

# Electronics Laboratory

Winter semester 2025

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## Lab 3 – MOSFETs

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Score and comments (only for tutors, please leave blank)

*Please fill out this cover sheet and submit it with your lab report.*



## 3 Lab 3 – MOSFETs

### Introduction

We will simulate and measure three MOSFET circuits: a common-source circuit, with which we can generate the typical input and output characteristics of a MOSFET; a simple inverter, for which the rise and fall times are of interest; and finally a dc step-up converter in which the MOSFET acts as a switch to allow generation of high dc voltage levels from low-voltage inputs.

### Notes

- Download the SPICE models `FDD86102.txt` and `10BQ040.txt` for the power MOSFET and the Schottky diode, respectively, used in the simulation of Section 3.4.1.
- The IRFML8244 used in Experiments 3.2 and 3.3 is a power MOSFET usually employed to switch over 5 A of current. We are using it here at atypically low current levels. To see its typical characteristics, model it in SPICE using the `.dc VDD 0 5 50m Vi 1 4 0.5` directive and  $R_D = 0$ .
- As you should recall from *Einführung in die Elektrotechnik*, rise and fall times are calculated using the 10% – 90% values. Thus the rise time for a signal which increases from 0 to 5 V is determined from the time required for it to rise from 0.5 to 4.5 V (10% of 5.0 V to 90% of 5.0 V); conversely, the fall time is then the time required for the signal to fall from 4.5 to 0.5 V.
- The propagation delay between two signals  $t_{pd}$ , for example between an input voltage  $v_i(t)$  and an output voltage  $v_o(t)$ , is defined using the 50% – 50% values: the time between when the input signal reaches 50% of its final value and when the output signal reaches 50% of its final value.



## 3.1 Preparation

Please answer the following questions *before* beginning with the simulations or experiments. Then complete the preparation quiz “Preparation 3.1 – MOSFETs” on Ilias using your results. The points for this preparation section will only be awarded via Ilias.

### 3.1.1 MOSFET parameters

You are given a n-channel FET with  $V_t = 1\text{ V}$ ,  $\mu_n = 600\text{ cm}^2/\text{Vs}$ ,  $t_{ox} = 30\text{ nm}$  and  $\epsilon_{ox} = 35\text{ pF/m}$ . You know that the channel width is  $15\text{ }\mu\text{m}$  and its length  $1\text{ }\mu\text{m}$ . Determine  $K_n$ . Pay attention to units.

### 3.1.2 Input characteristic

For the transistor given in Question 3.1.1, plot the  $I_D$  vs.  $V_{GS}$  characteristic for values for which the MOSFET is in saturation. Determine  $I_D$  for  $V_{GS} = 3\text{ V}$ .

### 3.1.3 Output current

Consider the circuit of Figure 3.2.1. Assuming that  $V_{DD} = 5\text{ V}$  and  $R_D = 200\text{ }\Omega$ , what is the maximum value for  $I_D$  (assuming that  $R_{DS(on)} \rightarrow 0$ )? Write down an equation for the output load line of this circuit. Using this equation, determine the values for  $V_{DS}$  when  $I_D = 3.3\text{ mA}$ ,  $6.6\text{ mA}$  and  $10.0\text{ mA}$ .

### 3.1.4 Transfer characteristic

Think about which operating regions a MOSFET inverter with a resistive load (such as that shown in Figure 3.3.1) goes through when then output switches from high to low ( $1 \rightarrow 0$ ).

### 3.1.5 Amplifying inverter

Consider using the circuit of Figure 3.3.1 as an amplifier; take  $R_D = 1\text{ k}\Omega$ . Given that the FET is made using the same technology as that of Question 3.1.1 except now  $W/L = 10/1$ , think about what conditions are necessary so that the small-signal model would apply and calculate the small-signal amplification around an operating point of  $I_{DS} = 2\text{ mA}$ .



## 3.2 MOSFET characteristics

In this first experiment, we will simulate and experimentally characterize the  $I_D/V_{GS}$  and  $I_D/V_{DS}$  characteristics of a MOSFET.

