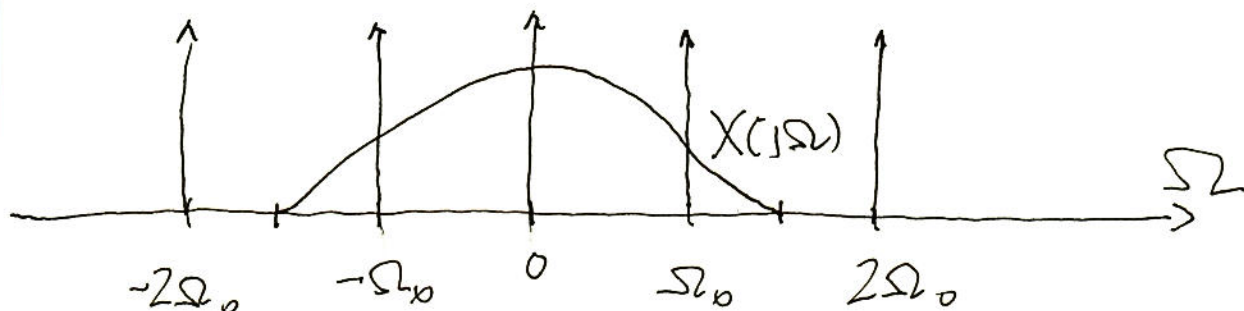


Problem 1

$$\begin{aligned}
 1) \text{ We have } y(t+T) &= \sum_{n=-\infty}^{\infty} x(t+T-nT) \\
 &= \sum_{m=-\infty}^{\infty} x(t-mT) \quad \text{where } m = n-1 \\
 &= y(t).
 \end{aligned}$$

2) The Fourier transform of $y(t)$, $Y(j\Omega)$ is given by

$$\begin{aligned}
 Y(j\Omega) &= \sum_{n=-\infty}^{\infty} e^{-jn\Omega T} X(j\Omega) \\
 &= \left(\sum_{n=-\infty}^{\infty} e^{-jn\Omega T} \right) X(j\Omega) \\
 &= \frac{2\pi}{T} \left(\sum_{k=-\infty}^{\infty} \delta(\Omega - 2\pi \frac{k}{T}) \right) X(j\Omega) \\
 &= \frac{2\pi}{T} \left(\sum_{k=-\infty}^{\infty} \delta(\Omega - k\Omega_0) \right) X(j\Omega)
 \end{aligned}$$



$$Y(j\Omega) = \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} X(jk\Omega_0) \delta(\Omega - k\Omega_0)$$

But since X is band limited on $(-1.5\Omega_0, 1.5\Omega_0)$

$X(jk\Omega_0) = 0$ except for $k = -1, 0, 1$. (see figure).

Hence

$$Y(j\Omega) = \frac{2\pi}{T} \left\{ X(j\Omega_0) \delta(\Omega - \Omega_0) + X(-j\Omega_0) \delta(\Omega + \Omega_0) + X(0) \delta(\Omega) \right\}$$

If we inverse Fourier transform this equality we get :

$$y(t) = \frac{1}{T} \left\{ X(j\Omega_0) e^{j\Omega_0 t} + X(-j\Omega_0) e^{-j\Omega_0 t} + X(0) \right\}$$

Since $x(t)$ is real $X(-j\Omega_0) = \int_{-\infty}^{\infty} x(t) e^{-j\Omega_0 t} dt$

$$= \left(\int_{-\infty}^{\infty} x(t) e^{j\Omega_0 t} dt \right)^*$$

If we write $X(j\Omega_0) = |X(j\Omega_0)| e^{j\angle X(j\Omega_0)}$

we have

$$y(t) = \frac{1}{T} X(0) + \frac{2}{T} |X(j\Omega_0)| \frac{1}{2} \left\{ e^{j\Omega_0 t + j\angle X(j\Omega_0)} + e^{-j\Omega_0 t + j\angle X(j\Omega_0)} \right\}$$

In other words

$$y(t) = \underbrace{\frac{1}{T} X(0)}_A + \underbrace{\frac{2}{T} |X(j\Omega_0)|}_B \cos\left(\underbrace{\Omega_0 t}_{\frac{2\pi}{T}} + \underbrace{\angle X(j\Omega_0)}_C\right)$$

Problem 2

1. Define $x_2(t) = x(t/2)$. The Fourier transform of x_2 is given by

$$X_2(j\Omega) = 2 X(j\Omega/2)$$

X_2 is band limited on the interval $[-\Omega_0/2, \Omega_0/2]$.

and can be sampled with $T_2 = \frac{\pi}{\Omega_0/2} = \frac{2\pi}{\Omega_0} = 2T$ without aliasing.

