

Lab 4: Preliminary IR Transmitter/Receiver Development

Prelab Questions:

Q1 - $\frac{12-1.35}{100\text{mA}} = 106.5\Omega$ are required for the given design specifications

Q2 - $P = I^2R = (100\text{mA})^2(106.5\Omega) = 1.065 \text{ Watts}$ will be absorbed by the resistor

Q3 - $\frac{12V}{RE} = \frac{12V}{50000\Omega} = I = 240 \mu\text{A}$ through the potentiometer

Q4 - $V@0\text{ ohm} = 12V$ $V@50\text{k ohm} = 0V$ range = 0V to 12V

Start 10/22

1. Construct Light meter circuit

$$RE = 1.35 \Omega$$

$$V_{out\ ambient} = 2.8675V + 11.394V = 14.276V$$

$$RE \text{ but covered} = \text{minimum drop } 0.33331V$$

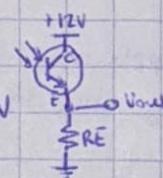
RE Vout partially covered = almost no change

RE Vout highest light intensity = significant voltage increase

$$RE = 0.31\text{k}\Omega \quad V_{out\ ambient} = 24.496\text{mV}$$

RE Vout ambient

$$1\text{M}\Omega \text{ ambient} = 11.394V$$



$$RE = 1.35\Omega$$

$RE = 0.39855 \text{ M}\Omega$ using Potentiometer

$$RE = 507.96 \text{ k}\Omega$$

Vout drops when the oscilloscope is connected

AC Coupling

Freq: 8.10 MHz

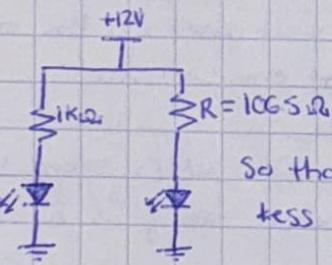
AC RMS - Cyc: 73.5mV

Period: 123.5ns

Avg - Cyc: -139.2mV

$$\frac{100}{73.5/(2 \times 139.2)} = 26.40\%$$

2. Construct IR Light Source



The 1kΩ will be exchanged for 5 200Ω

The 106.5Ω will be exchanged for 4.24Ω and 1 10Ω

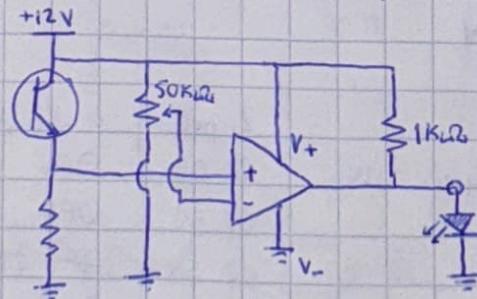
So that the power dissipated by each resistor is less than $\frac{1}{4}$ Watt

Distance

Voltage (Ambient Light)

0 in	11.9415V
1in	11.8310V
2in	11.8190V
3in	11.8118670V
4in	11.8101V
5in	11.2364V
6in	10.910V
1ft	10.0237V
2ft	9.7774V

3. Construct simple receiver



The voltage across the potentiometer is 10.3206V = Reference Voltage

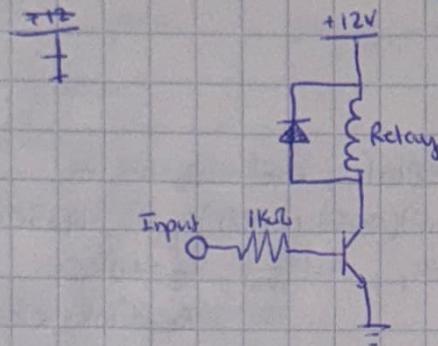
The range of the Receiver / Transmitter system is limited by the length of the wires connected to the power supply

The LED's brightness did not seem to be affected by the distance between the transmitter and receiver

Ambient Lighting may have influenced the reading of the phototransistor

Lab 4: Continued

Connecting the circuit to the Relay



The output of the OP Amp is connected to the Input of the BJT.