### 3.2.1 Simulation

Consider the common source MOSFET circuit shown in Figure 3.2.1 below.

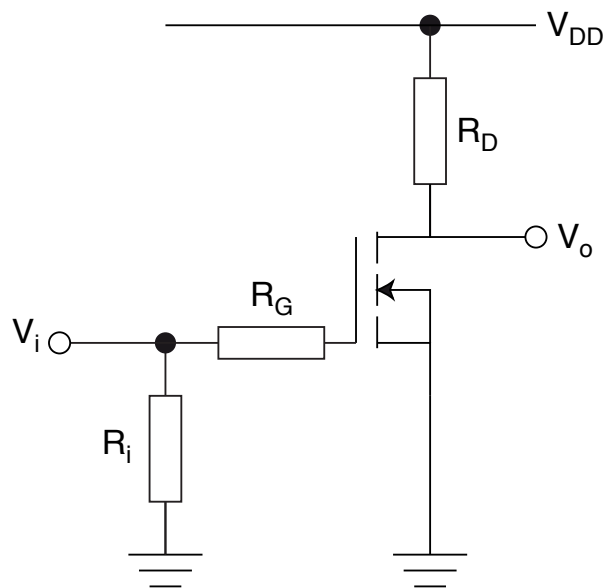


Figure 3.2.1

#### $I_D/V_{GS}$ characteristic

- Set up a SPICE simulation of this circuit with an IRFML8244 MOSFET and using the values  $R_i = 100\text{ k}\Omega$ ,  $R_G = 1\text{ k}\Omega$ ,  $R_D = 200\text{ }\Omega$  and  $V_{DD} = 5\text{ V}$ .
- Apply a voltage ramp at  $V_{in}$  of 1.8 V to 2.4 V in 1 mV steps.
- Plot  $I_D$  and  $V_o$  as a function of  $V_{GS}$ . Explain the form of the characteristic.
- Determine  $I_{D(max)}$  and compare this value with the calculated value from the preparation (Question 3.1.3).
- Determine the  $V_o$  values for which  $I_D = 3.3\text{ mA}$ , 6.6 mA and 10.0 mA. Compare with the calculated values from the preparation (Question 3.1.3).
- Determine and note down the  $V_i$  values for which  $I_D = 3.3\text{ mA}$ , 6.6 mA and 10.0 mA.



### $I_D/V_{DS}$ characteristic

- (g) Using the same circuit, remove the voltage ramp at  $V_{in}$  and replace it with a parameter sweep for  $V_i$  (using the `.step param` directive) with a `list` consisting of the three  $V_i$  values you determined in Step (f) above.
- (h) Define a voltage sweep  $0 \leq V_{DD} \leq 5\text{ V}$  in 50 mV steps, using the `.dc` directive.
- (i) Plot  $I_D$  as a function of  $V_{DS}$  in the range  $0 \leq V_{DS} \leq 5\text{ V}$ .
- (j) In this plot, fit a linear characteristic to the three points where  $V_{DD} = 5\text{ V}$  (i.e., the ends of the  $I_{DS}/V_{DS}$  curves). Extend to the  $V_{DS}$  and  $I_D$  axes. From the slope, determine  $R_D$  (assuming that  $R_{DS(on)} \rightarrow 0$ ).
- (k) Explain how and possibly why the simulated  $I_D/V_{DS}$  characteristic looks a bit different to the ones we considered in the lecture (Hint: look at the “Notes” at the beginning of this lab again).

### 3.2.2 Measurement

In the “MOSFET” section of the electronics board, we will use the *MOSFET characteristics* circuit, shown in Figure 3.2.2.