The sampled signal is given by

$$\sum_{n=-\infty}^{\infty} x_2(2nT) \delta(t - n2T) = \sum_{n=-\infty}^{\infty} x(nT) \delta(t - n2T)$$

which is exactly equal to $y(t)$.

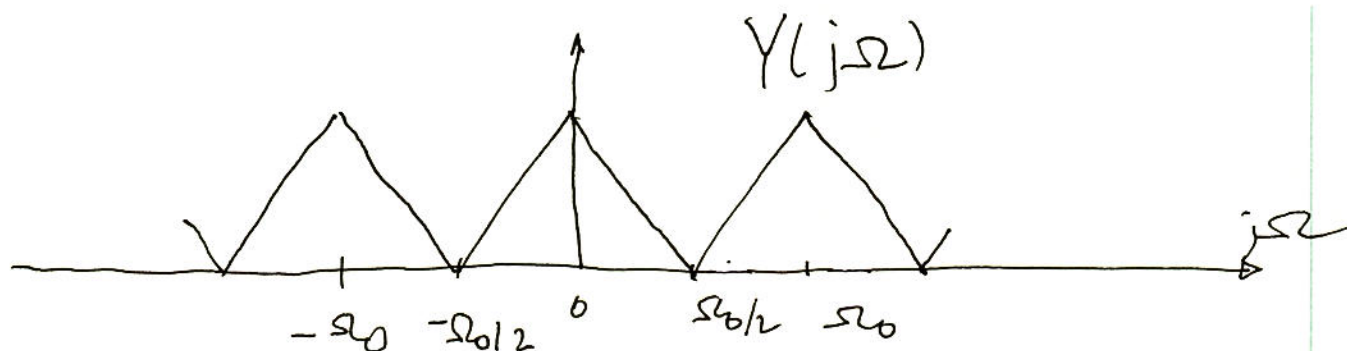
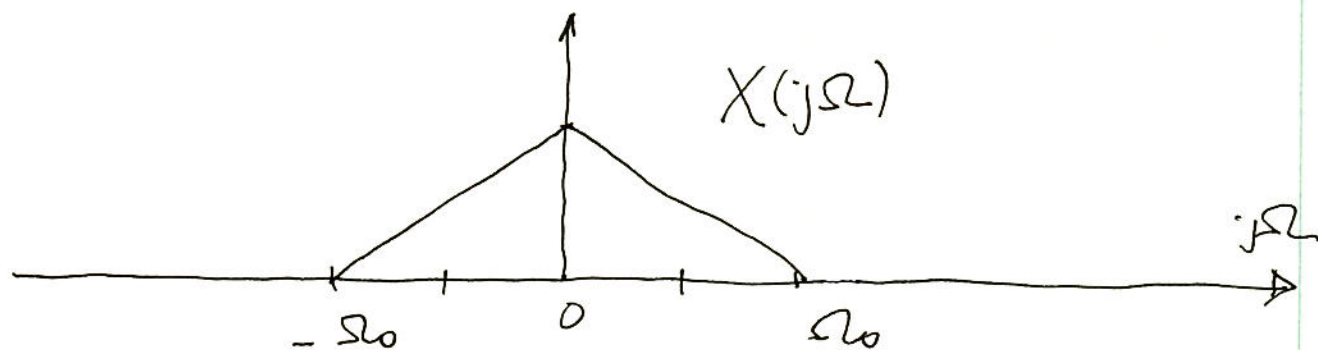
The Fourier transform of $y(t)$ is therefore the Fourier transform of the sampled signal $x_2(t)$ with period $2T$

$$Y(j\Omega) = \frac{1}{2T} \sum_{k=-\infty}^{\infty} X_2(j(\Omega - \frac{2\pi k}{2T}))$$

$$Y(j\Omega) = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} X_2(j(\Omega - k\frac{\pi}{T})) = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} 2X(2j(\Omega - k\frac{\pi}{T}))$$

