

Wavelets, Sparsity and their Applications

Pier Luigi Dragotti

Communications and Signal Processing Group
Imperial College London

Session Six: Wavelets from Iterated Filter-banks

Equivalent Filters and Functions

Let $g_0[n]$ and $g_1[n]$ denote low-pass and high-pass filters, respectively, and, for simplicity, assume that this is an orthogonal filter bank. The equivalent filters $g_0^{(i)}[n]$, $g_1^{(i)}[n]$ after i steps of iteration are given by:

$$G_0^{(i)}(z) = \prod_{k=0}^{i-1} G_0(z^{2^k})$$
$$G_1^{(i)}(z) = G_1(z^{2^{i-1}}) \prod_{k=0}^{i-2} G_0(z^{2^k}).$$

Let us define a continuous-time function associated with $g_0^{(i)}[n]$ and $g_1^{(i)}[n]$ in the following way:

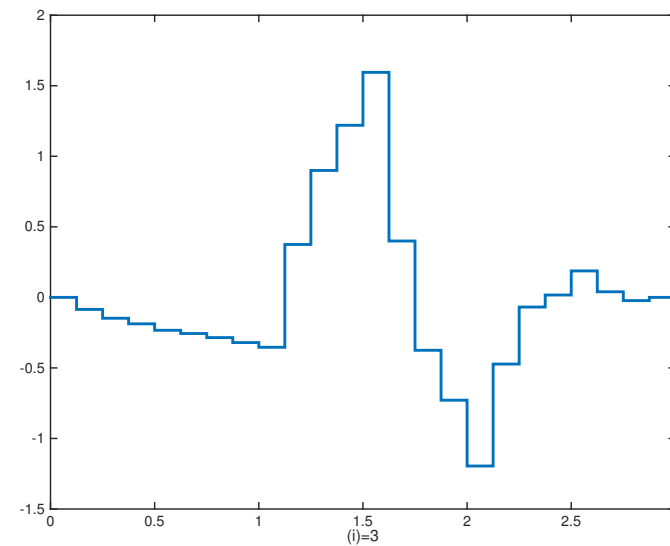
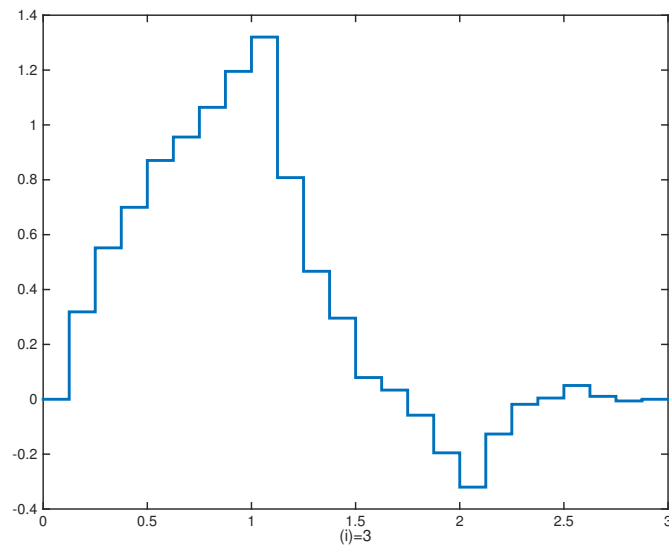
$$\varphi^{(i)}(t) = 2^{i/2} g_0^{(i)}[n], \quad n/2^i \leq t < (n+1)/2^i$$

$$\psi^{(i)}(t) = 2^{i/2} g_1^{(i)}[n], \quad n/2^i \leq t < (n+1)/2^i$$

Equivalent Filters and Functions (example)

