

# Recitation Class for Final Exam

## Chapter 10-12

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2021.08.03

# Outline

Chapter 10 - Fundamentals of the Metal–Oxide–Semiconductor Field-Effect Transistor

Chapter 11 - Metal–Oxide–Semiconductor Field-Effect Transistor: Additional Concepts

Chapter 12 - Bipolar Junction Transistor

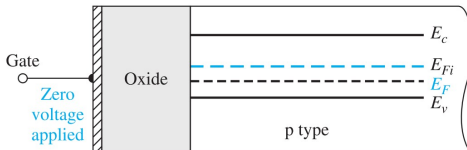
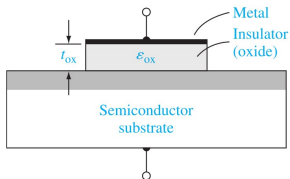
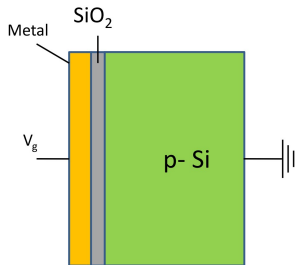
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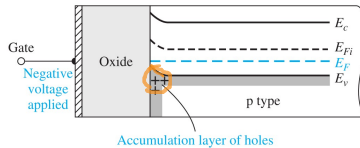
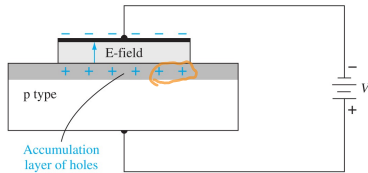
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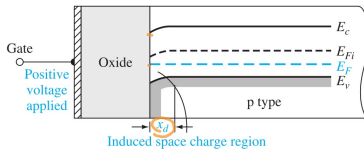
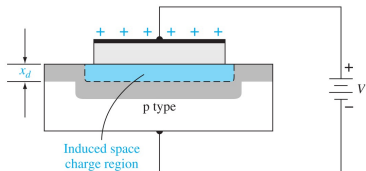
# Metal–Oxide–Semiconductor



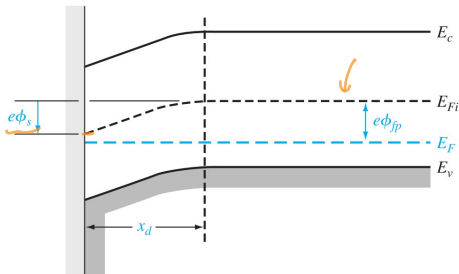
# Negative Gate Voltage



# Positive Gate Voltage



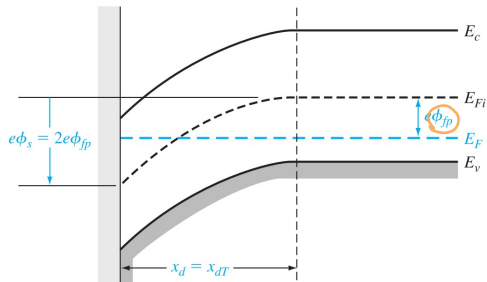
# Depletion Layer Thickness



$$\phi_{fp} = V_t \ln \left( \frac{N_a}{n_i} \right)$$
$$x_d = \left( \frac{2\epsilon_s \phi_s}{eN_a} \right)^{1/2}$$

$\phi_s$ : the surface potential, is the difference (in V) between  $E_{Fi}$  measured in the bulk semiconductor and  $E_{Fi}$  measured at the surface.

# Threshold Inversion Point



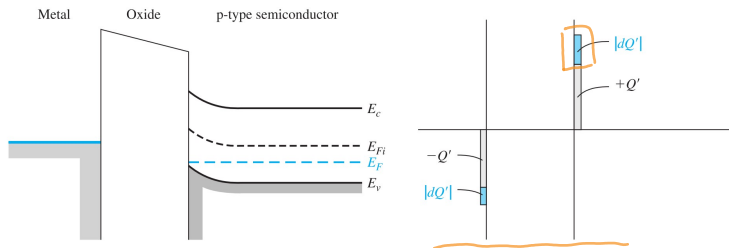
$$n_i = N_c \exp\left(\frac{E_F - E_{Fi}}{kT}\right)$$