10/27

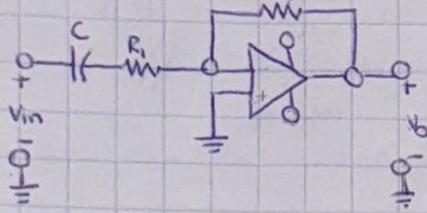
Lab 5: IR Remote Control Development Part 1

Prelab Questions:

$$Q1 - \frac{2V-0.3V}{75mA} = 116\Omega \approx R$$

Q2 - Since Period = $0.693CT(R_1 + 2R_2) = \frac{1}{f}$, $R_1 > 1\text{k}\Omega$, and $CT \geq \ln F$, the value for $R_1 = 2\text{k}\Omega$, $CT = \ln F$, and by the given equation $\frac{1}{5\text{kHz}} = 0.693(\ln F)(2\text{k}\Omega + 2(R_2))$, $R_2 = 14330\Omega \approx 15\text{k}\Omega$

Q3 -



Where $R_1 = 5\text{k}\Omega$, $R_2 = 150\text{k}\Omega$ and $C = 1.1\text{nF}$

$$f_0 = 1/(2\pi R_{eq}C), R_{eq} = R_E + R_1, A = \frac{R_2}{R_{eq}}, R_E = 10\text{k}\Omega$$

$$f_0 = 10\text{kHz}$$

$$R_{eq} = 10k + 40k = 50\text{k}\Omega$$

$$\text{If } R_2 = 500k \quad \uparrow$$

$$A = 10 = \frac{R_2}{10k + R_1} \rightarrow R_2 = 100k + 10R_1 \therefore 500k - 100k = 10R_1, 40k = R_1, \text{ Max}$$

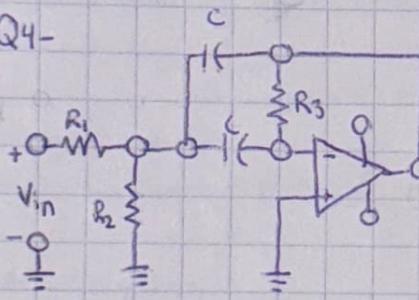
$$f_0 = 10\text{kHz} = 1/(2\pi(50k)C) \rightarrow C = 0.32\text{nF} < 0.5\text{nF} \times$$

$$f_0 = 10\text{kHz} = 1/(2\pi(10k)C) \rightarrow C = 1.1\text{nF} > 0.5\text{nF} \& < 0.1\text{nF} \checkmark$$

$$100k + 10(5k) = 150k = R_2 > 1\text{k}\Omega \& < 500\text{k}\Omega \checkmark$$

$$\left| \frac{150k}{10k + 50k} \right| = 10 = A \checkmark$$

Q4 -



$$10 = H_0 = H(f = f_0) = \frac{R_3}{2R_1}, Q = \frac{f_0}{B}, R_2 = \frac{Q}{\pi f_0 C}, R_1 = \frac{R_3}{2H_0}, R_2 = \frac{R_3}{4Q^2 - 2H_0}$$

$$Q = 10 \quad f_0 = 7.5\text{kHz} \quad \text{LET } C = 2\text{nF}$$

$$R_3 = \frac{10}{\pi(7.5\text{kHz})(2\text{nF})} = 212206.6\Omega$$

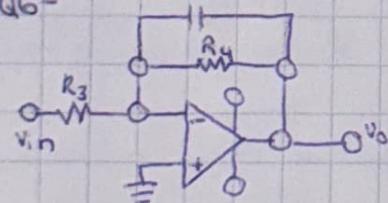
$$R_1 = \frac{212206.6}{2(10)} = 10610.3\Omega$$

$$R_2 = \frac{212206.6}{4(10)^2 - 2(10)} = 558.4\Omega$$

Where $R_1 = 10.6\text{k}\Omega$, $R_2 = 560\Omega$, $R_3 = 210\text{k}\Omega$, and $C = 2\text{nF}$

Q5 - Since $A = \frac{R_2}{R_1}$, a gain of 5 can be achieved by the rectifier by using an R_2 that is 5 times the value of R_1 . Using the resistor constraints ($1 < R < 100\text{k}$), $R_1 = 1\text{k}\Omega$ and $R_2 = 5\text{k}\Omega$

Q6 -



$$f_0 = 1/(2\pi R_4 C) \quad R_4 C < 0.1s, 5 < f_0 < 20\text{Hz}, A = \frac{R_4}{R_3}, R_4 \sim 1\text{M}\Omega$$

