

Channel Estimation Techniques for Multicarrier OFDM 5G Wireless Communication Systems

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Abstract

- We will discuss about channel estimation techniques in multicarrier orthogonal frequency-division-multiplexing (OFDM) 5G wireless communication systems in Rayleigh and Rician channels.
- Proposed M-estimator based channel estimation technique in comparison with classical least squares (LS) and linear minimum mean-squared error (LMMSE) estimation is studied and analyzed.
- We will compare the proposed technique over LS and LMMSE through simulations .

System Model

- OFDM is one of the efficient modulation formats used in present 5G wireless communication systems.

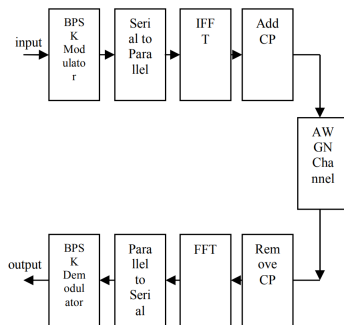


Figure: Block diagram of OFDM system with FFT

System Model

The received signal (at the receiver) may be described by

$$Y = HX + W \quad (1)$$

where

$$H = [H[0], H[1], \dots, H[N-1]]^T \quad (2)$$

$$W = [W[0], W[1], \dots, W[N-1]]^T \quad (3)$$

$$X = \begin{bmatrix} X[0] & 0 & \dots & 0 \\ 0 & X[1] & & \vdots \\ \vdots & & \ddots & 0 \\ 0 & \dots & 0 & X[N-1] \end{bmatrix} \quad (4)$$

M-Estimator

consider a signal model

$$r_j = \sum_{k=1}^K s_j^k \theta_k + n_j, \quad j = 1, 2, \dots, N \quad (5)$$

or in matrix notation

$$\underline{r} = \underline{S}\underline{\theta} + \underline{n} \quad (6)$$

where

$$\underline{S} \triangleq [\underline{s}_1, \underline{s}_2, \dots, \underline{s}_K] \quad (7)$$

$$\underline{\theta} \triangleq [\theta_1, \theta_2, \dots, \theta_K]^T \quad (8)$$

M-Estimator

Huber proposed to minimize a sum of a less rapidly increasing function ρ of the residuals

$$\hat{\underline{\theta}} = \arg \min_{\underline{\theta} \in \mathcal{R}^K} \sum_{j=1}^N \rho \left(r_j - \sum_{k=1}^K s_j^k \theta_k \right) \quad (9)$$

suppose that ρ has a derivative $\psi = \rho'$; then , the solution to (9) satisfies the equation

$$\sum_{j=1}^N \psi \left(r_j - \sum_{l=1}^K s_j^l \theta_l \right) s_j^k = 0, \quad k = 1, 2, \dots, K \quad (10)$$

or in vector form

$$S^T \psi(\underline{r} - \underline{S}\underline{\theta}) = \underline{0}_K \quad (11)$$

$$\text{where } \psi(\underline{x}) \triangleq [\psi(x_1), \dots, \psi(x_K)]^T \text{ for any } \underline{x} \in \mathcal{R}^K \quad (12)$$

M-Estimator

Assume that the penalty function $\rho(x)$ in (9) has a bounded second-order derivative i.e., $|\rho''(x)| = |\psi'(x)| \leq \mu$ for some $\mu > 0$. Then (10) can be solved iteratively by the following modified residual method . Let $\underline{\theta}^t$ be the estimated at the t th step; then it is updated according to

$$\underline{z}^t \triangleq \psi(\underline{r} - \underline{S}\underline{\theta}^t) \quad (13)$$

$$\underline{\theta}^{t+1} = \underline{\theta}^t + \frac{1}{\mu} \left(\underline{S}^T \underline{S} \right)^{-1} \underline{S}^T \underline{z}^t \quad (14)$$

where μ is a step-size parameter . $\left(\mu = \frac{1}{N} \sum_{j=1}^N \psi' \left(r_j - \sum_{k=1}^K s_j^k \theta_k^t \right) \right)$

For initial estimate $\underline{\theta}^0$, we can take

$$\underline{\theta}^0 = \frac{1}{\mu} \left(\underline{S}^T \underline{S} \right)^{-1} \underline{S}^T \underline{r}. \quad (15)$$

The iteration is stopped if $\|\underline{\theta}^t - \underline{\theta}^{t-1}\| \leq \epsilon$ for some small number ϵ .

