# Model Documentation of the:

## Modular Multilevel Converter (MMC)

### 1 Nomenclature

## 1.1 Nomenclature for Model Equations

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total stored energy (scaled by \frac{2}{3})
e_{s0}
        (vertical) difference between all upper and all lower arms (scaled by \frac{2}{2})
e_{d0}
        complex energy sum
\underline{e}_s
        complex energy difference
\underline{e}_d
        angle of rotating reference frame of \underline{e}_s and \underline{e}_d
        angular speed of the rotating reference frame
\omega
        DC voltage of the MMC
v_{DC}
        common-mode voltage
v_{y0}
        complex output voltage
\underline{v}_y
        voltage which drives the DC currents
v_{x0}
        voltage which drives the internal currents
\underline{v}_x
        grid voltage
        complex output current
        scaled version of DC current
        complex circulating current
        arm inductance
M_z
        mutual inductance
R
        resistance of the load
L
        inductance of the load
```

#### 1.2 Nomenclature for Derivation

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\begin{array}{ll} e_{z1},\, ...,\, e_{z6} & \text{arm energies} \\ i_{z1},\, ...,\, i_{z6} & \text{arm currents} \\ g_0,\, g_\alpha,\, g_\beta & \text{clark transform constants} \\ \underline{g}_{\alpha\beta} & \text{complex constant using } g_\alpha,\, g_\beta \text{ and } \theta \end{array}
```

## 2 Model Equations

State Vector and Input Vector:

$$\underline{x} = (x_1 \ x_2 \ \underline{x}_3 \ \underline{x}_4 \ \underline{x}_5 \ x_6 \ \underline{x}_7 \ x_8)^T = (e_{s0} \ e_{d0} \ \underline{e}_s \ \underline{e}_d \ \underline{i}_s \ i_{s0} \ \underline{i} \ \theta)^T$$

$$\underline{u} = (\underline{u}_1 \ u_2 \ \underline{u}_3 \ u_4)^T = (\underline{v}_y \ v_{y0} \ \underline{v}_x \ v_{x0})^T$$

Model Equations:

$$\dot{x}_1 = v_{DC} x_6 - \operatorname{Re}(\underline{x}_7 \underline{u}_1^*) \tag{1a}$$

$$\dot{x}_2 = -2u_2 x_6 - \operatorname{Re}(\underline{x}_5^* \underline{v}_{y\Delta}) \tag{1b}$$

$$\underline{\dot{x}}_3 = v_{DC} \underline{x}_5 - e^{-3jx_8} \underline{u}_1^* \underline{x}_7^* - 2\underline{x}_7 u_2 - j\omega \underline{x}_3$$
(1c)

$$\underline{\dot{x}}_4 = v_{DC} \, \underline{x}_7 - e^{-3jx_8} \, \underline{x}_5^* \, \underline{v}_{y\Delta}^* - 2\underline{x}_5 \, u_2 - 2x_6 \, \underline{v}_{y\Delta} - j\omega \underline{x}_2$$
 (1d)

$$\underline{\dot{x}}_5 = \frac{1}{L_z + M_z} (\underline{u}_3 - j\omega(L_z + M_z) \underline{x}_5) \tag{1e}$$

$$\underline{\dot{x}}_{5} = \frac{1}{L_{z} + M_{z}} (\underline{u}_{3} - j\omega(L_{z} + M_{z}) \underline{x}_{5})$$

$$\dot{x}_{6} = \frac{1}{L_{z} + M_{z}} u_{4}$$
(1e)

$$\underline{\underline{x}}_7 = \frac{1}{L}(\underline{u}_1 - (R + j\omega L)\underline{x}_7 - \underline{v}_g)$$
(1g)

$$\dot{x}_8 = \omega \tag{1h}$$

with

$$\underline{v}_{y\Delta} = \underline{u}_1 - M_z(j\omega\underline{i} + \underline{x}_7) \tag{2}$$

Parameters:  $v_{DC}$ ,  $\underline{v}_q$ ,  $\omega$ ,  $L_z$ ,  $M_z$ , R, L

Outputs:  $\underline{e}_s$ ,  $e_{s0}$ ,  $\underline{e}_d$ ,  $e_{d0}$ 

#### Assumptions 2.1

- 1. The cells of the arm k = 1, 2, ..., 6 are represented by one equivalent cell with the duty cycle  $q_k \in [0,1]$  and a voltage  $v_C k$  that accords the sum of the individual cells in the arm. This implies that the underlying problem of balancing the voltages within each arm has already been solved.
- 2. The load currents are assumed to be continuous, matched to the initial currents of the arm inductors, and satisfy the constraint  $i_1 + i_2 + i_3 = 0$ caused by junction N.

