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## Brushless motor

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## 3 Phase 250kW motor controller

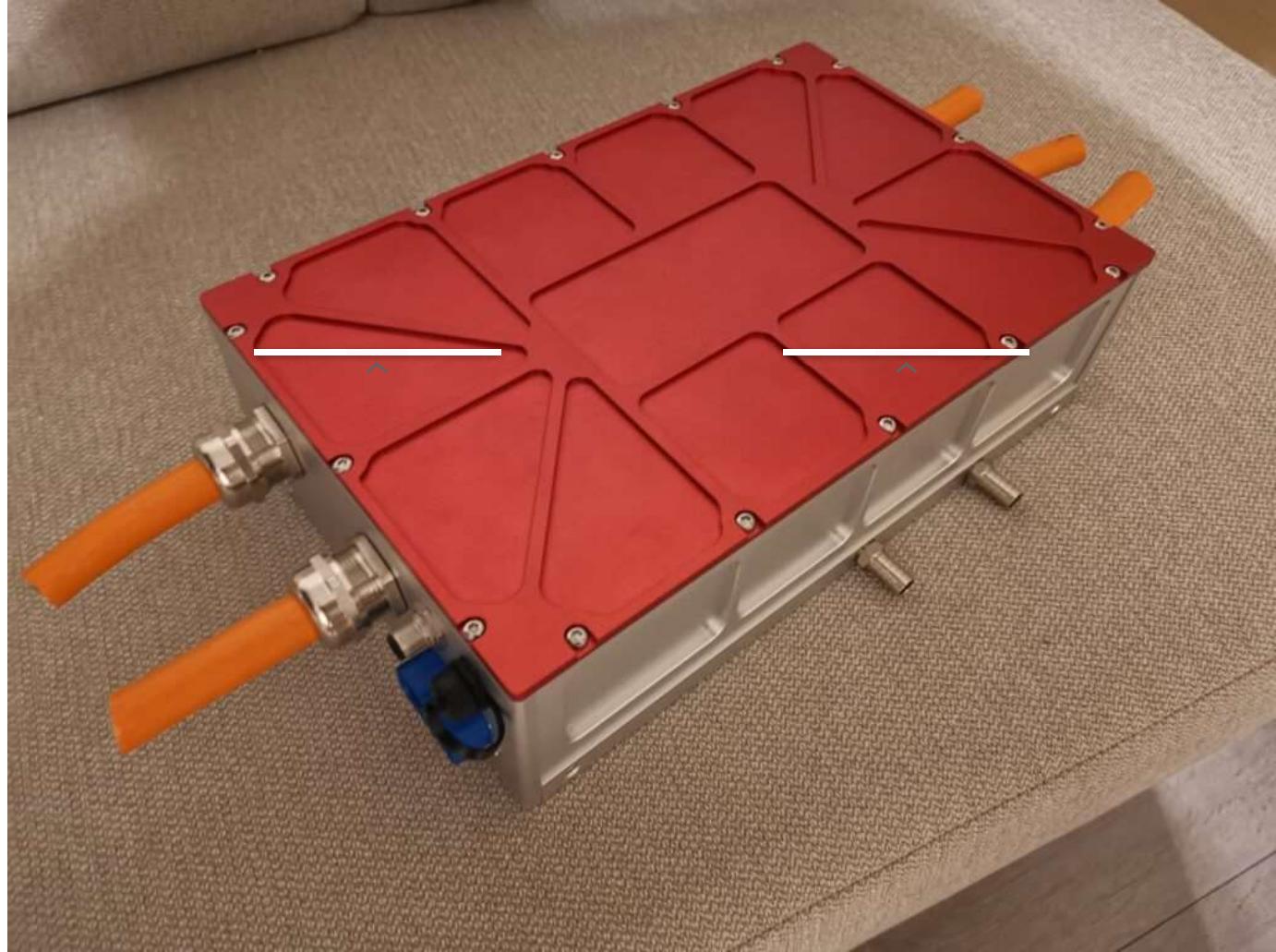
Posted: 10th May 2020 by [iulian207](#) in [Projects](#)

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Very lightweight design, water-cooling.

Specifications:

- Continuous power: 250kW
- Nominal Voltage: 650Vdc
- Peak Current: 420Arms
- Continuous output current: 385Arms
- Max Efficiency >98%
- Weight ~ 11kg
- SD card black box for controller data analysis, via CAN dongle
- Wi-fi telemetry via CAN dongle
- ✓ All settings can be user customized. Hardware and software can be also customized according to the client needs.

Generates maximum torque per amp across the full range of current *and* temperature by modifying the proportional and integral gains every cycle to match the direct and quadrature inductances as well as stator resistance in real time.

We are in the beta testing phase of our controller design. It is performing well and we can't wait to release it. But we won't release a fully tested product, so we ask for your patience. This isn't the first high powered controller we have made, but it is the first with a Joule Motors badge on it.

