

# STANFORD UNIVERSITY

## EE 102B Spring-2013

### Lecture 03

### Periodic Signals, Harmonics & Time-Varying Sinusoids April 5, 2013

## ASSIGNMENTS

- Reading for this Lecture:
  - SPF: Chapter 3 and .pdf of new Section 3.6
  - S&S: Review book sections on continuous-time Fourier series
- HW#1 and Lab #1 are posted
  - Both due on April 10

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## LECTURE OBJECTIVES

- Signals with **HARMONIC** Frequencies
  - Add Sinusoids with  $f_k = kf_0$

$$x(t) = A_0 + \sum_{k=1}^N A_k \cos(2\pi k f_0 t + \varphi_k)$$

**Second Topic:** FREQUENCY can change vs. TIME

Introduce Spectrogram Visualization

(*spectrogram.m*)

(*plotspec.m*)

Chirps:  $x(t) = \cos(\alpha t^2)$

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## Lab #01

$$x(t) = \sum_{k=-N}^N a_k e^{j2\pi f_k t}$$

```
function xx = syn_fourier(tt, ak, fk)
%SYN_FOURIER Function to synthesize a sum of complex
% exponentials over the time range given by tt
% usage:
% xx = syn_fourier( tt, ak, fk )
% tt = vector of times, for the time axis
% ak = vector of complex Fourier coefficients
% fk = vector of frequencies
% (usually contains both negative and positive freqs)
% xx = vector of synthesized waveform values
%
% Note: fk and ak must be the same length.
% ak(1) corresponds to frequency fk(1),
% ak(2) corresponds to frequency fk(2), etc.
%
% Note: the output might have a tiny imaginary part even if it
% is supposed to be purely real. If so, take the real part.
```

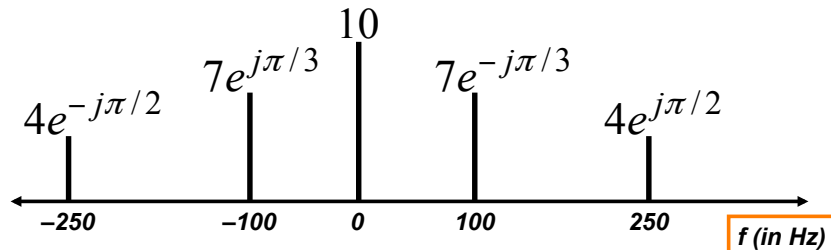
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## SPECTRUM PLOT

$$x(t) = \sum_{k=-N}^N a_k e^{j2\pi f_k t}$$

- Complex Amplitude vs. Frequency



$$x(t) = 10 + 7e^{-j\pi/3} e^{j2\pi(100)t} + 7e^{j\pi/3} e^{-j2\pi(100)t} + 4e^{j\pi/2} e^{j2\pi(250)t} + 4e^{-j\pi/2} e^{-j2\pi(250)t}$$

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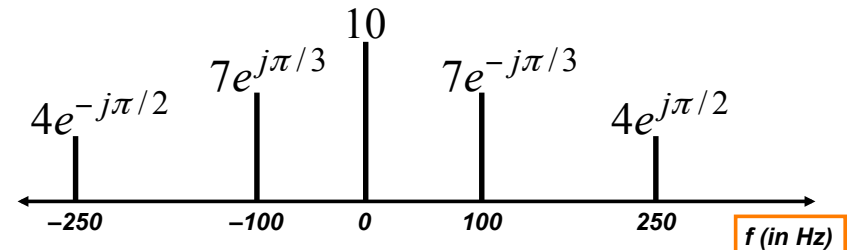
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## SPECTRUM PLOT

$$x(t) = \sum_{k=-N}^N a_k e^{j2\pi f_k t}$$

- Complex Amplitude vs. Frequency



$$x(t) = 10 + 14 \cos(2\pi(100)t - \pi/3) + 8 \cos(2\pi(250)t + \pi/2)$$

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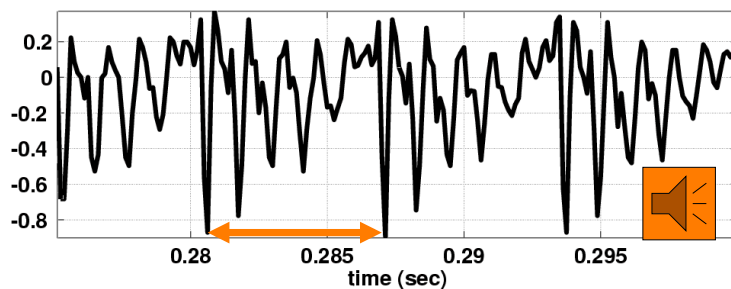
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## Example of a Periodic Signal

