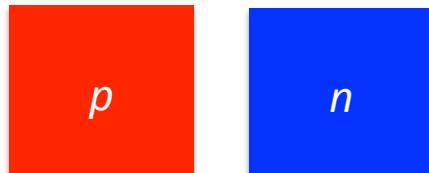


Overview

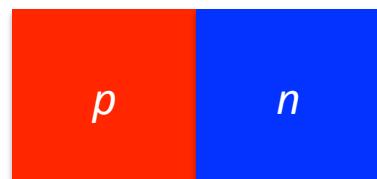
- Junctions between *n* and *p*-type doped semiconductors are called **p-n junctions**
- These junctions are at the heart of most semiconductor devices
- If the transition between n- and p-areas are abrupt we speak of an **abrupt p-n junction**
- Electrons and holes of both sides can combine via drift and diffusion, resulting in a **depletion or space charge zone**
- The redistribution of charge results in a **built-in potential V_{bi}**
- **Band structure, doping levels and profiles, ionization energies as well as generation and recombination rates** determine the performance of the junction and hence that of the devices
- Junctions between different semiconductors (band structure, doping levels, work functions and electron affinities) are called **heterojunctions**

Equilibrium case (no external fields)

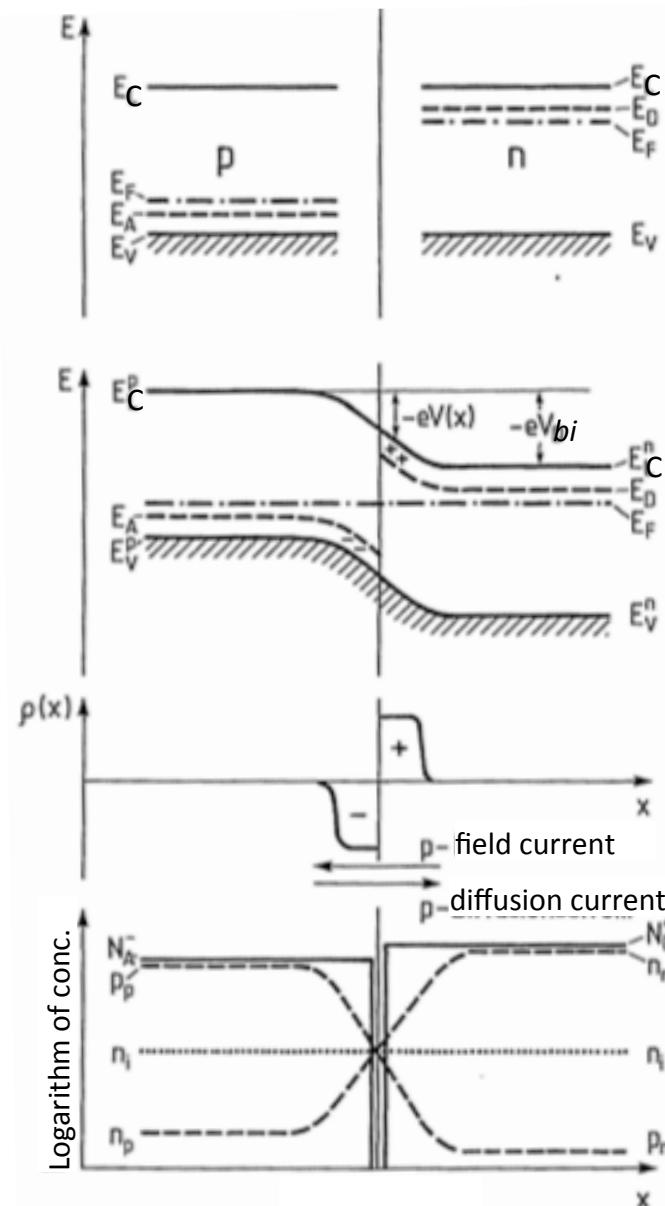
- p and n semiconductors separated



- p and n semiconductors in contact

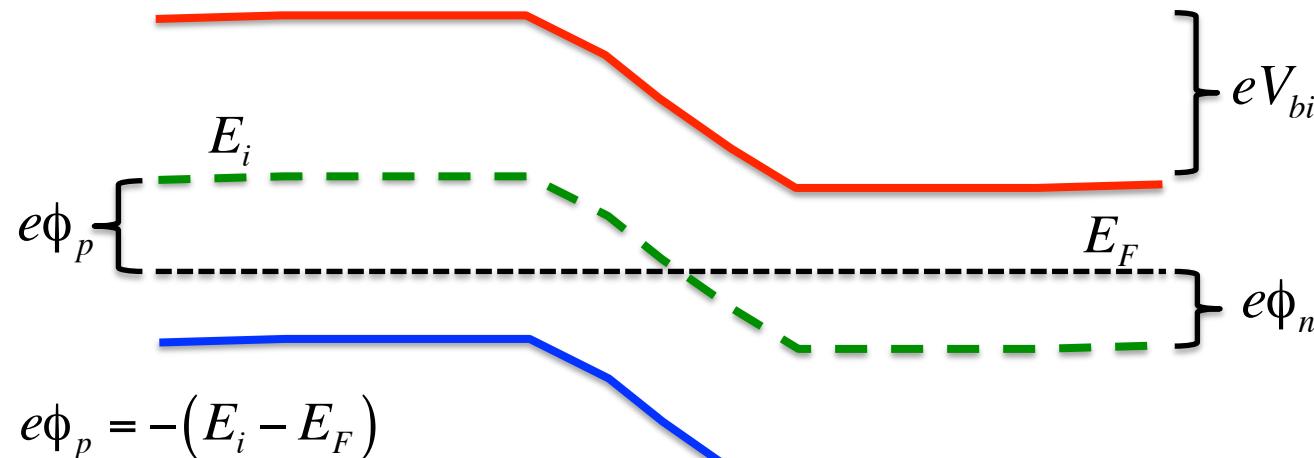


- Chemical potentials ($\sim E_F$) equilibrate at interface
- Charge carrier drift and diffusion fluxes equilibrate
- A built-in potential V_{bi} forms due to charge separation
- Charge carrier densities become spatially dependent



Equilibrium case (no external fields)

- The built-in potential V_{bi}



- Charge neutrality

$$N_D - N_A + p - n = 0 \quad \left| \quad p = p_0 \exp\left(\frac{E_i - E_F}{k_B T}\right) \quad \right| \quad p_0 \approx n_i \Rightarrow k_B T \ln\left(\frac{p}{n_i}\right) = E_i - E_F$$

- Full ionisation

$$p = N_A \Rightarrow e\phi_p = -k_B T \ln\left(\frac{N_A}{n_i}\right) \quad \& \quad e\phi_n = -k_B T \ln\left(\frac{N_D}{n_i}\right)$$