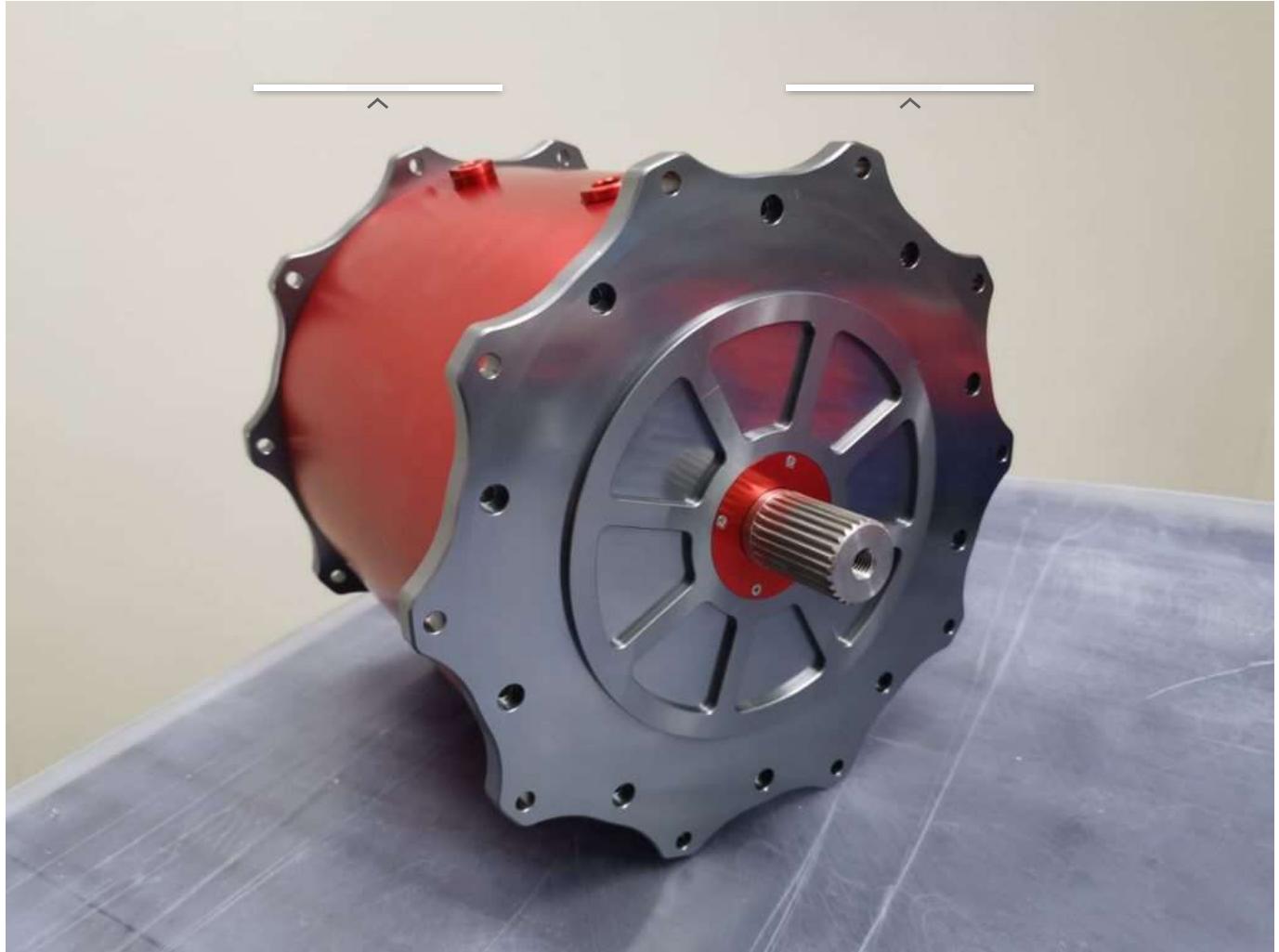
Testing motor controller with added Resolver function