- Nearly **Periodic** in the Vowel Region
  - Period is (Approximately)  $T = 0.0065$  sec

Speech: BAT



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## Harmonic Signal

Periodic signal :  $x(t) = x(t + T)$

Can only have **harmonic** freqs :  $f_k = k f_0$

$$x(t) = A_0 + \sum_{k=1}^N A_k \cos(2\pi k f_0 t + \varphi_k)$$

$x(t)$  will be periodic if

$$\cos(2\pi k f_0 (t + T) + \varphi_k) = \cos(2\pi k f_0 t + 2\pi k f_0 T + \varphi_k)$$

$$f_0 T = 1$$

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## Harmonic freqs : $f_k = k f_0$

$$x(t) = A_0 + \sum_{k=1}^N A_k \cos(2\pi k f_0 t + \varphi_k)$$

Phasor:  $X_k = A_k e^{j\varphi_k}$

$$f_0 = \frac{1}{T}$$

$$x(t) = X_0 + \sum_{k=1}^N \left\{ \frac{1}{2} X_k e^{j2\pi k f_0 t} + \frac{1}{2} X_k^* e^{-j2\pi k f_0 t} \right\}$$

$$x(t) = a_0 + \sum_{k=1}^N \left\{ a_k e^{j2\pi k f_0 t} + a_{-k} e^{-j2\pi k f_0 t} \right\}$$

## Define FUNDAMENTAL FREQ

$$x(t) = A_0 + \sum_{k=1}^N A_k \cos(2\pi k f_0 t + \phi_k) = \sum_{k=-N}^N a_k e^{j2\pi k f_0 t}$$

Largest  $f_0$  such that

$$f_k = k f_0 \quad (\omega_0 = 2\pi f_0)$$

$$f_0 = \frac{1}{T_0}$$

$f_0$  = fundamental Frequency

$f_k / f_0$  = integer, for all  $k$

$T_0$  = fundamental Period

**Main point:**

for periodic signals, all spectral components are integer multiples of the fundamental frequency

## General Periodic Signals

■ Synthesis

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{j\omega_0 k t}$$

■ Analysis

$$a_k = \frac{1}{T_0} \int_0^{T_0} x(t) e^{-j\omega_0 k t} dt$$

■ Fourier transform

$$X(j\omega) = \sum_{k=-\infty}^{\infty} 2\pi a_k \delta(\omega - k\omega_0)$$

## General Non-Periodic Signals

■ Synthesis

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{j\omega_k t}$$

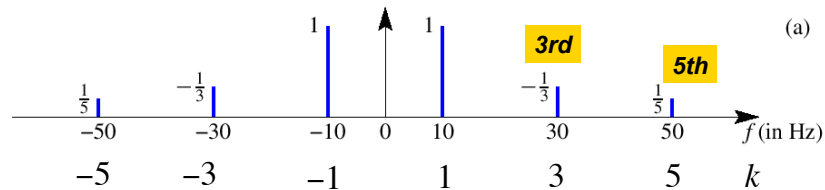
■ Analysis

■ Fourier transform

$$X(j\omega) = \sum_{k=-\infty}^{\infty} 2\pi a_k \delta(\omega - \omega_k)$$

## Harmonic Spectrum (3 Freqs)

$$x(t) = \sum_{k=-N}^N a_k e^{j2\pi k f_0 t}$$

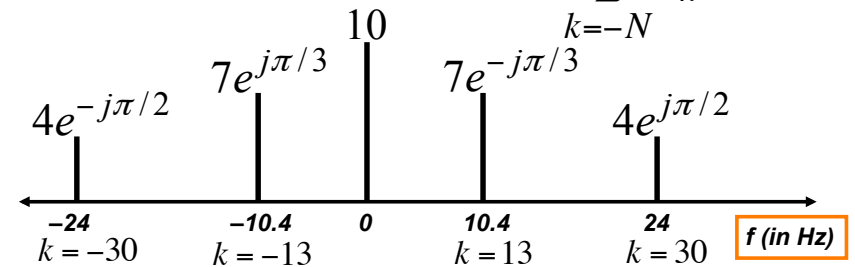


What is the fundamental frequency?

