

Linear time invariant systems

$$Y[n] = T\{X[n]\} \quad \text{general definition of a system}$$

A general sequence $X[n]$ can be expressed as a linear combination of delayed and scaled unit pulses, i.e. $X[n] = \sum_{k=-\infty}^{+\infty} X[k] \delta[n-k]$

$$Y[n] = T\left\{\sum_{k=-\infty}^{+\infty} X[k] \delta[n-k]\right\} = \sum_{k=-\infty}^{+\infty} X[k] \underbrace{T\{\delta[n-k]\}}_{\text{linear system}}$$

Let us define $h_x[n] = T\{\delta[n-k]\}$, impulse response of linear system.

If the system is also time invariant, and $h[n]$ is the response to $\delta[n]$, then the response to $\delta[n-k]$ is $h[n-k]$

$$Y[n] = \sum_{k=-\infty}^{+\infty} X[k] h[n-k] \quad \begin{array}{c} \uparrow \\ \text{linear,} \\ \text{time invariant} \end{array}$$

This is called convolution sum, and is also expressed as

$$Y[n] = X[n] * h[n]$$

Convolution is commutative

$$Y[n] = \sum_{k=-\infty}^{+\infty} X[k] h[n-k] = \sum_{k=-\infty}^{+\infty} h[k] X[n-k] = h[n] * X[n]$$

Commutative property of convolution

$$Y[n] = X[n] \otimes h[n] = h[n] \otimes X[n]$$

Proof.

$$Y[n] = \sum_{K=-\infty}^{+\infty} X[K] h[n-K] :$$

$$\downarrow$$
$$q = n - K \rightarrow \begin{cases} \text{if } K \rightarrow -\infty \Rightarrow q \rightarrow +\infty \\ \text{if } K \rightarrow +\infty \Rightarrow q \rightarrow -\infty \end{cases}$$

$$= \sum_{q=-\infty}^{+\infty} X[n-q] h[q] = \sum_{q=-\infty}^{+\infty} h[q] X[n-q] = h[n] \otimes X[n]$$

The impulse response of accumulator system is

$$h[n] = \sum_{K=-\infty}^n \delta[K]$$

$$\downarrow$$
$$Y[n] = \sum_{K=-\infty}^n X[K]$$

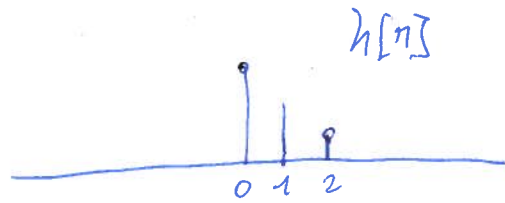
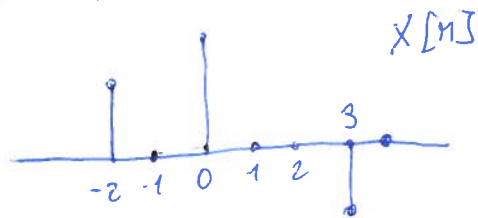
Demonstration

$$Y[n] = X[n] \otimes h[n] = \sum_{K=-\infty}^{+\infty} X[K] \sum_{q=-\infty}^{n-K} \delta[q]$$

$$\sum_{q=-\infty}^{n-K} \delta[q] = \begin{cases} 1 & \text{if } n-K \geq 0 \rightarrow n \geq K \rightarrow K \leq n \\ 0 & \text{if } n-K < 0 \rightarrow n < K \rightarrow K > n \end{cases}$$

$$\rightarrow Y[n] = \sum_{K=-\infty}^n X[K] \quad \leftarrow \text{definition seen in previous lecture}$$

Example



$$x[n] = x_{-2} \delta[n+2] + x_0 \delta[n] + x_3 \delta[n-3]$$

$$y[n] = \sum_{k=-\infty}^{+\infty} x[k] \mathcal{T}\{\delta[n-k]\} = \sum_{k=-\infty}^{+\infty} x[k] h[n-k]$$

we can calculate the response to the individual samples of the input

$$k = -2$$

$$y_{-2}[n] = x_{-2} h[n+2]$$

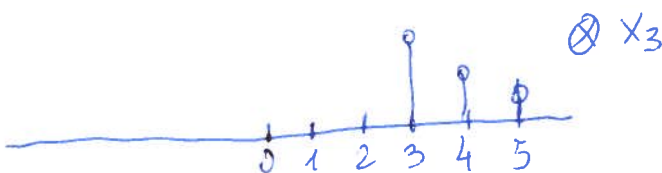
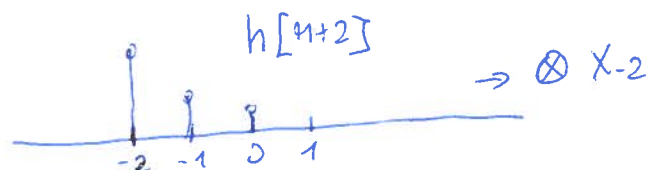
$$k = 0$$