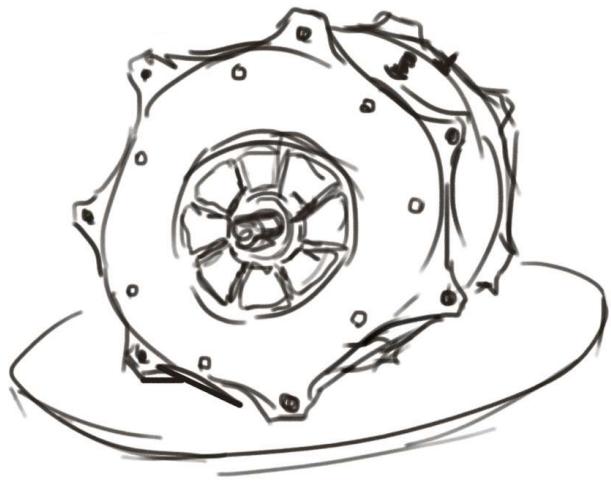


## [350kW 650V High performance Permanent magnet Synchronous motor 98% efficiency measured](#)

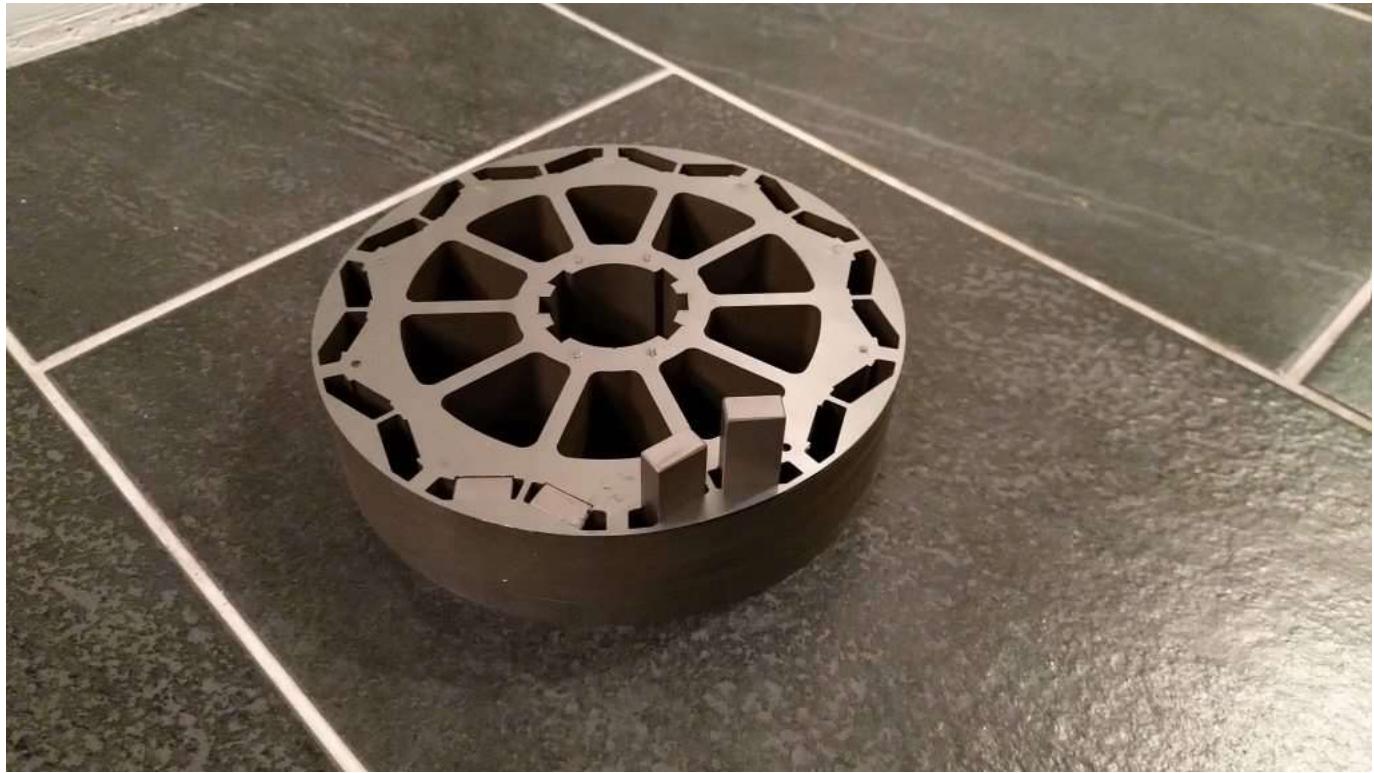
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Tags: [High efficiency Permanent magnet motor](#), [High performance 200kW motor](#), [high performance 300kW motor](#), [high performance 350kW motor](#), [PMSM motor 200kW](#), [14](#)

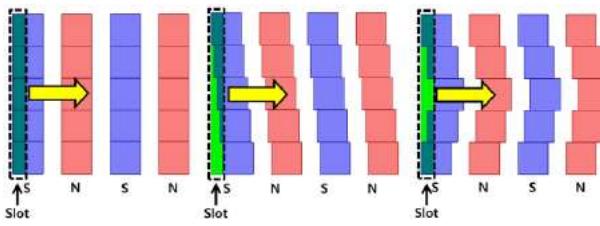




Motors with Next level of efficiency 98%



Rotor slice with custom magnet shape and grade N45UH



Magnets skewing is implemented for a perfect sine-wave Back EMF, and the best efficiency possible.  
The magnet optimum skewing angle formula is :

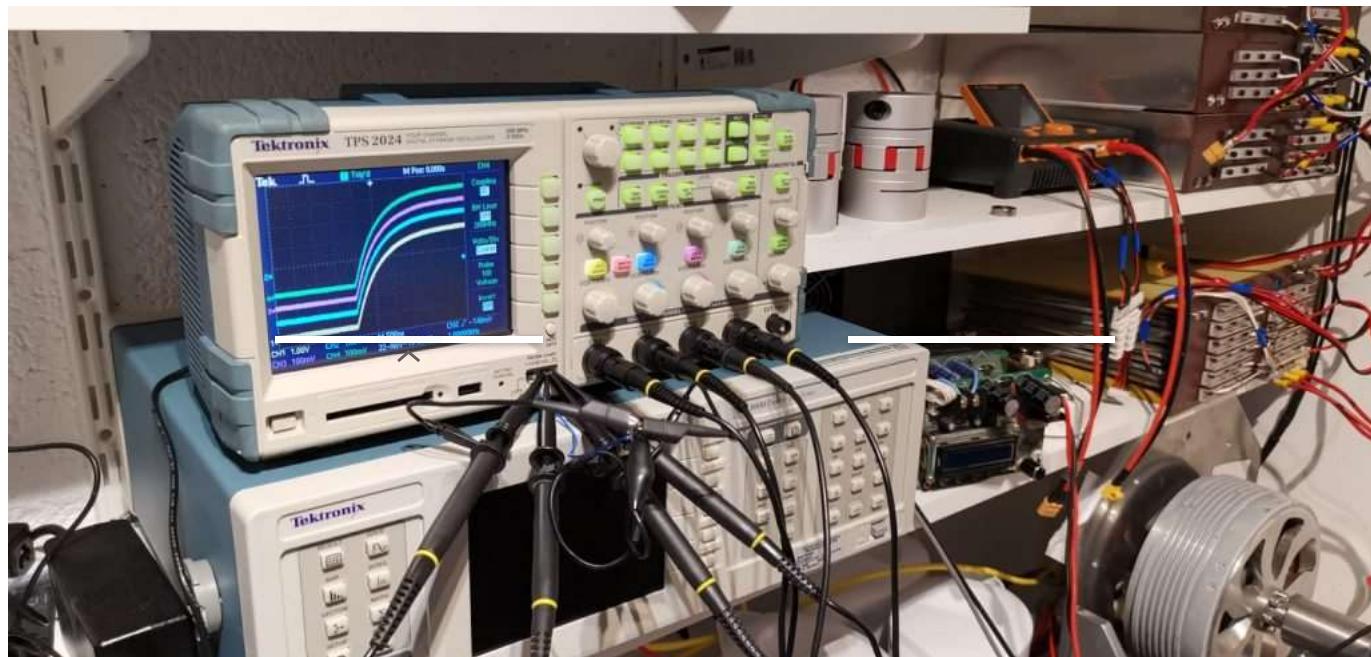
$$\Delta\alpha = \frac{360}{2P} - \frac{360}{Z}$$