$$\varphi^{(i)}(t) = 2^{i/2} g_0^{(i)}[n], \quad n/2^i \leq t < (n+1)/2^i$$

$$\psi^{(i)}(t) = 2^{i/2} g_1^{(i)}[n], \quad n/2^i \leq t < (n+1)/2^i$$



Equivalent Filters and Functions

Since

$$\varphi^{(i)}(t) = \sum_n g_0^{(i)}[n] 2^{i/2} \delta(2^i t - n) * \mathbf{1}[0, 2^{-i}],$$

in the Fourier domain, using $M_0(\omega) = G_0(e^{j\omega})/\sqrt{2}$ and $M_1(\omega) = G_1(e^{j\omega})/\sqrt{2}$, we have that

$$\hat{\varphi}^{(i)}(\omega) = \Theta^{(i)}(\omega) \prod_{k=1}^i M_0\left(\frac{\omega}{2^k}\right)$$

where

$$\Theta^{(i)}(\omega) = e^{-j\omega/2^{i+1}} \frac{\sin(\omega/2^{i+1})}{\omega/2^{i+1}}$$

and

$$\hat{\psi}^{(i)}(\omega) = M_1\left(\frac{\omega}{2}\right) \Theta^{(i)}(\omega) \prod_{k=2}^i M_0\left(\frac{\omega}{2^k}\right).$$

Equivalent Filters and Functions

Now assume that the limit for $i \rightarrow \infty$ of the two functions $\varphi^{(i)}(t)$, $\psi^{(i)}(t)$ exists and is well defined. Let $\varphi(t)$, $\psi(t)$ denote the two limit functions, that is

$$\varphi(t) = \lim_{i \rightarrow \infty} \varphi^{(i)}(t),$$

$$\psi(t) = \lim_{i \rightarrow \infty} \psi^{(i)}(t).$$

In the Fourier domain

$$\hat{\varphi}(\omega) = \lim_{i \rightarrow \infty} \hat{\varphi}^{(i)}(\omega) = \prod_{k=1}^{\infty} M_0 \left(\frac{\omega}{2^k} \right)$$

$$\hat{\psi}(\omega) = \lim_{i \rightarrow \infty} \hat{\psi}^{(i)}(\omega) = M_1 \left(\frac{\omega}{2} \right) \prod_{k=2}^{\infty} M_0 \left(\frac{\omega}{2^k} \right)$$

since $\Theta^{(i)}(\omega)$ tends to 1 as $i \rightarrow \infty$.

Equivalent Filters and Functions

We first need to establish some necessary conditions for the limit to exist.

Theorem 1. *For the limit $\varphi(t) = \lim_{i \rightarrow \infty} \varphi^{(i)}(t)$ to exist, it is necessary that $G_0(e^{j\omega}) = \sqrt{2}$ for $\omega = 0$ and $G_0(e^{j\omega}) = 0$ for $\omega = \pi$.*

Claim: Given that the limits exist, the so obtained functions are indeed a scaling function and a wavelet.

Proof:

We show that $\varphi(t)$ satisfies the criteria of a valid scaling function. First of all it satisfies the two scale equation. In frequency domain the equation can be written as:

$$\varphi(t) = \sqrt{2} \sum_n g_0[n] \varphi(2t - n) \Leftrightarrow \frac{1}{\sqrt{2}} G_0(e^{j\omega/2}) \hat{\varphi}\left(\frac{\omega}{2}\right).$$

By construction

$$\hat{\varphi}(\omega) = \prod_{k=1}^{\infty} M_0\left(\frac{\omega}{2^k}\right) = M_0\left(\frac{\omega}{2}\right) \prod_{k=2}^{\infty} M_0\left(\frac{\omega}{2^k}\right) = M_0\left(\frac{\omega}{2}\right) \hat{\varphi}\left(\frac{\omega}{2}\right) = \frac{1}{\sqrt{2}} G_0(e^{j\omega/2}) \hat{\varphi}\left(\frac{\omega}{2}\right).$$

So the two scale relationship is satisfied.