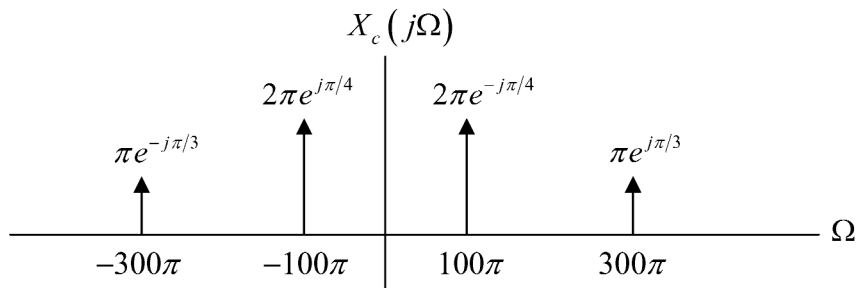
$$Y(j\Omega) = \frac{1}{\pi} \sum_{n=-\infty}^{\infty} X(j2(\Omega - n\Omega_0))$$

- 2) As explained before there is no aliasing. This is also clear from the Fourier transform of $y(t)$, which is the same as the Fourier transform of the sampled $x(t)$ with sampling period T when the frequency is scaled by a factor 2.

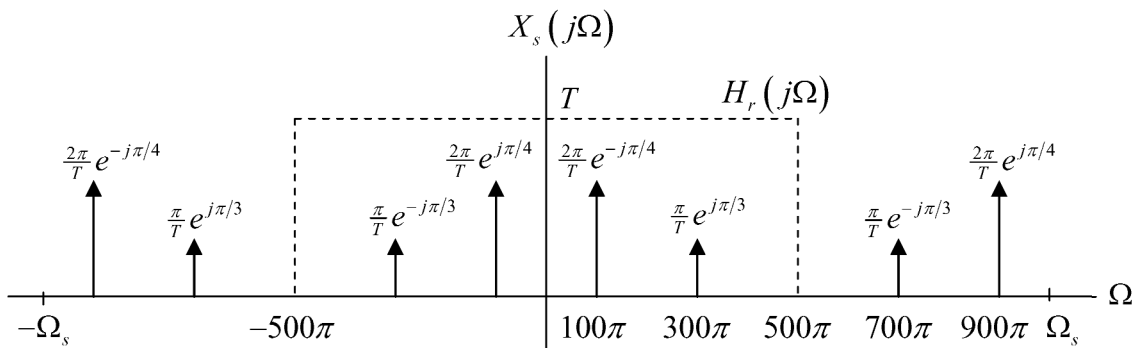


4.28. A.

$$X_c(j\Omega) = 2\pi e^{-j\pi/4} \delta(\Omega - 100\pi) + 2\pi e^{j\pi/4} \delta(\Omega + 100\pi) \\ + \pi e^{j\pi/3} \delta(\Omega - 300\pi) + \pi e^{-j\pi/3} \delta(\Omega + 300\pi).$$



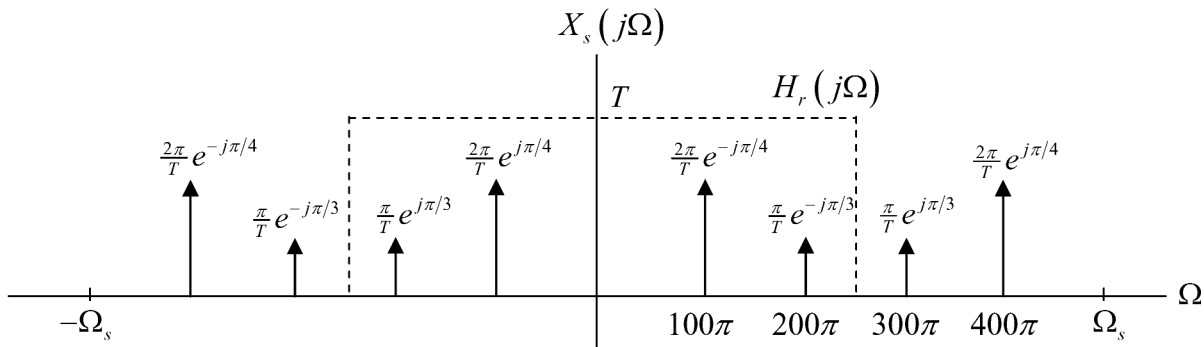
B. If $f_s = 1/T = 500$ samples/s then $\Omega_s = 2\pi/T = 1000\pi$ rad/s.



