VE373 Recitation Class

Week 6

2022.06.18

L9 — Output Compare and PWM

1. PWM Signal (Pulse-width modulation)

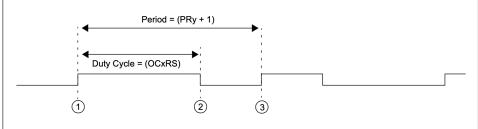
Information is encoded into the signal width.

- Advantages of PWM
 - Average value proportional to duty cycle
 - Low power used in transistors used to switch the signal
 - Fast switching possible due to MOSFETS and power transistors at speeds in excess of 100 kHz
 - Digital signal is resistant to noise less
 - Heat dissipated versus using resistors for intermediate voltage values
- Disadvantages of PWM
 - Cost Complexity of circuit
 - Radio Frequency Interference
 - Voltage spikes
 - Electromagnetic noise

2. PWM Operation Mode

In PWM mode, the <code>OCxR</code> register is a **read-only** slave duty cycle register and <code>OCxRS</code> is a buffer register that is written by the user to update the PWM duty cycle.

Figure 16-17: PWM Output Waveform



- (1) Timery is cleared and the new duty cycle value is loaded from OCxRS into OCxR.
- (2) Timer value equals the value in the OCxR register; OCx Pin is driven low.
- Timer overflow; value from OCxRS is loaded into OCxR; OCx pin is driven high. TylF interrupt flag is asserted.

3. Calculations

$$T_{\rm PWM} = (PR+1) \times T_{\rm PB} \times {\rm Timer~Prescale~Value}$$

$$2^{\rm PWMResolution} = \frac{T_{\rm PWM}}{T_{\rm timer}}$$

$${\rm PWMResolution~(bits)} = \log_2 \left(\frac{F_{\rm PB}}{F_{\rm PWM} \times {\rm Timer~Prescale~Value}} \right)$$

4. Comfiguration

- 1. Set the PWM period by writing to the selected timer period register (PRy).
- 2. Set the PWM duty cycle by writing to the OCxRS register.

- 3. Write the OxCR register with the initial duty cycle.
- 4. Enable interrupts, if required, for the timer and Output Compare modules. The output compare interrupt is required for PWM Fault pin utilization.
- 5. Configure the Output Compare module for one of two PWM Operation modes by writing to the Output Compare mode bits, OCM<2:0> (OCxCON<2:0>).
- 6. Set the TMRy prescale value and enable the time base by setting TON (TxCON<15>) = 1.

L10 — Analog-to-Digital Conversion (ADC)

1. Nyquist Theorem

A bandlimited analog signal that has been sampled can be perfectly reconstructed from an infinite sequence of samples if the sampling rate f_s exceeds $2f_{max}$ samples per second, where f_{max} is the highest frequency in the original signal.

In short:

$$f_{\text{sampling}} > 2f_{\text{original}}$$

2. Aliasing

Aliasing is when the digital signal appears to have a different frequency than the original analog signal.

If the analog signal does contain frequency components larger than $f_s/2$, then there will be an aliasing error.

Aliasing example:

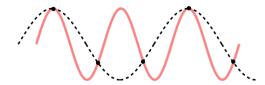
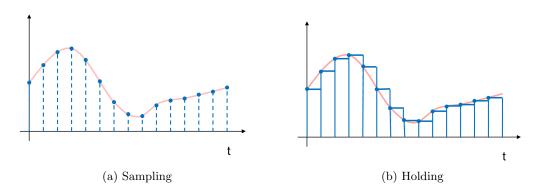


Figure 1: The samples of two sine waves can be identical when at least one of them is at a frequency above half the sample rate.