Equivalent Filters and Functions

Proof (cont'd):

Note that we started with an orthogonal filter bank. Therefore:

$$|G_0(e^{j\omega})|^2 + |G_0(e^{j\omega+\pi})|^2 = 2.$$

Moreover, for a function satisfying a two scale relationship, the Riesz criterion for orthogonality:

$$\sum_l |\hat{\varphi}(\omega + 2k\pi)|^2 = 1$$

is satisfied if and only if

$$|G_0(e^{j\omega})|^2 + |G_0(e^{j\omega+\pi})|^2 = 2.$$

Thus our function satisfies the criterion.

Finally, by using Poisson summation formula, we write partition of unity as follows:

$$\sum_{n=-\infty}^{\infty} \varphi(t - n) = \sum_{k=-\infty}^{\infty} \hat{\varphi}(2\pi k) e^{j2\pi k t} = 1,$$

which is satisfied since we assumed $G_0(1) = \sqrt{2}$ and $G_0(-1) = 0$.

Regularity

Given a filter $G_0(e^{j\omega})$, the limit function $\varphi(t)$ depends on the behaviour of the product

$$\prod_{k=1}^i M_0\left(\frac{\omega}{2^k}\right)$$

as $i \rightarrow \infty$.

Daubechies studied the regularity of iterated filter banks in detail and provided sufficient conditions for regularity.

Factor $M_0(\omega)$ as follows

$$M_0(\omega) = \left(\frac{1 + e^{j\omega}}{2} \right)^N R(\omega).$$

Because of the necessary condition, we know that N must be at least equal to 1. Define B as

$$B = \sup_{\omega \in [0, 2\pi]} |R(\omega)|$$

Regularity

Theorem 2. *If*

$$B < 2^{N-1}$$

then the limit $\lim_{i \rightarrow \infty} \varphi_i(t)$ converges pointwise to a continuous function $\varphi(t)$ with Fourier transform

$$\hat{\varphi}(\omega) = \prod_{k=1}^{\infty} M_0 \left(\frac{\omega}{2^k} \right).$$

Moreover, if

$$B < 2^{N-1-n} \quad n = 1, 2, \dots$$

then $\varphi(t)$ is n -times continuously differentiable.

Regularity

Proof:

- First remember that if a constant K and $\epsilon > 0$ exists such that $|\hat{f}(\omega)| \leq \frac{K}{1+|\omega|^{n+1+\epsilon}}$, then $f(t)$ is n -times continuously differentiable.
- Just need to prove $|\hat{\varphi}(\omega)| \leq \frac{K}{1+|\omega|^{n+1+\epsilon}}$

•

$$\hat{\varphi}(\omega) = \prod_{k=1}^{\infty} \left(\frac{1 + e^{j\omega/2^k}}{2} \right)^N \prod_{k=1}^{\infty} R\left(\frac{\omega}{2^k}\right)$$

- The first product converges to $\text{sinc}\left(\frac{\omega}{2}\right)^N$ which decays like $(1 + |\omega|^{-N})$
- The second product is upper bounded by a constant when $|\omega| < 1$. Therefore for $2^{J-1} < \omega < 2^J$, we have:

$$\prod_{k=1}^{\infty} \left| R\left(\frac{\omega}{2^k}\right) \right| = \prod_{k=1}^J \left| R\left(\frac{\omega}{2^k}\right) \right| \prod_{k=1}^{\infty} \left| R\left(\frac{\omega}{2^k 2^J}\right) \right|$$

Regularity

Proof (cont'd):

- The first term is smaller or equal to B^J while the second term is upper bounded by a constant, so we can write

$$\prod_{k=1}^{\infty} \left| R \left(\frac{\omega}{2^k} \right) \right| \leq c_0 B^J \leq c_1 2^{J(N-1-n-\epsilon)} < c_2 (1 + |\omega|)^{N-1-n-\epsilon},$$

where we have used the fact that $B < 2^{N-1-n}$.

- Putting all together gives us the desired decay.