Proposed M-Estimator

Penalty, influence and weight functions of the proposed M-estimator for channel estimation in 5G multicarrier wireless communications in Rayleigh and Rician fading channels are given by

$$\rho_{PROPOSED}(x) = \begin{cases} \frac{x^2}{2} & |x| \leq a \\ a^2 - a|x| & a < |x| \leq b \\ \frac{-ab}{2} \exp\left(1 - \frac{x^2}{b^2}\right) + d & |x| > b \end{cases} \quad (16)$$

$$\psi_{PROPOSED}(x) = \begin{cases} x & |x| \leq a \\ asgn(x) & a < |x| \leq b \\ \frac{a}{b}x \exp\left(1 - \frac{x^2}{b^2}\right) & |x| > b \end{cases} \quad (17)$$

Proposed M-Estimator

$$w_{PROPOSED}(x) = \begin{cases} 1 & |x| \leq a \\ \frac{asgn(x)}{x} & a < |x| \leq b \\ \frac{a}{b} \exp\left(1 - \frac{x^2}{b^2}\right) & |x| > b \end{cases} \quad (18)$$

where a and b are any constants and x is any data. From an influence function, robustness measures are derived and $a(= kv^2)$ and $b(= 2kv^2)$ are selected (where k is any constant).

where $\rho(x)$ is penalty function, $\psi(x) = \frac{d\rho(x)}{dx}$ is the influence function and $w(x) = \frac{\psi(x)}{x}$ is the weight function.

Proposed M-Estimator

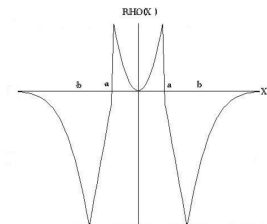


Figure: Proposed M-estimator penalty function.

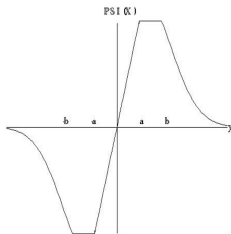


Figure: Proposed M-estimator influence function

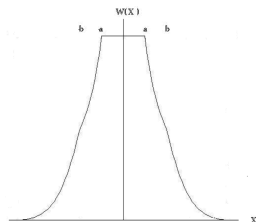


Figure: Proposed M-estimator weight function

Channel Estimation

LS channel estimation

$$\hat{H}_{LS} = \left(X^H X\right)^{-1} X^H Y = X^{-1} Y \quad (19)$$

where estimate of H is \hat{H} and X^H is the Hermitian matrix of X

MMSE channel estimation

$$\hat{H}_{MMSE} = R_{HH} \left[R_{HH} + \left(X X^H \right)^{-1} W v^2 \right]^{-1} \hat{H}_{LS} \quad (20)$$

where R_{HH} is the matrix form of auto-covariance of X and v^2 is variance of noise vector.

proposed estimation

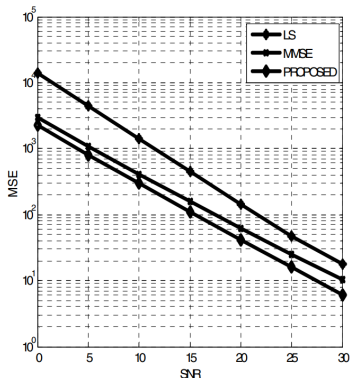
Proposed technique estimates channel parameters for 5G multicarrier wireless communications in Rayleigh and Rician fading channels using

$$H^{t+1} = H^t + \frac{1}{\mu} \left(X^H X \right)^{-1} X^H \psi(Y - HX) \quad (21)$$

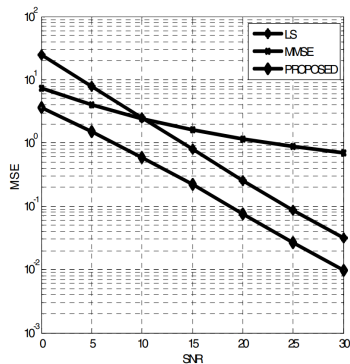
for some $\mu > 0$ where $H^0 = \frac{1}{\mu} \hat{H}_{LS}$

SIMULATION RESULTS

In this simulation , the comparison shows the performance gains achieved by the proposed technique over LS and MMSE estimation in multicarrier OFDM 5G wireless communication systems



(a)



(b)

Figure: MSE vs. SNR Performance of LS, MMSE and Proposed channel estimators in (a) Rayleigh and (b) Rician fading

CONCLUSION

- M-estimator based channel estimation technique for 5G multicarrier OFDM wireless communication systems in Rayleigh and Rician fading channels is analyzed in this paper.
- Observations from simulation results imply that the proposed M-estimator based channel estimation technique for 5G multicarrier OFDM wireless communication systems offers better performance than LS and MMSE techniques in Rayleigh and Rician fading channels.