

# 1) MULTI-RATE SIGNAL PROCESSING

(a)

$$(1) X(z)H(z^{-1})$$

$$(2) \frac{1}{2} \left( X(z^{1/2})H(z^{-1/2}) + X(-z^{1/2})H(-z^{-1/2}) \right)$$

$$(3) \frac{1}{2} \left( X(z)H(z^{-1}) + X(-z)H(-z^{-1}) \right)$$

$$(4) \frac{1}{2} H(z) \left[ X(z)H(z^{-1}) + X(-z)H(-z^{-1}) \right]$$

IN FOURIER DOMAIN

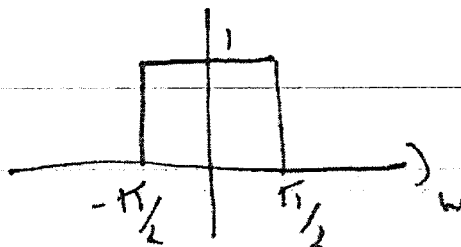
$$(1) X(e^{j\omega})H(e^{-j\omega})$$

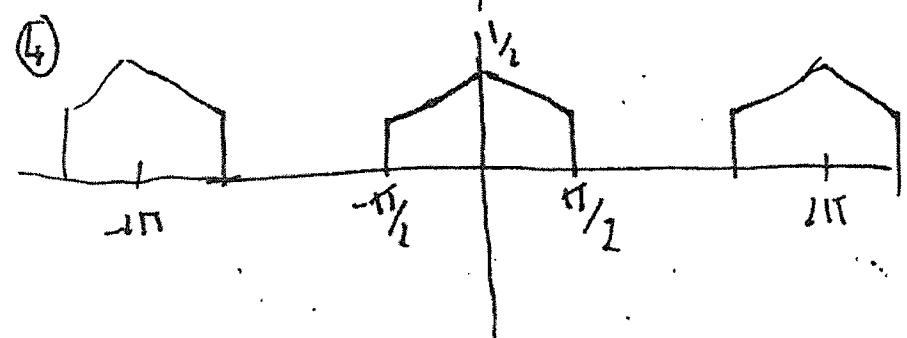
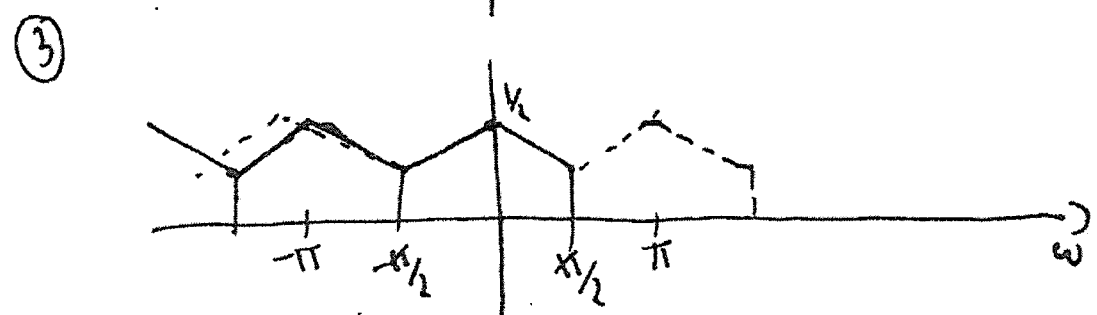
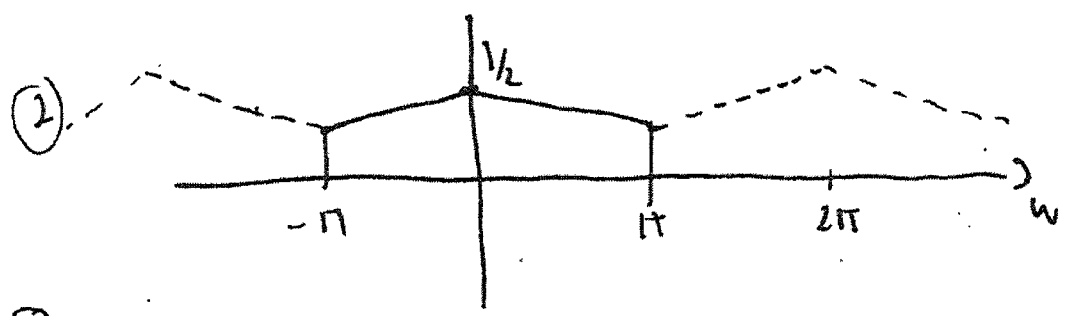
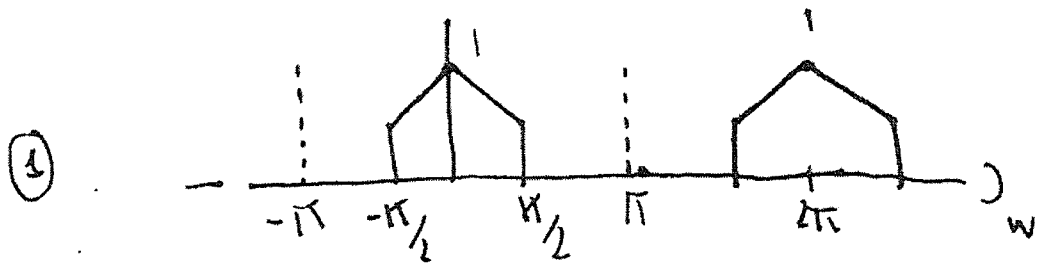
$$(2) \frac{1}{2} \left( X(e^{j\omega/2})H(e^{-j\omega/2}) + X(e^{j(\omega/2+\pi)})H(e^{-j(\omega/2+\pi)}) \right)$$