- Built-in potential

$$V_{bi} = \phi_n - \phi_p = \frac{k_B T}{e} \ln\left(\frac{N_A N_D}{n_i^2}\right)$$

Equilibrium case (no external fields)

- Charge carrier densities

$$n = n_0 e^{-\frac{E_C - \mu - eV(x)}{k_B T}}$$

\Rightarrow

$$\frac{\partial n}{\partial x} = n \frac{e}{k_B T} \frac{\partial V(x)}{\partial x}$$

$$p = p_0 e^{-\frac{\mu - E_V + eV(x)}{k_B T}}$$

\Rightarrow

$$\frac{\partial p}{\partial x} = -p \frac{e}{k_B T} \frac{\partial V(x)}{\partial x}$$

- Diffusion current

$$j^{diff} = j_n^{diff} + j_p^{diff}$$

$$= e \left(D_n \frac{\partial n}{\partial x} - D_p \frac{\partial p}{\partial x} \right) = e^2 \left(n \frac{D_n}{k_B T} + p \frac{D_p}{k_B T} \right) \frac{\partial V(x)}{\partial x}$$

- Drift current

$$j^{field} = j_n^{field} + j_p^{field}$$

$$= e(n\mu_n + p\mu_p) E_x = -e(n\mu_n + p\mu_p) \frac{\partial V(x)}{\partial x}$$

Equilibrium case (no external fields)

- Equilibrium:

$$j_n^{diff} = j_n^{field}$$

$$j_p^{diff} = j_p^{field}$$

- Diffusivities:

$$\Rightarrow D_n = \frac{k_B T}{e} \mu_n, D_p = \frac{k_B T}{e} \mu_p$$

Einstein-relation

- Mobilities:

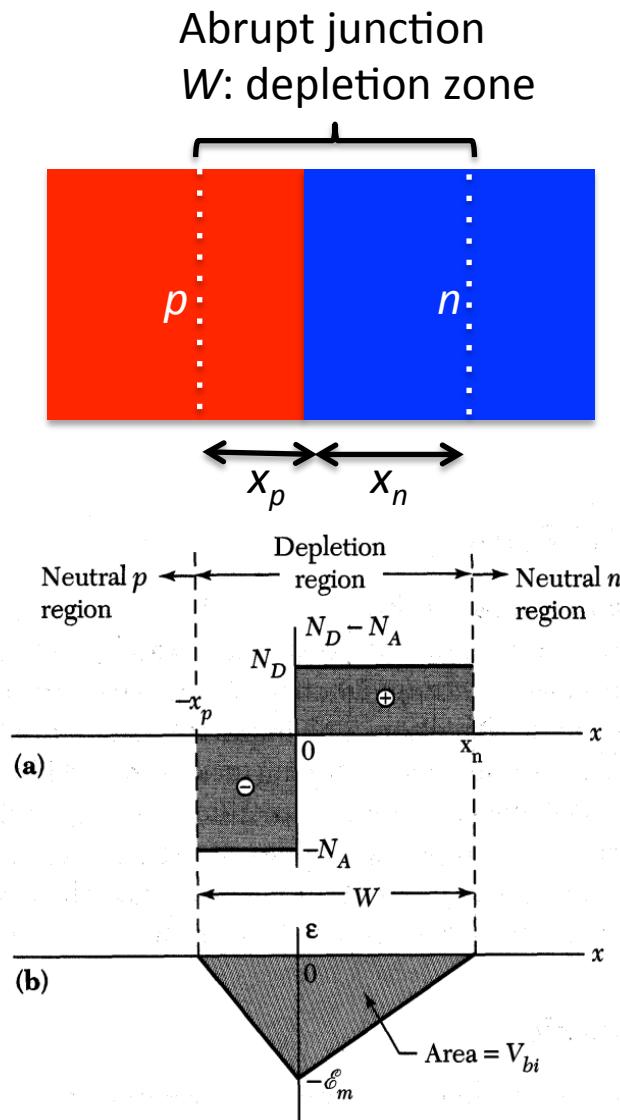
$$\mu_n = e \frac{\tau_n}{m_e^*}, \mu_p = e \frac{\tau_p}{m_h^*}$$

τ_n, τ_p : lifetimes of electrons and holes, respectively

m_e^*, m_h^* : effective masses of electrons in the CB and holes in the VB, respectively

3.1 The p-n junction

Equilibrium case (no external fields)



- Charge neutrality: $N_A x_p = N_D x_n$

- n-side:
$$\frac{d^2V}{dx^2} = -\frac{eN_D}{\epsilon_s} \quad 0 < x \leq x_n$$

$$\Rightarrow E_x = -\frac{dV}{dx} = -E_{\max} + \frac{eN_D}{\epsilon_s}(x - x_n)$$

- p-side:
$$\frac{d^2V}{dx^2} = \frac{eN_A}{\epsilon_s} \quad -x_p \leq x < 0$$

$$\Rightarrow E_x = -\frac{dV}{dx} = -\frac{eN_A}{\epsilon_s}(x + x_p)$$

- Maximum field at $x = 0$:

$$E_{\max} = \frac{eN_D x_p}{\epsilon_s} = \frac{eN_A x_n}{\epsilon_s}$$

- Built-in potential and depletion layer width W:

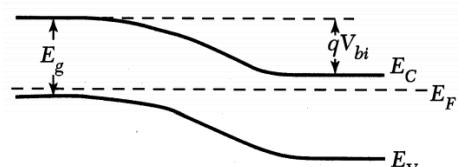
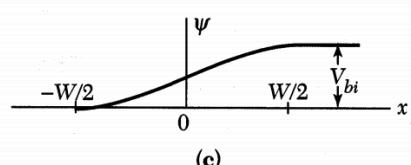
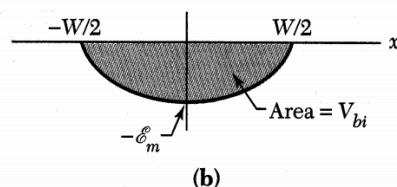
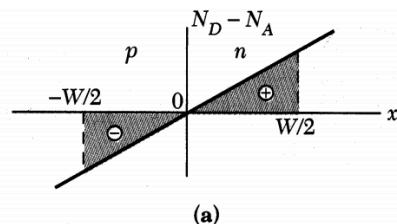
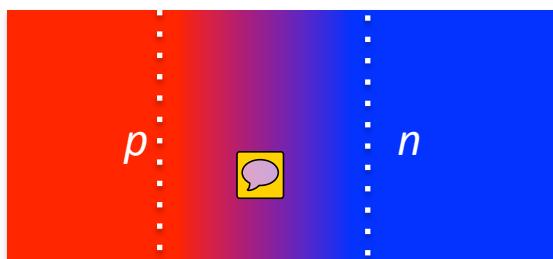
$$V_{bi} = - \int_{-x_p}^{x_n} E_x dx$$

$$\Rightarrow V_{bi} = \frac{1}{2} E_{\max} W$$

$$W = \sqrt{\frac{2\epsilon_s}{e} \left(\frac{N_A + N_D}{N_A N_D} \right) V_{bi}}$$