Properties of the Wavelet series

A key property of the wavelet transform is that of *vanishing moments*. We know that $g_0[n]$ has at least one zero at $\omega = \pi$ and thus $g_1[n]$ has at least one zeros at $\omega = 0$. Since $\hat{\varphi}(0) = 1$ (from the normalization of $M_0(\omega)$), it follows that

$$\int_{-\infty}^{\infty} \psi(t) dt = \hat{\psi}(0) = \underbrace{\frac{G_1(1)}{\sqrt{2}}}_{=0} \hat{\varphi}(0) = 0.$$

In general, if $g_0[n]$ has a zero of order N at π then $\hat{\psi}(\omega)$ has N zeros at $\omega = 0$ and using the moment property of the Fourier transform we have that

$$\int_{-\infty}^{\infty} t^k \psi(t) dt = 0 \quad k = 0, \dots, N - 1. \quad (1)$$

Properties of the Wavelet series

As a consequence of the vanishing moment property, scaling functions reproduce polynomials.



Properties of the Wavelet series

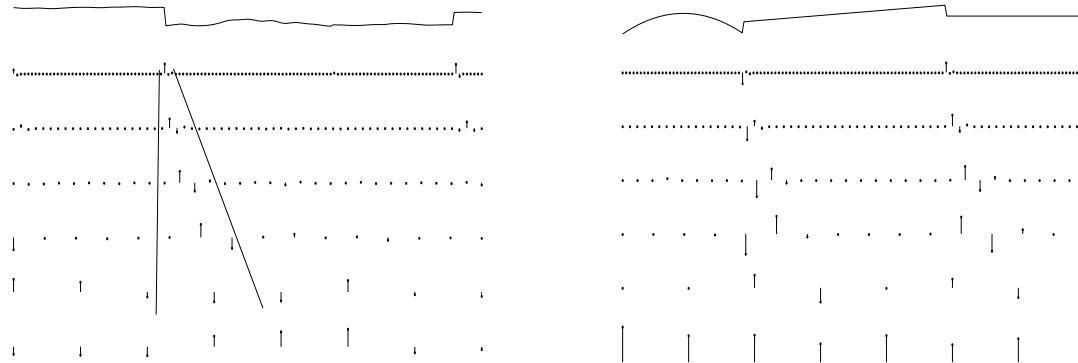
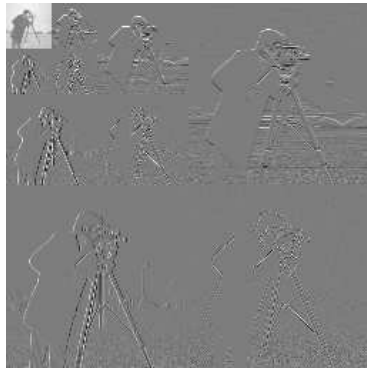
The restriction of $f(t)$ to $[a, b]$ is uniformly Lipschitz $\alpha \geq 0$ over $[a, b]$ if there exists $K > 0$ such that for all $\nu \in [a, b]$ there exists a polynomial $p_\nu(t)$ of degree $m = \lfloor \alpha \rfloor$ such that

$$\forall t \in (a, b), \quad |f(t) - p_\nu(t)| \leq K|t - \nu|^\alpha.$$

Assume that $f(t)$ is uniformly α -Lipshitz around t_0 and that $\psi(t)$ has at least $\lfloor \alpha \rfloor + 1$ vanishing moments. Assume that $\psi(t)$ is of compact support C , then:

$$\begin{aligned} \langle f, \psi_{m,n} \rangle &= \underbrace{\langle p_{t_0}(t), \psi_{m,n}(t) \rangle}_{=0} + \langle \epsilon(t), \psi_{m,n}(t) \rangle \\ &\leq K 2^{-m/2} \int_{-\infty}^{\infty} |t - t_0|^\alpha \psi(2^{-m}t - n) dt \\ &= K 2^{m/2} \int_{-\infty}^{\infty} |x 2^m + n 2^m - t_0|^\alpha \psi(x) dx \\ &\leq K C 2^{m(\alpha+1/2)} \underbrace{\int_{-\infty}^{\infty} (|x| + |C|)^\alpha \psi(x) dx}_{=A} \\ &= C_1 2^{m(\alpha+1/2)}. \end{aligned}$$

Signals of Interest and Wavelet Representations



- Wavelet coefficients around smooth parts of the signal are small and have fast decay ($\sim 2^{-j(\alpha+1/2)}$).
- Wavelet coefficients around polynomial parts of the signal are exactly zero.
- Discontinuities generate a **finite** number of large wavelet coefficients.