#### 2.2Exemplary parameter values

Parameter Name	Symbol	Value	Unit
DC voltage	$v_{DC}$	300	V
grid voltage	$v_g$	235	V
angular speed	$\omega$	$100\pi$	Hz
arm inductance	$L_z$	1.5	mH
mutual inductance	$M_z$	0.94	mH
load resistance	R	26	$\Omega$
load inductance	L	3	mH

## 3 Derivation and Explanation

The six arm energies  $e_{z1}$ , ...,  $e_{z6}$  transformed into

$$e_{s0} = 2g_0 \left[ (e_{z1}, e_{z3}, e_{z5})^T + (e_{z2}, e_{z4}, e_{z6})^T \right]$$
 (3a)

$$e_{d0} = 2g_0 \left[ (e_{z1}, e_{z3}, e_{z5})^T - (e_{z2}, e_{z4}, e_{z6})^T \right]$$
 (3b)

$$\underline{e}_s = 2g_{\alpha\beta} \left[ (e_{z1}, e_{z3}, e_{z5})^T + (e_{z2}, e_{z4}, e_{z6})^T \right]$$
(3c)

$$\underline{e}_d = 2\underline{g}_{\alpha\beta} \left[ (e_{z1}, e_{z3}, e_{z5})^T - (e_{z2}, e_{z4}, e_{z6})^T \right]$$
(3d)

with use of the Clark Transform

$$T_{0\alpha\beta} = \frac{1}{3} \begin{pmatrix} 1 & 1 & 1 \\ 2 & -1 & -1 \\ 0 & \sqrt{3} & -\sqrt{3} \end{pmatrix} = \begin{pmatrix} g_0 \\ g_\alpha \\ g_\beta \end{pmatrix}$$
(4)

and  $\underline{g}_{\alpha\beta} = e^{-j\theta}(g_{\alpha} + jg_{\beta}).$ 

The currents can be transformed as

$$i_{s0} = 2g_0 \left[ (i_{z1}, i_{z3}, i_{z5})^T + (i_{z2}, i_{z4}, i_{z6})^T \right]$$
(5a)

$$0 = g_0(i_1, i_2, i_3)^T (5b)$$

$$\underline{i}_s = \underline{g}_{\alpha\beta} \left[ (i_{z1}, i_{z3}, i_{z5})^T + (i_{z2}, i_{z4}, i_{z6})^T \right]$$
 (5c)

$$\underline{i} = \underline{g}_{\alpha\beta} (i_1, i_2, i_3)^T. \tag{5d}$$

The voltages are transformed as

$$v_{x0} = g_0 \left[ v_{DC}(1, 1, 1)^T - (v_{q1}, v_{q3}, v_{q5})^T - (v_{q2}, v_4, v_{q6})^T \right]$$
(6a)

$$v_{y0} = g_0 (v_{y1}, v_{y2}, v_{y3})^T \tag{6b}$$

$$\underline{v}_x = \underline{g}_{\alpha\beta} \left[ v_{DC}(1, 1, 1)^T - (v_{q1}, v_{q3}, v_{q5})^T - (v_{q2}, v_4, v_{q6})^T \right]$$
(6c)

$$\underline{v}_y = \underline{g}_{\alpha\beta} (v_{y1}, v_{y2}, v_{y3})^T \tag{6d}$$

$$= \underline{g}_{\alpha\beta} \left[ (v_{q2}, v_4, v_{q6})^T - (v_{q1}, v_{q3}, v_{q5})^T - (L_z - M_z) \frac{d}{dt} (i_1, i_2, i_3)^T \right].$$
 (6e)

### References

[1] Fehr, H.; Gensior, A.: Improved Energy Balancing of Grid-Side Modular Multilevel Converters by Optimized Feedforward Circulating Currents and Common-Mode Voltage, IEEE 2018