$$\phi_s = 2\phi_{fp}$$

$$x_{dT} = \left( \frac{4\epsilon_s \phi_{fp}}{eN_a} \right)^{1/2}$$

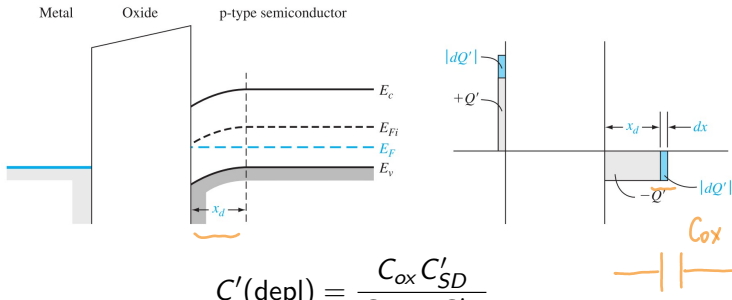


# Accumulation



$$C'(\text{acc}) = C_{ox} = \frac{\epsilon_{ox}}{t_{ox}}$$

# Depletion

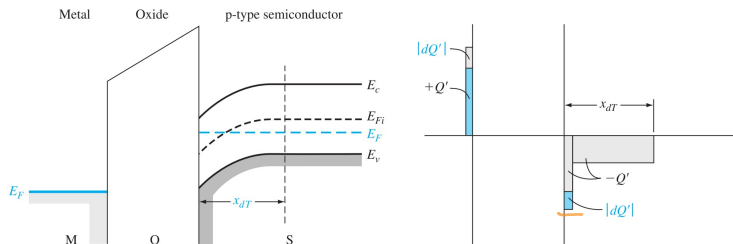


$$C'(\text{depl}) = \frac{C_{ox} C'_{SD}}{C_{ox} + C'_{SD}}$$

$$= \frac{\epsilon_{ox}}{t_{ox} + \left( \frac{\epsilon_{ox}}{\epsilon_s} \right) x_d}$$

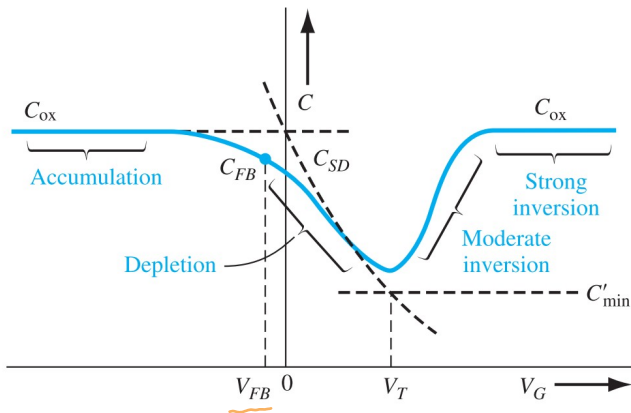
$$C'_{min} = \frac{\epsilon_{ox}}{t_{ox} + \left( \frac{\epsilon_{ox}}{\epsilon_s} \right) x_{dT}}$$

# Inversion



$$C'(\text{inv}) = C_{ox} = \frac{\epsilon_{ox}}{t_{ox}}$$

# Ideal Low-Frequency C-V Curve

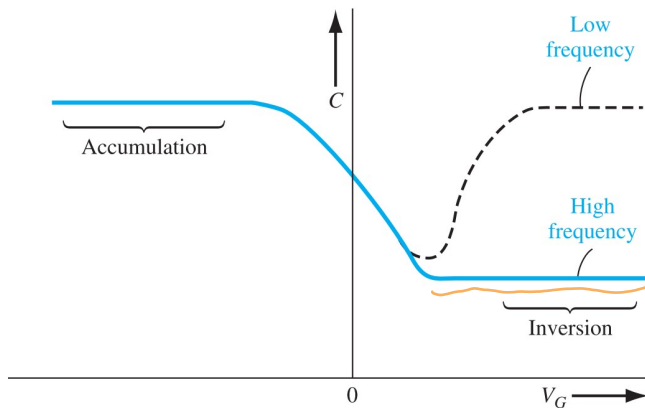


$$C'_{FB} = \frac{\epsilon_{ox}}{t_{ox} + \left( \frac{\epsilon_{ox}}{\epsilon_s} \right) \sqrt{\left( \frac{kT}{e} \right) \left( \frac{\epsilon_s}{eN_a} \right)}}$$

