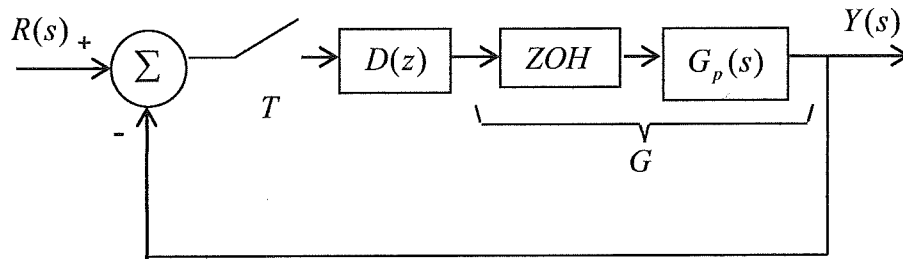


Example of phase lag and phase lead design using Bode diagrams

①



Figuur / Figure 1

Die stelsel in figuur 1 het die volgende oordragsfunksie: /
The system in figure 1 has the following transfer function:

$$G_p(s) = \frac{20K}{s(s+1)(s+4)}$$

Vir $K = 1$, is die diskrete oordragsfunksie van die stelsel soos volg: /
For $K = 1$, the discrete transfer function of the system is as follows:

$$G(z) = \frac{3.292 \cdot 10^{-6} z^2 + 1.3 \cdot 10^{-5} z + 3.211 \cdot 10^{-6}}{(z-1)(z-0.99)(z-0.9608)}, \quad T = 0.01s$$

Figuur 2 toon die bodediagram van $G(j\omega)$ vir $K = 1$. /
Figure 2 shows the bode diagram of $G(j\omega)$ for $K = 1$.

4.1 Hou $K = 1$ en ontwerp 'n fasealoopnetwerk $D(z)$ wat 'n fasegrens van 45° tot gevolg sal hê, maar nie die bestendige gedrag van die stelsel sal verander nie. /

Keep $K = 1$ and design a phase lag compensator $D(z)$ that will give a phase margin of 45° for the system without changing the system's steady state performance.

4.2 Hou $K = 1$ en ontwerp 'n fasevoorloopnetwerk $D(z)$ wat 'n fasegrens van 30° tot gevolg sal hê, maar nie die bestendige gedrag van die stelsel sal verander nie. /

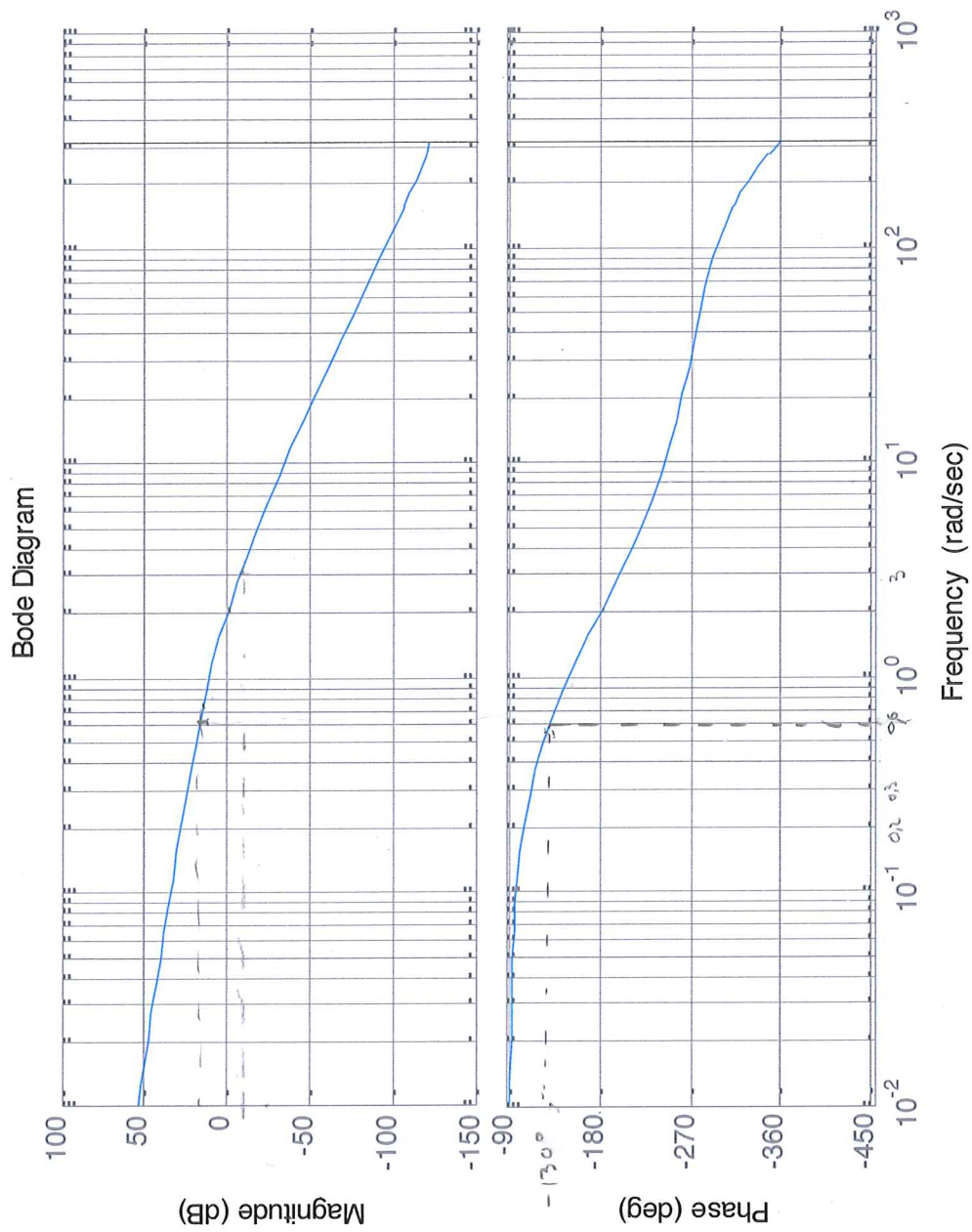
Keep $K = 1$ and design a phase lead compensator $D(z)$ that will give a phase margin of 30° for the system without changing the system's steady state performance.