Wavelets vs Fourier

Preview 1: Approximation

Consider signals in the vector space \mathbb{R}^N

There are many orthonormal expansions

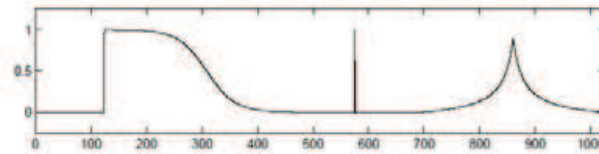
$$x = \sum_{i=1}^N \alpha_i \varphi_i$$

How do they differ for *approximation*?

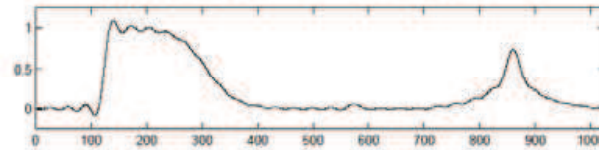
$$x \approx \hat{x}_K = \sum_{i=1}^K \beta_i \varphi_{j_i}, \quad K < N$$

- Different bases $\{\varphi_i\}_{i=1}^N$
- Different orders j_1, j_2, j_3, \dots

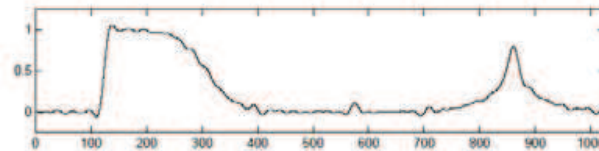
Wavelets vs Fourier



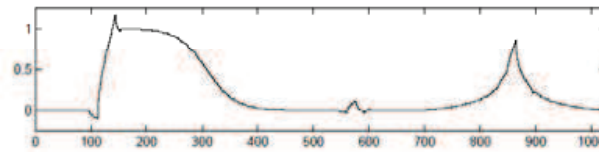
Original signal, N=1024



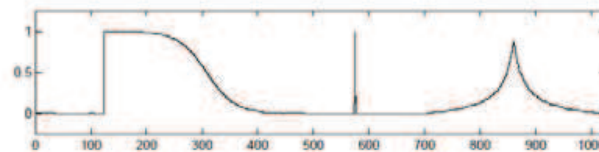
M=64, linear
Fourier D= 2.7



M=64, nonlinear
Fourier D= 2.4



M=64, linear
Daubechies wavelet D=3.5



M=64, nonlinear
Daubechies wavelet D=0.01

5

Wavelets vs Fourier

Preview 2: Estimation

Consider signal x observed through noisy observation

$$y = x + d \quad d \sim \mathcal{N}(0, \sigma^2 I)$$

MMSE estimate $\hat{x}_{\text{MMSE}} = E[x \mid y]$ can be complicated (if even known)

For simplicity, we often insist on a diagonal estimator in a transform domain:

$$\hat{x} = T^{-1} D(Ty), \quad D(\cdot) \text{ diagonal}$$

What are good choices for T and D ?

