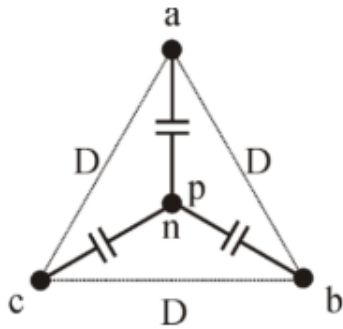


# Introduction to Electrical Power Engineering

## Final Exam

1. (20) Assume that 1. conductors are equally spaced,  $D$ , and have equal radii  $r$ . 2.  $q_a + q_b + q_c = 0$  ( $c_a = c_b = c_c = c$ ,  $v_a + v_b + v_c = 0$ ) 3.  $i_a + i_b + i_c = 0$ . Find (a)  $c = ?$  (b)  $\lambda_a = ?$



### 1. a.

Assume that

1. Conductors are equally spaced,  $D$ , and have equal radii  $r$ .
2.  $q_a + q_b + q_c = 0$  ( $c_a = c_b = c_c = c$ ,  $v_a + v_b + v_c = 0$ ).

$$v_a = \frac{1}{2\pi\epsilon} \left( q_a \ln \frac{1}{r} + q_b \ln \frac{1}{D} + q_c \ln \frac{1}{D} \right) = \frac{1}{2\pi\epsilon} \left( q_a \ln \frac{1}{r} - q_a \ln \frac{1}{D} \right) = \frac{1}{2\pi\epsilon} \left( q_a \ln \frac{D}{r} \right)$$

$$C = \frac{q}{v}, \quad c_a = c_b = c_c = c = \frac{2\pi\epsilon}{\ln(D/r)} \quad (\text{F/m}) \text{ to neutral}$$

$$c_a = c_b = c_c = 2\pi\epsilon \ln \frac{D_m}{r} \text{ for one line transposition}$$

$$c_a = c_b = c_c = 2\pi\epsilon \ln \frac{D_m}{R_{GMR}} \text{ for conductor bundling transposition}$$

$$D_m = (D_{12}D_{23}D_{13})^{1/3}$$

$$R_b^c = R_{GMR} = (rd_{12}d_{13}d_{1b})^{1/b}, \quad b > 1; \quad R_b^c = R_{GMR} = r, \quad \text{when } b = 1$$


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## 1. b.

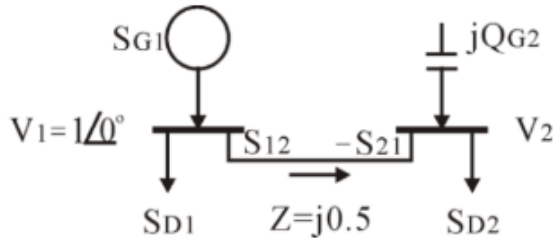
Assume that

1. Conductors equally spaced  $D$  and have equal radii  $r$ .
2.  $i_a + i_b + i_c = 0$ .

$$\begin{aligned}
 \lambda_a &= \left( \frac{\mu_0}{2\pi} \right) \{ i_a [\mu_r/4 + \ln(1/r)] + i_b \ln(1/D) + i_c \ln(1/D) \} \\
 &= \left( \frac{\mu_0}{2\pi} \right) \{ i_a [\mu_r/4 + \ln(1/r)] - i_a \ln(1/D) \} \\
 &= \left( \frac{\mu_0}{2\pi} \right) [\mu_r/4 + \ln(1/r) - \ln(1/D)] \times i_a \\
 &= \left( \frac{\mu_0}{2\pi} \right) \left[ \ln e^{(\mu_r/4)} + \ln(1/r) - \ln(1/D) \right] \times i_a \\
 &= \left( \frac{\mu_0}{2\pi} \right) \left[ \ln(1/r e^{-(\mu_r/4)}) - \ln(1/D) \right] \times i_a \\
 &= \left( \frac{\mu_0}{2\pi} \right) [\ln(1/r') - \ln(1/D)] \times i_a \\
 &= \left( \frac{\mu_0}{2\pi} \right) [\ln(D/r')] \times i_a \\
 &= l_a \times i_a
 \end{aligned}$$


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3. (20)  $V_1 = 1 \angle 0^\circ$ ,  $jQ_{G2} = j1.0$ ,  $Z_L = j0.5$ ,  $S_{D2} = P_{D2} + j1.0$ . Find  $S_1$  and  $V_2$ . We consider the solution as a function of  $P_{D2}$  for  $P_{D2} \geq 0$ . (a) If  $P_{D2} > 1 \Rightarrow V_2 = ?$  (b) If  $P_{D2} = 1 \Rightarrow V_2 = ?$  (c) If  $P_{D2} = 0.5 \Rightarrow V_2 = ?$