Equilibrium case (no external fields)

Linearly graded junction



$$\frac{d^2V}{dx^2} = \frac{-\rho_s}{\epsilon_s} = \frac{-eax}{\epsilon_s} \quad -\frac{W}{2} \leq x < +\frac{W}{2}$$

$$\Rightarrow E_x = - \int_{-W/2}^{+W/2} \frac{d^2V}{dx^2} dx$$

$$\Rightarrow E_x = - \frac{ea}{\epsilon_s} \left(\frac{(W/2)^2 - x^2}{2} \right)$$

- Maximum field at $x = 0$:

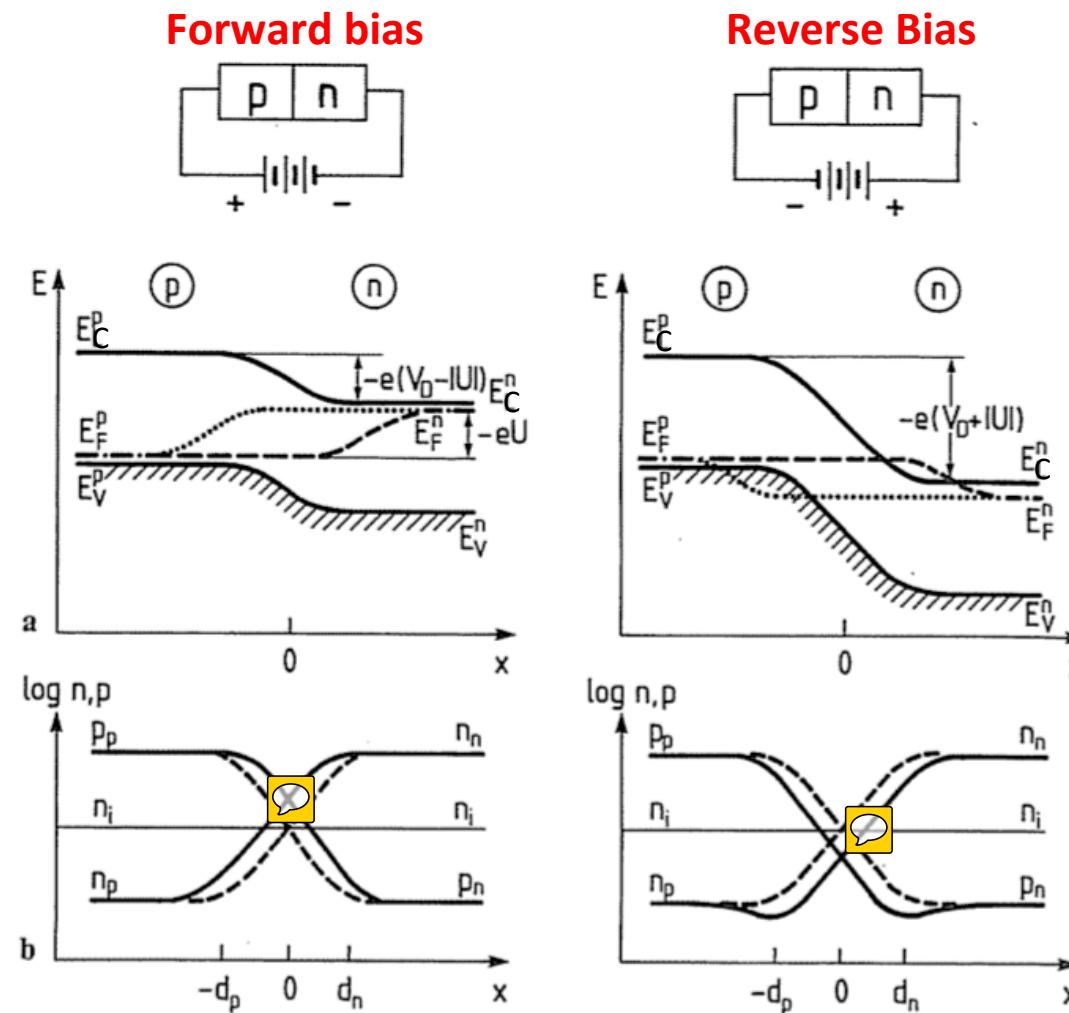
$$E_{\max} = \frac{eaW^2}{8\epsilon_s}$$

- Built-in potential and depletion zone width:

$$V_{bi} = \frac{k_B T}{e} \ln \left(\frac{aW}{2n_i} \right)$$

$$W = \sqrt[3]{\frac{12\epsilon_s V_{bi}}{ea}}$$

Effect of an external electric field in p-n-junctions



Effect of an external electric field in p-n-junctions

- Charge carrier concentrations not in equilibrium:

$n \rightarrow n_p$ Minority
 charge
 $p \rightarrow p_n$ carriers

- Calculation of j_n , j_p , n_p and p_n for the stationary state:

$$\left. \begin{aligned} j_n &= en_p \mu_n \frac{\partial V}{\partial x} + eD_n \frac{\partial n_p}{\partial x} \\ j_p &= ep_n \mu_p \frac{\partial V}{\partial x} - eD_n \frac{\partial p_n}{\partial x} \end{aligned} \right\}$$

Field- and
Diffusion currents

- Electron-hole recombination

$$\frac{\partial j_n}{\partial x} = -\frac{\partial j_p}{\partial x} = eR$$

$$R = \frac{n_p - n_{p0}}{\tau_n} = \frac{p_n - p_{n0}}{\tau_p}$$

Excess charge carrier current density, τ_n , τ_p : lifetime of electrons and holes, n_{p0} , p_{n0} : equilibrium electron and hole concentrations on the p- and n-side, respectively

Effect of an external electric field in p-n-junctions

- Poisson-Equation for abrupt junction:

$$\frac{\partial^2 V(x)}{\partial x^2} = -\frac{e}{\epsilon_s} (p + N_D^+ - n - N_A^-)$$

- No general analytical solution for Poisson-Equation

- Approximation $x_{n,p} \ll$ diffusion lengths L_n, L_p :