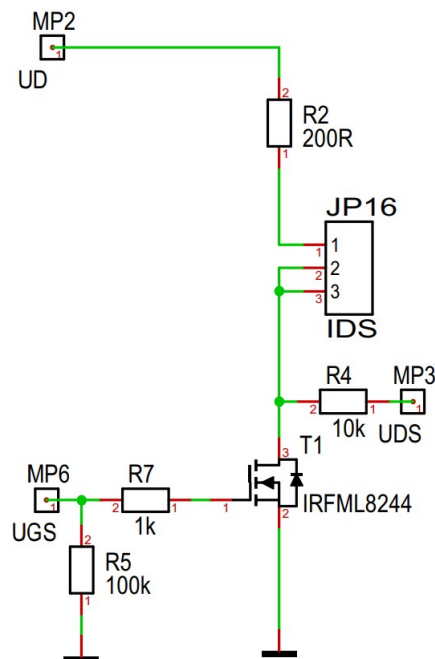


Figure 3.2.2: Schematic of the *MOSFET characteristics* circuit

### $I_D/V_{GS}$ characteristic

- (a) Determine the actual value of  $R_2$  using the handheld multimeter.

- (b) Short-circuit *pin 1* and *pin 2* of jumper JP16.
- (c) Connect measurement points *MP2* and *MP3* to the differential amplifier<sup>1</sup> and connect it to *Channel 2* (*MP75*) to measure the voltage across  $R_2$ .
- (d) Connect the measurement point *MP6* to *Channel 1* (*MP72*) to measure  $V_{GS}$ .
- (e) Connect the DC input (*MP52* or *MP32*) with  $V_{cc} = 5\text{ V}$  to the measurement point *MP2*.
- (f) Connect the first wave generator *W1* (*MP69*) to the measurement point *MP6* ( $V_{GS}$ ).
- (g) Create new channels (using *Add Channel* in *WaveForms - Scope*): 1) with the name  $R$ , the unit  $\Omega$  and the value you measured for  $R_2$  (decimal point); 2) with the name  $V_g$ , the unit V and the command  $C1$ ; 3) with the name  $V_r$ , the unit V and the command  $C2$ ; and 4) with the name  $I_d$ , the unit A and the command  $V_r/R$ .
- (h) Set a triangular voltage (Triangle, 1 V – 4 V, 50 %, 50 Hz) as  $V_{GS}$ .
- (i) Measure and plot  $I_D$  as a function of  $V_{GS}$ .
- (j) Determine the voltages  $V_{GS,3.3\text{ mA}}$ ,  $V_{GS,6.6\text{ mA}}$  and  $V_{GS,10\text{ mA}}$  where the current reaches  $I_{D,1} = 3.3\text{ mA}$ ,  $I_{D,2} = 6.6\text{ mA}$  and  $I_{D,3} = 10.0\text{ mA}$  respectively.

#### $I_D/V_{DS}$ characteristic

- (k) Measure and note the values to be set for the voltage generator *W1* (*WaveGen - DC*, *MP69*) so that the voltages  $V_{GS,3.3\text{ mA}}$ ,  $V_{GS,6.6\text{ mA}}$  and  $V_{GS,10.0\text{ mA}}$  (from the previous task) are actually applied to the board.
- (l) Short-circuit *pin 1* and *pin 2* of jumper JP16.
- (m) Connect measurement points *MP2* and *MP3* to the differential amplifier and connect it to *Channel 2* (*MP75*) to measure the voltage across  $R_2$ .
- (n) Connect the measurement point *MP3* to *Channel 1* (*MP72*) to measure  $V_{DS}$ .
- (o) Connect the DC input (*MP52* or *MP32*) with  $V_{cc} = 5\text{ V}$  to the measurement point *MP2* and connect the first wave generator *W1* (*MP69*) to the measurement point *MP6* ( $V_{GS}$ ).
- (p) Create new channels (using *Add Channel* in *WaveForms - Scope*): 1) with the name  $R$ , the unit  $\Omega$  and the value you measured for  $R_2$  (decimal point); 2) with the name  $V_{ds}$ , the unit V and the command  $C1$ ; 3) with the name  $V_r$ , the unit V and the command  $C2$ ; and 4) with the name  $I_d$ , the unit A and the command  $V_r/R$ .
- (q) Set  $V_{cc} = 5\text{ V}$  and set  $V_{GS,10.0\text{ mA}}$  such that  $I_D \cong 10.0\text{ mA}$ . Start your search at the

<sup>1</sup>Recall that the input of the differential amp is between *MP55* & *MP65* and its output is between *MP66* and ground.



value you measured earlier for  $W1$ . Note the final value and justify any deviations.

