

STANFORD UNIVERSITY

EE 102B Spring-2013

Lecture 18

The DFT as a Sampled DTFT and Block Processing

May 13, 2013

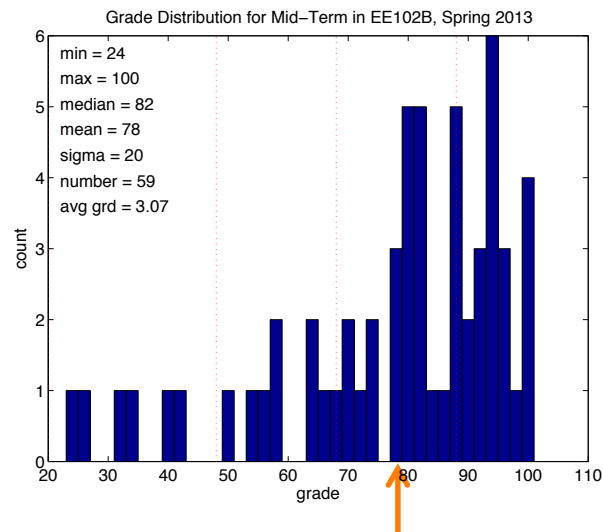
ASSIGNMENTS

- Reading for this Lecture:
 - SPF: Section 12-3, Chapter 66-6 thru 66-9
 - S&S:
 - HW#06 is due by 5pm Wednesday, May 15, in Packard 263. No late penalty if handed in by 5pm Friday, May 17.
 - Lab #05 is due by 5pm, Friday, May 17, in Packard 263.

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Office Hours for Course Staff – Come see us.

- Ron Schafer: Mon. and Weds. 2:00-4:00 pm, Packard 211. (Available today only 4:00-5:00 pm.)
- Dookun Park: Tues. 1:30 ~ 3:30 pm in Packard 107
- Ruo Yu Gu (Roy): Mon. 4:00 ~ 6:00 pm in Packard 106

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Lecture Objective

- A DFT example
- Sampling theorem for the DTFT
 - The DFT is a sampled DTFT
 - Derivation of result for recovery of $x[n]$
- Block processing with the DFT
 - Segmenting a signal
 - Convolution with a segmented signal

DFT is
sampled
version of
DTFT

Comparison: DFT and DTFT

DFT

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn} \quad k = 0, 1, \dots, N-1$$

Inverse DFT

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn} \quad n = 0, 1, \dots, N-1$$

DTFT

$$X(e^{j\hat{\omega}}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\hat{\omega}n} \quad -\pi \leq \hat{\omega} < \pi$$

Inverse DTFT

$$x[n] = \frac{1}{2\pi} \int_0^{2\pi} X(e^{j\hat{\omega}}) e^{j\hat{\omega}n} d\hat{\omega} \quad -\infty < n < \infty$$

A DFT EXAMPLE

A DFT Example

- The DFT of an impulse is a constant

$$P[k] = \sum_{n=0}^{N-1} \delta[n] e^{-j(2\pi/N)kn} = 1, \quad k = 0, 1, \dots, N-1$$

$$p[n] = \frac{1}{N} \sum_{k=0}^{N-1} e^{j(2\pi/N)kn} = \begin{cases} 1 & n = rN \\ 0 & \text{otherwise} \end{cases}$$

- Showing the implicit periodicity explicitly:

$$p[n] = \sum_{r=-\infty}^{\infty} \delta[n - rN]$$

A SAMPLING THEOREM FOR THE DTFT

Sampling the DTFT

- The DTFT of finite-length $x[n]$

$$X(e^{j\hat{\omega}}) = \sum_{n=0}^{N-1} x[n]e^{-j\hat{\omega}n} \quad 0 \leq \hat{\omega} < 2\pi$$

- The DFT is the sampled DTFT

$$X[k] = \sum_{n=0}^{N-1} x[n]e^{-j(2\pi/N)kn} = X(e^{j\omega_k}) \quad k = 0, 1, \dots, N-1$$

