

Figure 1: Sketch of the system with mass $m = 1$ kg, length to center of mass $l = 0.5$ m, and a friction given by $\tau_f = b\dot{\theta}$ with $b = 0.1$ Nm/(rad/s).

1. System Modeling

- Setup a model (nth-order differential equation) of the system.
- Reformulate the system model as a system of 1st order differential equations.
- Derive a linearized model of the system at an angle of $\pi/3$ rad.

Equation: (Rotierende Newtons 2. lov)

$$I\alpha = u - \tau_f + \sin(\theta) \cdot g \cdot l \cdot m$$

$$I = ml^2 \text{ (Punktmass)}$$

$$I\ddot{\theta} = u - b\dot{\theta} + \sin(\theta) \cdot m \cdot g \cdot l$$

$$\ddot{\theta} = \frac{u}{I} - \frac{b\dot{\theta}}{I} + \frac{\sin(\theta) \cdot m \cdot g \cdot l}{I}$$

As 2 first order equations;

$$x = \begin{bmatrix} \theta \\ \dot{\theta} \end{bmatrix}$$

$$\begin{aligned} \dot{x} &= f(x, u) = \begin{bmatrix} x_2 \\ \frac{u}{I} - \frac{b x_2}{I} + \frac{\sin(x_1) \cdot m \cdot g \cdot l}{I} \end{bmatrix} \\ &= \begin{bmatrix} x_2 \\ \frac{u}{I} - \frac{b x_2}{I} + \frac{\sin(x_1) \cdot g}{l} \end{bmatrix} = \begin{bmatrix} f_1 \\ f_2 \end{bmatrix} \end{aligned}$$

Linearize at $\theta = \frac{\pi}{3}$, $\dot{\theta} = 0$, $\ddot{\theta} = 0$:

$$\bar{x}_1 = \frac{\pi}{3}, \quad \bar{x}_2 = 0, \quad \dot{x} = 0$$

$$\bar{u} = -m l g \sin\left(\frac{\pi}{3}\right) \text{ (counteracts gravity at the point)}$$

$$\hat{x} = x - \bar{x}, \quad \hat{u} = u - \bar{u}$$

Linearizing by first order Taylor expansion:

$$\dot{x} \approx \underbrace{f(\bar{x}, \bar{u})}_0 + \left. \frac{\partial f}{\partial x} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}} \hat{x} + \left. \frac{\partial f}{\partial u} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}} \hat{u}$$

$$\frac{\partial f}{\partial x} = \begin{bmatrix} 0 & 1 \\ \frac{\cos(x_1) \cdot g}{L} & -\frac{b}{I} \end{bmatrix}$$

$$\frac{\partial f}{\partial u} = \begin{bmatrix} 0 \\ \frac{1}{I} \end{bmatrix}$$

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ \frac{\cos(x_1) \cdot g}{L} & -\frac{b}{I} \end{bmatrix} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} \hat{x} + \begin{bmatrix} 0 \\ \frac{1}{I} \end{bmatrix} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} \hat{u}$$

$$= \begin{bmatrix} 0 & 1 \\ \frac{\cos(\frac{\pi}{3}) \cdot 9,82}{0,5} & -\frac{0,1}{0,25} \end{bmatrix} \cdot \hat{x} + \begin{bmatrix} 0 \\ \frac{1}{0,25} \end{bmatrix} \cdot \hat{u}$$

$$= \underbrace{\begin{bmatrix} 0 & 1 \\ 9,82 & -0,4 \end{bmatrix}}_A \cdot \left(x - \frac{\pi}{3}\right) + \underbrace{\begin{bmatrix} 0 \\ 4 \end{bmatrix}}_B \cdot (u - 4,2522)$$

$$y = \underbrace{\begin{bmatrix} 1 & 0 \end{bmatrix}}_C \begin{bmatrix} \theta \\ \dot{\theta} \end{bmatrix} + \underbrace{0}_D \cdot u$$

We measure position as output.