10 Hz

## Another Example

$$x(t) = \sum_{k=-N}^N a_k e^{j2\pi k f_0 t}$$



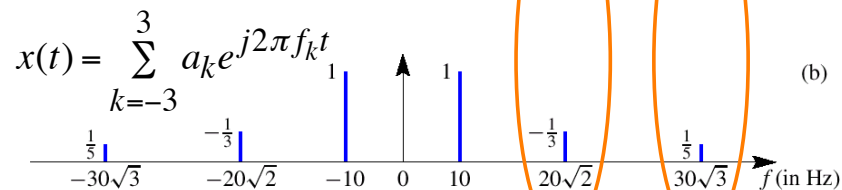
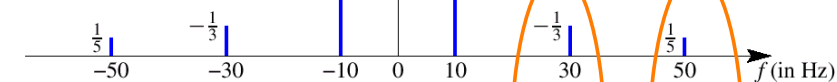
What is the fundamental frequency?

$$(0.1)\text{GCD}(104, 240) = (0.1)(8) = 0.8 \text{ Hz}$$

## Irrational Frequencies

**SPECIAL RELATIONSHIP**  
to get a **PERIODIC SIGNAL**

$$x(t) = \sum_{k=-5}^5 a_k e^{j2\pi k f_0 t}$$

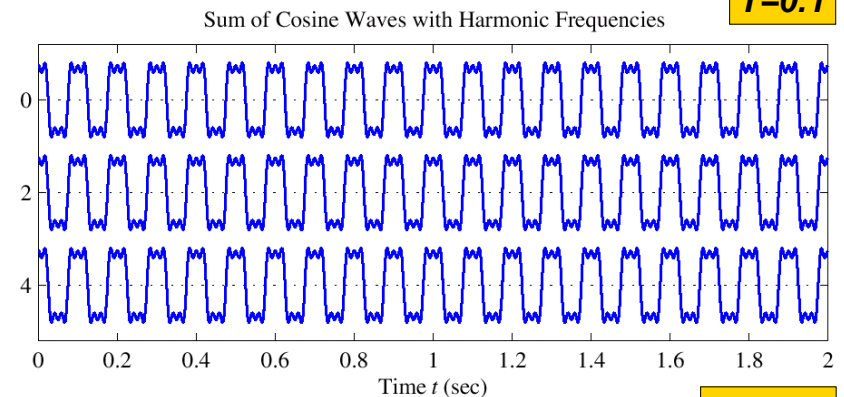


**NON-PERIODIC SIGNAL**

28.28... 51.96...

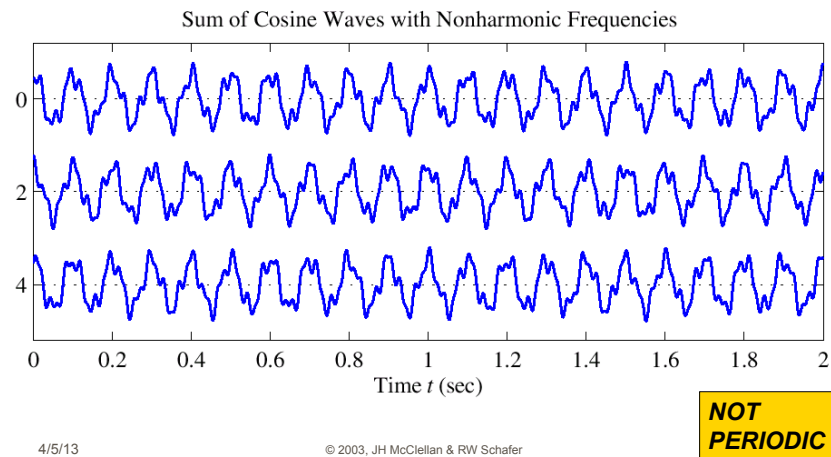
## Harmonic Signal (3 Freqs)