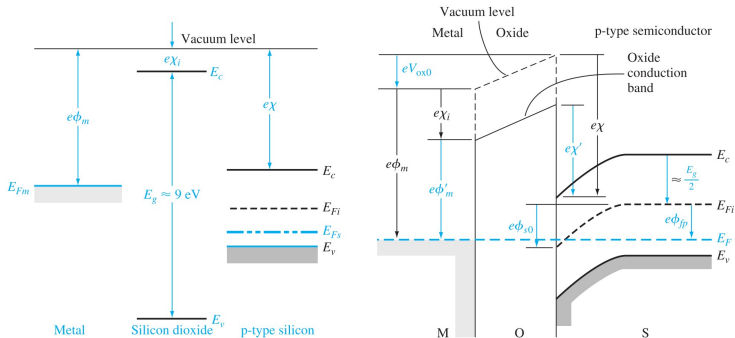
# Frequency Effects

Two sources of electrons

1. Diffusion of minority carrier electrons.
2. Thermal generation of electron-hole pairs within the space charge region.



# Work Function Difference



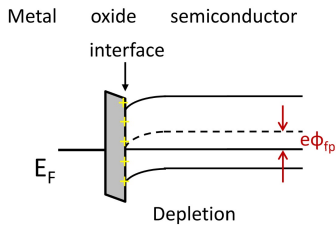
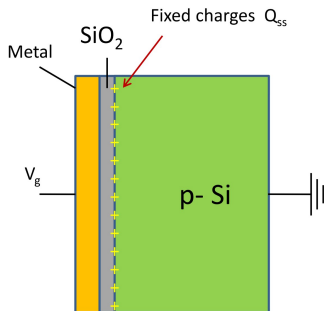
$$\underline{\phi_{ms}} = \phi'_m - \left( \chi' + \frac{E_g}{2e} + \phi_{fp} \right)$$

$$\phi_{ms} = \phi'_m - \left( \chi' + \frac{E_g}{2e} - \phi_{fn} \right)$$

$V = 0$

Not required.

# Fixed Charge



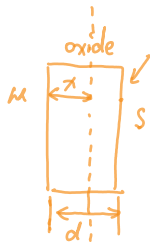
## Adjustment on $V_T$

$$|Q'_{SD}(\max)| = eN_a x_{dT} = 2\sqrt{e\epsilon_s N_a \phi_{fp}}$$

$$V_{TN} = \frac{|Q'_{SD}(\max)|}{C_{ox}} + V_{FB} + 2\phi_{fp}$$

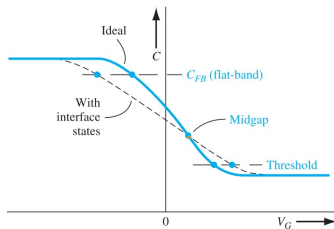
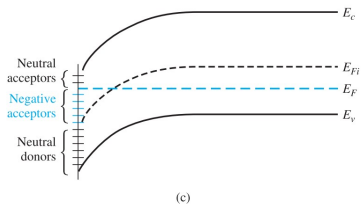
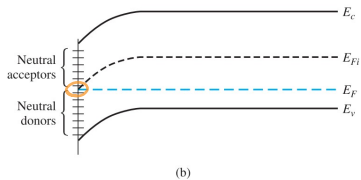
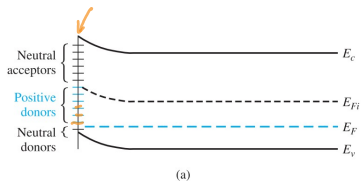
$$V_{TP} = -\frac{|Q'_{SD}(\max)|}{C_{ox}} + V_{FB} - 2\phi_{fn}$$