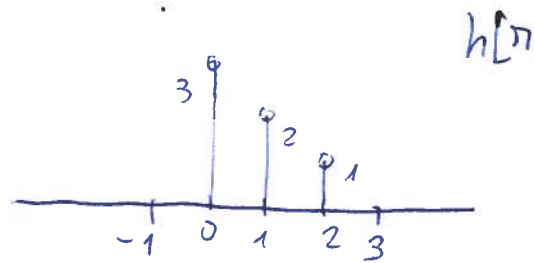
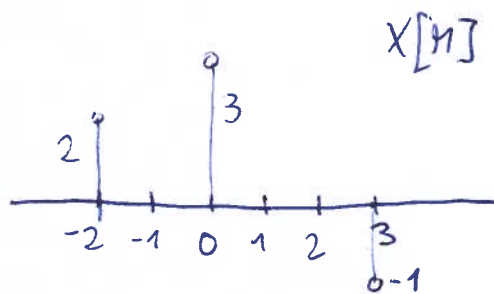
$$y_0[n] = x_0 h[n]$$

$$k = 3$$

$$y_3[n] = x_3 h[n-3]$$

$$y[n] = y_{-2}[n] + y_0[n] + y_3[n] = x_{-2} h[n+2] + x_0 h[n] + x_3 h[n-3]$$





$$x[n] = 2\delta[n+2] + 3\delta[n] - \delta[n-3]$$

$$h[n] = 3\delta[n] + 2\delta[n-1] + \delta[n-2]$$

$$y[n] = \sum_{k=-\infty}^{+\infty} x[k] h[n-k]$$

$$= x[-2]h[n+2] + x[0]h[n] + x[3]h[n-3] =$$

$$= 2h[n+2] + 3h[n] - h[n-3]$$

$$h[n+2] = 3\delta[n+2] + 2\delta[n+1] + \delta[n]$$

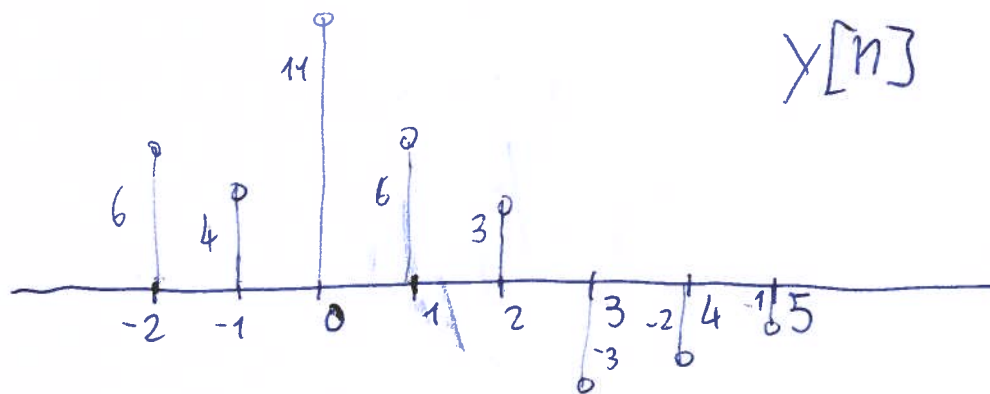
$$h[n] = 3\delta[n] + 2\delta[n-1] + \delta[n-2]$$