**T=0.1**



**PERIODIC**

## NON-Harmonic Signal



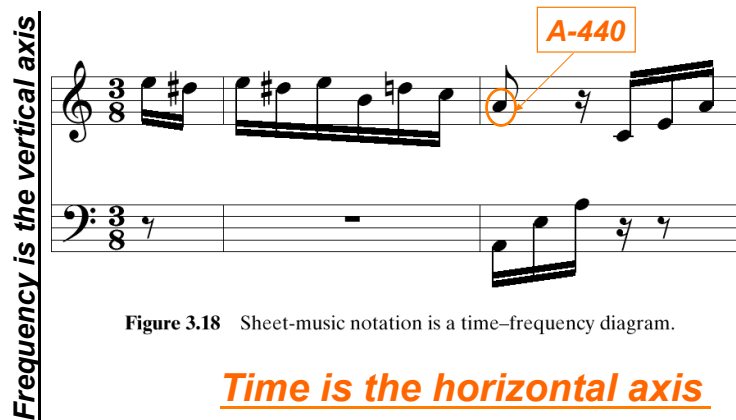
## TIME-VARYING SPECTRUM

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## Time-Varying FREQUENCIES Diagram



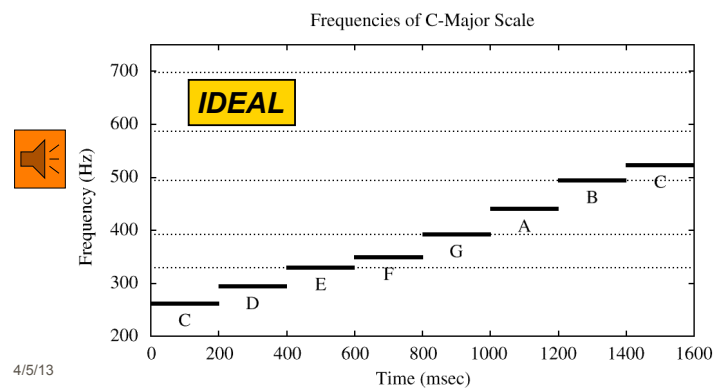
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## A Simple Time-Varying Signal

- C-major SCALE: stepped frequencies
- Frequency is constant for each note



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# SPECTROGRAM

- SPECTROGRAM Tool
  - MATLAB function is `spectrogram.m`
  - SP-First has `plotspec.m` & `spectgr.m`
- **ANALYSIS** program
  - Takes  $x(t)$  as input
  - Produces spectrum values  $X_k$
  - Breaks  $x(t)$  into **SHORT TIME SEGMENTS**
    - Then uses the FFT (Fast Fourier Transform)

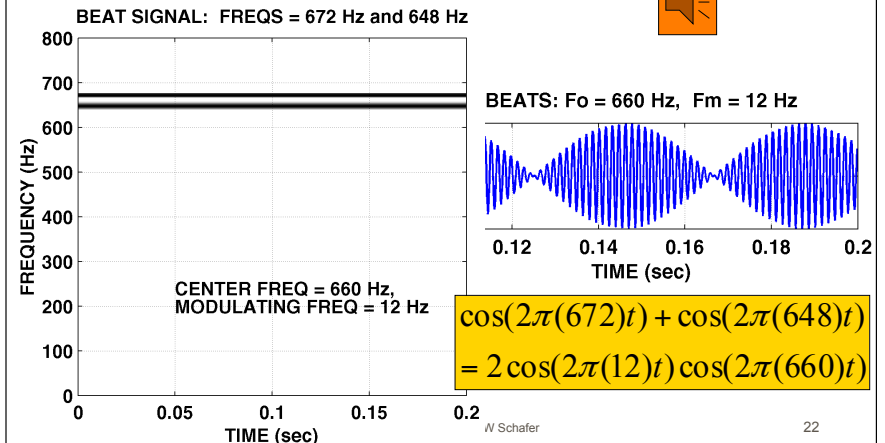
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# SPECTROGRAM EXAMPLE

- Two **Constant** Frequencies: Beats



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# Like a AM DSB-SC Radio Signal

- Same as BEAT Notes



$$\cos(2\pi(660)t) \sin(2\pi(12)t)$$

$$\frac{1}{2} \left( e^{j2\pi(660)t} + e^{-j2\pi(660)t} \right) \frac{1}{2j} \left( e^{j2\pi(12)t} - e^{-j2\pi(12)t} \right)$$

$$\frac{1}{4j} \left( e^{j2\pi(672)t} - e^{-j2\pi(672)t} - e^{j2\pi(648)t} + e^{-j2\pi(648)t} \right)$$

$$\frac{1}{2} \cos(2\pi(672)t - \frac{\pi}{2}) + \frac{1}{2} \cos(2\pi(648)t + \frac{\pi}{2})$$

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# SPECTRUM of AM (Amplitude Modulation)

- **SUM** of 4 complex exponentials:



What is the fundamental frequency?