$\Delta\alpha$  = optimum magnet skewing angle

P= nr of magnets

Z= nr. of slots.

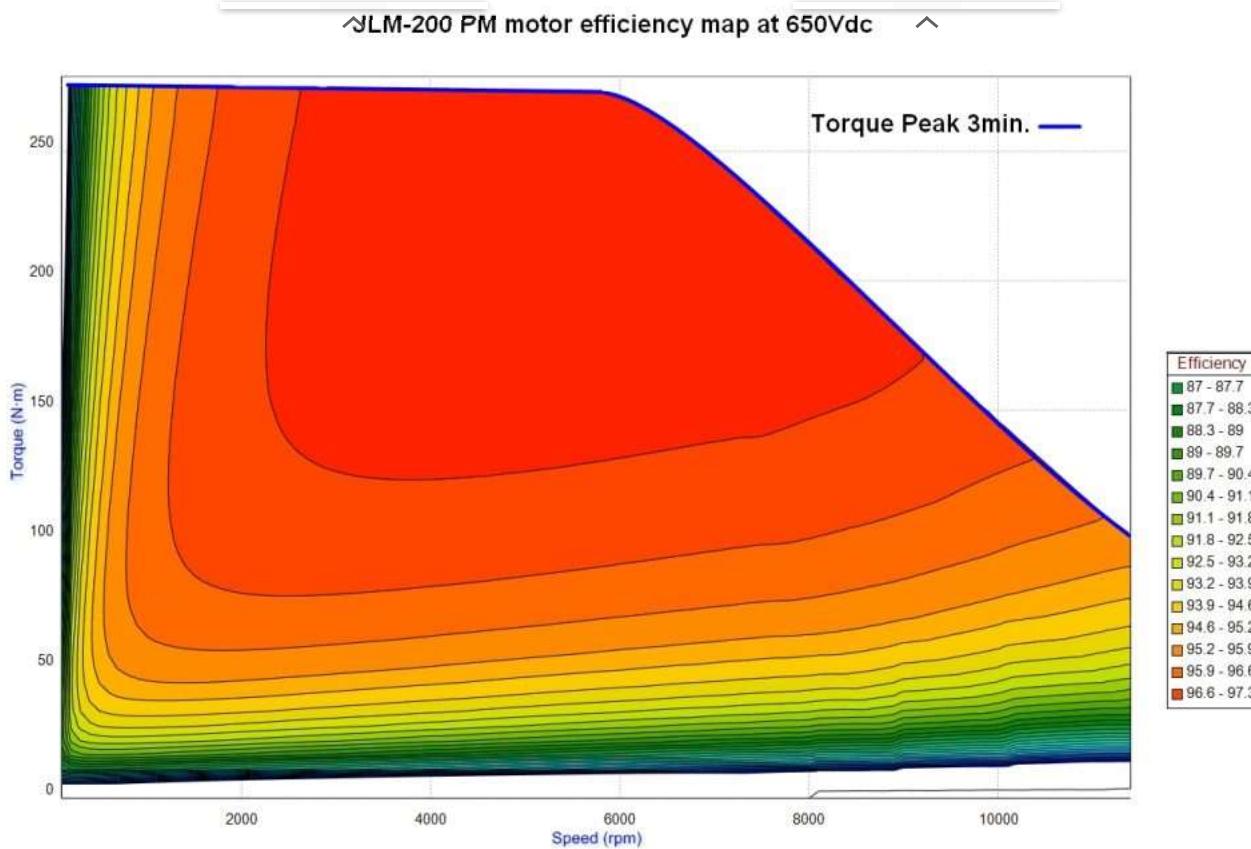
Example for my motor 60 slot 10 pole:  $\Delta\alpha = (360/20)-(360/60) = 18-6 = 6$  degree skewing is ideal for zero cogging torque.