Addisionele inligting / Additional information:

$$D(w) = a_0 \left[\frac{1 + w/(a_0/a_1)}{1 + w/(1/b_1)} \right]$$

$$a_1 = \frac{1 - a_0 |G(j\omega_{w1})| \cos \theta}{\omega_{w1} |G(j\omega_{w1})| \sin \theta}, \quad b_1 = \frac{\cos \theta - a_0 |G(j\omega_{w1})|}{\omega_{w1} \sin \theta} \quad [20]$$

$$K_d = a_0 \left[\frac{\omega_{wp}(\omega_{w0} + 2/T)}{\omega_{w0}(\omega_{wp} + 2/T)} \right], \quad z_0 = \left[\frac{2/T - \omega_{w0}}{2/T + \omega_{w0}} \right], \quad z_p = \left[\frac{2/T - \omega_{wp}}{2/T + \omega_{wp}} \right]$$



Figuur / Figure 2

(3)

Memo

4.1 Determine ω_{w1} at which the phase angle is :

$$\phi \approx (-180^\circ + \phi_m + 5^\circ) \\ \approx -130^\circ$$

$$\therefore \omega_{w1} \approx 0,63 \text{ rad/sec.}$$

$$\text{Choose } \omega_{w0} = 0,1 \cdot \omega_{w1} = 0,063 \text{ rad/sec.}$$

$$\text{Then } \omega_{wp} = \frac{0,1 \omega_{w1}}{a_0 |G(j\omega_{w1})|}$$

To keep steady state response requirement

$$a_0 = 1$$

$$\text{Then } 20 \log |G(j\omega_{w1})| = 18 \text{ dB}$$

$$|G(j\omega_{w1})| = 10^{18/20} = 7,94 \approx 8$$

$$\therefore \omega_{wp} = \frac{0,1 \cdot 0,63}{8} = 0,008 \text{ rad/sec.}$$

$$k_d = a_0 \left[\frac{\omega_{wp}(\omega_{w0} + \frac{2}{T})}{\omega_{w0}(\omega_{wp} + \frac{2}{T})} \right] = 1 \left[\frac{0,008(0,063 + \frac{2}{0,01})}{0,063(0,008 + \frac{2}{0,01})} \right]$$

$$= 0,13$$

$$z_0 = \frac{\frac{2}{T} - \omega_{w0}}{\frac{2}{T} + \omega_{w0}} = \frac{\frac{2}{0,01} - 0,063}{\frac{2}{0,01} + 0,063} = 0,9994$$

(4)

$$z_p = \frac{z/T - \omega_{wp}}{z/T + \omega_{wp}} = \frac{z/0,01 - 0,008}{z/0,01 + 0,008} = 0,9999$$

$$\therefore D(z) = \frac{0,13 (z - 0,9994)}{(z - 0,9999)} \quad \square$$

4.2. We start with the requirements for a phase lead:

$$\angle G(j\omega_{w1}) < -180^\circ + 30^\circ \\ < -150^\circ$$

and $|G(j\omega_{w1})| < 1$

So choose $\omega_{w1} > 1 \text{ rad/sec}$

\therefore choose $\omega_{w1} = 3 \text{ rad/sec}$

and then check.

$$\cos \theta > |G(j\omega_{w1})| = 0,464$$

$$\theta = 180^\circ + 30^\circ + 198^\circ = 48^\circ$$

$$\cos 48^\circ = 0,67 > 0,464$$

$$a_1 = \frac{1 - a_0 |G(j\omega_{w1})| \cos \theta}{\omega_{w1} |G(j\omega_{w1})| \sin \theta} \\ = \frac{1 - 1 \cdot (0,464) \cdot 0,67}{3 \cdot 0,464 \cdot \sin 48^\circ} = 0,666$$

$$b_1 = \frac{0,67 - 0,464}{3 \sin 48^\circ} = 0,0924$$

$$\omega_{w0} = \frac{1}{a_1} = 1,5 \text{ rad/sec}$$

$$\omega_{wp} = \frac{1}{b_1} = 10,8 \text{ rad/sec}$$

$$K_d = \frac{10,8 (1,5 + 200)}{1,5 (10,8 + 200)} = 6,88$$

$$z_0 = \frac{200 - 1,5}{200 + 1,5} = 0,985$$

$$z_p = \frac{200 - 10,8}{200 + 10,8} = 0,898$$

$$D(z) = \frac{6,88 (z - 0,985)}{(z - 0,898)}$$

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