$$V_{FB} = \phi_{ms} - \frac{Q'_{ss}}{C_{ox}} \cdot \frac{x}{d} = 1$$

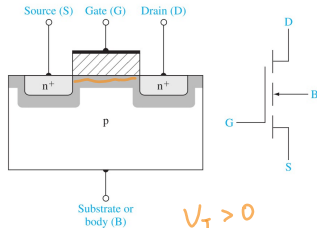




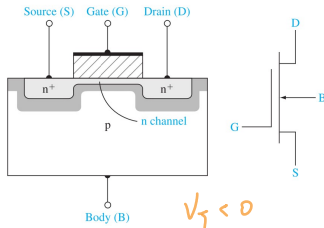
# Surface States



# MOSFET



(a) n-channel enhancement MOSFET



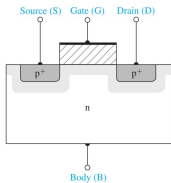
(b) n-channel depletion MOSFET

Enhancement mode: the semiconductor substrate is not inverted directly under the oxide with zero gate voltage.

Depletion mode: a p-channel region exists under the oxide with 0V applied to the gate.

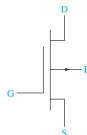
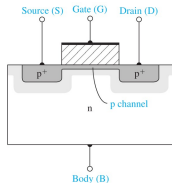
# MOSFET

$$V_T < 0$$



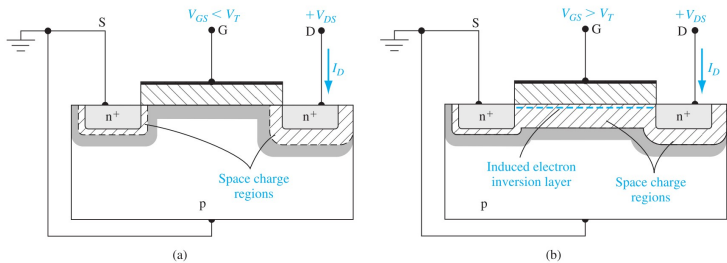
(a) p-channel enhancement MOSFET

$$V_T > 0$$



(b) p-channel depletion MOSFET

$$V_{GS} - V_T$$



- ▶  $V_{GS} < V_T$ : no inversion layer, no current.
- ▶  $V_{GS} > V_T$ : inversion layer created, current flow from drain to source.

$V_{DS}$  when  $V_{GS} > V_T$

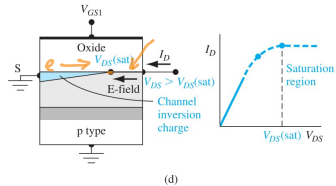
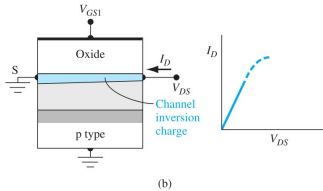
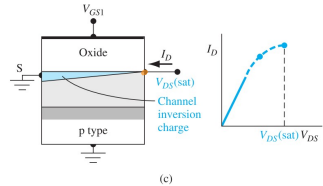
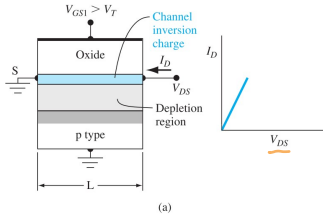
- ▶  $V_{DS}$  low ( $V_{DS} < V_{GS} - V_T$ ): act as a controllable resistor.

$$I_D = \mu_n C_{ox} \frac{W}{L} \left( V_{GS} - V_T - \frac{V_{DS}}{2} \right) V_{DS}$$

- ▶  $V_{DS}$  high ( $V_{DS} \geq V_{GS} - V_T$ ): saturation

$$I_D = \mu_n C_{ox} \frac{W}{2L} (V_{GS} - V_T)^2 \quad (\text{saturation})$$

$$V_{GS} - V_T$$

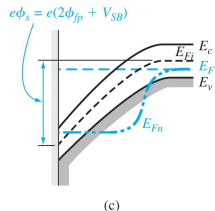
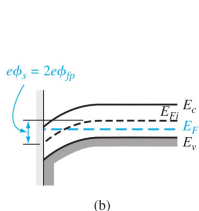
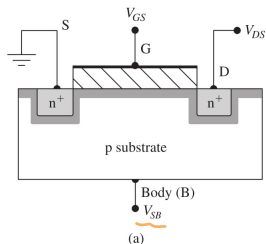


After pinch off: the electrons are injected into the space charge region where they are swept by the E-field to the drain contact.