$$h[n-3] = 3\delta[n-3] + 2\delta[n-4] + \delta[n-5]$$

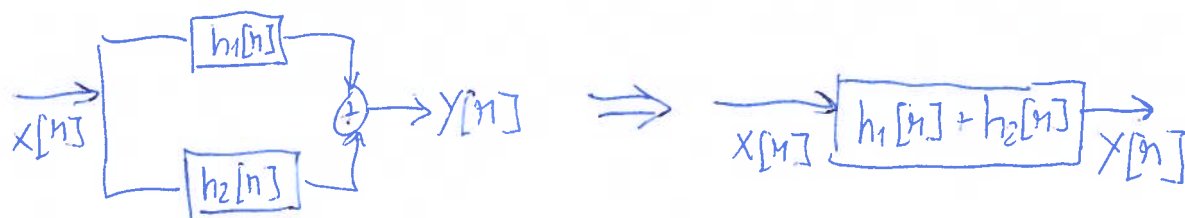
$$\begin{aligned} y[n] &= 2(3\delta[n+2] + 2\delta[n+1] + \delta[n]) + \\ &\quad + 3(3\delta[n] + 2\delta[n-1] + \delta[n-2]) + \\ &\quad - (3\delta[n-3] + 2\delta[n-4] + \delta[n-5]) = \end{aligned}$$

$$\begin{aligned} &= 6\delta[n+2] + 4\delta[n+1] + 2\delta[n] + 9\delta[n] + 6\delta[n-1] + \\ &\quad + 3\delta[n-2] - 3\delta[n-3] - 2\delta[n-4] - \delta[n-5] = \end{aligned}$$

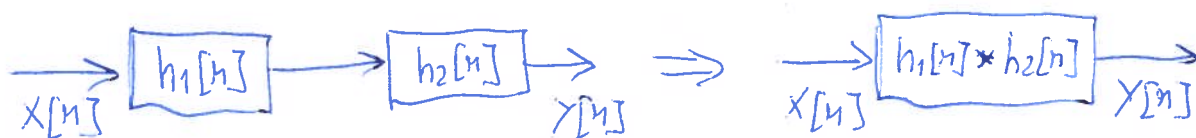
$$\begin{aligned} &= 6\delta[n+2] + 4\delta[n+1] + 11\delta[n] + \cancel{6\delta[n-1]} + \\ &\quad + 3\delta[n-2] - 3\delta[n-3] - 2\delta[n-4] - \delta[n-5] \end{aligned}$$



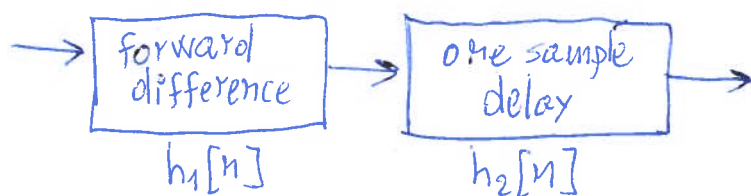
Parallel combination of LTI systems



Cascade combination of LTI systems



Example



$$h_1[n] = \delta[n+1] - \delta[n]$$

$$h_2[n] = \delta[n-1]$$

$$h_1[n] \otimes h_2[n] = (\delta[n+1] - \delta[n]) \otimes \delta[n-1] = \delta[n-1] \otimes (\delta[n+1] - \delta[n])$$

$$\delta[n-1] \otimes \delta[n+1] = \sum_{k=-\infty}^{+\infty} \delta[k-1] \delta[n+1-k]$$

$$\delta[k-1] = \begin{cases} 1 & k=1 \\ 0 & k \neq 1 \end{cases}$$

$$\delta[n+1-k] \Big|_{k=1} = \delta[n]$$

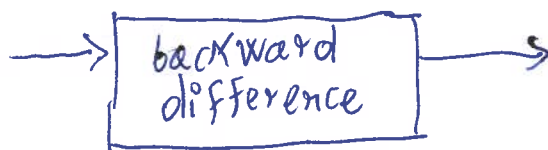
$$\Rightarrow \delta[n-1] \otimes \delta[n+1] = \delta[n]$$

$$\delta[n-1] \otimes \delta[n] = \sum_{k=-\infty}^{+\infty} \delta[k-1] \delta[n-k]$$

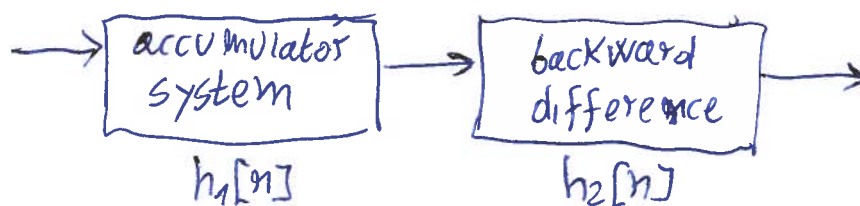
$$\delta[n-k] \Big|_{k=1} = \delta[n-1]$$

