# **Agents mathematical modeling**

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Here we focus on the problem of deriving the mathematical model of N identical LTI agents  $S_i$ , i = 1, 2, ..., N described as

• Continuous-time LTI systems:

$$\dot{x}_i(t) = Ax_i(t) + Bu_i(t), \ y_i(t) = Cx_i(t), \ t \in \mathbb{R}^+$$
 (state space description) (1)

$$H(s) = C(sI - A)^{-1}B + D$$
 (transfer function description) (2)

Discrete-time LTI systems:

$$x_i(k+1) = Ax_i(k) + Bu_i(k), \ y_i(k) = Cx_i(k), \ k \in \mathbb{N}$$
 (state space description) (3)

$$H(z) = C(zI - A)^{-1}B + D$$
 (transfer function description) (4)

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#### Which kind of model?

Different approaches can be used to derive the model of a system from available physical insight and/or collected experimental data:

- 1 First-principle modeling
  - derived by applying fundamental principles of Physics
  - ▶ leads to white-box models
  - detailed structure of the equations derived from Physics
  - physical parameters values a-priori known and/or derived by suitable (dedicated) experimental procedures

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#### Which kind of model?

Different approaches can be used to derive the model of a system from available physical insight and/or collected experimental data:

#### 2 System identification

- derived from a set of input-output data experimentally collected, by applying suitable mathematical techniques/algorithms
- ▶ leads to *black-box models*
- structure of the equations based on some (mild) a-priori information/assumptions
- parameters values provided as output of the identification/estimation procedure

parameters do not have (in general) a physical meaning

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#### Which kind of model?

Different approaches can be used to derive the model of a system from available physical insight and/or collected experimental data:

- 3 Mixed approach
  - ▶ leads to gray-box models
  - structure of the equations (partially) from Physics
  - physical parameters values provided as output of the identification/estimation procedure

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#### generalities

- Objective: to obtain a mathematical model from input/output experimentally collected data (measurements)
- Collected data (being the output of a measurement procedure) are (typically) affected by uncertainty (measurements noise/errors)
- Model class/structure selected on the basis of a-priori physical insight (e.g., linearity/nonlinearity, (upper bound) on the system order, etc.)
- Since system identification (SysID) is typically performed using input-output data, the most natural description is by means of transfer function
- State-space models can be obtained from transfer functions by applying basic results on *realization theory*

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#### generalities

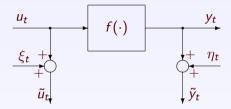
- Since experimental measurement procedure typically provides samples of input-output sequences, we start by considering identification of discrete-time models
- procedures for identifying continuous-time models from sampled input-output sequences will also be discussed
- we assume (without loss of generality) that the (nonlinear) system can be modeled by means of the following *regression form*

$$y(k) = f(y(k-1), y(k-2), \dots, y(k-n), u(k), u(k-1), \dots, u(k-m))$$
 (5)

- *n* is the *system order*
- $m \le n$  (always true for physical systems)
- u(k), y(k), k = 1, 2, ..., H are the (noise-free) samples of the input and output sequences

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#### General black-box EIV set-up



- *Errors-in-variables* (EIV) problems refer to the most general case where both the input and the output collected samples are affected by noise
- a-priori assumption on the model:  $f \in \mathcal{F}$  where  $\mathcal{F}$  is a given class of functions
- a-priori assumption on the noise are available (e.g. statistical distribution, boundedness, etc.)

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#### A naive noiseless example

• Let us consider the following second order LTI discrete-time system (agent):

$$y(k) = f(y(k-1), y(k-2)), u(k), u(k-1), u(k-2))$$

$$= -\theta_1 y(k-1) - \theta_2 y(k-2) + \theta_3 u(k) + \theta_4 u(k-1) + \theta_5 u(k-2)$$
(6)

where  $\theta = [\theta_1, \theta_2, \dots, \theta_5]$ , is the array of parameters to be estimated.