There is no aliasing, so $x_r(t) = x_c(t)$; that is,

$$x_r(t) = 2 \cos(100\pi t - \pi/4) + \cos(300\pi t + \pi/3).$$

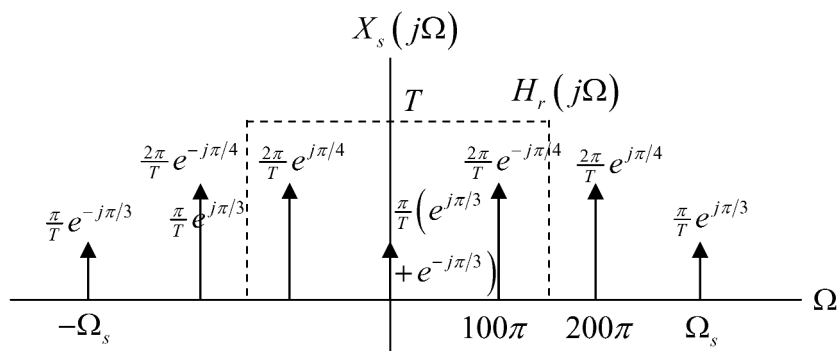
C. If $f_s = 1/T = 250$ samples/s then $\Omega_s = 2\pi/T = 500\pi$ rad/s.



Now there is aliasing and

$$x_r(t) = 2 \cos(100\pi t - \pi/4) + \cos(200\pi t - \pi/3).$$

D. We want to sample the component at 300π rad/s exactly once per cycle, so that all the samples have the same value. At $\Omega_s = 300\pi$ rad/s we have

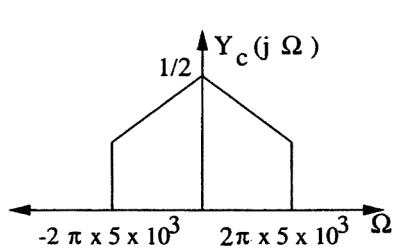
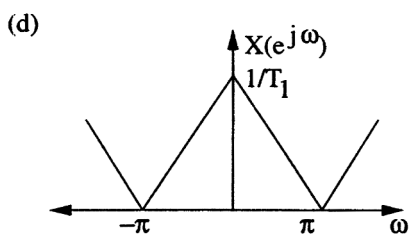
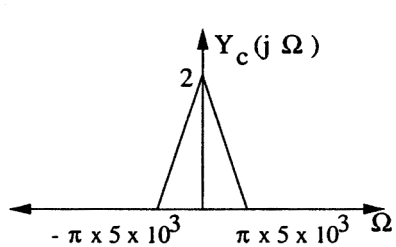
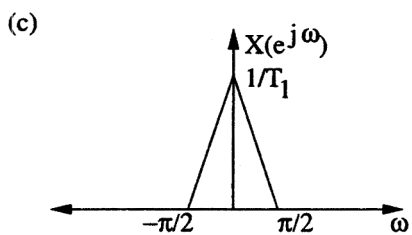
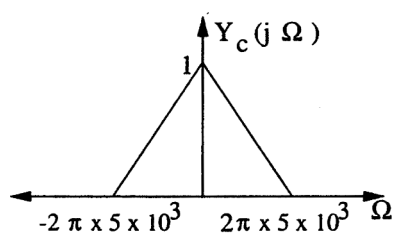
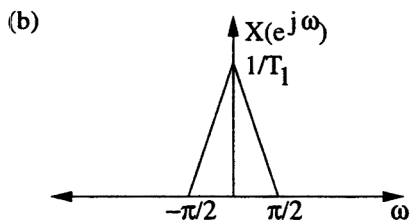
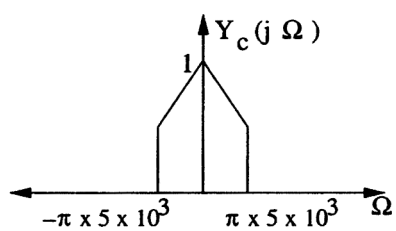
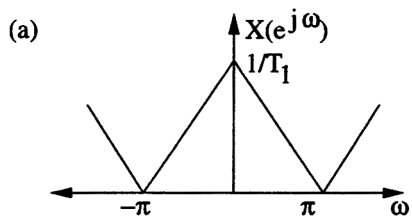