648 Hz ?

24 Hz ?

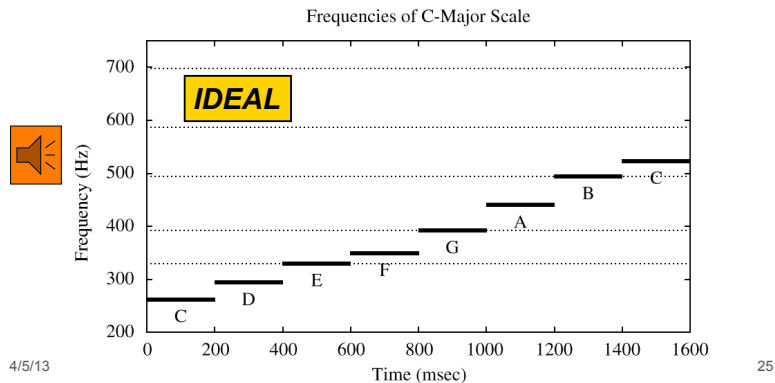
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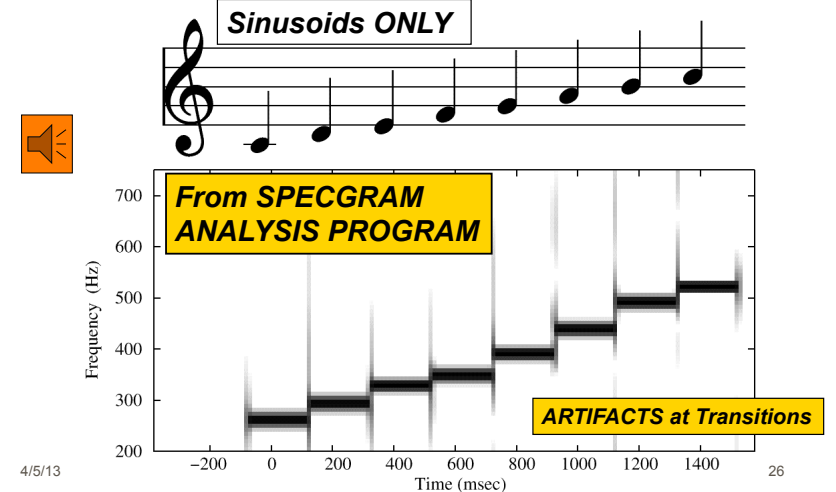
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## STEPPED FREQUENCIES

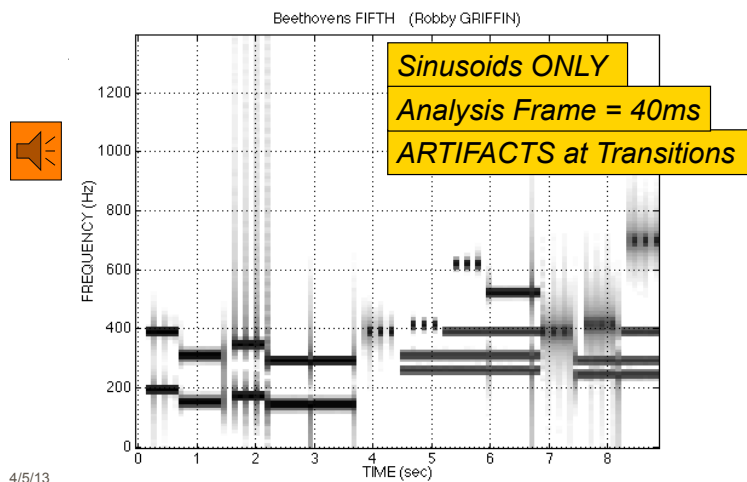
- C-major SCALE: successive sinusoids
  - Frequency is constant for each note