Conduction

$$j = j_s \left(e^{\frac{eV}{k_B T}} - 1 \right)$$

Reverse bias

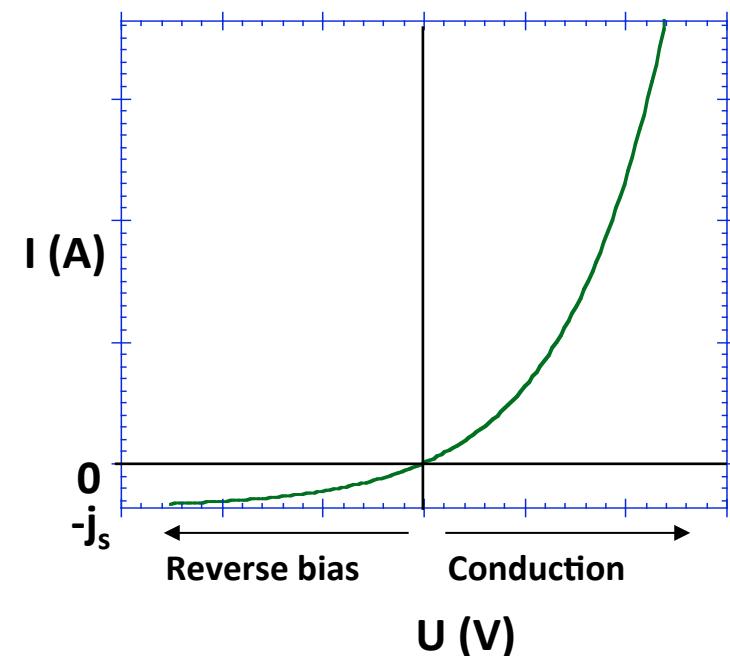
$$j = -j_s$$

With saturation current density

$$j_s = e \left(\frac{D_p p_{n0}}{L_p} + \frac{D_n n_{p0}}{L_n} \right)$$

n_O, p_V : minority charge carrier densities

$$j_s = e \sum_i D_i \frac{dc_i}{dx} \approx e \left(D_p \frac{p_{n0}}{L_p} + D_n \frac{n_{p0}}{L_n} \right)$$



Effect of an external electric field in p-n-junctions

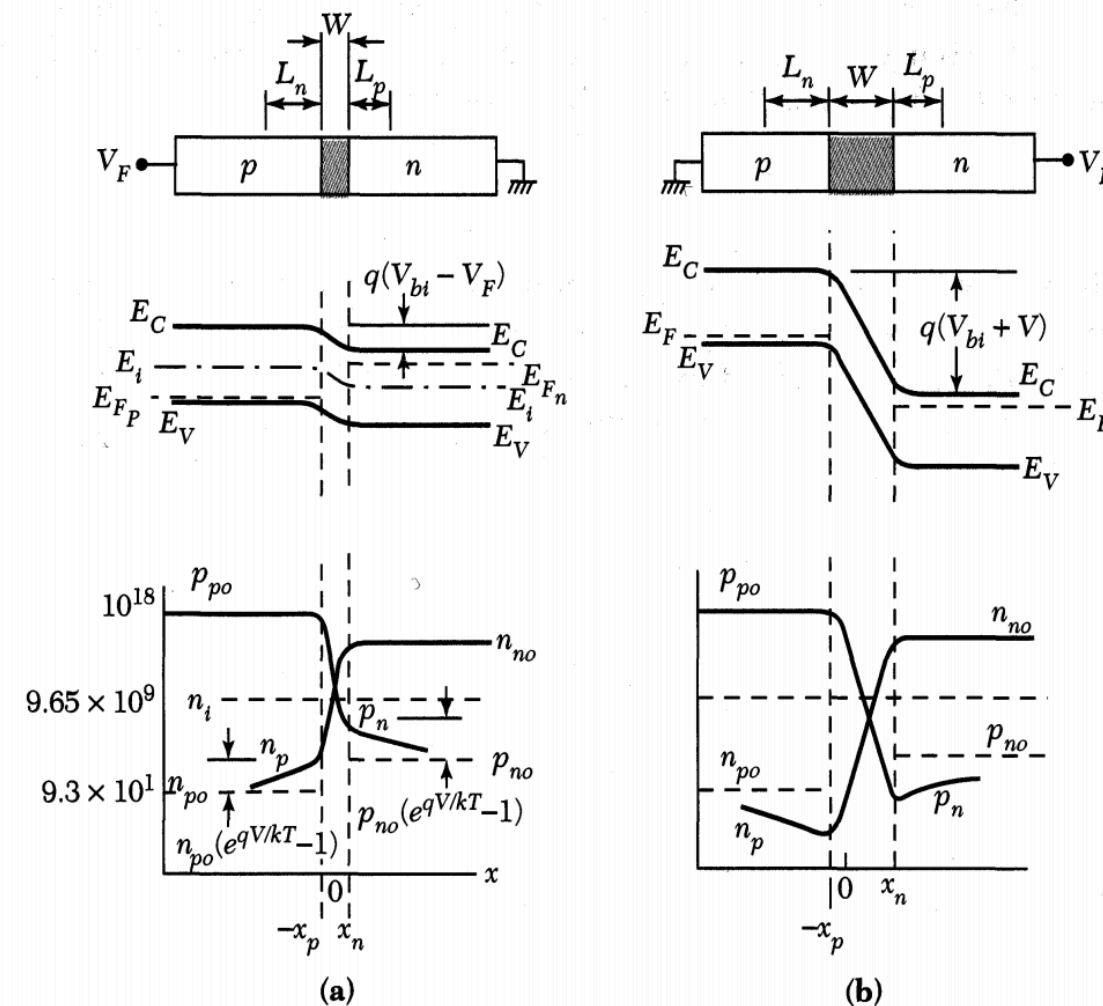


Fig. 16 Depletion region, energy band diagram and carrier distribution. (a) Forward bias. (b) Reverse bias.