• By introducing the backward shift operator  $q^{-r}(q^{-r}s(t) = s(k-r))$  we can rewrite the equation as

$$y(k) = -\theta_1 q^{-1} y(k) - \theta_2 q^{-2} y(k) + \theta_3 u(k) + \theta_4 q^{-1} u(k) + \theta_5 q^{-2} u(k)$$
 (7)

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#### A naive noiseless example

• Solving the equation in the y(k) we obtain:

$$y(k) = \frac{\theta_3 + \theta_4 q^{-1} + \theta_5 q^{-2}}{1 + \theta_1 q^{-1} + \theta_2 q^{-2}} u(k)$$
 (8)

• By applying properties of the *Z*-transform it is possible to show that the system transfer function G(z) can be obtained by simply replacing  $q^{-1}$  with  $z^{-1}$ :

$$G(z) \doteq \frac{Y(z)}{U(z)} = \frac{\theta_3 z^2 + \theta_4 z + \theta_5}{z^2 + \theta_1 z + \theta_2}$$
 (9)

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#### A naive noiseless example: parameter estimation

• Given *H* experimentally collected input-output samples equation (7) leads to the following system of linear equations:

$$y = A\theta \tag{10}$$

where  $y = [y(3) \ y(4) \ \dots \ y(H)]'$ ,  $\theta = [\theta_1 \ \theta_2 \ \dots \ \theta_5]$ , and

$$A = \begin{bmatrix} -y(2) & -y(1) & u(3) & u(2) & u(1) \\ -y(3) & -y(2) & u(4) & u(3) & u(2) \\ \vdots & \vdots & \ddots & \vdots & \ddots & \vdots \\ -y(H-1) & -y(H-2) & u(H) & u(H-1) & u(H-2) \end{bmatrix}$$
(11)

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#### A naive noiseless example: parameter estimation

- From equations (10) and (11) it appears clear that by collecting H=7 samples we obtain a set of 5 linear equations in 5 unknown
- Assuming that the input sequence u(k), k = 1, 2, ..., 7 is such that A is invertible we can easily compute the parameter  $\theta$  as:

$$\theta = A^{-1}y \tag{12}$$

- How to guarantee that A is invertible?
  - ▶ One possible solution: apply to the system a **random input** sequence *u*

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#### A naive noiseless example: the role of the a-priori information on the system

- In the proposed solution we have started from the a-priori assumption that our agent can be modeled as a LTI discrete-time system of order 2
- This information is crucial since the model cannot be built by only using the collected input-output data
- In fact, the same 7 input-output sample can be perfectly fitted by an infinite number of different models depending of 5 unknown parameters
- For example, the model given by:

$$y(k) = \theta_1 u(k) + \theta_2 u(k)^2 + \theta_3 u(k)^3 + \theta_4 u(k)^4 + \theta_5 u(k)^5$$
 (13)

provided that the corresponding matrix A is still invertible.

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#### Noise effect

- Let us know assume a more realistic situation where the collected data are corrupted by noise
- How can we attenuate the effect of the noise on the computed (estimated) parameter values?
  - ldea: let's collected more data (H >> 2n + 1, n = system order)
- Matrix A in equation (11) is no more square (no more invertible)
- Naive idea: rely on the pseudo-inverse A\* of A

$$\hat{\theta} = (A^T A)^{-1} A^T y \tag{14}$$

where 
$$A^* = (A^T A)^{-1} A^T$$

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#### Least-square approach

• The parameter estimate  $\hat{\theta}$  computed in equation (14) is, in fact, the so-called Least-square solution  $\theta_{LS}$  of the system of equation in (10) and (11), i.e.

$$\theta_{LS} = \arg\min_{\theta} = \|y - A\theta\|_2 \tag{15}$$

- In the context of system identification (but not only)  $\theta_{LS}$  as what we refer to as the *Least square estimate* of the system parameter vector.
- Computation burden of the Least square algorithms is quite low also for large H
- The Least square estimate can be (also) computed recursively (i.e. online).

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### Least square estimate: Consistency property

- The LS estimate enjoy an interesting property if the following assumptions are satisfied:
  - 1 The effect of the uncertainties corrupting the collected data can be taken into account by introducing an additive term *e* (called *equation error*) as follows:

$$y(k) = -\theta_1 y(k-1) - \theta_2 y(k-2) + \theta_3 u(k) + \theta_4 u(k-1) + \theta_5 u(k-2) + e(k)$$
 (16)

**2** e(k), k = 1, 2, ..., H are independent and identically distributed (iid) random variables (typically it is assumed that e can be modeled as a white, zero-mean Gaussian noise)

#### Consistency property

Under assumptions 1 and 2:

$$\lim_{H\to\infty} E[\theta_{LS}] = \theta$$

where  $E[\cdot]$  is the expected value

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#### Least square estimate: limits of application

 What are the performance of the LS estimator if assumptions 1 and 2 are not satisfied? (⇒ Lab activity 1, next week)

What can be done if assumptions 1 and 2 are not satisfied? (⇒ Set-membership identification/estimation)

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