$$0.01s = 1\text{M}\Omega(C) \rightarrow C = 10\text{nF}$$

$$f_0 = 1/(2\pi(0.01)) = 159\text{Hz} < 20\text{Hz} \& > 5\text{Hz} \checkmark$$

$$A = 3 = \frac{1\text{M}\Omega}{R_3} \rightarrow R_3 = 333\text{k}\Omega$$

Where $R_3 = 333\text{k}\Omega$, $R_4 = 1\text{M}\Omega$, and $C = 10\text{nF}$

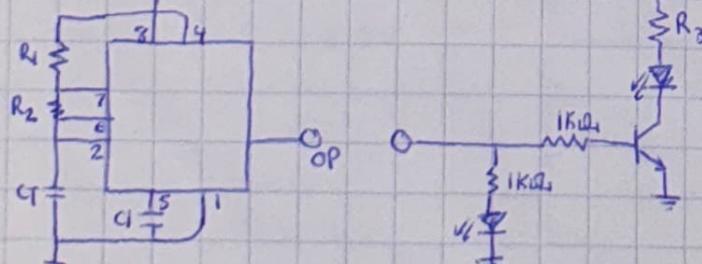
4.895 kHz

4.78 kHz

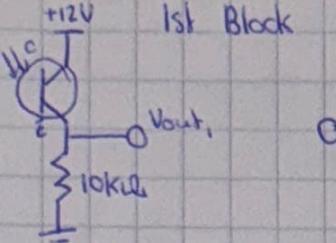
The observed frequency is 5.34 kHz and the LED remains lit. R_3 is hot. Constructed on Circuit Board 2211#4

Where $VCC = 9V$, $R_1 = 1K$, $R_2 = 4.7K$, $CT = 25\text{nF}$, $C1 = 0.1\mu\text{F}$

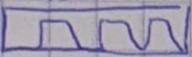
$$\frac{22}{6.2K}$$



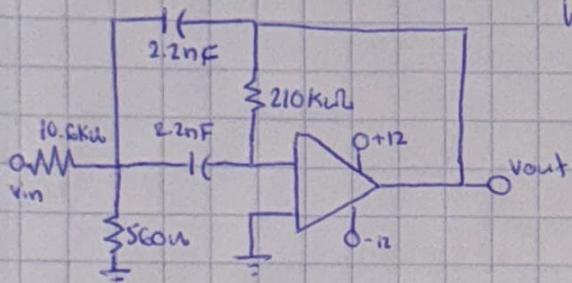
1st Block



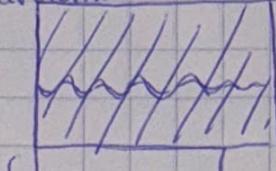
No recognizable waveform



3rd Block: Bandpass



Waveform:



$f_0 = 4.9 \text{ kHz}$
 $f_0 = 7.5 \text{ kHz}$

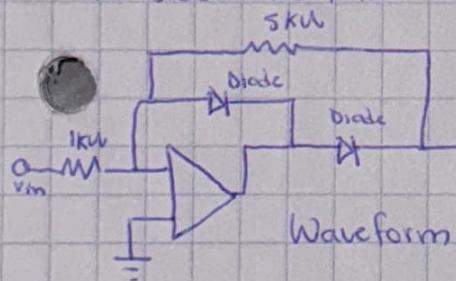
from oscilloscope

The bandpass filter is further from V_{out_2} resulting in a cleaner sinusoidal waveform. $H_0 = \frac{R_3}{2R_1} = 9.91$

$$f_0 = 4.9 \text{ kHz}, R_3 = 293 \text{ k}\Omega, R_1 = 14.7 \text{ k}\Omega, R_2 = 777 \text{ }\Omega$$

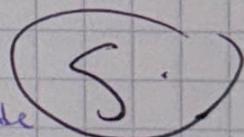
Matlab: The bandpass.m and setup-equip.m files are now obsolete as Matlab has switched from supporting the visa interface to the visadev interface upon converting, the setup-equip.m requires an unknown argument.

4th Block: Rectifiers



Waveform

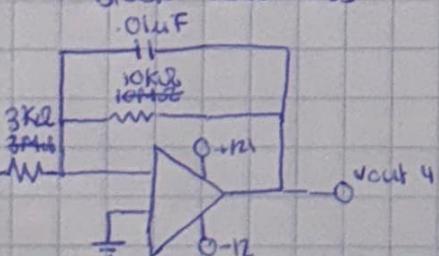
The rectifier ~~transforms~~ smooths the sinusoidal waveform from further ~~smoothed to square~~. The frequency of the waveform is the same as the waveform observed leaving the bandpass filter. As the transmitter was moved away from the receiver the amplitude increased.



As the transmitter was moved away from the receiver the amplitude increased. The rectifier is clipped for higher frequencies since the waveform has passed through the highpass filter.

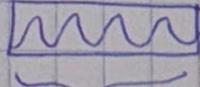
$\square\square @ R_2 = 5 \text{ k}\Omega$
 $\wedge\wedge @ R_2 = 3.3 \text{ k}\Omega$

5th Block: Low Pass



Waveform:

$$f_0 = 5 \text{ Hz} = 5 \text{ kHz}$$



272 μs

$$f_0 = 3.6765 \text{ kHz}$$

from oscilloscope

Similarly the low pass amp filter smooths out the sinusoidal waveform and has the same frequency as the previous waveforms. The gain is $\frac{R_4}{R_3} = 3.3$. It would not matter if the waveform was clipped for higher intensities because the low pass filter only allows frequencies lower than its frequency.

