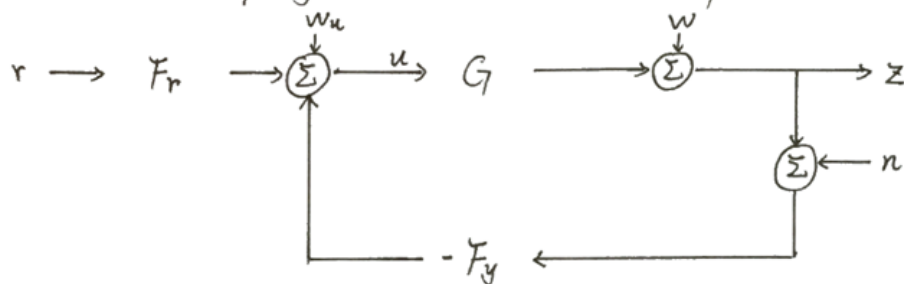


Lecture 02: The closed-loop system

- Goals:
- Closed-loop is characterized by 7 transfer functions.
 - several objectives, but trade-offs exist, e.g. $S+T=1$.
 - internal stability requires all 7 to be stable.
 - Determine, analyze, and design desired sensitivity functions

1. The closed-loop system (SISO this time)



Controller: feedback F_y feedforward F_r

Disturbances: w, w_u

Measurement noise: n .

aim: find z follow r , despite w, w_u, n .

(and model uncertainty Δ)

• For Output:

$$z = w + G(w_u + F_r(r) - F_y(z + n))$$

$$\Rightarrow z = \underbrace{\frac{1}{1 + G F_y}}_S \cdot w + \underbrace{\frac{G}{1 + G F_y}}_{SG} w_u + \underbrace{\frac{G F_r}{1 + G F_y}}_{G_c} \cdot r - \underbrace{\frac{G F_y}{1 + G F_y}}_T \cdot n$$

S : The sensitivity function.

$$\Rightarrow S + T = 1 \quad (\text{trade-off})$$

T : The complementary sensitivity function.

G_c : The closed-loop transfer function.

Thus, make $|S|$ small for disturbance.

make $|T|$ small for noise.

make $|T - G_c|$ small for set points.

• For Input.

$$u = w_u + F_r r - F_y (n + w + G u)$$

$$\Rightarrow u = \underbrace{\frac{1}{1 + G F_y}}_S w_u + \underbrace{\frac{F_r}{1 + G F_y}}_{S \cdot F_r} \cdot r - \underbrace{\frac{F_y}{1 + G F_y}}_{S \cdot F_y} (n + w)$$

make $|S|$, $|S F_r|$, $|S F_y|$ small.

• Nominal stability = internal stability.

closed loop is internally stable if input-output stable for all γ transfer functions.

S , $S G$, G_c , T , $S F_r$, $S F_y$, F_r are stable. (test poles)

Note: - if S stable, then T stable.

- if S and F_r stable, $S F_r$ stable.

- if $S G$, F_r stable, G_c stable.

\Uparrow
check S , $S G$, F_r , $S F_y$ is enough.

$$\text{EX: } G = \frac{1}{s-1}, \text{ unstable, } F_r = \frac{s-1}{s}$$

$$\Rightarrow S = \frac{s}{s+1}, \text{ stable, } S G = \frac{s}{(s+1)(s-1)}, \text{ unstable.}$$

* Never cancel poles and zeros!

• The sensitivity S : (why to make S small).

1) Effect of disturbance w of on z :

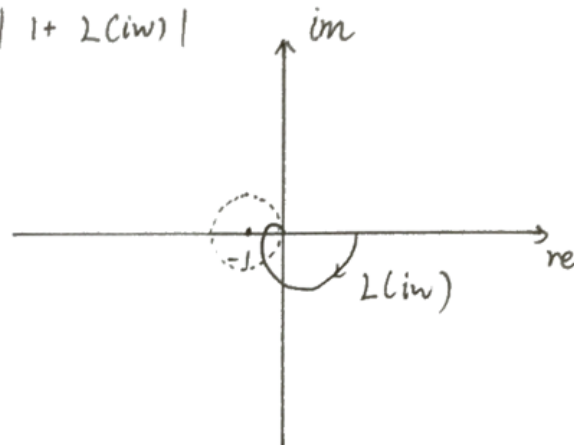
- if without feedback $\Delta z = 1 \cdot w$.