3. Steps to convert analog signal to digital signal

1. Sampling & Holding



2. Quantization

- \bullet Separating the input signal into a discrete states with K increments
- $K = 2^N$
- N is the number of bits of the ADC

3. Coding

Assigning a unique digital code to each state for input into the microprocessor

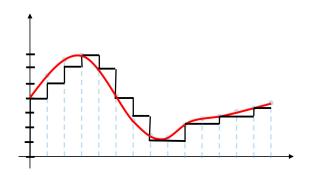


Figure 3: Quantization and Coding

4. Resolution

Amount of information the digital code is able to represent

$$Q = \frac{V_{\text{max}} - V_{\text{min}}}{2^N}$$

Larger range means lower resolution given fixed number of bits

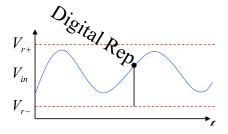
5. Quantization error

How far off discrete value is from actual value

Quantization error =
$$\frac{1}{2}Q$$

Larger range means larger quantization error

6. Digital representation



Digital Rep =
$$2^N \cdot \frac{V_{in} - V_{r-}}{V_{r+} - V_{r-}}$$

 $V_{in} = \text{Digital Rep} \cdot \frac{V_{r+} - V_{r-}}{2^N} + V_{r-}$

7. ADC implementation

1. Method 1: Voltage Divider

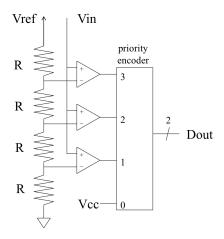


Figure 4: Voltage Divider

2. Method 2: Successive Approximation

Most used in PIC MCUs.

Example: 4-bit resolution, $2^4 = 16$ discrete representations

- B3 = 1 (upper half of 1/2)
- B2 = 0 (lower half of 3/4)
- B1 = 0 (lower half of 5/8)
- B0 = 1 (upper half of 9/16)

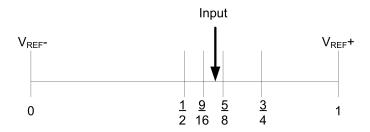
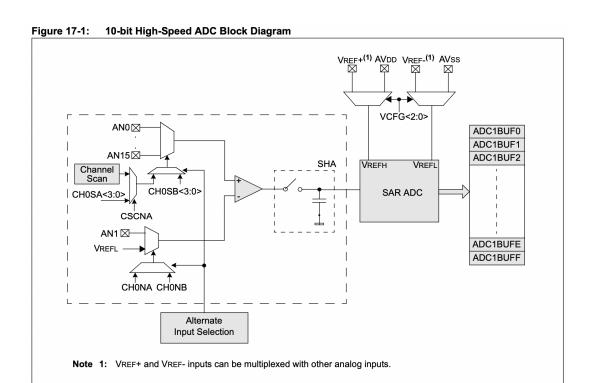


Figure 5: Successive Approximation

8. ADC of PIC32

- 10 bits, i.e. N = 10.
- 16 analog input channels
- 8 conversion result formats
- Up to 16 words storage for converted digital value
- Selectable conversion trigger source

9. ADC Block Diagram



10. ADC time taken

1. Acquisition

Analog signal sampled and hold SHA: Sample and Hold Amplifier Enough time must be given, variable

Set by SAMC (AD1CON3<12:8>)

2. Conversion

Sampled analog signal converted to digital

Requires $12T_{AD}$ time to finish

- One T_{AD} per bit + 2 additional T_{AD}
- T_{AD} is configurable, > 65 ns

 T_{AD} source can be selected from internal RC source or prescaled PBCLK by configuring ADRC (AD1CON3<15>)

ADC Total Sample Time

Acquisition Time

Conversion Time

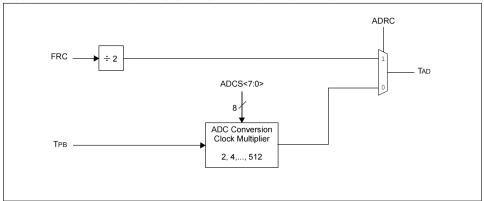
Analog-to-digital conversion complete, result is written into the ADC result buffer. Optionally generate interrupt.