Effect of an external electric field in p-n-junctions

Built-in potential

$$V_{bi} = \frac{k_B T}{e} \ln \left(\frac{p_{p0} n_{n0}}{n_i^2} \right) = \frac{k_B T}{e} \ln \left(\frac{n_{n0}}{n_{p0}} \right)$$

$$\text{Mass action law: } p_{p0} n_{p0} = n_i^2$$

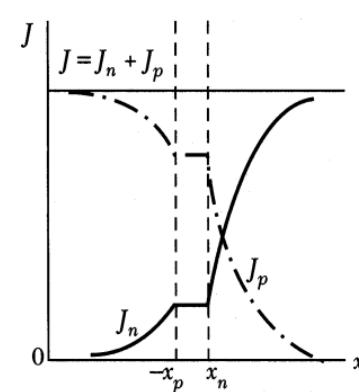
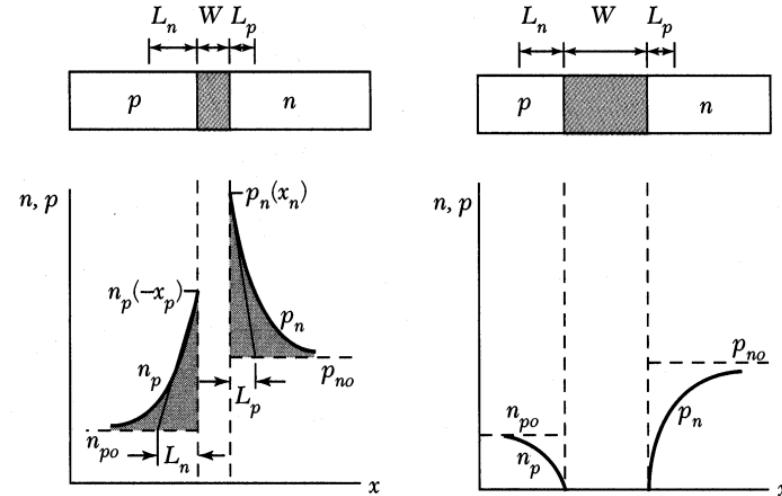
$$\Rightarrow n_{n0} = n_{p0} \exp \left(\frac{e V_{bi}}{k_B T} \right)$$

$$\text{Bias Voltage } V: n_n = n_p \exp \left(\frac{e(V_{bi} - V)}{k_B T} \right)$$

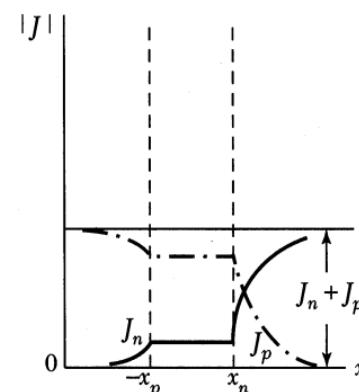
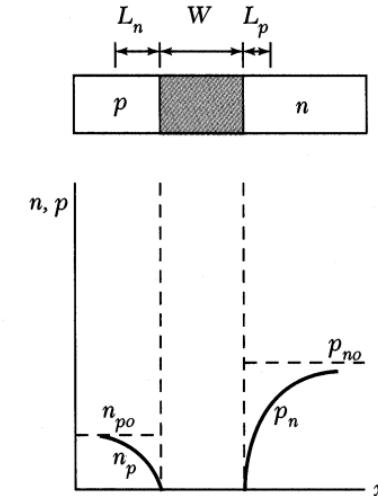
$$\text{Low injection: } n_n \approx n_{n0}$$

$$\Rightarrow n_p - n_{p0} = n_{p0} \left[\exp \left(\frac{eV}{k_B T} \right) - 1 \right]$$

$$p_n - p_{n0} = p_{n0} \left[\exp \left(\frac{eV}{k_B T} \right) - 1 \right]$$



(a)

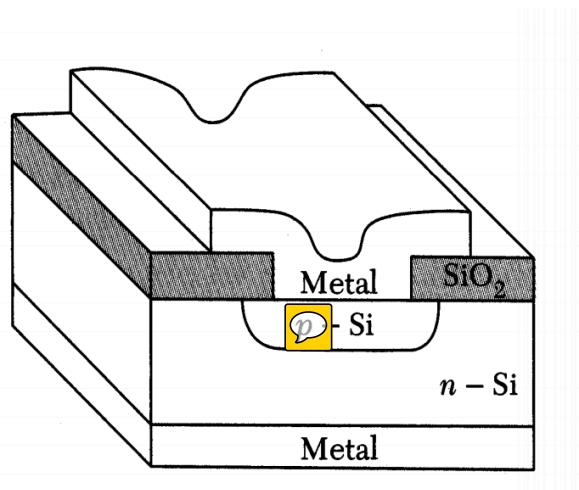


(b)

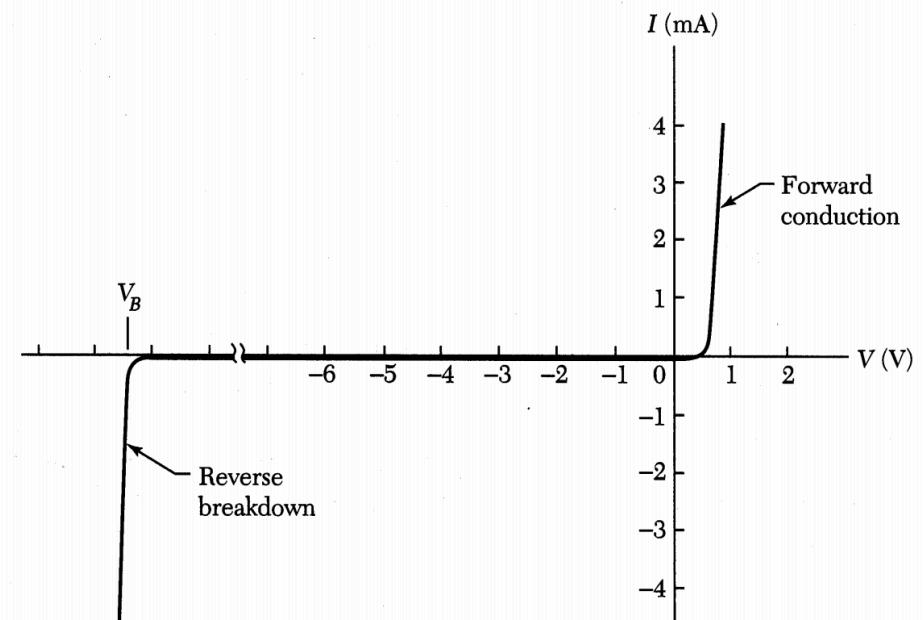
Fig. 17 Injected minority carrier distribution and electron and hole currents. (a) Forward bias. (b) Reverse bias. The figure illustrates idealized currents. For practical devices, the currents are not constant across the space charge layer.

