$X(k) = \{3, 2+j, 1, 2-j\} \Rightarrow X^*(k) = \{3, 2-j, 1, 2+j\}$$



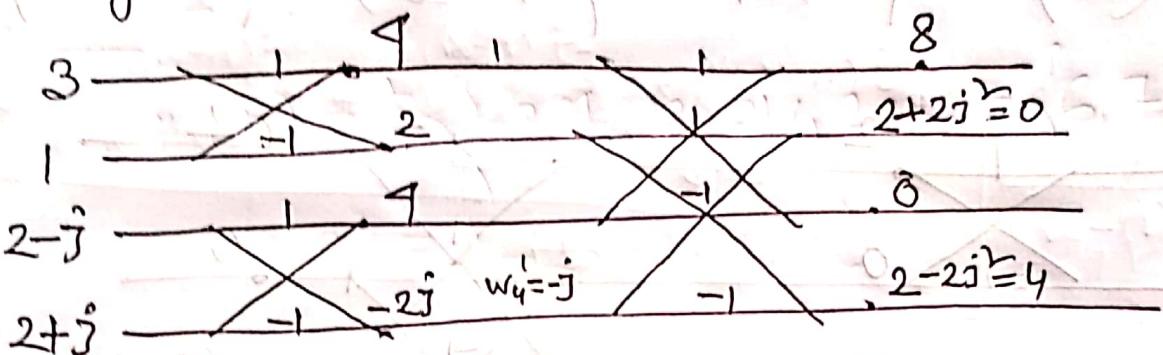
$$\therefore X(n) = \frac{1}{4} \{ 8, 0, 0, 4 \} = \{ 2, 0, 0, 1 \} \quad \underline{\text{Ans}}$$

2016

6(b)  $X(k) = \{3, 2+j, 1, 2-j\}$  Find IDFT using DIT

$$X^*(k) = \{3, 2-j, 1, 2+j\}$$

using DIT:



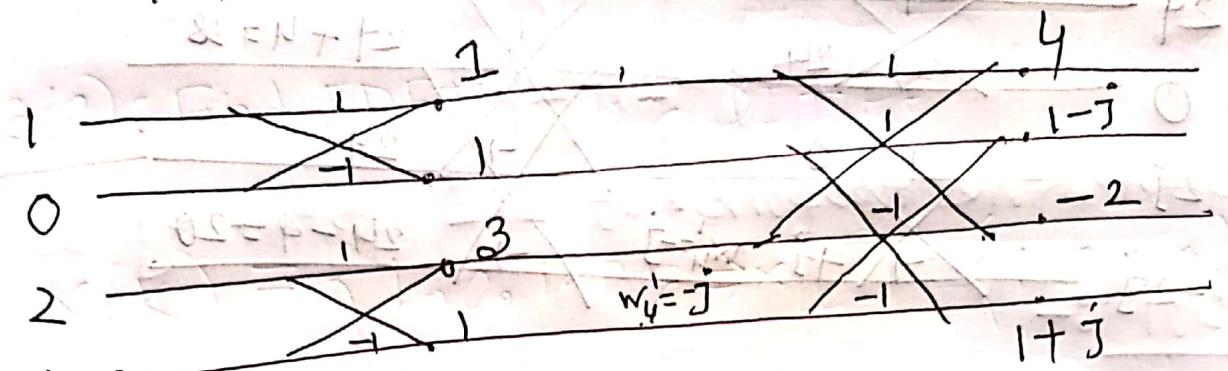
$$\therefore h(n) = \frac{1}{4} \{8, 0, 0, 4\} = \{2, 0, 0, 1\}$$