$$(3) \frac{1}{2} \left[ X(e^{j\omega})H(e^{-j\omega}) + X(e^{j\omega+\pi})H(e^{-j\omega+\pi}) \right]$$

$$(4) \frac{1}{2} H(e^{j\omega}) \left[ X(e^{j\omega})H(e^{-j\omega}) + X(e^{j(\omega+\pi)})H(e^{-j(\omega+\pi)}) \right]$$

$$H(e^{j\omega}) = H(e^{-j\omega})$$





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QUESTION 1.2  
~~QUESTION~~

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(a)

$$\hat{X}(z) = \frac{1}{2} \left[ G(z^{1/2}) H(z^{1/2}) + G(-z^{1/2}) H(-z^{1/2}) \right] X(z)$$

PR CONDITION :

$$G(z) H(z) + G(-z) H(-z) = 2 \quad \text{OR}$$

$$P(z) + P(-z) = 2 \quad *$$

(b)

$$G(z) = a \quad \text{DOES NOT WORK}$$

$$\text{TRY} \quad G(z) = a z^{-1} + b + a z$$

$$\begin{aligned} P(z) &= (z^{-2} + z^{-1} + 1 + z + z^2)(a z^{-1} + b + a z) = \\ &= a z^{-3} + (a+b) z^{-2} + (2a+b) z^{-1} + (2a+b) + (2a+b) z + \\ &\quad (a+b) z^1 + a z^3. \end{aligned}$$

$$P(z) + P(-z) = 2 \Rightarrow \begin{cases} a+b=0 \\ 2a+b=1 \end{cases} \Rightarrow \begin{matrix} a=1 \\ b=-1 \end{matrix}$$

$$G(z) = (z^{-1} - 1 + z)$$

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$$(c) \quad \sum_{k=0}^{\infty} P(k) = 0 \quad (\Leftrightarrow) \quad P(7) + P(-7) = 0$$