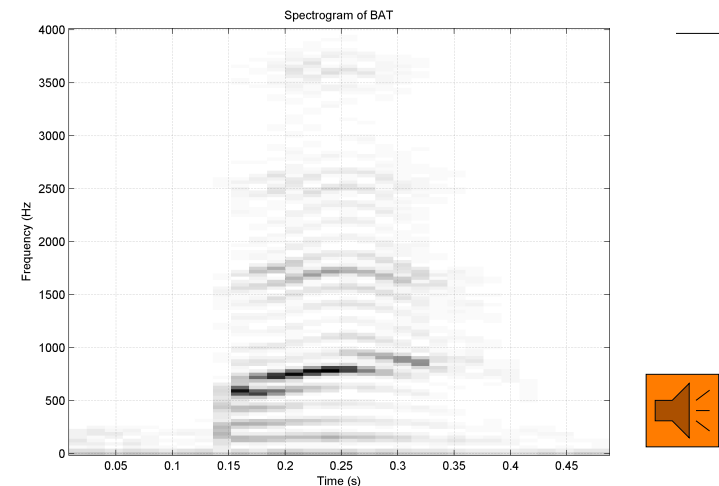
## SPECTROGRAM of C-Scale



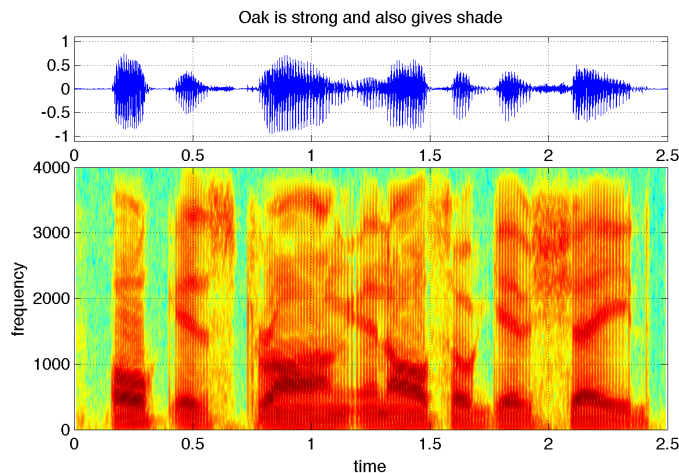
## Spectrogram of LAB SONG



## Spectrogram of BAT (plotspec)



## Speech Spectrogram



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## Time-Varying Frequency

- Frequency can change **vs. time**
  - Continuously, not stepped
- FREQUENCY MODULATION (FM)**

$$x(t) = \cos(2\pi f_c t + v(t))$$

VOICE

- CHIRP SIGNALS 
- Linear Frequency Modulation (LFM)

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## New Signal: Linear FM

- Called **Chirp** Signals (LFM)
  - Quadratic phase

QUADRATIC

$$x(t) = A \cos(\alpha t^2 + 2\pi f_0 t + \varphi)$$

- Freq will change **LINEARLY** vs. time
  - Example of Frequency Modulation (FM)
  - Define “instantaneous frequency”

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## INSTANTANEOUS FREQ

- Definition

$$x(t) = A \cos(\psi(t))$$

$$\Rightarrow \omega_i(t) = \frac{d}{dt} \psi(t)$$

Derivative  
of the “Angle”

- For Sinusoid:

$$x(t) = A \cos(2\pi f_0 t + \varphi)$$

$$\psi(t) = 2\pi f_0 t + \varphi$$

Makes sense

$$\Rightarrow \omega_i(t) = \frac{d}{dt} \psi(t) = 2\pi f_0$$

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## INSTANTANEOUS FREQ of the Chirp