6(c):  $x_1(n) = \delta(n) + 2\delta(n-1) + \delta(n-3)$

$$\Rightarrow h_1(n) = \{1, 2, 0, 1\}$$

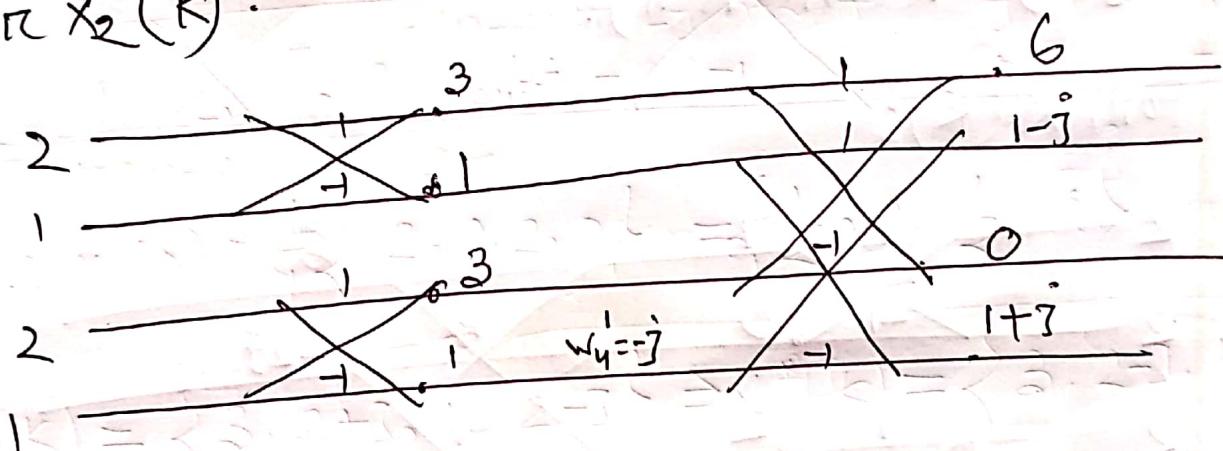
$$\therefore h_2(n) = \{2, 2, 1, 1\}$$

For  $x_1(k)$ : using DIT



$$\therefore x_1(k) = \{4, 1-j, -2, 1+j\}$$

For  $x_2(k)$ :