- DFT samples the DTFT at frequencies

$$\omega_k = (2\pi/N)k \quad k = 0, 1, \dots, N-1$$

Sample the DTFT and then reconstruct $x[n]$ by inverse - I

$$X[k] = X(e^{j\hat{\omega}}) \Big|_{\hat{\omega}=(2\pi/N)k} = \sum_{m=0}^{N-1} x[m]e^{-j(2\pi/N)km}$$

Inverse DFT
of sampled
DTFT

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k]e^{j(2\pi/N)kn}$$

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \left(\sum_{m=0}^{N-1} x[m]e^{-j(2\pi/N)km} \right) e^{j(2\pi/N)kn}$$

$$\tilde{x}[n] = \sum_{m=0}^{N-1} x[m] \left(\frac{1}{N} \sum_{k=0}^{N-1} e^{j(2\pi/N)k(n-m)} \right) = \sum_{m=0}^{N-1} x[m]p[n-m]$$

Sample the DTFT and then reconstruct - II

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k]e^{j(2\pi/N)kn} = \sum_{m=0}^{N-1} x[m]p[n-m]$$

$$\tilde{x}[n] = \sum_{m=0}^{N-1} x[m] \left(\sum_{r=-\infty}^{\infty} \delta[n-m-rN] \right)$$

$$\tilde{x}[n] = \sum_{r=-\infty}^{\infty} \left(\sum_{m=0}^{N-1} x[m]\delta[n-m-rN] \right)$$

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k]e^{j(2\pi/N)kn} = \sum_{r=-\infty}^{\infty} x[n-rN]$$

Time-aliased
signal $x[n]$

Summary of DTFT Sampling Theorem

- Sample DTFT to get DFT

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn} = X(e^{j\omega_k}) \quad k = 0, 1, \dots, N-1$$

- Reconstruction by IDFT

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn} = \sum_{r=-\infty}^{\infty} x[n - rN]$$