# Transconductance

$$g_m = \frac{\partial I_D}{\partial V_{GS}}$$
$$= \begin{cases} \mu_n C_{ox} \frac{W}{L} V_{DS}, & 0 < V_{DS} < V_{GS} - V_T \\ \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_T), & V_{DS} > V_{GS} - V_T \end{cases}$$

# Substrate Bias Effects



$$V_{SB} = 0 : Q'_{SD}(\text{max}) = -\sqrt{2e\epsilon_s N_a (2\phi_{fp})}$$

$$V_{SB} > 0 : Q'_{SD} = -\sqrt{2e\epsilon_s N_a (2\phi_{fp} + V_{SB})}$$

$$\Delta V_T = -\frac{\Delta Q'_{SD}}{C_{ox}} = \frac{\sqrt{2e\epsilon_s N_a}}{C_{ox}} \left[ \sqrt{2\phi_{fp} + V_{SB}} - \sqrt{2\phi_{fp}} \right]$$

where  $\Delta V_T = V_T(V_{SB} > 0) - V_T(V_{SB} = 0)$  for NMOS.



# Substrate Bias Effects

$$\Delta V_T = -\frac{\Delta Q'_{SD}}{C_{ox}} = \frac{\sqrt{2e\epsilon_s N_a}}{C_{ox}} \left[ \sqrt{2\phi_{fp} + V_{SB}} - \sqrt{2\phi_{fp}} \right]$$
$$= \gamma \left[ \sqrt{2\phi_{fp} + V_{SB}} - \sqrt{2\phi_{fp}} \right], \quad \gamma = \frac{\sqrt{2e\epsilon_s N_a}}{C_{ox}}$$

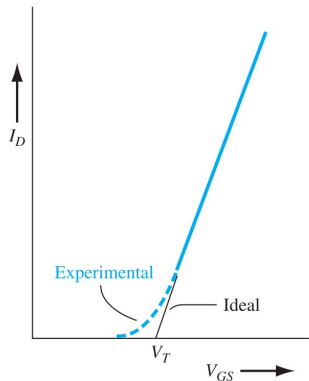
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# I - Subthreshold Conduction (Leakage Current)

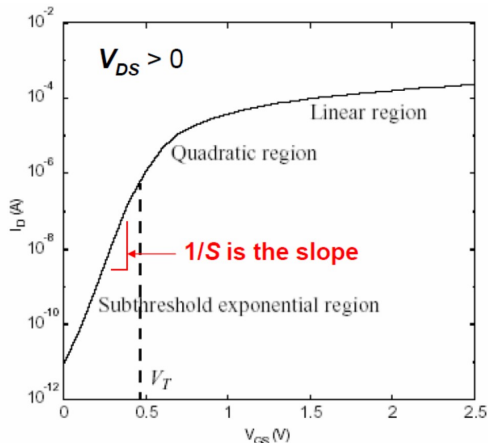


**Figure:** Comparison of ideal and experimental plots of  $\sqrt{I_D}$  versus  $V_{GS}$ .

When  $V_{GS} < V_T$ ,  $I_D \propto \exp\left(\frac{qV_{GS}}{nkT}\right)$

# I - Subthreshold Conduction (Leakage Current)

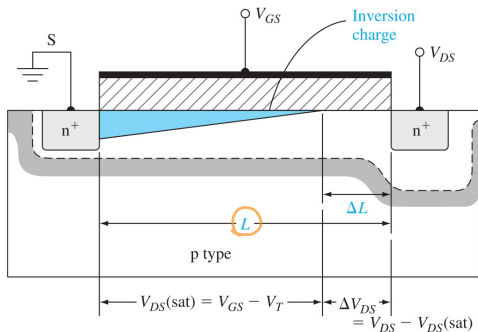
Slope Factor: defined to be the inverse slope of the  $\log(I_D)$  vs.  $V_{GS}$  characteristic in the subthreshold region.



$$S = n \left( \frac{kT}{q} \ln(10) \right) \text{ (Volts per decade)}$$

$S \geq 60 \text{ mV/dec}$  at room temperature.