$$\delta[n-1] \otimes \delta[n] = \delta[n-1]$$

$$\Rightarrow h_1[n] \otimes h_2[n] = \delta[n] - \delta[n-1]$$



example



$$h_1[n] = \sum_{k=-\infty}^{+\infty} \delta[k] = \begin{cases} 1 & n \geq 0 \\ 0 & n < 0 \end{cases} \Rightarrow h_1[n] = u[n]$$

$$h_2[n] = \delta[n] - \delta[n-1]$$

$$h_1[n] \otimes h_2[n] = u[n] \otimes (\delta[n] - \delta[n-1])$$

$$u[n] \otimes \delta[n] = \sum_{k=-\infty}^{+\infty} u[k] \delta[n-k]$$

$$\delta[n-k] = \begin{cases} 1 & n=k \\ 0 & n \neq k \end{cases}$$

$$\Rightarrow = u[n]$$

$$u[n] \otimes \delta[n-1] = u[n-1]$$



$$h_1[n] \otimes h_2[n] = u[n] - u[n-1] = \delta[n]$$

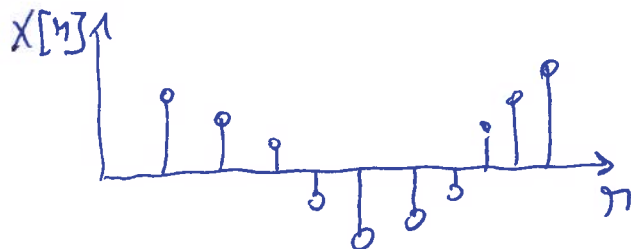
Backward difference is then the inverse system of the accumulator system

Fourier Transform of a continuous signal



$$X(\omega) = \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt \quad \left| \begin{array}{l} \text{physical interpretation:} \\ \text{spectrum of a signal} \end{array} \right.$$

What about the spectrum of a discrete signal, obtained by sampling $x(t)$?



It can be demonstrated that the spectrum of a discrete time signal can be calculated as

$$X(\omega) = \sum_{n=-\infty}^{+\infty} x[n] e^{-j\omega n} \quad \begin{array}{l} \text{discrete time} \\ \text{Fourier transform} \\ (\text{DTFT}) \end{array}$$

A common notation for DTFT is $X(e^{j\omega})$

$X(e^{j\omega})$ is periodic of period 2π :

$$\begin{aligned} X(e^{j(\omega+2\pi)}) &= \sum_{n=-\infty}^{+\infty} x[n] e^{-j(\omega+2\pi)n} = \\ &= \sum_{n=-\infty}^{+\infty} x[n] e^{-j\omega n} \underbrace{e^{-j2\pi n}}_{=1} = \sum_{n=-\infty}^{+\infty} x[n] e^{-j\omega n} = X(e^{j\omega}) \end{aligned}$$

Since DTFT is 2π periodic, the inverse Fourier transform can be calculated by integrating over $[-\pi, \pi]$:

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

DTFT of impulse response

$$H(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} h[n] e^{-j\omega n}$$

$$h[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega$$

What is the condition for existence of Fourier transform? Which signals can be represented

as $x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$?

Fourier Transform exists if $\sum_{n=-\infty}^{+\infty} x[n] e^{-j\omega n}$ series converges.

$$\begin{aligned} |X(e^{j\omega})| &= \left| \sum_{n=-\infty}^{+\infty} x[n] e^{-j\omega n} \right| \leq \sum_{n=-\infty}^{+\infty} |x[n] e^{-j\omega n}| = \\ &= \sum_{n=-\infty}^{+\infty} |x[n]| \underbrace{|e^{-j\omega n}|}_{=1} = \sum_{n=-\infty}^{+\infty} |x[n]| < \infty \end{aligned}$$

Series $x[n]$ must be absolutely summable!

Example

$$x[n] = a^n u[n]$$

$$X(e^{j\omega}) = \sum_{n=0}^{\infty} a^n e^{-j\omega n} = \sum_{n=0}^{\infty} (a e^{-j\omega})^n$$

$$\left| \sum_{n=0}^{\infty} z^n = \begin{cases} \infty & \text{if } z \geq 1 \\ \frac{1}{1-z} & \text{if } z < 1 \end{cases} \right.$$

$$= \frac{1}{1 - a e^{-j\omega}} \quad \text{if } |a e^{-j\omega}| < 1 \rightarrow \text{if } |a| < 1$$

Fourier Transform exists if $|a| < 1$