New arrived tool with isolated inputs to allow to measure switching losses for each transistor. This is very important, I'm using it to fine tune the controller switching timing, and measure the switching losses.

	A	B	C	D	E	F	G	H	I	J
1	Tektronix PWRVIEW measurement snap-shot									
2	Created	Tuesday	24 November	20:52:25						
3	'PA3000(0'S/N C010122'	Ver.3.1.1								
4	Meas	'PA3000(0122)	'PA3000(0122)	'PA3000(0122)'PA3000(0122) 1 Sum'	'PA3000(0122) 4'	Aux	'PA3000(0122) Aux Values'	Formula		
5	Vrms	130.22 V	129.14 V	131.05 V	225.52 V	Torque	121.20 Nm			
6	Arms	106.12 A	106.49 A	105.74 A	106.10 A					
7	Watts	11.330 kW	11.356 kW	11.328 kW	34.020 kW	34.246 kW			InvertorEfficiency(%)	
8	VA	13.819 kVA	13.752 kVA	13.856 kVA	41.445 kVA					99.34
9	Freq	218.10 Hz	217.75 Hz	217.75 Hz						
10	PF	819.85 m	825.77 m	817.50 m	820.86 m					
11	Vdc				284.42 V					
12	Adc				120.41 A		Speed			
13	Vrmn	124.45 V	123.42 V	125.11 V						2613.00
14	Vh1m	108.59 V	108.45 V	108.33 V						
15	Vh1p	0.0000 Å°	-118.67 Å°	120.27 Å°			ShaftPower			
16	Vh2m	649.47 mV	637.18 mV	1.0762 V						33166.00
17	Vh2p	168.74 Å°	60.638 Å°	-49.231 Å°						
18	Vh3m	2.5005 V	2.5815 V	2.7940 V			MotorEfficiency			
19	Vh3p	61.050 Å°	59.447 Å°	38.911 Å°						97.49
20	Vh4m	254.99 mV	303.10 mV	199.68 mV						
21	Vh4p	86.026 Å°	-28.752 Å°	-132.60 Å°			Torqueper/Amp			
22	Vh5m	3.0810 V	2.2200 V	2.8037 V						1.15
23	Vh5p	179.43 Å°	-57.300 Å°	44.717 Å°						
24	Vh6m	1.3110 V	1.5255 V	1.4435 V			MortorElectricalPower			
25	Vh6p	136.03 Å°	146.97 Å°	145.80 Å°						34020.00

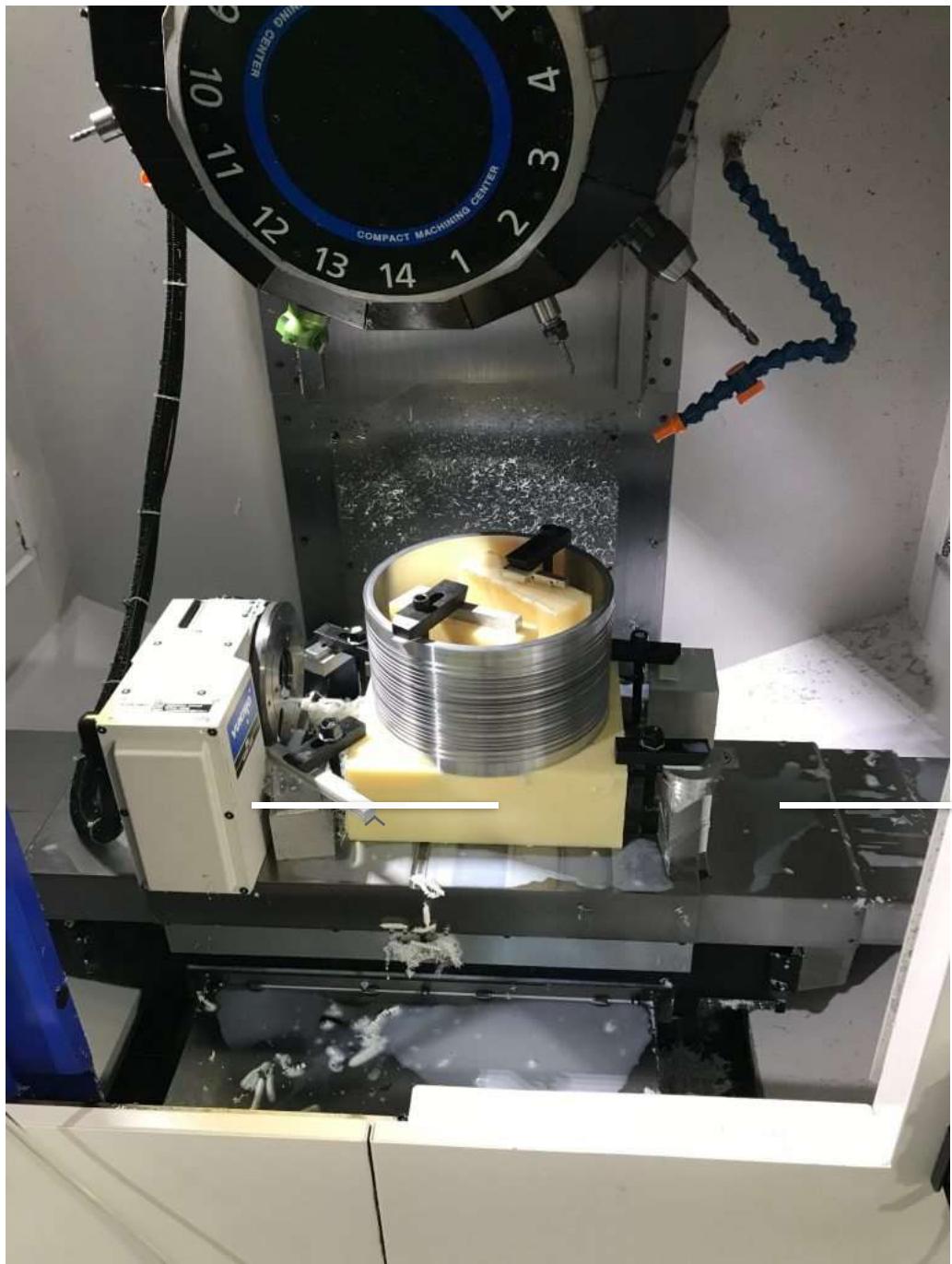
Measured motor efficiency matches within 0.2% of the simulations value