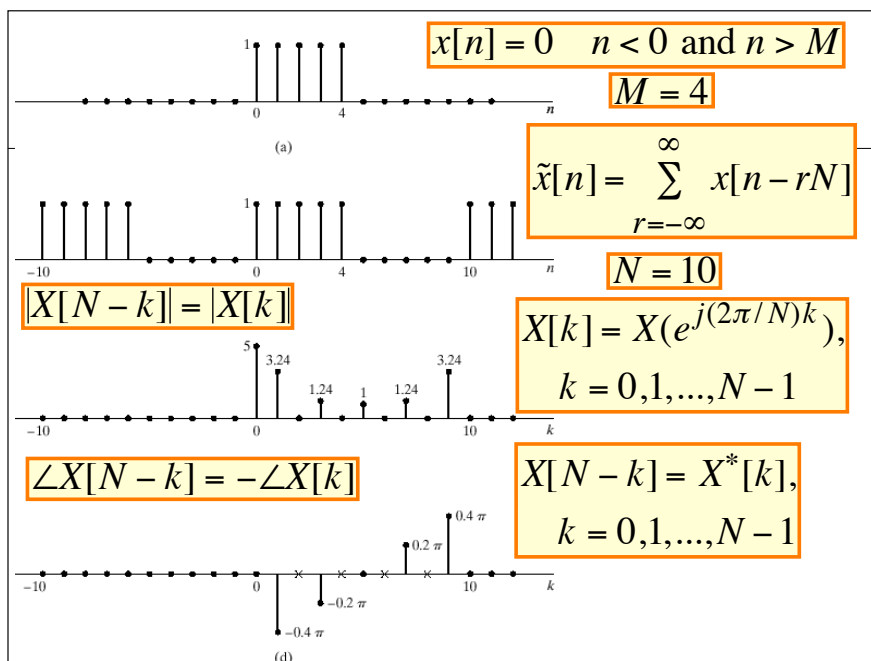
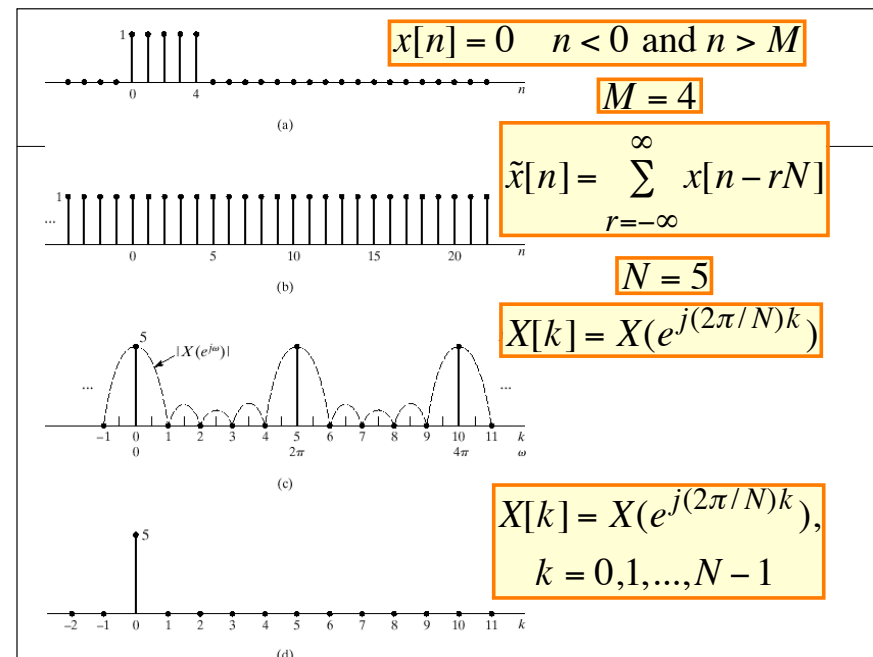
- Exact reconstruction if no overlap; i.e.,

$$\tilde{x}[n] = x[n] \quad 0 \leq n \leq N-1, \text{ if } x[n] \neq 0 \text{ only for } 0 \leq n \leq N-1$$

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USING THE DTFT SAMPLING THEOREM FOR TIME-DOMAIN CONVOLUTION

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Convolution of Finite-Length Sequences

- Let $h[n]$ be of length P and $x[n]$ of length L

$$y[n] = x[n] * h[n] = \sum_{m=0}^{L-1} x[m]h[n-m]$$

- $y[n]$ also has finite length, $L+P-1$ samples

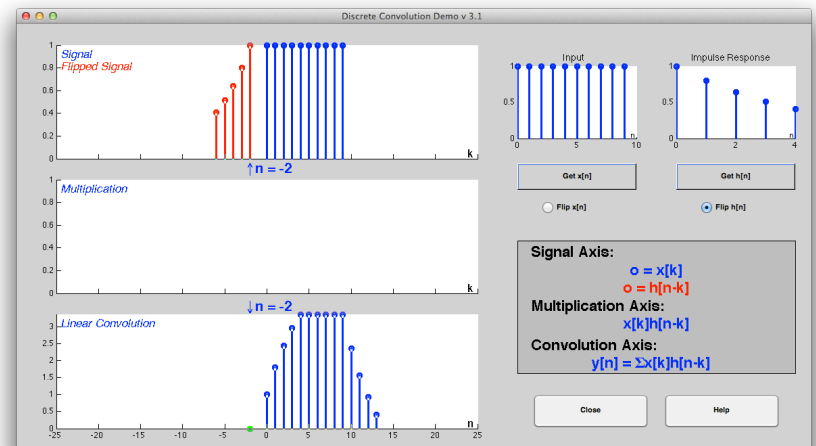
$$y[n] = 0 \quad \text{for } n < 0 \text{ and for } n > L + P - 2$$