THIS IS ACHIEVED WHEN  $G(t) = (1-t^{-1})$

IN THIS CASE WE HAVE

$$P(7) = (1 + 7 + 7^2 + 7^3)(1-t^{-1}) = 7^3 - 7^{-1}$$

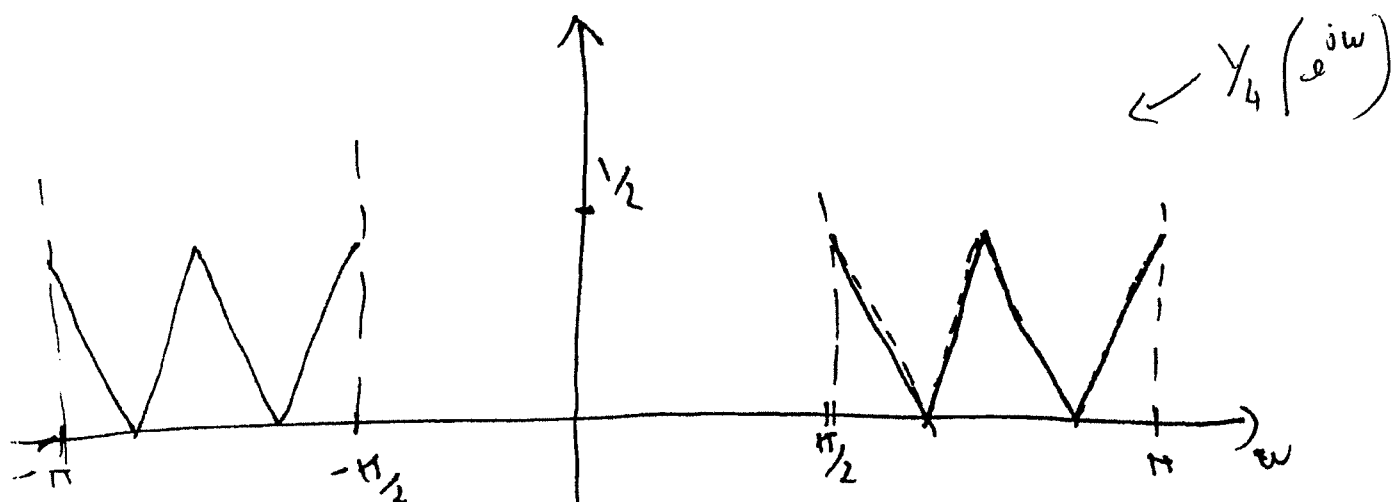