Wavelets vs Fourier

Original image

Original image +
i.i.d. Gaussian noise

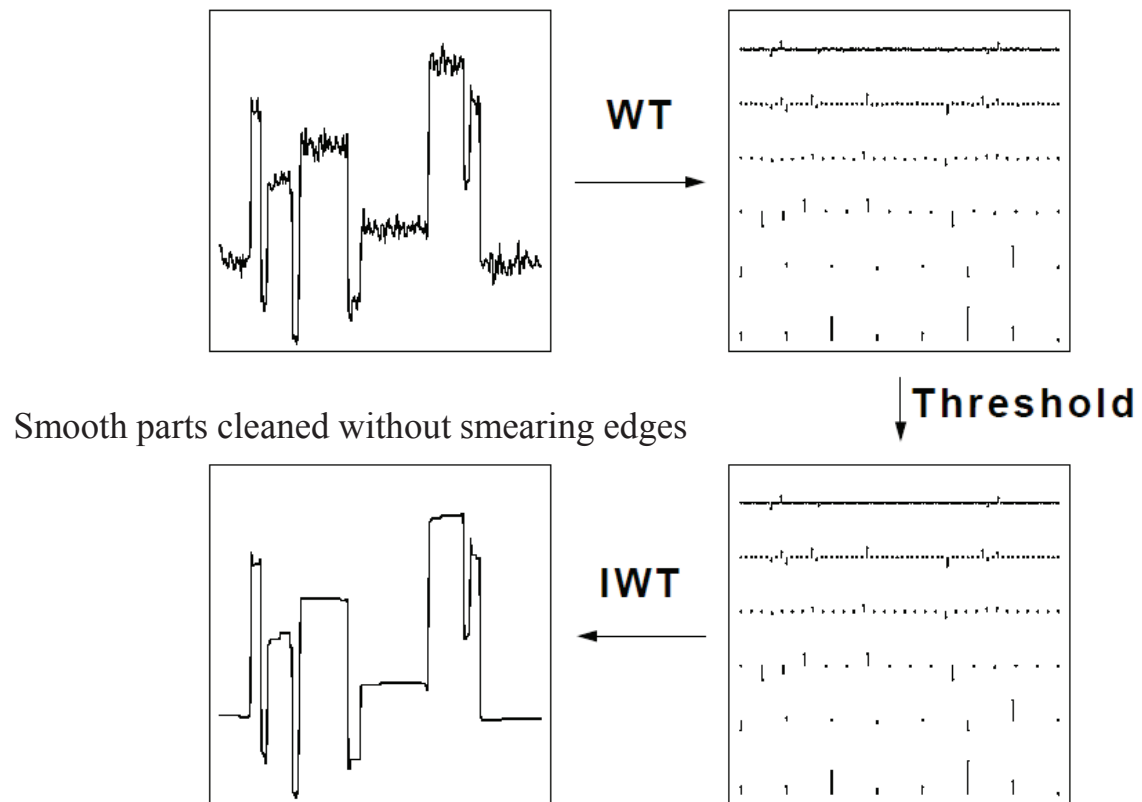
Denoised in
Fourier domain

Denoised in
Wavelet domain



Wavelets vs Fourier

1-D Example

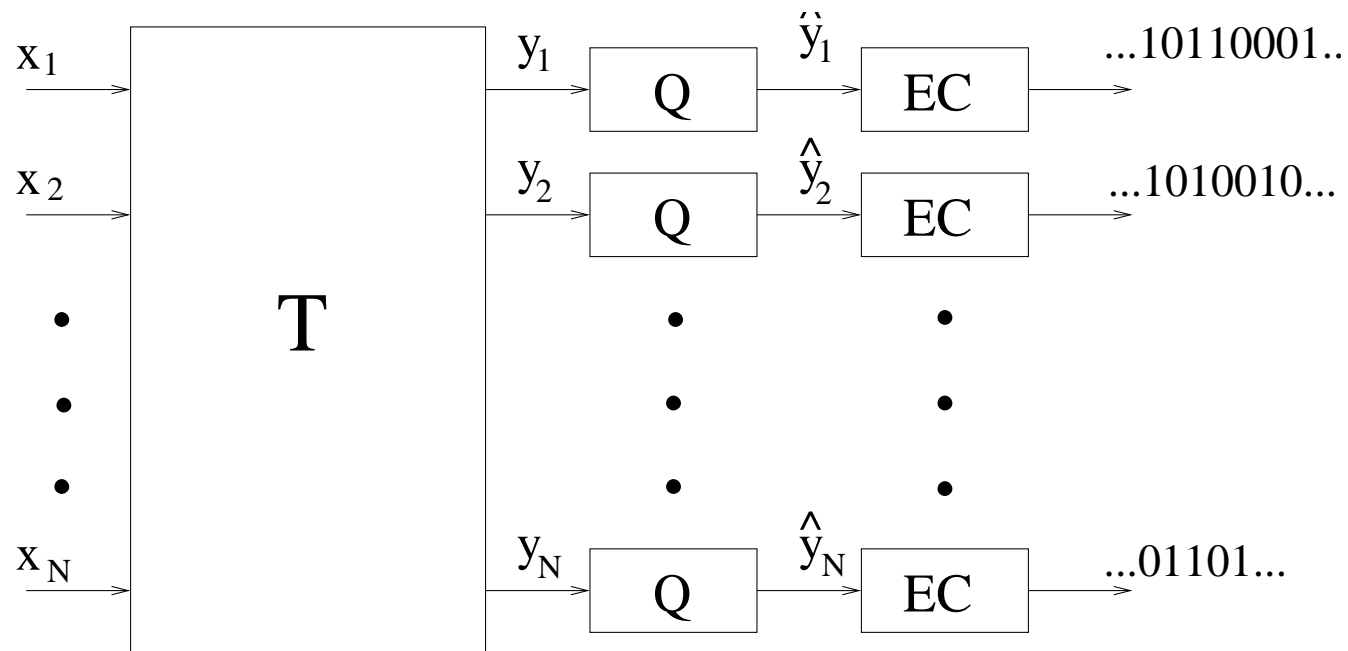


Wavelets vs Fourier

| Fourier | Wavelet |
|------------------------|----------------------------------|
| linear, time invariant | not linear, time invariant |
| Gaussian | non-Gaussian |
| stationary | non-stationary |