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$n < 0$
 $L = 10, P = 5$

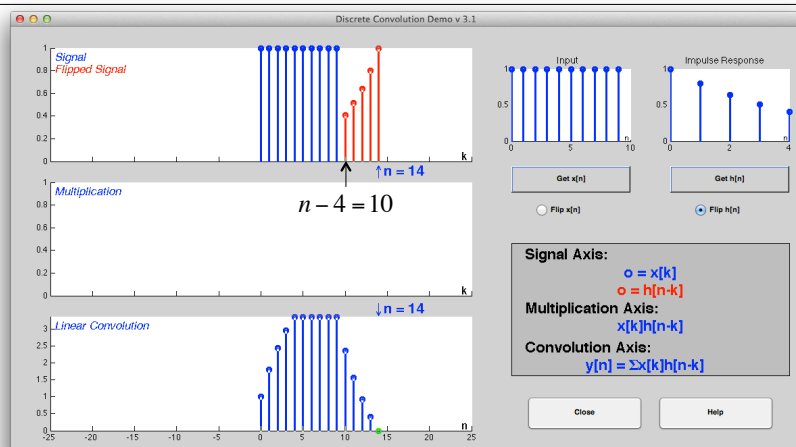


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$n > L + P - 2$
 $L = 10, P = 5$



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Frequency Domain

- DTFT of a convolution

$$y[n] = h[n] * x[n] \Leftrightarrow Y(e^{j\hat{\omega}}) = H(e^{j\hat{\omega}})X(e^{j\hat{\omega}})$$

- Sample the DTFT

$$Y(e^{j(2\pi/N)k}) = H(e^{j(2\pi/N)k})X(e^{j(2\pi/N)k})$$

$$Y[k] = H[k]X[k]$$

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Sampling the DTFT of a Convolution

$$y[n] = h[n] * x[n] \Leftrightarrow Y(e^{j\hat{\omega}}) = H(e^{j\hat{\omega}})X(e^{j\hat{\omega}})$$

$$Y(e^{j(2\pi/N)k}) = H(e^{j(2\pi/N)k})X(e^{j(2\pi/N)k})$$

Compute the sampled DTFT by computing the DFTs

$$X[k] = \sum_{n=0}^{L-1} x[n]e^{-j(2\pi/N)kn} \quad H[k] = \sum_{n=0}^{P-1} h[n]e^{-j(2\pi/N)kn}$$

$$\tilde{y}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \underbrace{H[k]X[k]}_{Y[k]} e^{j(2\pi/N)kn} = \sum_{r=-\infty}^{\infty} y[n-rN]$$

$$\tilde{y}[n] = y[n] \quad 0 \leq n \leq N-1 \quad \text{if } N \geq L+P-1$$

BLOCK PROCESSING OF LONG SIGNALS

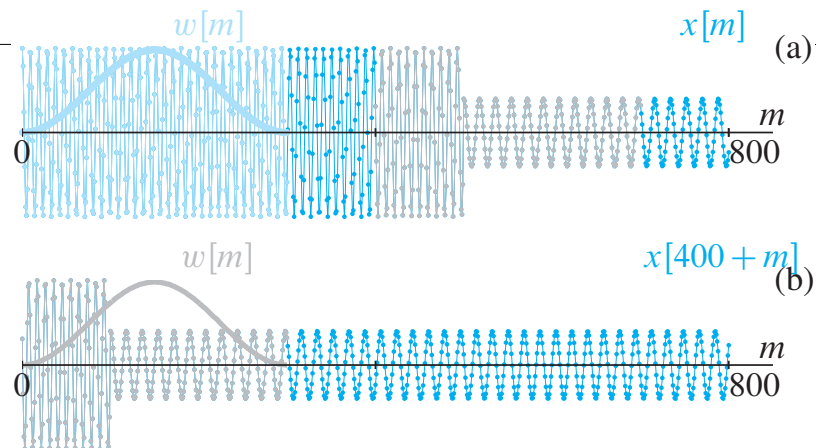
Time-Dependent (Short Time) DTFT and DFT