SHA is disconnected from input and holds the signal.

Analog-to-digital conversion is started by the conversion trigger source.

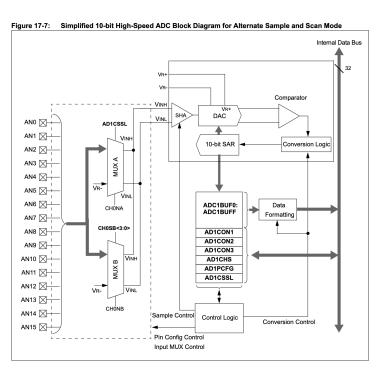
SHA is connected to the analog input pin for sampling.

FIGURE 22-2: ADC CONVERSION CLOCK PERIOD BLOCK DIAGRAM



11. ADC Modes

- Manual
 - Manually start acquisition
 - Manually start conversion
- Automatic
 - Continuous conversion sequence
 - Multiple sample and conversions
- Scan mode
 - Scan through multiple inputs
 - Inputs selectable
 - MUX A only
- Alternate scan mode
 - MUX A and MUX B
 - Input selectable



12. ADC Interrupt

Configured by SMPI<2:0> (AD1CON2<5:2>). Can generate interrupt after 1-16 samples.

The next sequence starts filling the buffer from the top even if the number of samples in the previous sequence was less than 16.

13. ADC Setup

- 1. Configure the analog port pins in AD1PCFG<15:0>.
- 2. Select the analog inputs to the ADC multiplexers in AD1CHS<32:0>.
- 3. Select the format of the ADC result using FORM<2:0> (AD1CON1<10:8>).
- 4. Select the sample clock source using SSRC<2:0> (AD1CON1<7:5>).
- 5. Select the voltage reference source using VCFG<2:0> (AD1CON2<15:13>).
- 6. Select the Scan mode using CSCNA (AD1CON2<10>).
- 7. Set the number of conversions per interrupt SMP<3:0> (AD1CON2<5:2>), if interrupts are to be used.
- 8. Set Buffer Fill mode using BUFM (AD1CON2<1>).
- 9. Select the MUX to be connected to the ADC in ALTS AD1CON2<0>.
- 10. Select the ADC clock source using ADRC (AD1CON3<15>).
- 11. Select the sample time using SAMC<4:0> (AD1CON3<12:8>), if auto-convert is to be used.
- 12. Select the ADC clock prescaler using ADCS<7:0> (AD1CON3<7:0>).
- 13. Turn the ADC module on using AD1CON1<15>.
- 14. To configure ADC interrupt (if required):
 - a) Clear the AD1IF bit (IFS1<1>).
 - b) Select ADC interrupt priority AD1IP<2:0> (IPC<28:26>) and subpriority AD1IS<1:0> (IPC<24:24>) if interrupts are to be used.
- 15. Start the conversion sequence by initiating sampling.
 - Manual mode: setting 1 to SAMP bit (AD1CON1<1>). Software must manually manage the start and end of the acquisition period by setting and then clearing the SAMP bit after the desired acquisition period has elapsed.
 - Auto-sample mode: settings 1 to ASAM bit (AD1CON1<2>)
 Acquisition is automatically started after a conversion is completed. Auto-Sample mode can be used with any trigger source other than manual.

Note: Steps 1 through 12, above, can be performed in any order, but Step 13 must be the final step in every case.

14. DONE bit

The DONE bit (AD1CON1<0>) is set when a conversion sequence is complete.

• In Manual mode:

The DONE bit is persistent.

Remains set until it is cleared by software.

Can be polled to determine when the conversion has completed.

• In all automatic sample modes (ASAM bit = 1):

the DONE bit is not persistent.

Set at the end of a conversion sequence and cleared by hardware when the next acquisition is started.

Polling the DONE bit is not recommended when operating the ADC in automatic modes.

The AD1IF flag bit (IFS1<1>) is latched after a conversion sequence is completed and can therefore be polled.