- (r) Keeping  $V_{GS}$  constant, vary  $V_{cc}$  (0, 5, 0.5) Record  $V_{DS}$  and  $I_D$  for each measurement point.
- (s) Repeat the last three steps for  $V_{GS,3.3\text{mA}}(I_D = 3.3\text{mA})$  and  $V_{GS,6.6\text{mA}}(I_D = 6.6\text{mA})$ .
- (t) Plot the characteristic curves (for each  $V_{GS}$ ) in a single diagram.
- (u) Determine the Early voltage of each trace.
- (v) Fit a linear trend line to the three measurement points for  $V_{cc} = 5\text{V}$ . This fit represents the load line for  $R_2$  and the voltage  $V_{cc}$ . Calculate from this data the value for  $R_2$  assuming that  $R_{DS(on)} \rightarrow 0$ .



### 3.3 MOS logic gates

We will analyze the time-dependent turn-on and turn-off behavior of a simple MOSFET inverter by simulation and measurement.

#### 3.3.1 Simulation

Consider the simple MOSFET inverter shown in Figure 3.3.1 below.

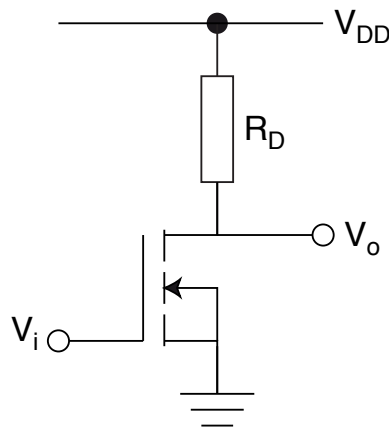


Figure 3.3.1

- Set up a SPICE simulation of this circuit using  $R_D = 200\ \Omega$  and an IRFML8244 MOSFET.
- Define the voltage source  $V_i$  as an asymmetric square wave signal from 0 V to 5 V at 50 Hz using the source definition `pulse (0 5 10m 1u 1m 10m 20m)`.
- Set the voltage source  $V_{DD} = 5\text{ V}$ .
- Place a terminal for the output  $V_o$  at the drain.
- Perform a time-based simulation (using the `.tran` directive), and analyze  $V_i$  and  $V_o$ . Fiddle with the parameters of the `.tran` directive and perform the simulation numerous times to look at the low  $\rightarrow$  high and high  $\rightarrow$  low transitions in detail.
- Determine the maximum and minimum voltages of both signals at their steady state.
- Using an appropriate magnification on the time axis, make a plot of the rising edge transitions of the input and output voltages, together in a single plot. Make a second plot of the falling edge transitions. Explain what you see.
- Determine the rise and fall times for  $V_i$  and  $V_o$ . Use the appropriate definitions.
- Try to determine the propagation delays ( $t_{pd(rise)}$  and  $t_{pd(fall)}$ ) for the rising and





falling edges of  $V_i$ . Why might this determination be difficult?

### 3.3.2 Measurement

In the “MOSFET” section of the electronics board, we will continue to use the *MOSFET characteristics* circuit, shown in Figure 3.2.2 above.