$$x_2(k) = \{3, -1-j, 0, 1+j\}$$

$$\therefore x(k) = x_1(k) * x_2(k)$$

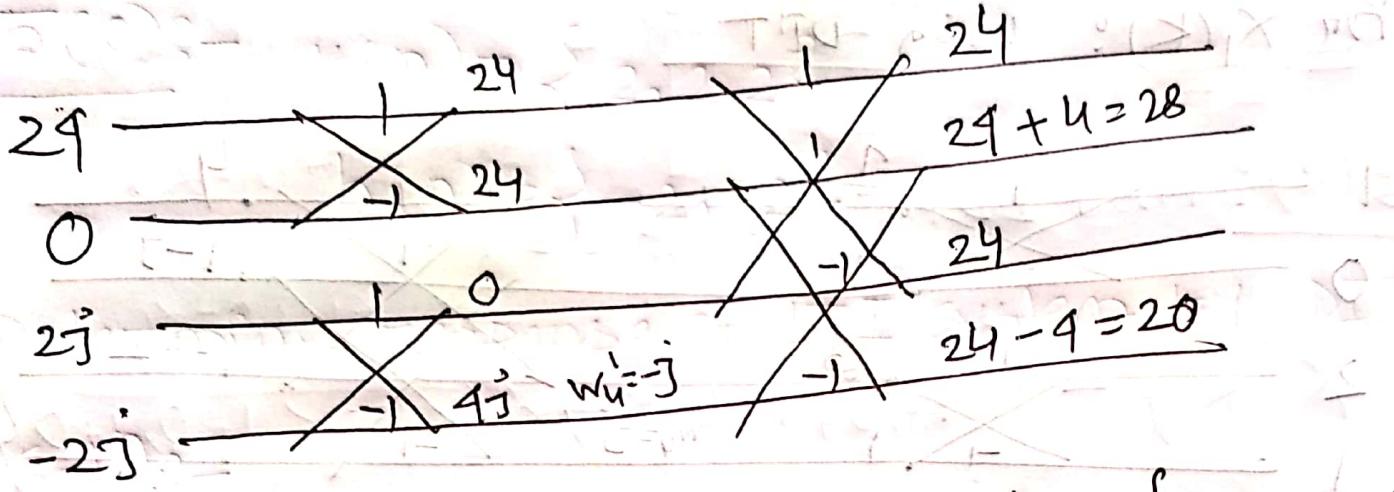
$$= \{24, -2j, 0, 2j\}$$

$$\begin{aligned} (-j)(1-j) \\ = 1 - j - j + j^2 \\ = -2j \end{aligned}$$

$$\begin{aligned} (1+j)(1+j) \\ = (1+j)^2 \\ = 1+2j+j^2 \\ = 2j \end{aligned}$$

$$\therefore x^*(k) = \{24, 2j, 0, -2j\}$$

$$= 2j$$



$$\therefore \chi(5) = \frac{1}{4} \{ 24, 28, 24, 20 \} = \{ 6, 7, 6, 5 \}$$

Ans

(2)  $\text{Ex 5.10}$

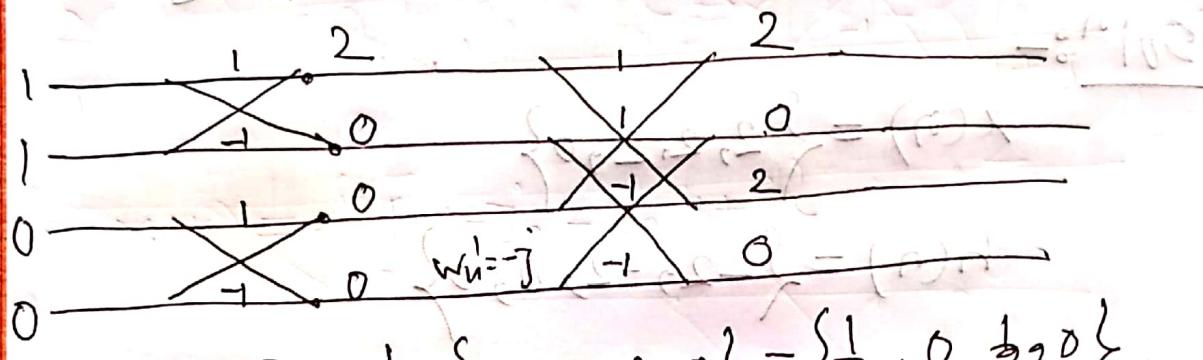
(3)  $\text{Ex}$

2076 (BL)

7(b) :- IDFT by DIT FFT ( $\omega = e^{j\pi/4}$ )

$$X(k) = \{1, 0, 1, 0\}$$

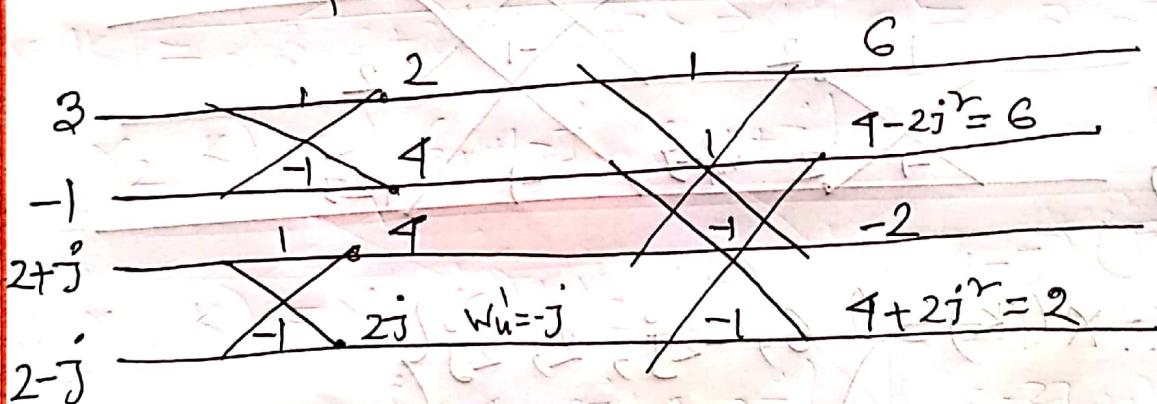
$$X^*(k) = \{1, 0, 1, 0\}$$



$$\therefore x(n) = \frac{1}{4} \{ 1/2, 0, 1/2, 0 \} = \{ \frac{1}{2}, 0, \frac{1}{2}, 0 \}$$

$$X(k) = \{3, 2-j, -1, 2+j\}$$

$$X^*(k) = \{3, 2+j, -1, 2-j\}$$



$$\therefore x(n) = \frac{1}{4} \{ 6, 6, -2, 2 \}$$

$$= \left\{ \frac{3}{2}, \frac{3}{2}, -\frac{1}{2}, \frac{1}{2} \right\}$$

$$5(c):- \quad x(n) = \{2, 2, 2\}$$

$$h(n) = \{-2, -2, 0, 0\}$$

Determine response of LTI system by DIT;