## 2. a.

$z = r + j\omega l$  = series impedance per meter

$y = g + j\omega c$  = shunt admittance per meter to neutral

$$\frac{dV}{dx} = zI ; \frac{dI}{dx} = yV$$

$$\frac{d^2V}{dx^2} = yzV = \gamma^2 V ; \frac{d^2I}{dx^2} = yzI = \gamma^2 I$$

Propagation constant:  $\gamma = (yz)^{0.5} = \alpha + j\beta$

$$V = k_1 e^{\gamma x} + k_2 e^{-\gamma x} = (k_1 + k_2) \frac{(e^{\gamma x} + e^{-\gamma x})}{2} + (k_1 - k_2) \frac{(e^{\gamma x} - e^{-\gamma x})}{2}$$

$$V = K_1 \cosh \gamma x + K_2 \sinh \gamma x ; \quad K_1 = (k_1 + k_2) , \quad K_2 = (k_1 - k_2)$$

$$\frac{dV}{dx} = K_1 \gamma \sinh \gamma x + K_2 \gamma \cosh \gamma x \quad (4.8)$$

When  $x = 0$ ,  $V = V_2 \Rightarrow V = k_1 + k_2 = K_1 = V_2$ ,

When  $x = 0$ ,  $I = I_2 \Rightarrow \frac{dV(0)}{dx} = zI_2 = K_2 \gamma = (z/\gamma)I_2 = (z/y)^{0.5} I_2$

$Z_c = (z/y)^{0.5}$  (characteristic impedance)

$$V = V_2 \cosh \gamma x + Z_c I_2 \sinh \gamma x ; \quad I = I_2 \cosh \gamma x + (V_2/Z_c) \sinh \gamma x$$

**When**  $x = l$ ,  $V = V_1$ ,  $I = I_1 \Rightarrow$

$$V_1 = V_2 \cosh \gamma l + Z_c I_2 \sinh \gamma l;$$

$$I_1 = I_2 \cosh \gamma l + (V_2/Z_c) \sinh \gamma l$$


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## 2. b.

$$z = r + j\omega l = 0.169 + j0.789 = 0.807 \angle 77.9^\circ \Omega/\text{mile}$$

$$y = g + j\omega c = j5.38 \times 10^{-6} = 5.38 \times 10^{-6} \angle 90^\circ \text{ mho/mile}$$

$$\Rightarrow Z_c = (z/y)^{0.5} = 387.3 \angle -6.05^\circ \Omega$$

$$\Rightarrow \gamma l = 225(yz)^{0.5} = 0.4688 \angle 83.95^\circ = 0.0494 + j0.466$$

$$2 \sinh \gamma l = e^{\gamma l} - e^{-\gamma l} = e^{0.0494} e^{j0.466} - e^{-0.0494} e^{-j0.466} = 1.051 \angle 0.466 \text{ rad} - 0.952 \angle -0.466$$

$$\sinh \gamma l = 0.452 \angle 84.4^\circ$$

$$2 \cosh \gamma l = e^{\gamma l} + e^{-\gamma l} = 1.051 \angle 0.466 \text{ rad} + 0.952 \angle -0.466 \text{ rad} = 1.790 \angle 1.42^\circ$$

$$\cosh \gamma l = 0.895 \angle 1.42^\circ$$

$$|V_2| = 132 \times 10^3 / \sqrt{3} = 76.2 \text{ kV} \Rightarrow V_2 = 76.2 \angle 0^\circ \text{ kV}$$

$$1\Phi P_{\text{load}} = 40 \times 10^6 / 3 = 13.33 \text{ MW} = 0.95 |V_2| |I_2| \Rightarrow |I_2| = 184.1, I_2 = 184.1 \angle -18.195^\circ$$

$$V_1 = V_2 \cosh \gamma l + Z_c I_2 \sinh \gamma l \Rightarrow V_1 = 89.28 \angle 19.39^\circ \text{ kV} \Rightarrow V_{LL} = \sqrt{3} \times 89.28 = 154.64 \text{ kV}$$

$$I_1 = I_2 \cosh \gamma l + (V_2/Z_c) \sinh \gamma l = 162.42 \angle 14.76^\circ \text{ A}$$

$$P_{12} = \text{real}(V_1 I_1^*) = 89.28 \times 10^3 \times 162.42 \cos(19.39^\circ - 14.76^\circ) = 14.45 \text{ MW}$$

$$\text{Efficiency} = \frac{13.33}{14.45} = 0.92 = 92\%$$


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2. (20) A 60-Hz 138-kV 3 $\Phi$  transmission line is 225 mile long. The distributed line parameters are  $r = 0.169 \Omega/\text{mile}$ ,  $l = 2.093 \text{ mH/mile}$ ,  $c = 0.01427 \mu\text{F/mile}$ ,  $g = 0$ . The transmission line delivers 40 MW at 132 kV with 95% power factor lagging. (a) Find the sending-end voltage and current. (b) Find the transmission line efficiency.