Torque Vs Speed combined with efficiency map  
I use 0.2mm thick lamination to achieve highest efficiency on the market, above 97,3%.



Motor housing ready first 4 motors



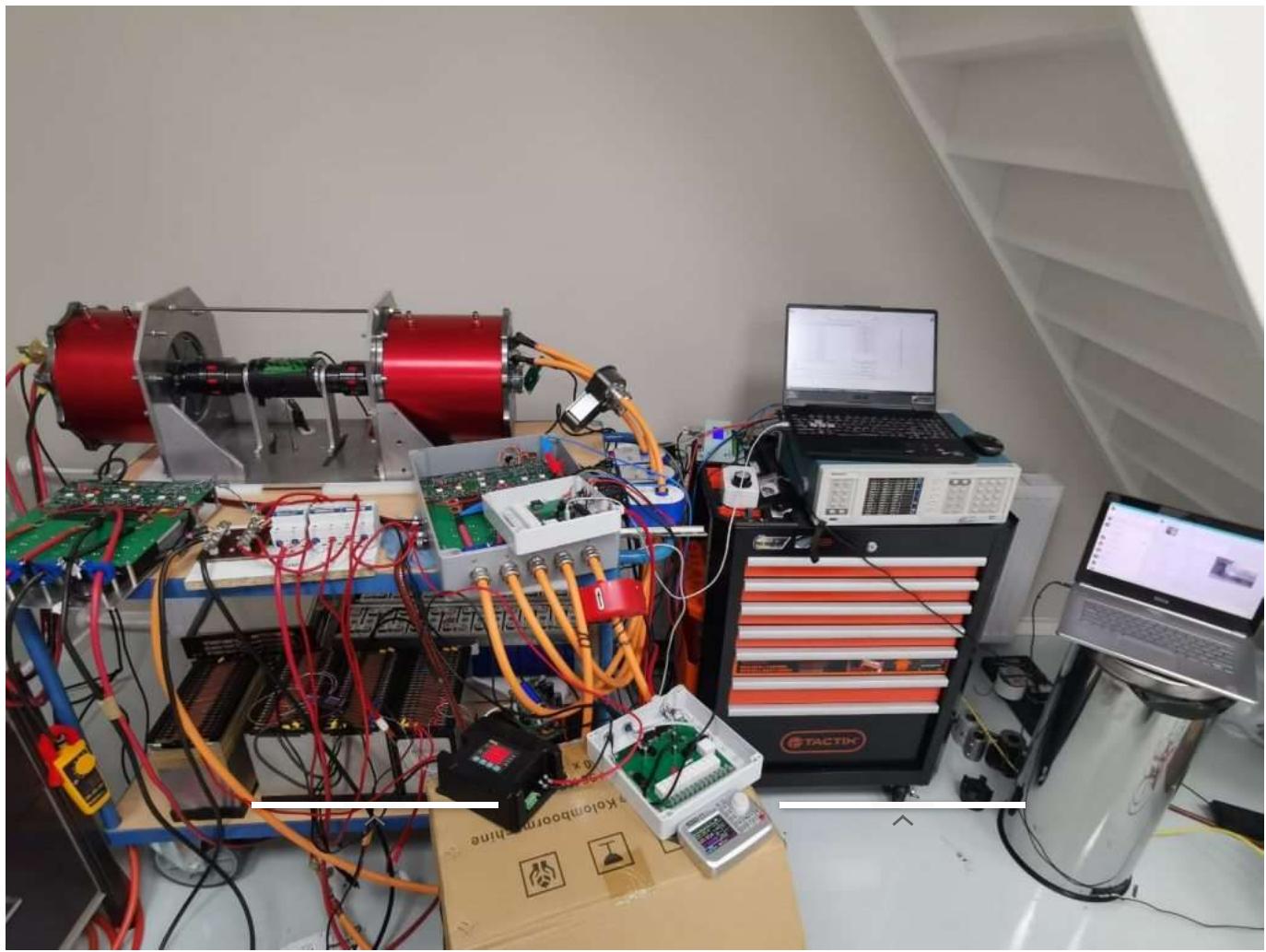
Inner cooling ring being machined

Balancing a motor rotor with an oscilloscope