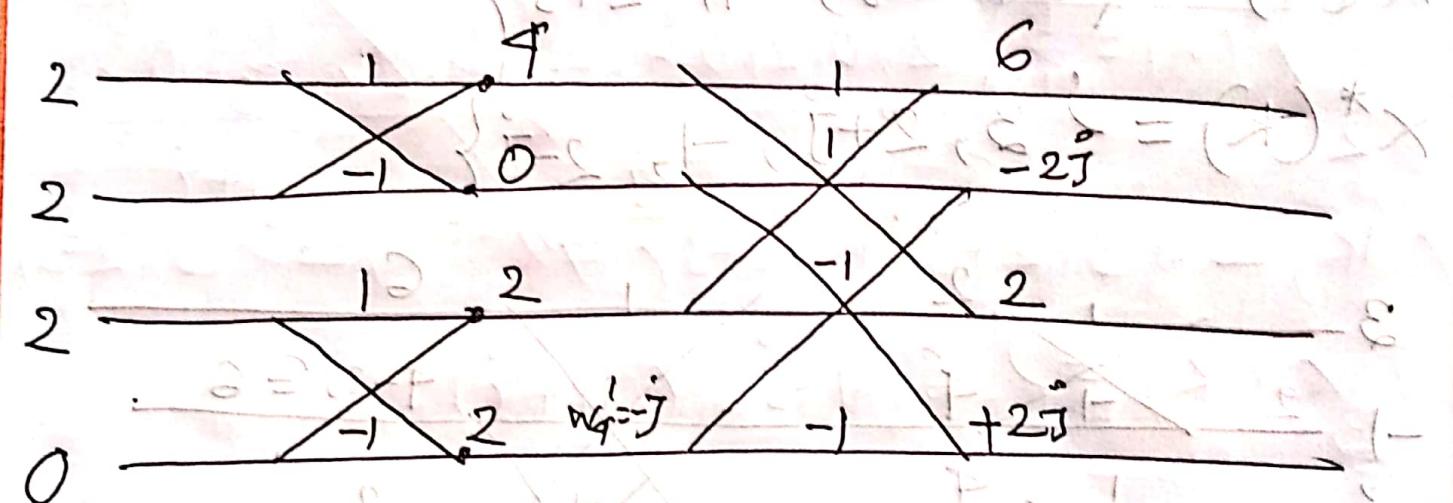
Sol:-

$$x(n) = \{2, 2, 2, 0\}$$

$$h(n) = \{-2, -2, 0, 0\}$$

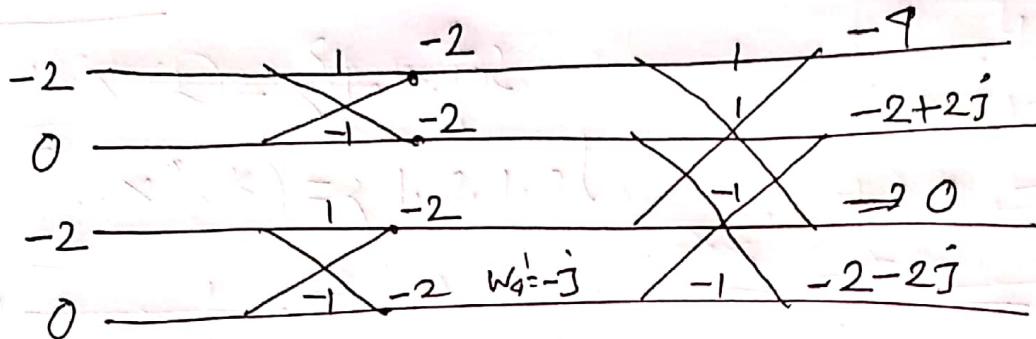
$$y(n) = x(n) * h(n)$$

For,  $x(k)$ :



$$\therefore x(k) = \{6, -2j, 2, 2j\}$$

For  $H(k)$ :

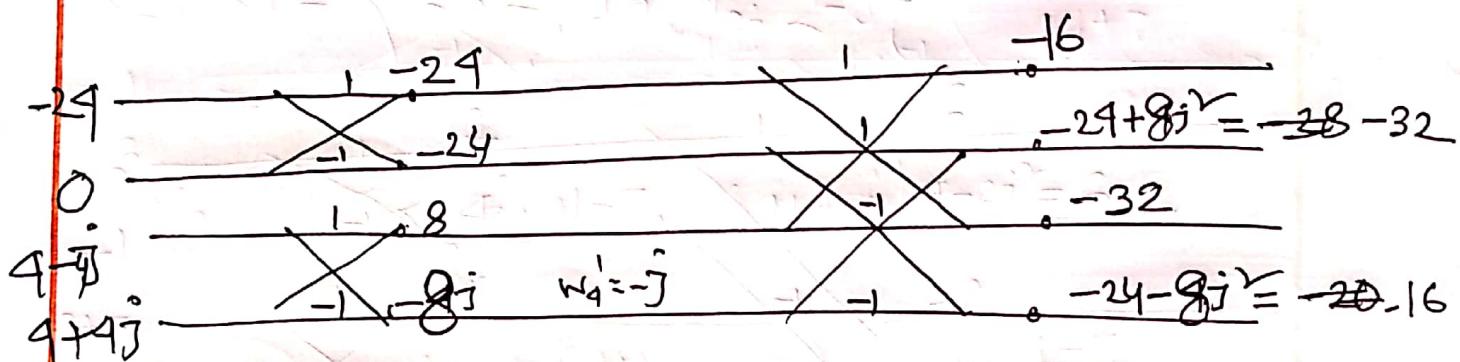


$$\cancel{H(k)} = \{-4, -2+2j, 0, -2-2j\}$$

$$\therefore Y(k) = x(k) * H(k)$$

$$= \{-24, +4j+4, 0, -4j+4\}$$

$$\therefore Y^*(k) = \{-24, -4-4j, 0, 4+4j\}$$



$$\therefore Y(k) = \frac{1}{4} \{-16, -32, -32, -16\}$$

$$= \{-4, -8, -8, -4\}$$

$$= \{-4, -8, -8, -4\}$$