| continuous | discontinuous |
| smooth | containing edges |
| known structure | (partly) unknown structure |
| global | locally “adaptive” |
| single resolution | multiple resolutions |
| modeled by sinusoids | modeled by piecewise polynomials |

10

Application in Compression: Transform Coding



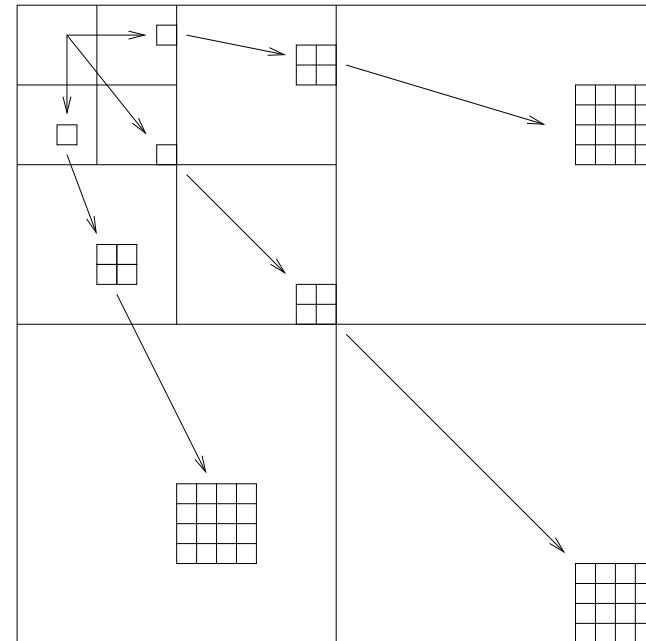
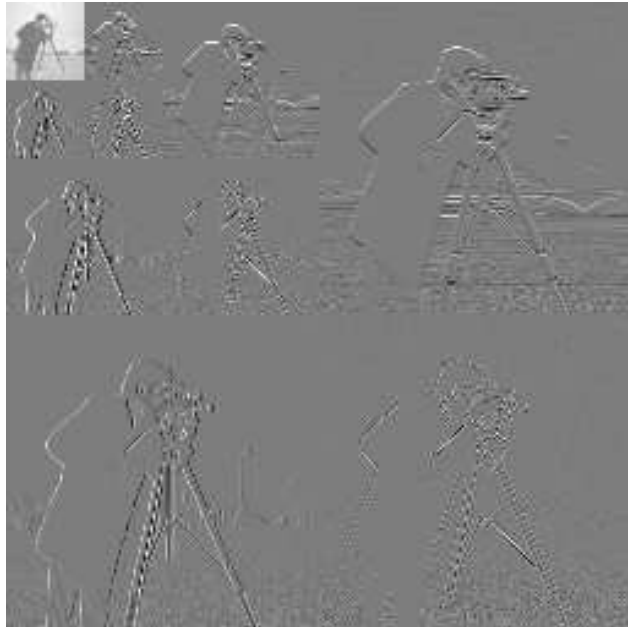
A transform coder is made of a linear transform, scalar quantisers and entropy encoders.