Rotor flanges



200kW motor +130kW motor on the test bench



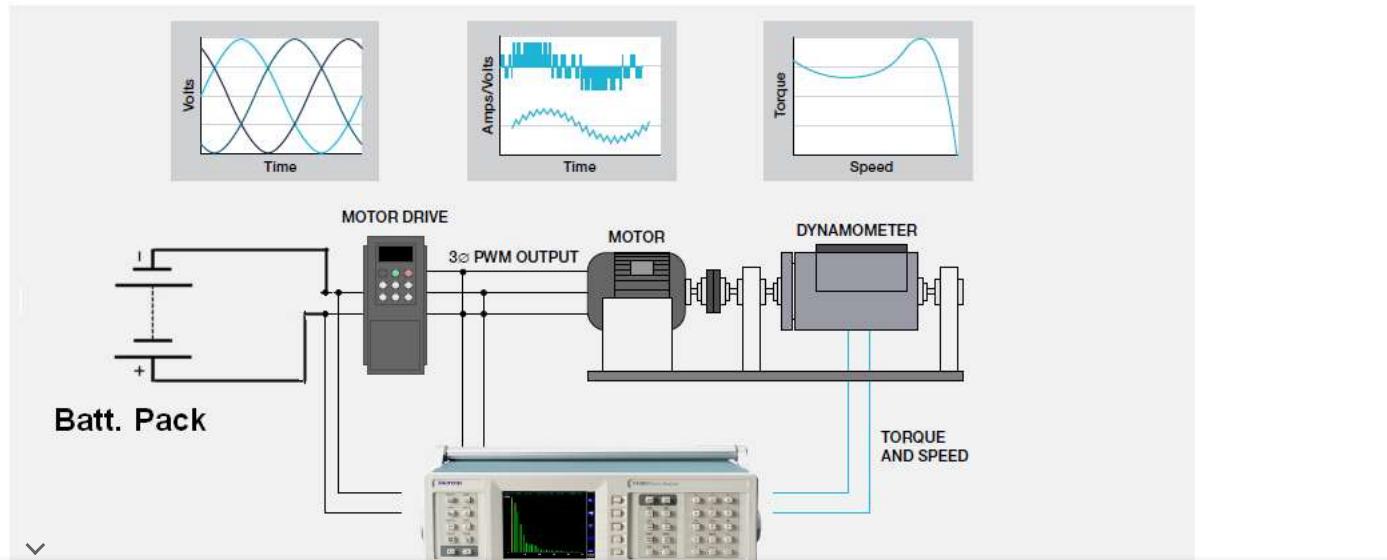
## >97% Motor efficiency measured. Motor/ controller performance evaluation, test bench setup.

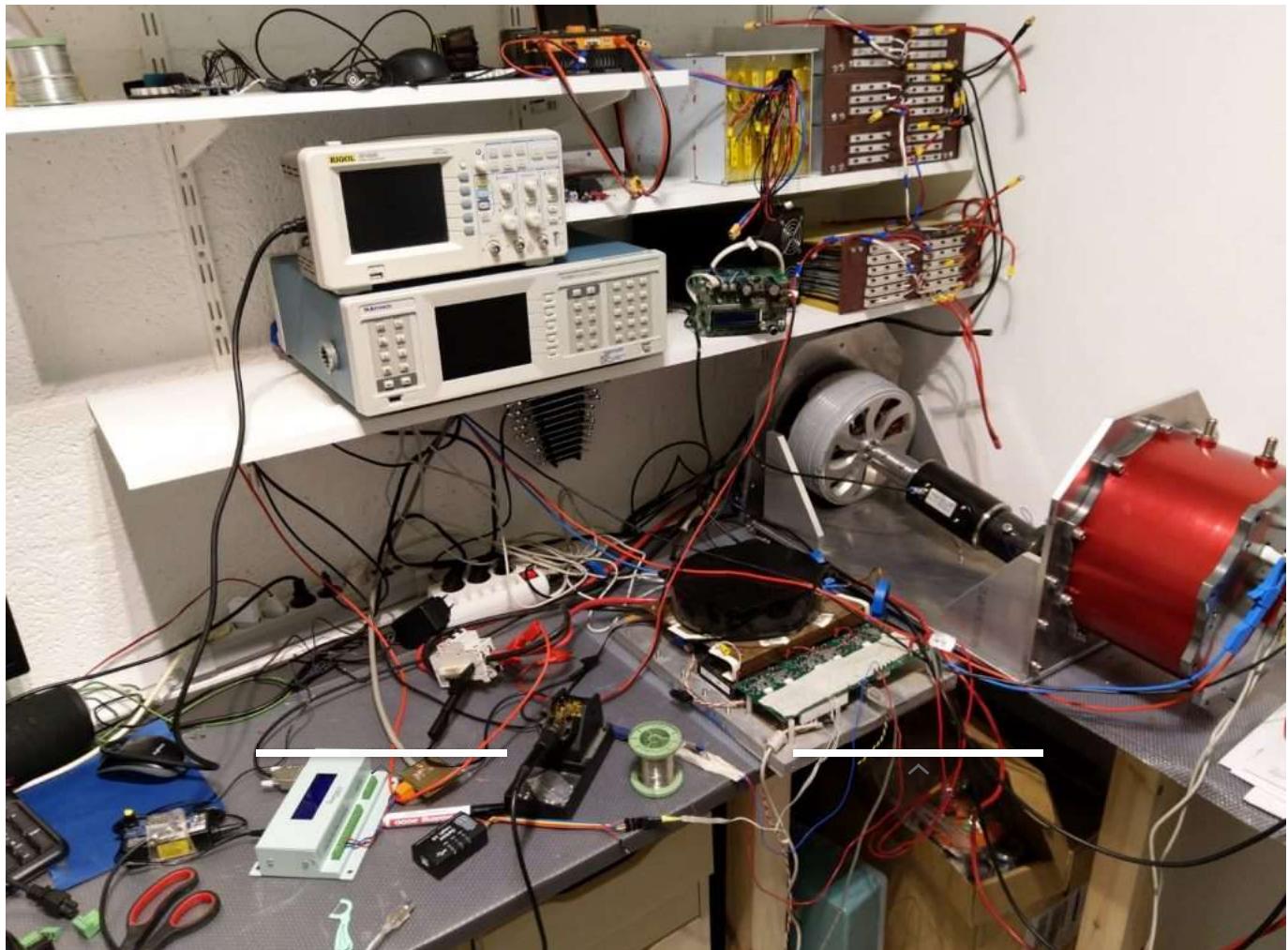
Posted: 5th July 2019 by iulian207 in [Projects](#)