- (a) Short-circuit *pin 1* and *pin 2* of jumper JP16.
- (b) Transfer the potential from the measurement point *MP32* to the measurement point *MP2* and measure the resulting supply voltage there.
- (c) Connect a square wave signal (0 V - 5 V, 50 Hz, 50 %) to *MP6*.
- (d) Measure the voltage  $V_i$  at *MP6* and  $V_o$  at *MP3*.
- (e) Determine the maximum and minimum voltages of both signals in their steady states.<sup>2</sup>
- (f) Using an appropriate magnification on the time axis, make a plot of the rising edge transitions of  $V_i$  and  $V_o$ , together in a single plot. Make a second plot of the falling edge transitions. Explain what you see.
- (g) Determine the propagation delays ( $t_{pd(rise)}$  and  $t_{pd(fall)}$ ) for the rising and falling edges of  $V_i$ .
- (h) Compare your measurement results with the simulation results of Section 3.3.1. Discuss and explain any significant differences.

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<sup>2</sup>To obtain enough samples, you can use the following measurement settings: 1) use repeated measurement mode; 2) for the rising edge transition use Source: Channel 1; Condition: Rising; Time axis scale: 1  $\mu$ s/div, position: 4  $\mu$ s; Level: 0.5 V (adjust if necessary); 3) for the falling edge transition use Source: Channel 1; Condition: Falling; Time axis scale: 1  $\mu$ s/div, position: 4  $\mu$ s; Level: 4 V (adjust if necessary).



### 3.4 Step-up converter

A step-up (or boost) converter is a circuit which generates a higher dc voltage from a lower dc voltage. The transformer, which you know from *Einführung in die Elektrotechnik*, only increases the voltage of **ac** signals, but the step-up converter is a useful circuit for generating higher **dc** voltages.

A simple step-up voltage converter circuit is shown in Figure 3.4.1. The MOSFET is essentially a switch which alternately generates a short or an open circuit. When the MOSFET is on (a short circuit), a current from the input flows through the coil and the transistor, establishing a magnetic field in the inductor. When the MOSFET switches off (an open circuit), the stored energy in the coil results in a current flow through the Schottky diode, charging the capacitor. When the MOSFET switches on again, the cycle repeats. The increasing charge on the capacitor generates the higher output voltage  $V_{dc-out}$ .

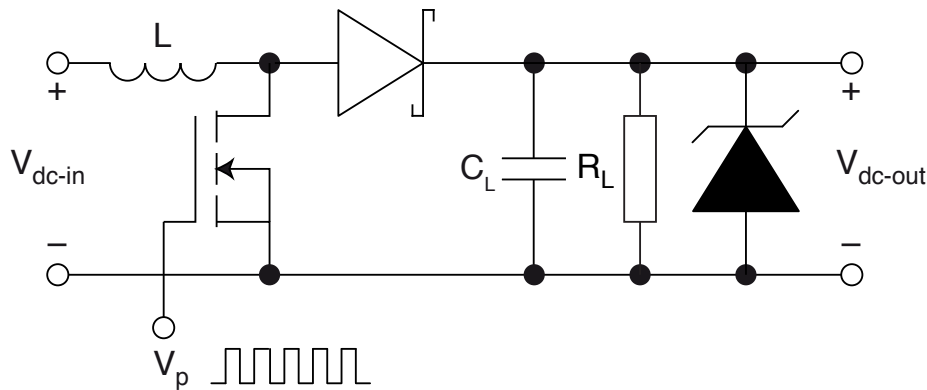


Figure 3.4.1: A basic step-up converter

As an interesting application for MOSFETs, we will now simulate and measure an MOS-based step-up converter.

#### 3.4.1 Simulation

- Set up a SPICE simulation of the step-up converter circuit of Figure 3.4.1, which corresponds to the circuit on the electronics board (Figure 3.4.2). Use the FDD86102 power MOSFET (download the model file). Take  $L = 100 \mu\text{H}$ ,  $C_L = 22 \mu\text{F}$  and  $R_L = 5 \text{k}\Omega$  in parallel to the capacitor. For the Schottky diode, use a 10BQ040 (download the model file) and for the Zener diode a BZX84B36VL (with  $V_Z = 36 \text{ V}$ , in the LTSpice library).
- Apply a dc bias of  $5 \text{ V}$  to  $V_{dc-in}$ .
- For  $V_p$ , use the **pulse** source definition and apply a  $10 \text{ kHz}$  square wave with a duty cycle of  $50\%$  and a magnitude of  $5 \text{ V}$ .