Now

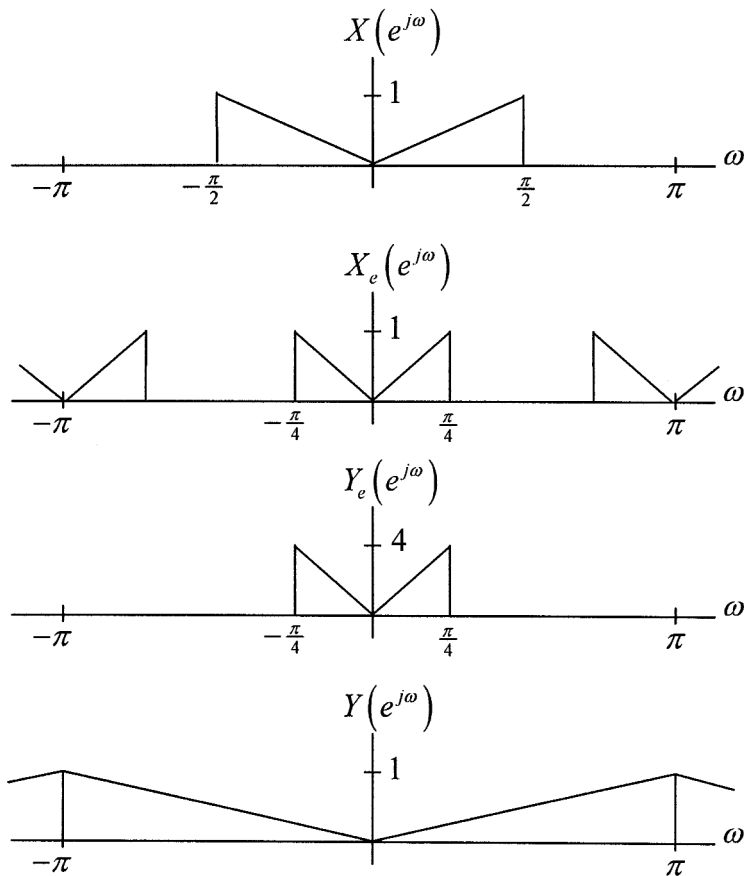
$$\begin{aligned} x_r(t) &= \cos(\pi/3) + 2 \cos(100\pi - \pi/4) \\ &= 1/2 + 2 \cos(100\pi - \pi/4). \end{aligned}$$

We have $A = 1/2$.

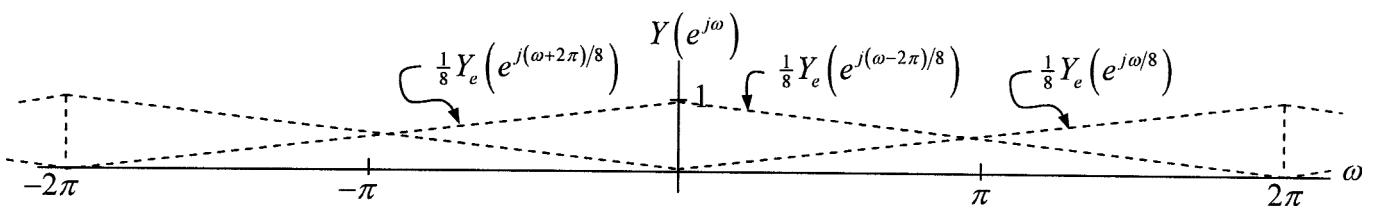
4.30. The Fourier transform of $y_c(t)$ is sketched below for each case.