- Definition: short-time DTFT

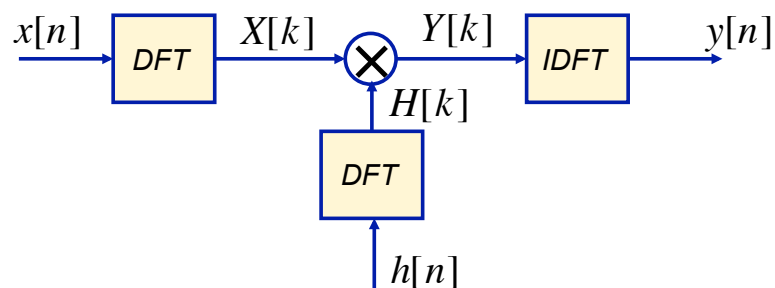
$$X(e^{j\hat{\omega}}, n) = \sum_{m=0}^{L-1} w[m]x[m+n]e^{-j\hat{\omega}n} \quad 0 \leq \hat{\omega} < 2\pi$$

- Definition: short-time DFT

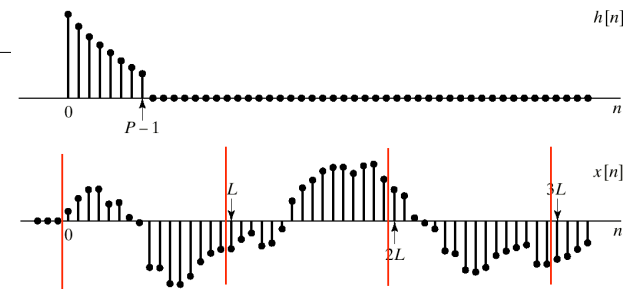
$$X[k, n] = \sum_{m=0}^{L-1} w[m]x[m+n]e^{-j(2\pi/N)kn} \quad k = 0, 1, \dots, N-1$$



Convolution Using the DFT



Block Convolution



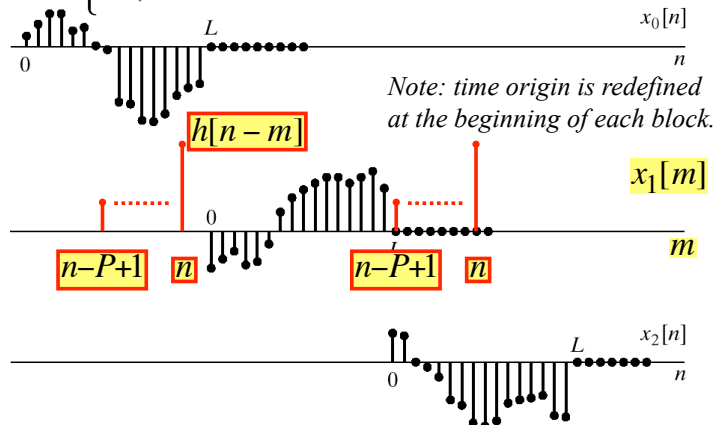
$$x[n] = \sum_{r=0}^{\infty} x_r[n - rL] \rightarrow y[n] = x[n] * h[n] = \sum_{r=0}^{\infty} y_r[n - rL]$$

$$x_r[m] = \begin{cases} x[m + rL], & 0 \leq m \leq L-1 \\ 0, & \text{otherwise} \end{cases} \rightarrow y_r[n] = x_r[n] * h[n]$$

Rectangular window

Segmenting the Input

$$x_r[n] = \begin{cases} x[n + rL], & 0 \leq n \leq L-1 \\ 0, & \text{otherwise} \end{cases} \quad x[n] = \sum_{r=0}^{\infty} x_r[n - rL]$$



Putting the Output Pieces Together

$$y[n] = \sum_{r=0}^{\infty} y_r[n - rL] \quad y_r[n] = x_r[n] * h[n]$$