Effect of an external electric field in p-n-junctions

- Structure of a p-n junction



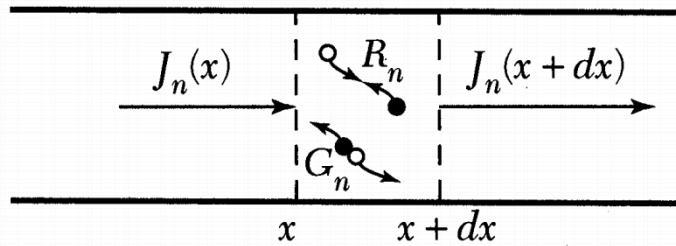
- Current/Voltage characteristic



p-n junction disrupted at a critical voltage V_B when reversed biased

Role of recombination centers

- Recombination centers in the band gap will constantly produce holes and electrons under bias conditions:



Net recombination rate:

$$U = \frac{v_{th} \sigma_n \sigma_p N_t (p_n n_n - n_i^2)}{\sigma_p \left(p_n + n_i \exp\left(\frac{E_i - E_t}{k_B T}\right) \right) + \sigma_n \left(n_n + n_i \exp\left(\frac{E_t - E_i}{k_B T}\right) \right)}$$

Assume $p_n < n_i$, $n_n < n_i$,

$\sigma_n = \sigma_p = \sigma_0$:

$$G = -U = \frac{v_{th} \sigma_0 N_t n_i}{2 \cosh\left(\frac{E_t - E_i}{k_B T}\right)} = \frac{n_i}{\tau_g}$$

Generation life time τ_g has pronounced maximum at $E_t = E_i$: only centers in the middle of the band gap contribute significantly to charge carrier generation

Role of recombination centers

- Current due to generation under reverse bias conditions:

$$J_G = e \int_0^W G dx \approx eGW = \frac{en_i W}{\tau_g}$$

Assume p⁺-n junction ($N_A \gg N_D$) and $V_R > 3k_B T/e$, the total current under reverse bias is then:

$$J_R = J_{diff} + J_G = e \sqrt{\frac{D_p}{\tau_p}} \frac{n_i^2}{N_D} + \frac{en_i W}{\tau_g} \quad \text{using} \quad L_p = \sqrt{D_p \tau_p}$$

Semiconductors with large n_i (e.g. Ge) are dominated by diffusion current and follow ideal diode characteristics, whereas materials with low n_i (e.g. Si, GaAs) will have a significant contribution of generation centers in the band gap.

Role of recombination centers

- For the forward biased current we find:

$$J_F = e \sqrt{\frac{D_p}{\tau_p}} \frac{n_i^2}{N_D} \exp\left(\frac{eV}{k_B T}\right) + \frac{en_i W}{2\tau_r} \exp\left(\frac{eV}{2k_B T}\right)$$

With the effective recombination lifetime $\tau_r = 1/(\sigma_0 v_{th} N_t)$

Similar relationship between n_i and contribution of recombination centers found as for the reverse bias case

Introducing an *ideality factor* η experiments can be empirically represented by:

$$J_F = \exp\left(\frac{eV}{\eta k_B T}\right)$$

$\eta = 1$: diffusion current dominates

$\eta = 2$: recombination current dominates

Role of recombination centers

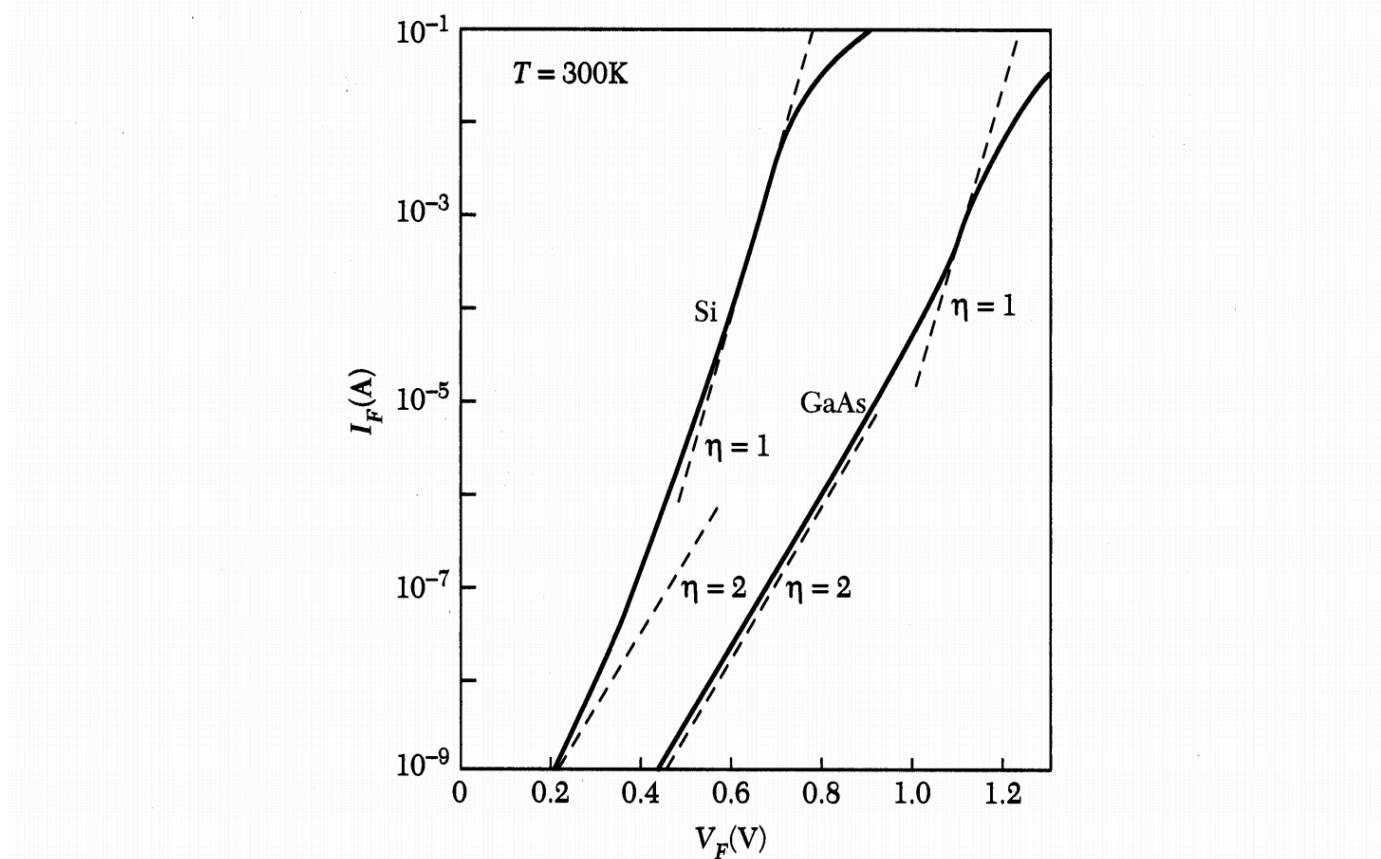


Fig. 19 Comparison of the forward current-voltage characteristics of Si and GaAs diodes² at 300 K. Dashed lines indicate slopes of different ideality factors η .