- if with feedback : $\Delta z = S \cdot w$

\Rightarrow disturbance attenuation for frequencies where $|S(i\omega)| < 1$,
amplification if $|S(i\omega)| > 1$.

Note: must have $|S(i\omega)| > 1$ for some ω , if ~~$G(s)$ has~~
 $L(s) = G(s)F_y(s)$
pole excess of at least 2.

pseudo-proof: $|1 + L(i\omega)|$



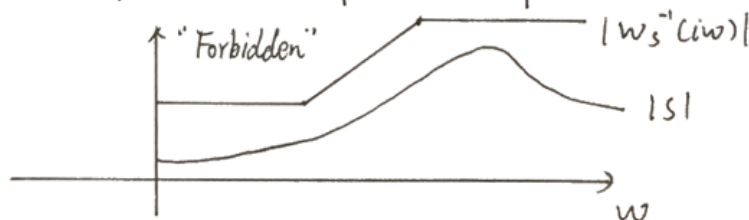
$$|1 + L(i\omega)| \leq 1 \Rightarrow |S(i\omega)| = \frac{1}{|1 + L(i\omega)|} > 1$$

- shaping the sensitivity S .

* can not make $|S(i\omega)|$ small at all ω .

* reasonable design specification.

$$|S(i\omega)| \leq |W_S^{-1}(i\omega)| \quad \forall \omega \Leftrightarrow |W_S \cdot S| < 1 \quad \forall \omega$$



$$\Leftrightarrow \|W_S \cdot S\|_\infty < 1$$

2) Effect of uncertainty on closed-loop G_c .

$$G_c = \frac{G F_r}{1 + G F_y} \Rightarrow \frac{dG_c}{dG} = \frac{\cancel{F_r}}{1 + \cancel{F_y}} \cdot \frac{F_r(1 + G \cancel{F_y}) - \cancel{F_y} G F_r}{(1 + G F_y)^2}$$

$$= \frac{F_r}{(1 + G F_y)^2} = S \cdot \frac{G_c}{G} \Rightarrow \frac{d G_c / G_c}{d G / G} = S$$

Thus, relative unc in model G is attenuation in G_c by a factor S .

• The complementary sensitivity: T

1) Noise: $\Delta z = T \cdot n \Rightarrow$ make $|T|$ small where n large.

2) Robust stability: assume true system is

$$\tilde{G} = G(1 + \Delta G)$$

how large can ΔG be

employ small gain theorem (SGT)



identify M : $u_\Delta = -G F_y (u_\Delta + y_\Delta)$

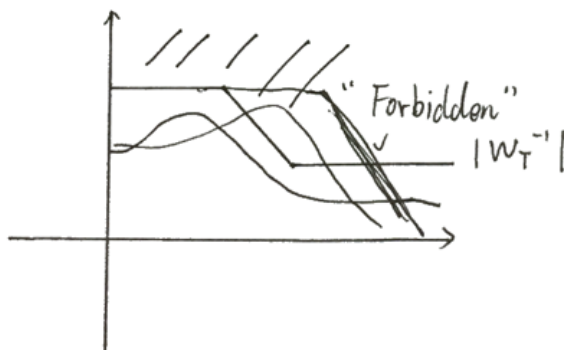
$$\Rightarrow M = \frac{G F_y}{1 + G F_y} = T$$

stable if M is stable,
 ΔG stable, and

$$\|M \cdot \Delta G\|_\infty < 1 \Rightarrow \|T \cdot \Delta G\|_\infty < 1$$

$$\Leftrightarrow |T(i\omega) \cdot \Delta G(i\omega)| < 1 \Leftrightarrow |T| < \frac{1}{|\Delta G|} \quad \forall \omega$$

Thus, make $|T|$ small where n and ΔG large.



recall: $S + T = 1$, can not make $|W_s|$ and $|W_T|$ large at same time.