- Chirp Signals have Quadratic phase
- Freq will change **LINEARLY** vs. time

$$x(t) = A \cos(\alpha t^2 + \beta t + \varphi)$$

$$\Rightarrow \psi(t) = \alpha t^2 + \beta t + \varphi$$

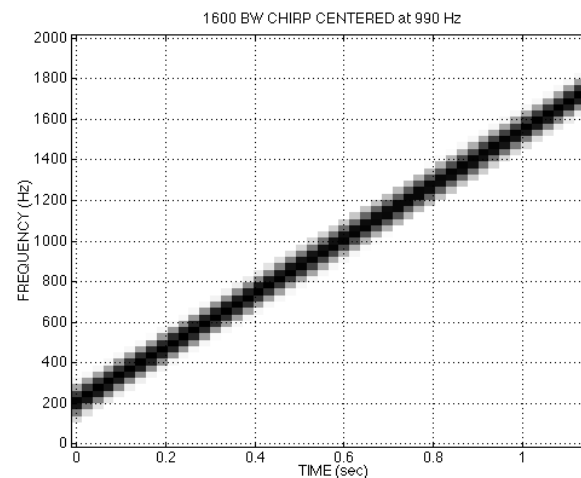
$$\Rightarrow \omega_i(t) = \frac{d}{dt} \psi(t) = 2\alpha t + \beta$$

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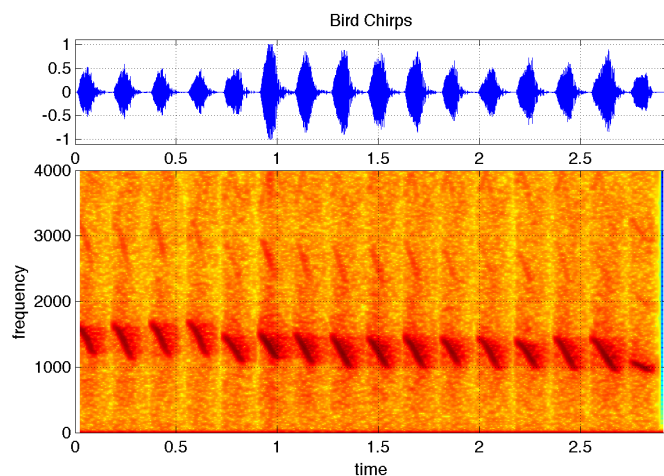
## CHIRP SPECTROGRAM



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## Bird Chirp



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## SAMPLING CONTINUOUS- TIME SIGNALS

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## SYSTEMS Process Signals



### PROCESSING GOALS:

- Change  $x(t)$  into  $y(t)$ 
  - For example, more BASS, pitch shifting
- Improve  $x(t)$ , e.g., image deblurring
- Extract Information from  $x(t)$
- Digital code for transmission and storage

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## System IMPLEMENTATION

### ANALOG/ELECTRONIC:

- Circuits: resistors, capacitors, op-amps



### DIGITAL/MICROPROCESSOR

- Convert  $x(t)$  to **numbers** stored in memory



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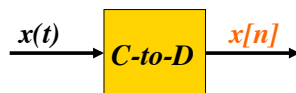
## SAMPLING $x(t)$

### SAMPLING PROCESS

- Convert  $x(t)$  to **numbers**  $x[n]$ 
  - " $n$ " is an integer;  $x[n]$  is a sequence of values
  - Think of " $n$ " as the storage address in memory

### UNIFORM SAMPLING at $t = nT_s$

- IDEAL:  $x[n] = x(nT_s)$



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## SAMPLING RATE, $f_s$

### SAMPLING RATE ( $f_s$ )

- $f_s = 1/T_s$ 
  - NUMBER of SAMPLES PER SECOND
- $T_s = 125 \text{ microsec} \rightarrow f_s = 8000 \text{ samples/sec}$ 
  - UNITS ARE HERTZ: 8000 Hz