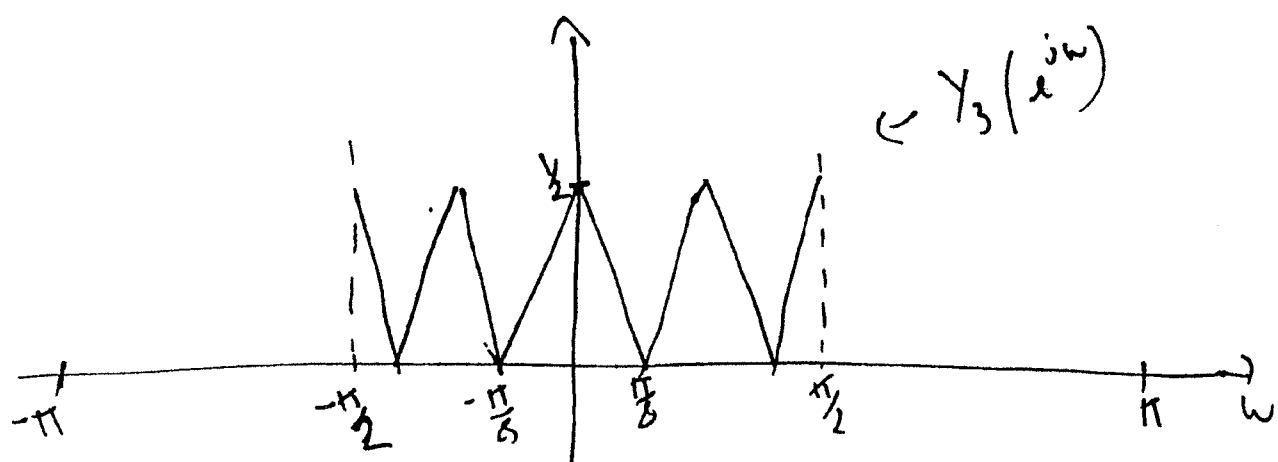
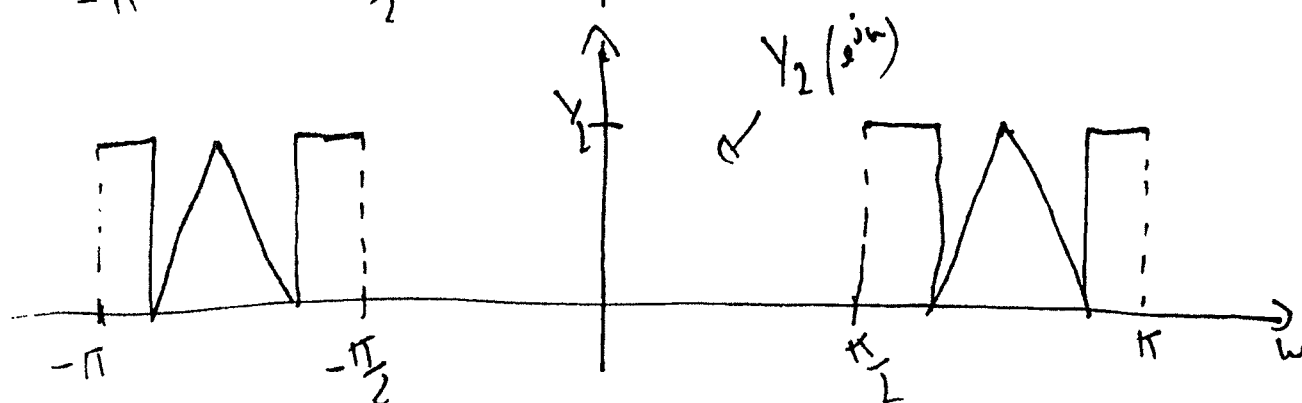
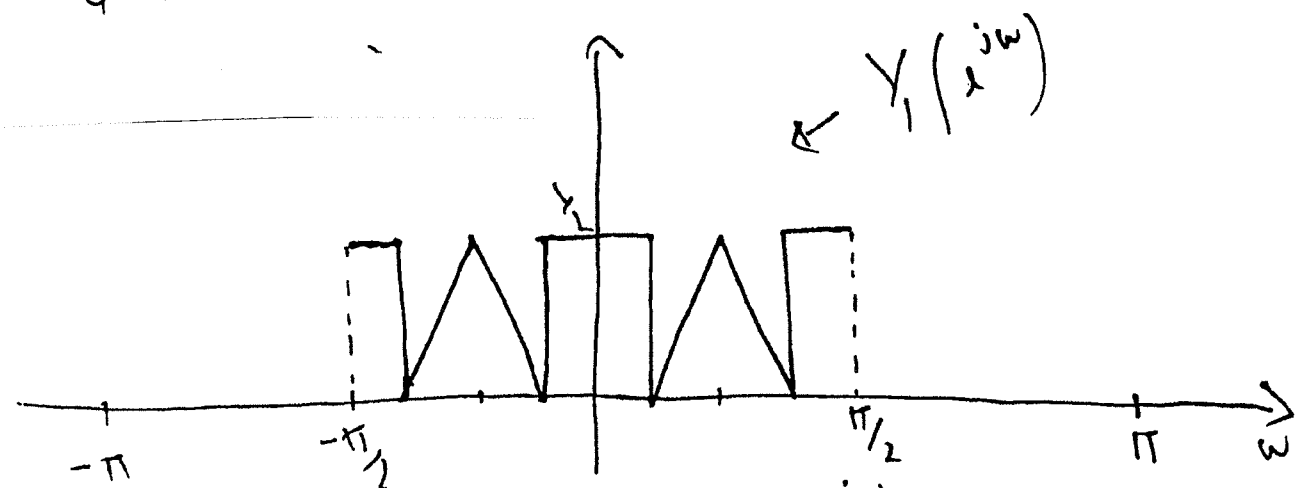
THUS