4.32. A. With $L=2$ and $M=4$,

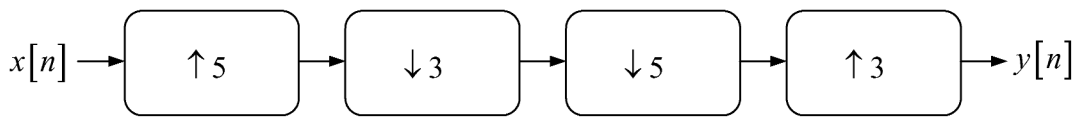


B. With $L=2$ and $M=8$, $X_e(e^{j\omega})$ and $Y_e(e^{j\omega})$ remain as in part A, except that $Y_e(e^{j\omega})$ now has a peak value of 8. After expanding we have

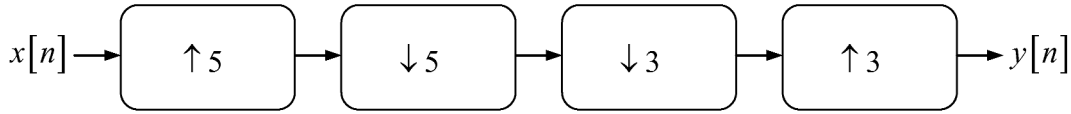


We see that $Y(e^{j\omega})=1$ for all ω . Inverse transforming gives $y[n]=\delta[n]$ in this case.

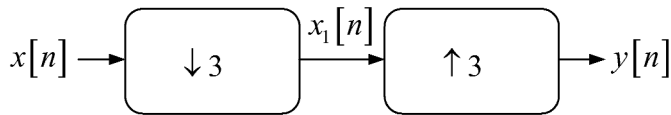
4.33.



Since 3 and 5 are relatively prime, the order of the two operations in the center can be interchanged. This gives



Expanding by 5 and immediately compressing by 5 produces no net effect. We have



Compressing by 3 produces

$$x_1[n] = x[3n].$$

Expanding by 3 now gives

$$y[n] = \begin{cases} x_1[n/3], & n = 3k, \quad k \text{ any integer} \\ 0, & \text{otherwise.} \end{cases}$$

That is,

$$y[n] = \begin{cases} x[n], & n = 3k, \quad k \text{ any integer} \\ 0, & \text{otherwise.} \end{cases}$$

- (a) There is one output sample generated for every pair of input samples. Even input samples require 3 multiplies and odd input samples require 2 multiplies. Thus each pair requires 5 multiplies.
- (b) Applying the compressor identity to the previous structure results in:

$$H(z) = H_0(z^2) + z^{-1}H_1(z^2).$$

From the difference equations in the previous part we have:

$$H_0(z) = \frac{-\frac{1}{3} + \frac{1}{8}z^{-1}}{1 - \frac{1}{4}z^{-1}},$$

and

$$H_1(z) = \frac{\frac{1}{12}}{1 - \frac{1}{4}z^{-1}}.$$

Thus,

$$H(z) = \frac{-\frac{1}{3} + \frac{1}{8}z^{-2} + \frac{1}{12}z^{-1}}{1 - \frac{1}{4}z^{-2}} = \frac{-\frac{1}{3}(1 - \frac{3}{4}z^{-1})(1 + \frac{1}{2}z^{-1})}{(1 - \frac{1}{2}z^{-1})(1 + \frac{1}{2}z^{-1})} = \frac{-\frac{1}{3} + \frac{1}{4}z^{-1}}{1 - \frac{1}{2}z^{-1}}.$$

Therefore, $a = 1/2$, $b = -1/3$ and $c = 1/4$.

- (c) In this implementation 3 multiplies are required for every input sample. For every output sample we need to calculate 2 values of $v[n]$. Altogether we need 6 multiplies per output sample.