- (d) Using the `.tran` directive, plot  $V_{dc-out}$  for 320 ms (the simulation may take a few minutes). Describe the form of the characteristic and determine the output voltage increase factor.
- (e) Vary the duty cycle (5%, 20%, 60%) and note what changes for  $V_{dc-out}$ . Adapt the simulation conditions if necessary.
- (f) Finally, apply square waves for the frequencies 2 kHz, 5 kHz, 20 kHz, 50 kHz, and 100 kHz, all with a duty cycle of 50 %, and describe the resulting output.

### 3.4.2 Measurement

In the “MOSFET” section of the electronics board, we will use the *step-up converter* circuit, shown in Figure 3.4.2.

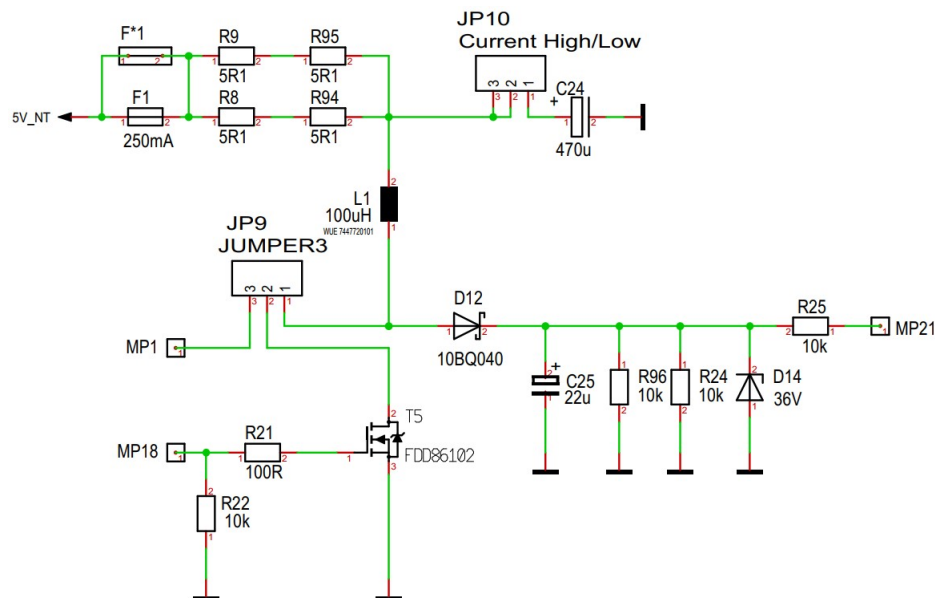


Figure 3.4.2: Schematic of the *step-up converter* circuit

- (a) Short-circuit *pin 1* and *pin 2* of the jumper *JP9/JUMPER3* (this is labeled *BOOST* on the board). Open *pin 1* and *pin 2* of the jumper *JP10*.
- (b) Connect the input signal of the signal generator to the measurement point *MP18*.
- (c) Attach a PWM signal (Square, 0 – 5 V, 50 %) to *MP18* and describe the output signal you measure at *MP21*.
- (d) Measure the average value for  $V_{out}$  for each of the frequencies: 100 Hz, 200 Hz, 500 Hz, 1 kHz, 2 kHz, 5 kHz, 10 kHz, 20 kHz, 50 kHz.
- (e) Plot  $V_{out}$  as a function of  $f$  in a graph with a logarithmic abscissa.

- (f) Set up another PWM signal (Square, 0 – 5 V, 10 kHz).
- (g) Measure the average value for  $V_{out}$  in each case for the duty cycle  $D$  (0, 100, 10%)<sup>3</sup>.
- (h) Plot  $V_{out}$  as a function of  $D$ .

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<sup>3</sup>You can set the duty cycle by changing the symmetry of a square wave in the *WaveForms* software.