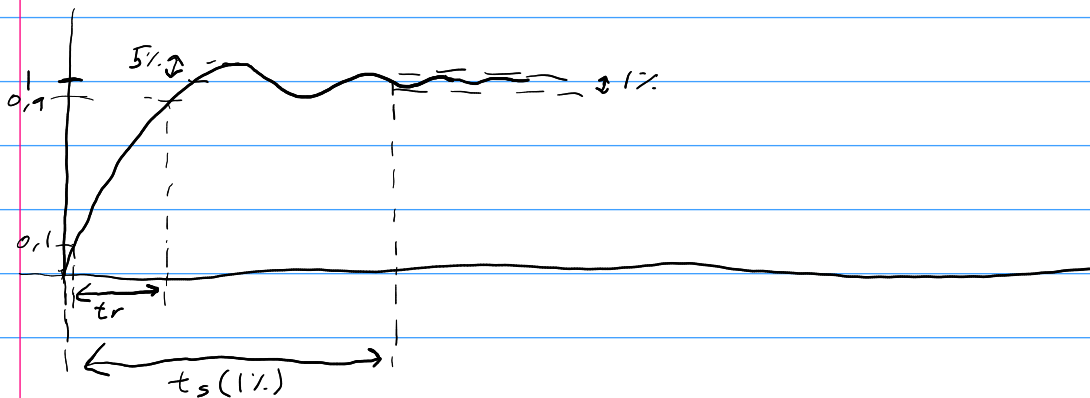
2. Performance Specification

- (a) Specify a desired performance of the system in time-domain.
- (b) Set up a corresponding performance specification in frequency-domain.

a) $M_p = 5\%$ (Overshoot)

$$t_r = 1s$$

$$t_s(1\%) = 4s$$



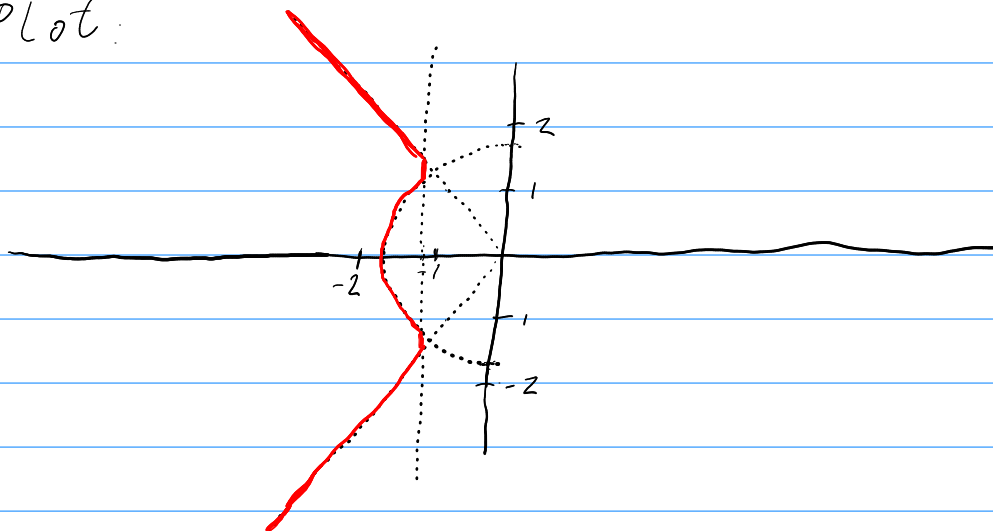
Frequency domain:

$$\omega_n = \frac{1.8}{t_r} = \frac{1.8}{1} = 1.8$$

$$\zeta = \sqrt{\frac{\left(\frac{\ln(M_p)}{-\pi}\right)^2}{1 + \left(\frac{\ln(M_p)}{-\pi}\right)^2}} = \sqrt{\frac{\left(\frac{\ln(0.05)}{-\pi}\right)^2}{1 + \left(\frac{\ln(0.05)}{-\pi}\right)^2}} = 0.6901$$

$$\sigma = \frac{-\ln(1/100)}{4} = 1.1513$$

Plot:



3. Controller Design

- (a) Design a PID controller for the linearized model of the system such that it attains the desired performance. The tuning procedure should be described (the template for the solution indicates that the controller type should be found by iterating over controller types starting with a P-controller).

sys =

$$\frac{4}{s^2 + 0.4s + 17}$$

characteristic equation:

$$1 + K(s)G(s) = 0$$

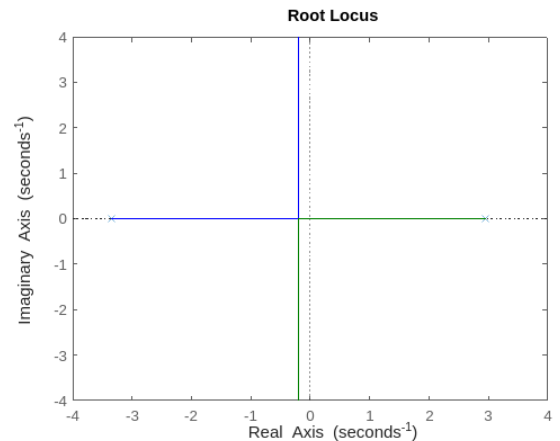
$$G(s) = \frac{4}{s^2 + 0.4s + 17}$$

P-controller:

$$K(s) = k_p \rightarrow 1 + K(s)G(s) = 0$$

$$1 + k_p G(s) = 0$$

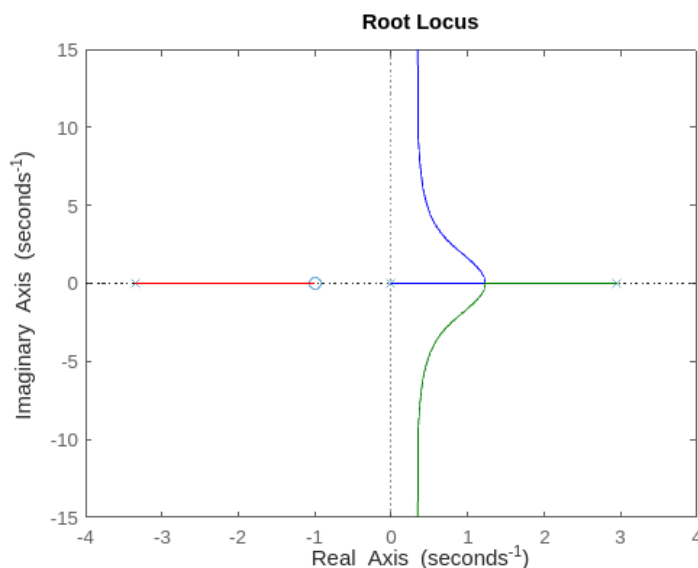
Not enough



PI-controller
(Automatic reset)

$$K(s) = k_p \left(1 + \frac{1}{T_i s}\right)$$

$$1 + K(s)G(s) = 0 \rightarrow 1 + k_p \left(1 + \frac{1}{T_i s}\right) \left(\frac{4}{s^2 + 0.4s + 17}\right)$$