## II - Channel Length Modulation



$L \sim \Delta L$

$$I'_D = \mu_n C_{ox} \frac{W}{2L} \left[ (V_{GS} - V_T)^2 (1 + \lambda V_{DS}) \right]$$



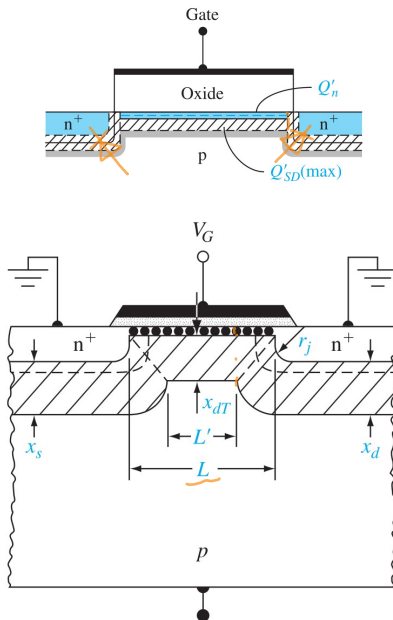
### III - Velocity Saturation

$$E = \frac{V}{d} \quad \text{MOS}$$

$$I_{DSAT} = WC_{ox} \left[ V_{GS} - V_T - \frac{V_{DSAT}}{2} \right] v_{sat}$$

where  $V_{DSAT} = \frac{L}{\mu_n} v_{sat}$ .

## IV - Short Channel Effect



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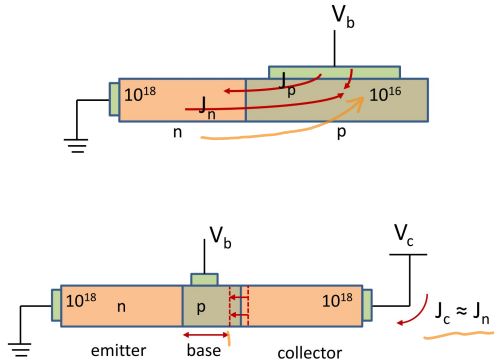
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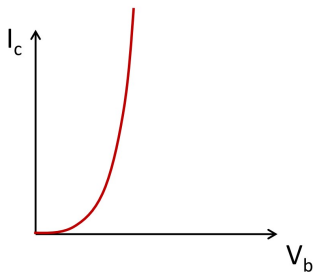
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# Bipolar Junction Transistor



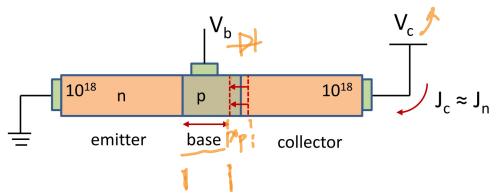
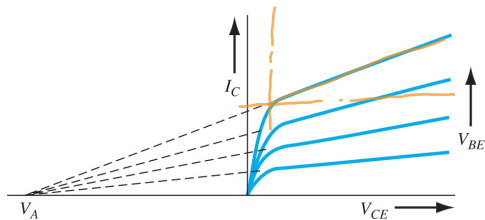
# Current Volatage Relationship



$$I_c = \beta I_s \left( e^{\frac{eV_b}{nkT}} - 1 \right)$$

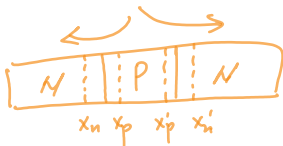
1. Narrower base  $\rightarrow$  larger gain
2.  $\beta \approx \underline{N_D/N_A}$ , higher emitter-to-base ratio  $\rightarrow$  higher gain

# Early Effect



Trade off:

Gain and Early effect.



Good luck to your final exam!