At higher forward bias diffusion current dominates

Break-through voltage

- The critical voltage for the break through of a p-n junction under reverse bias operation caused by tunnelling ($V > 4E_g/e$) or avalanche multiplication ($V > 6E_g/e$)

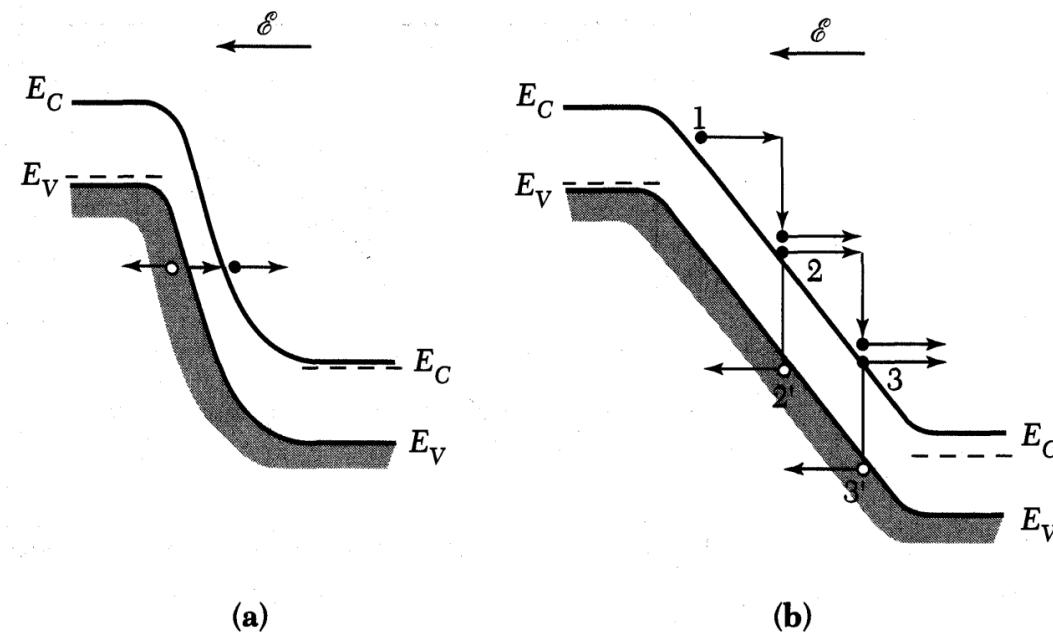


Fig. 24 Energy band diagrams under junction-breakdown conditions. (a) Tunneling effect. (b) Avalanche multiplication.

- Requires fields $> 10^6 \text{ V/m}$
- Needs sufficient doping levels ($> 5 \cdot 10^{17} \text{ cm}^{-3}$)

Break-through voltage

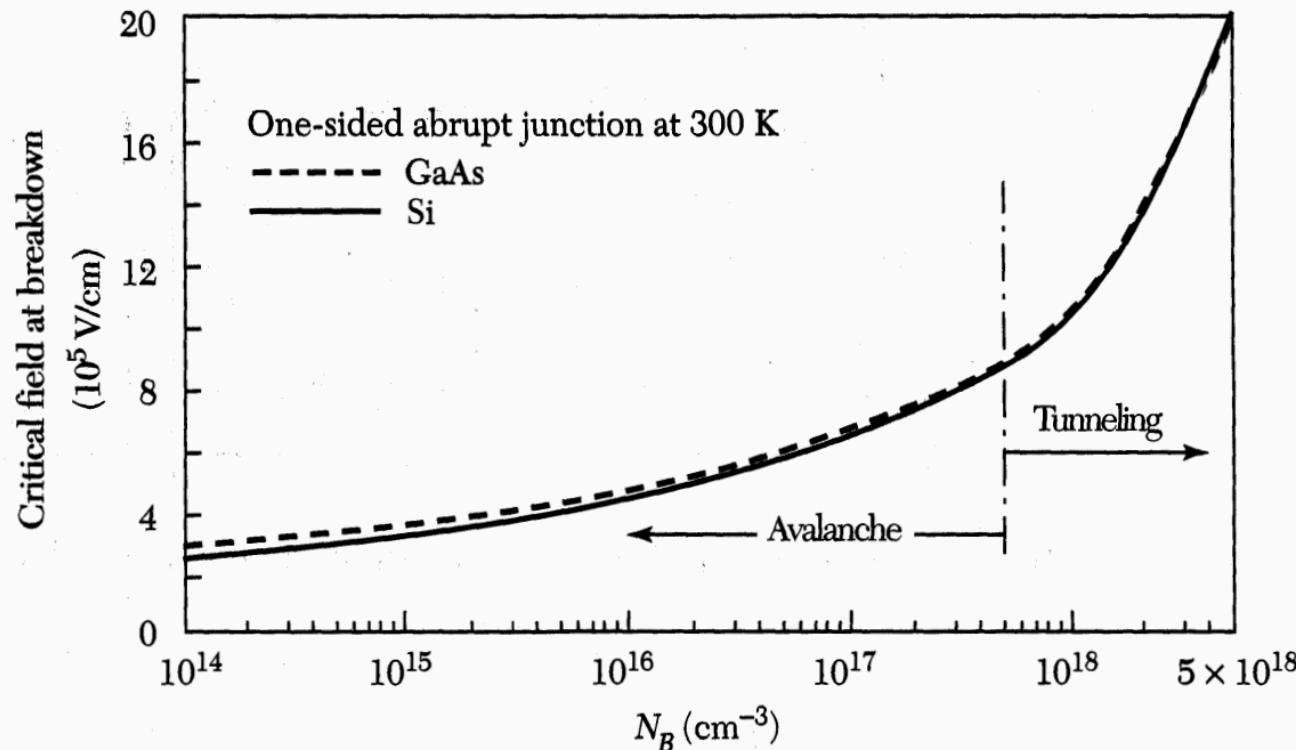
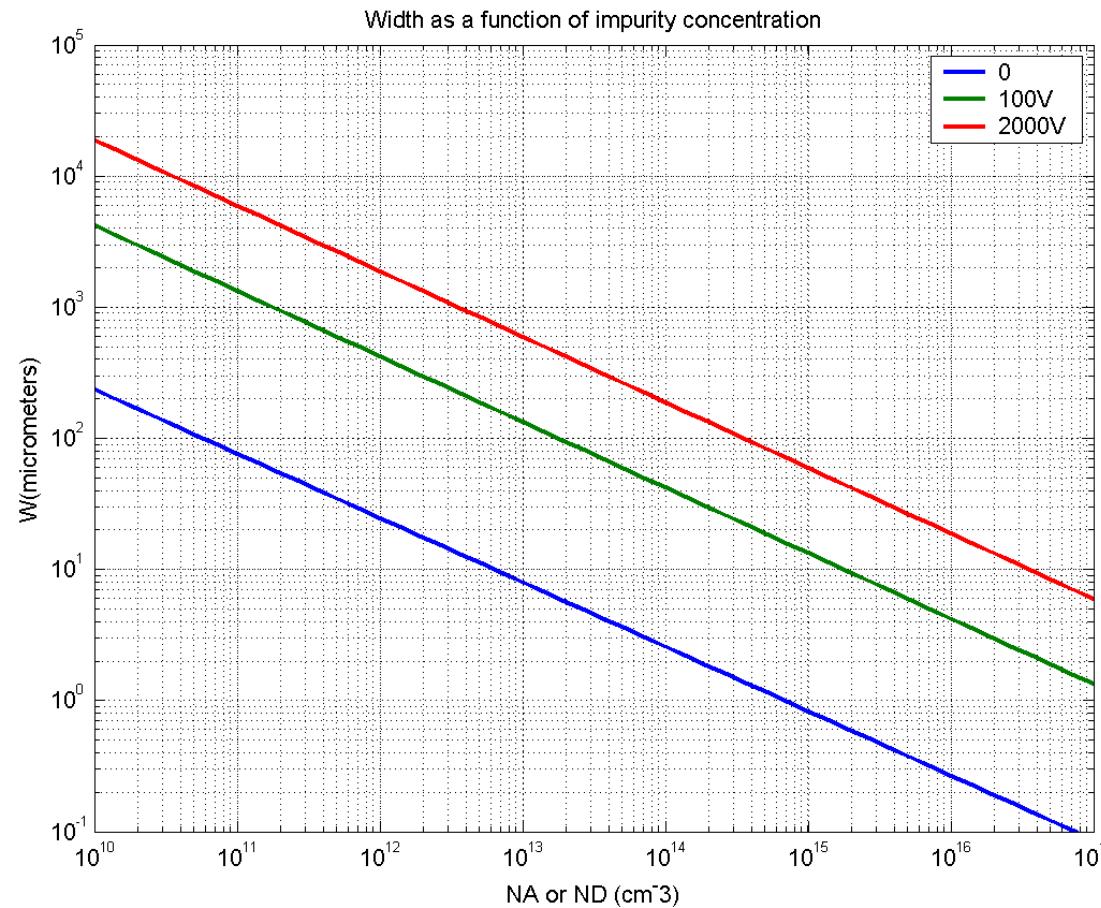