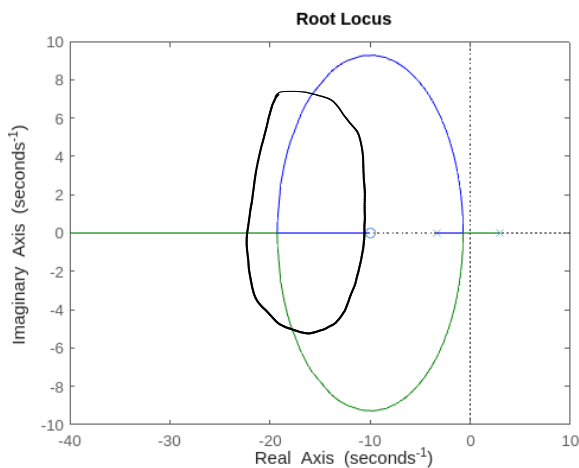
Also
wrong

PD-controller

$$K(s) = K_p(1 + T_d s)$$

$$1 + K(s)G(s) = 0$$

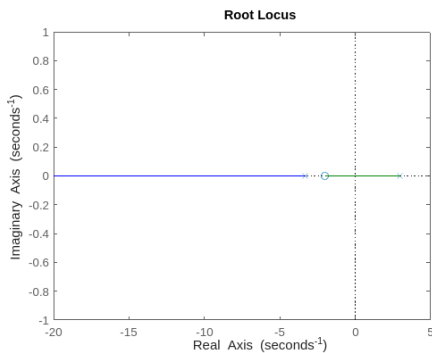
$$(1 + K_p(1 + T_d s)) \left(\frac{4}{s^2 + 0.45s - 9.82} \right) = 0$$



$$T_d = \frac{1}{10}$$

Hvis man vælger et polpar indenfor den sorte cirkel, troede jeg det ville virke, men stepresponsen ser forkert ud.

$$T_d = 1/2$$



```
index = 52;
poles(:,index)
```

```
ans = 2x1
    -708.4895
    -1.9995
```

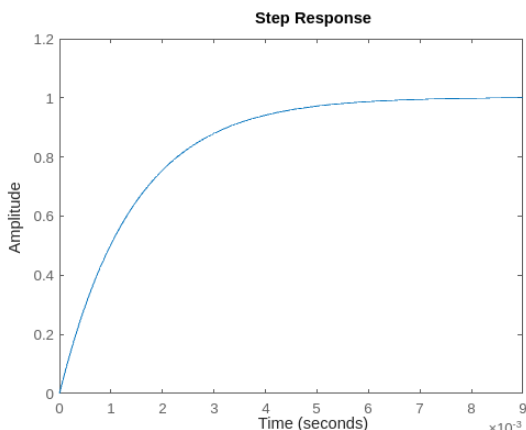
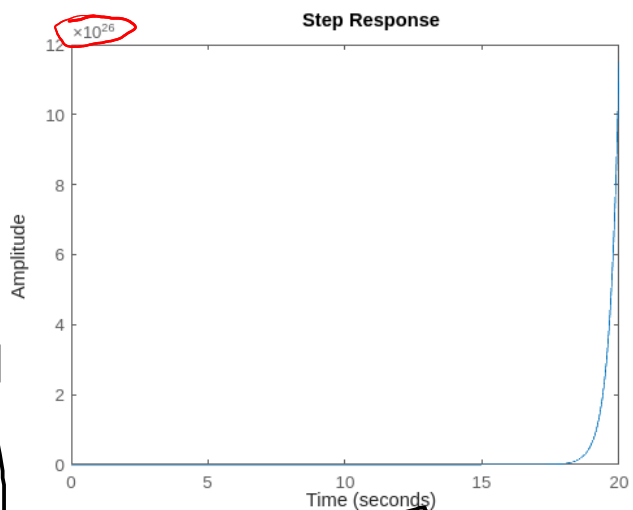
```
gains(1,index)
```

```
ans = 351.0000
```

```
G = gains(1,index)*(1+T_d*s);
sys1 = G*sys / (1+G*sys)
```

```
sys1 =
```

```
702 s^3 + 1685 s^2 - 6332 s - 1.379e04
-----
s^4 + 702.8 s^3 + 1665 s^2 - 6340 s - 1.369e04
```



```
stepinfo(sys1)
```

```
ans = struct with fields:
    RiseTime: 0.0032
    TransientTime: 0.0060
    SettlingTime: 0.0060
    SettlingMin: 0.9066
    SettlingMax: 1.0026
    Overshoot: 0
    Undershoot: 0
    Peak: 1.0026
    PeakTime: 0.0400
```