### UNIFORM SAMPLING at $t = nT_s = n/f_s$

- IDEAL:  $x[n] = x(nT_s) = x(n/f_s)$



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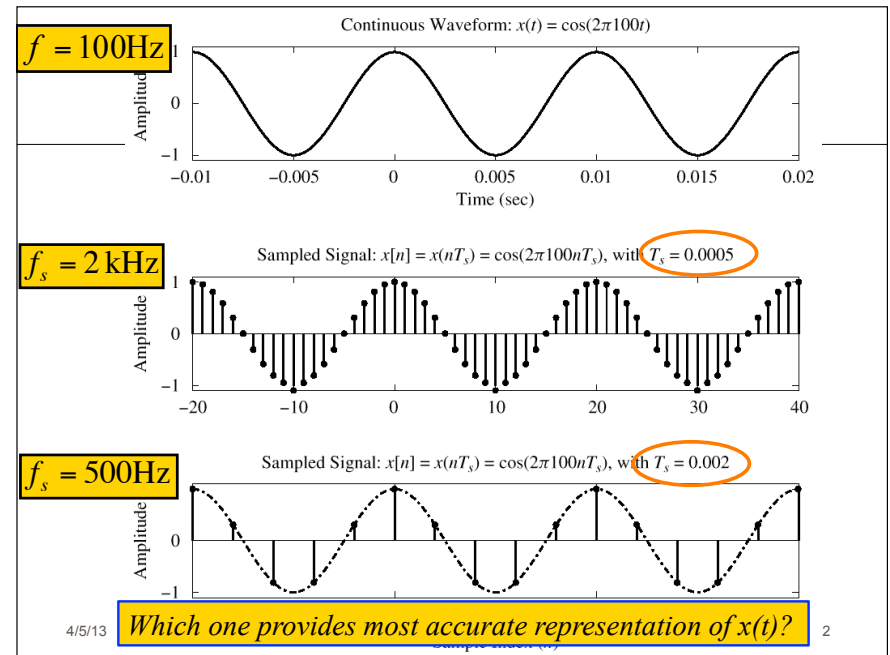
# STORING DIGITAL SOUND

- $x[n]$  is a SAMPLED SINUSOID
  - A list of numbers stored in memory
- EXAMPLE: audio CD
- CD rate is 44,100 samples per second
  - 16-bit samples
  - Stereo uses 2 channels
- Number of bytes for 1 minute is
  - $2 \times (16/8) \times 60 \times 44100 = 10.584$  Mbytes

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# SAMPLING THEOREM

- HOW OFTEN DO WE NEED TO SAMPLE?
  - DEPENDS on FREQUENCY of SINUSOID
  - ANSWERED by SHANNON/NYQUIST Theorem
  - ALSO DEPENDS on **“RECONSTRUCTION”**

## Shannon Sampling Theorem

A continuous-time signal  $x(t)$  with frequencies no higher than  $f_{\max}$  can be reconstructed exactly from its samples  $x[n] = x(nT_s)$ , if the samples are taken at a rate  $f_s = 1/T_s$  that is greater than  $2f_{\max}$ .

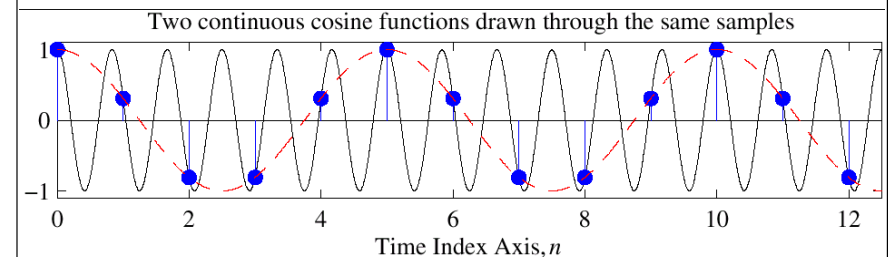
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# Reconstruction? Which One?

Given the samples, draw a sinusoid through the values



$$x[n] = \cos(0.4\pi n)$$

When  $n$  is an integer  
 $\cos(0.4\pi n) = \cos(2.4\pi n)$

Reconstruction picks lowest frequency sinusoid

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## Spatial Aliasing



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## Spatial Aliasing



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## DISCRETE-TIME SINUSOID

- Change  $x(t)$  into  $x[n]$

$$x(t) = A \cos(\omega t + \varphi)$$

$$x[n] = x(nT_s) = A \cos(\omega nT_s + \varphi)$$

$$x[n] = A \cos((\omega T_s)n + \varphi)$$

$$x[n] = A \cos(\hat{\omega}n + \varphi)$$

$$\hat{\omega} = \omega T_s = \frac{\omega}{f_s} \quad \text{DEFINE DIGITAL FREQUENCY}$$

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## DIGITAL FREQUENCY $\hat{\omega}$

$$\hat{\omega} = \omega T_s = \frac{2\pi f}{f_s}$$

- $\hat{\omega}$  VARIES from **0** to  **$2\pi$** , as  $f$  varies from 0 to the sampling frequency
- UNITS are radians, **not** rad/sec
  - DIGITAL FREQUENCY is NORMALIZED

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