### 3. a. b. c.

In this case, the capacitor injects a specified power, while the voltage is not controlled. Thus bus2 is a P, Q bus. In fact,

$$S_2 = S_{21} = S_{G2} - S_{D2} = j1.0 - (P_{D2} + j1.0) = -P_{D2}$$

$$S_{12} = |V_1|^2 e^{j\angle Z} / |Z| - |V_1| |V_2| e^{j\angle Z} e^{j\theta_{12}} / |Z|$$

$$S_{21} = |V_2|^2 e^{j\angle Z} / |Z| - |V_1| |V_2| e^{j\angle Z} e^{-j\theta_{12}} / |Z|$$

$$R = 0, \quad Z = jX = j0.5 \Rightarrow \angle Z = 90^\circ, \quad e^{j\angle Z} = j, \quad |V_1| = 1$$

$$S_{21} = j2|V_2|^2 - j2|V_2|e^{-j\theta_{12}} = -P_{D2}$$

We draw a receiving-end circle:

$$(2|V_2|^2)^2 + (P_{D2})^2 = (2|V_2|)^2 \Rightarrow 4|V_2|^4 - 4|V_2|^2 + (P_{D2})^2 = 0$$

$$|V_2|^2 = [1 \pm (1 - (P_{D2})^2)^{0.5}] / 2$$

**If  $P_{D2} > 1 \Rightarrow |V_2|$  has no solution,**

**If  $P_{D2} = 1 \Rightarrow |V_2| = 0.707, \theta_{12} = 45^\circ$**

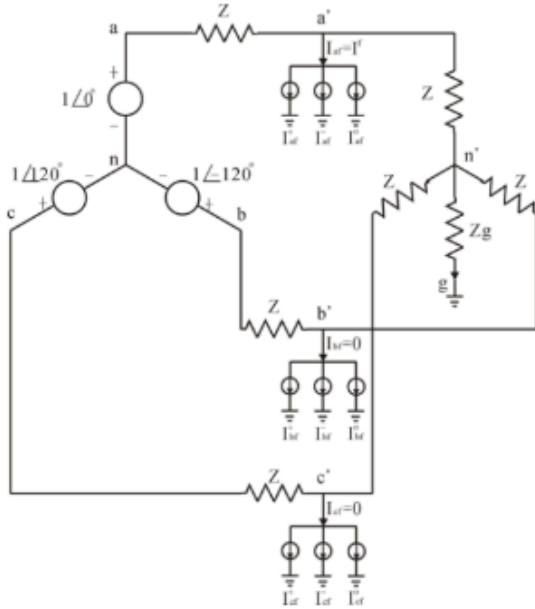
**If  $0 \leq P_{D2} \leq 1 \Rightarrow |V_2|$  has two solutions**

We can find  $\theta_{12}$  and  $S_1 = S_{12}$

**If  $P_{D2} = 0.5 \Rightarrow V_2 = 0.97 \angle -15^\circ$  (OK) and  $V_2 = 0.26 \angle -75^\circ$  (not OK)**

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4. (20) Fault current:  $I^f = [I_{af} \ I_{bf} \ I_{cf}] = [I^f \ 0 \ 0]$ . Find the symmetrical components of single line-to-ground faults currents.



4.

$$(13.5) \quad [I_{af} \ I_{bf} \ I_{cf}] = A [I_{af}^0 \ I_{af}^+ \ I_{af}^-]$$

$$(13.6) \quad A = \begin{bmatrix} 1 & 1 & 1 \\ 1 & \alpha & \alpha^2 \\ 1 & \alpha^2 & \alpha \end{bmatrix}$$

Fault current:

$$I^f = [I_{af} \ I_{bf} \ I_{cf}] = [I^f \ 0 \ 0]$$

$$I_{af} = I_{af}^0 + I_{af}^+ + I_{af}^- = I^f$$

$$I_{bf} = I_{bf}^0 + I_{bf}^+ + I_{bf}^- = \left(\frac{I^f}{3}\right) [1 \ \alpha^2 \ \alpha] = 0$$

$$I_{cf} = I_{cf}^0 + I_{cf}^+ + I_{cf}^- = \left(\frac{I^f}{3}\right) [1 \ \alpha \ \alpha^2] = 0$$

$$A^{-1} = \left(\frac{1}{3}\right) \begin{bmatrix} 1 & 1 & 1 \\ 1 & \alpha & \alpha^2 \\ 1 & \alpha^2 & \alpha \end{bmatrix}$$

$$\begin{bmatrix} I_{af}^0 & I_{af}^+ & I_{af}^- \end{bmatrix} = A^{-1} \begin{bmatrix} I^f & 0 & 0 \end{bmatrix} = \left( \frac{I^f}{3} \right) \begin{bmatrix} 1 & 1 & 1 \end{bmatrix}$$

$$I_{af}^+ = I_{af}^- = I_{af}^0 = \left( \frac{I^f}{3} \right)$$

Using Superposition:

- Positive Sequence
- Negative Sequence
- Zero Sequence

5.(20)  $v_a = 180 \cos \omega t$ ,  $v_b = 180 \cos(\omega t - 120^\circ)$ ,  $v_c = 180 \cos(\omega t + 120^\circ)$  (a) Find abc Reference Frame to Stationary Reference Frame (b) Stationary Reference Frame to Synchronous Reference Frame

**5. a.**

$$\begin{bmatrix} v_{qs} \\ v_{ds} \end{bmatrix} = \begin{bmatrix} 2/3 & -1/3 & -1/3 \\ 0 & -1/\sqrt{3} & 1/\sqrt{3} \end{bmatrix} \begin{bmatrix} v_a \\ v_b \\ v_c \end{bmatrix}$$

$$\begin{bmatrix} v_a \\ v_b \\ v_c \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ -1/2 & -\sqrt{3}/2 \\ -1/2 & \sqrt{3}/2 \end{bmatrix} \begin{bmatrix} v_{qs} \\ v_{ds} \end{bmatrix}$$

3-phase  $220V_{rms}$  (60Hz, line-to-line),  $\omega = 2\pi f = 120\pi = 377 \text{ rad/sec}$ .

$v_a + v_b + v_c = 0$ , balanced system, ( $v_a + v_b + v_c \neq 0$ , unbalanced system)

$$v_a = 180 \cos \omega t, \quad v_b = 180 \cos(\omega t - 120^\circ), \quad v_c = 180 \cos(\omega t + 120^\circ)$$

**abc Reference Frame to Stationary Reference Frame**

$$v_{qs} = \left( \frac{2}{3} \right) v_a + \left( -\frac{1}{3} \right) v_b + \left( -\frac{1}{3} \right) v_c = \left( \frac{2}{3} \right) v_a + \left( \frac{1}{3} \right) v_a = v_a = 180 \cos \omega t$$

$$v_{ds} = \left(-\frac{1}{\sqrt{3}}\right) v_b + \left(\frac{1}{\sqrt{3}}\right) v_c = \left(-\frac{1}{\sqrt{3}}\right) [v_a + 2v_b] = -180 \sin \omega t$$

Stationary Reference Frame to abc Reference Frame

$$v_a = v_{qs}, \quad v_b = \left(-\frac{1}{2}\right) v_{qs} + \left(-\frac{\sqrt{3}}{2}\right) v_{ds} = 180 \cos(\omega t - 120^\circ)$$

$$v_c = \left(-\frac{1}{2}\right) v_{qs} + \left(\frac{\sqrt{3}}{2}\right) v_{ds} = 180 \cos(\omega t + 120^\circ)$$


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**5. b.**

$$\begin{bmatrix} v_{qe} \\ v_{de} \end{bmatrix} = \begin{bmatrix} \cos \omega t & -\sin \omega t \\ \sin \omega t & \cos \omega t \end{bmatrix} \begin{bmatrix} v_{qs} \\ v_{ds} \end{bmatrix}$$

$$\begin{bmatrix} v_{qs} \\ v_{ds} \end{bmatrix} = \begin{bmatrix} \cos \omega t & \sin \omega t \\ -\sin \omega t & \cos \omega t \end{bmatrix} \begin{bmatrix} v_{qe} \\ v_{de} \end{bmatrix}$$

3-phase 220V<sub>rms</sub> (60Hz, line-to-line),  $\omega = 2\pi f = 120\pi = 377$  rad/sec.

$$v_{qs} = 180 \cos \omega t, \quad v_{ds} = -180 \sin \omega t$$

**Stationary Reference Frame to Synchronous Reference Frame**

$$v_{qe} = \cos \omega t v_{qs} - \sin \omega t v_{ds} = 180$$

$$v_{de} = \sin \omega t v_{qs} + \cos \omega t v_{ds} = 0$$

**Synchronous Reference Frame to Stationary Reference Frame**

$$v_{qs} = \cos \omega t v_{qe} + \sin \omega t v_{de} = 180 \cos \omega t$$

$$v_{ds} = -\sin \omega t v_{qe} + \cos \omega t v_{de} = -180 \sin \omega t$$