$$P(7) + P(-7) = 0$$

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# QUESTION 1.3

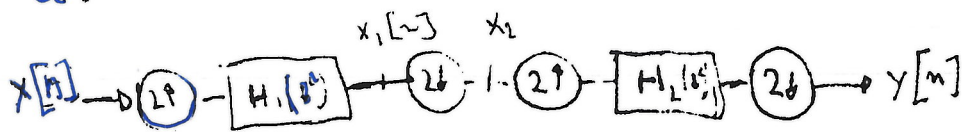
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# EXERCISE 1.4

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a.



$$X_1(z) = H_1(z) X(z)$$

$$X_2(z) = H_1(z) X(z)$$

$$Y(z) = H_2(z) H_1(z) X(z) \Rightarrow$$

$$H_{eq}(z) = H_2(z) H_1(z)$$

(b). (~~NEW EXAMPLE~~)

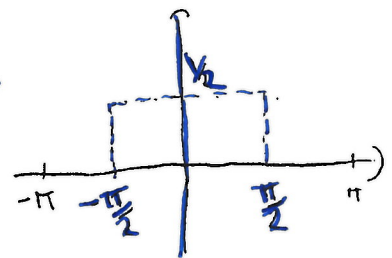
WHEN  $M=2$

$$Y(z) = X(z) \frac{1}{2} (H(z^{1/2}) + H(-z^{1/2}))$$

SINCE  $H(z)$  IS A LOW-PASS FILTER WITH CUT-OFF AT  $\pi/4$

$$\frac{1}{2} (H(z^{1/2}) + H(-z^{1/2})) \Leftrightarrow$$

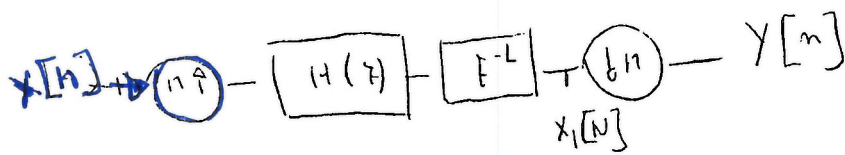
IS EQUIVALENT TO A LOW-PASS FILTER WITH CUT-OFF AT  $\pi/2$  AND AMPLITUDE  $Y_2$



WHEN  $M=4$

WE GET AN ALL PASS -FILTER WITH AMPLITUDE  $Y_4$

(C) (~~NEW EXAMPLE~~)



$$X_1(z) = H(z) z^{-L} X(z^M)$$

$$Y(z) = X(z) \cdot \frac{1}{M} \sum_{k=0}^{M-1} H\left(W_M^k z^{1/M}\right) z^{-L/M} W_M^{kL}$$

SINCE  $H(z)$  IS IDEAL LOW PASS WITH CUT-OFF  
AT  $\frac{\pi}{M}$ , ALL THE ALIASED COMPONENT  
DON'T CONTRIBUTE AND  $H(z^{1/M})$  IS  
AN ALL-PASS FILTER  $\Rightarrow$   
 $\frac{1}{M} \sum_{k=0}^{M-1} H\left(W_M^k z^{1/M}\right) z^{-L/M} W_M^{kL} = \frac{1}{M} z^{-L/M}$

THUS

$$H_{EQ}(z) = \frac{1}{M} z^{-L/M}$$