Fig. 26 Critical field at breakdown versus background doping for Si and GaAs one-sided abrupt junctions.⁵

N_B : background doping of the lightly doped side of the junction, α : ionisation rate

- Condition for breakdown:
$$\frac{I_n(W) - I_n(0)}{I} = \int_0^W \alpha dx = 1 \quad I_n: \text{electron current}$$

Break-through voltage



- Depletion zone width as a function of dopant concentration

3.1 The p-n junction

Heterojunctions

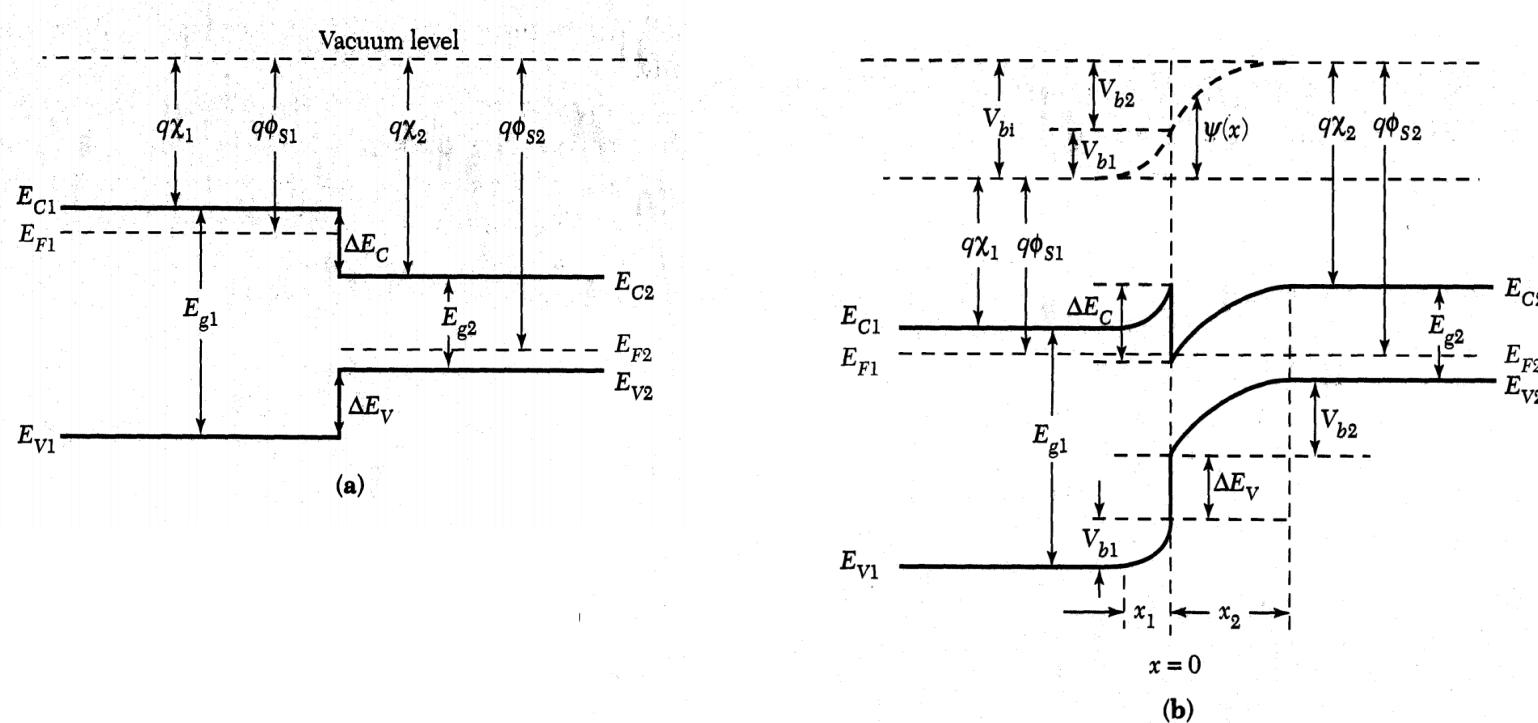


Fig. 32 (a) Energy band diagram of two isolated semiconductors. (b) Energy band diagram of an ideal $n-p$ heterojunction at thermal equilibrium.

Heterojunctions important for devices using confinement effects such as laser diodes