Approximation and Compression

More formally:

- Non-linear approximation of piecewise smooth signals
 - With Fourier $MSE \sim M^{-1}$
 - With Wavelets $MSE \sim M^{-2\alpha}$
- Compression of piecewise smooth signals
 - With Fourier $D(R) \leq c_0 R^{-1}$
 - With Wavelets $D(R) \leq c_1 R^{-2\alpha} + c_2 \sqrt{R} 2^{-c_2 \sqrt{R}}$ ([CohenDGO:02])
 - With Footprints $D(R) \leq c_3 R^{-2\alpha} + c_4 2^{-c_5 R}$ ([DragottiV:03])
- Compression of piecewise polynomial signals
 - With Wavelets $D(R) \leq c_2 \sqrt{R} 2^{-c_2 \sqrt{R}}$
 - With Footprints $D(R) \leq c_4 2^{-c_5 R}$

Application in Compression



Zerotree algorithm.

Application in Compression

| | | | | | | | |
|-----|-----|-----|-----|---|----|-----|----|
| 63 | -34 | 49 | 10 | 7 | 13 | -12 | 7 |
| -31 | 23 | 14 | -13 | 3 | 4 | 6 | -1 |
| 15 | 14 | 3 | -12 | 5 | -7 | 3 | 9 |
| -9 | -7 | -14 | 8 | 4 | -2 | 3 | 2 |
| -5 | 9 | -1 | 47 | 4 | 6 | -2 | 2 |
| 3 | 0 | -3 | 2 | 3 | -2 | 0 | 4 |
| 2 | -3 | 6 | -4 | 3 | 6 | 3 | 6 |
| 5 | 11 | 5 | 6 | 0 | 3 | -4 | 4 |

| Coefficient | Symbol | Reconstruction |
|-------------|--------|----------------|
| 63 | POS | 48 |
| -34 | NEG | -48 |
| -31 | IZ | 0 |
| 23 | ZTR | 0 |
| 49 | POS | 48 |
| 10 | ZTR | 0 |
| 14 | ZTR | 0 |
| -13 | ZTR | 0 |
| 15 | ZTR | 0 |
| 14 | IZ | 0 |
| -9 | ZTR | 0 |
| -7 | ZTR | 0 |
| 7 | Z | 0 |
| 13 | Z | 0 |
| 3 | Z | 0 |
| 4 | Z | 0 |
| -1 | Z | 0 |
| 47 | POS | 48 |

A Success Story

Wavelets are in the new image compression standard (JPEG2000)



Original Lena Image
(256 × 256 pixels)



JPEG (Compression Ratio
43:1)



JPEG2000 (Compression
Ratio 43:1)

Note: images courtesy of dspworx.com