Tags: [motor efficiency measurement](#), [tektronix pa3000 measurements](#), [Tektronix pa3000 motor performance evaluation](#)

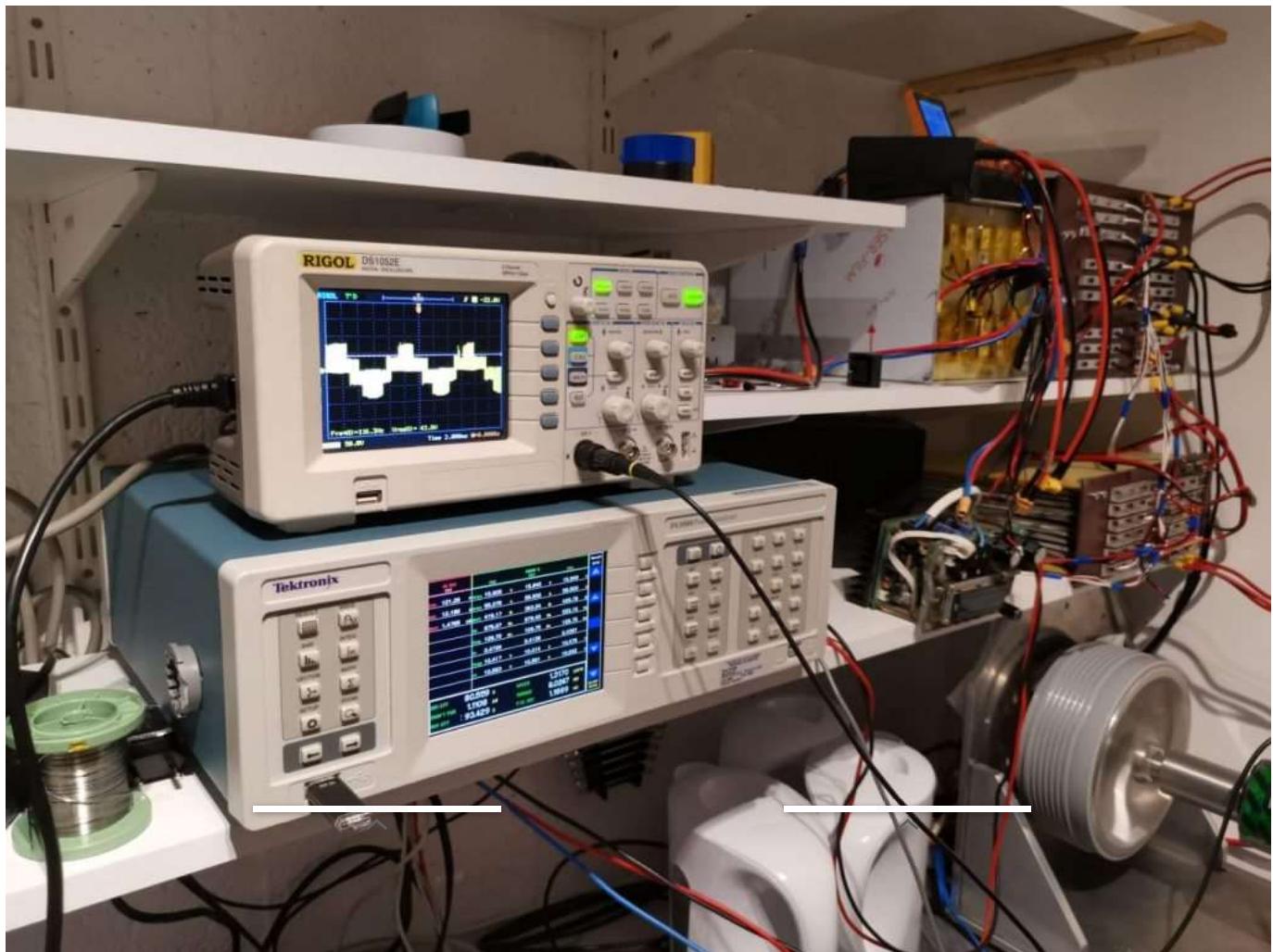
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Efficiency analysis between an IGBT controller and a mosfet controller

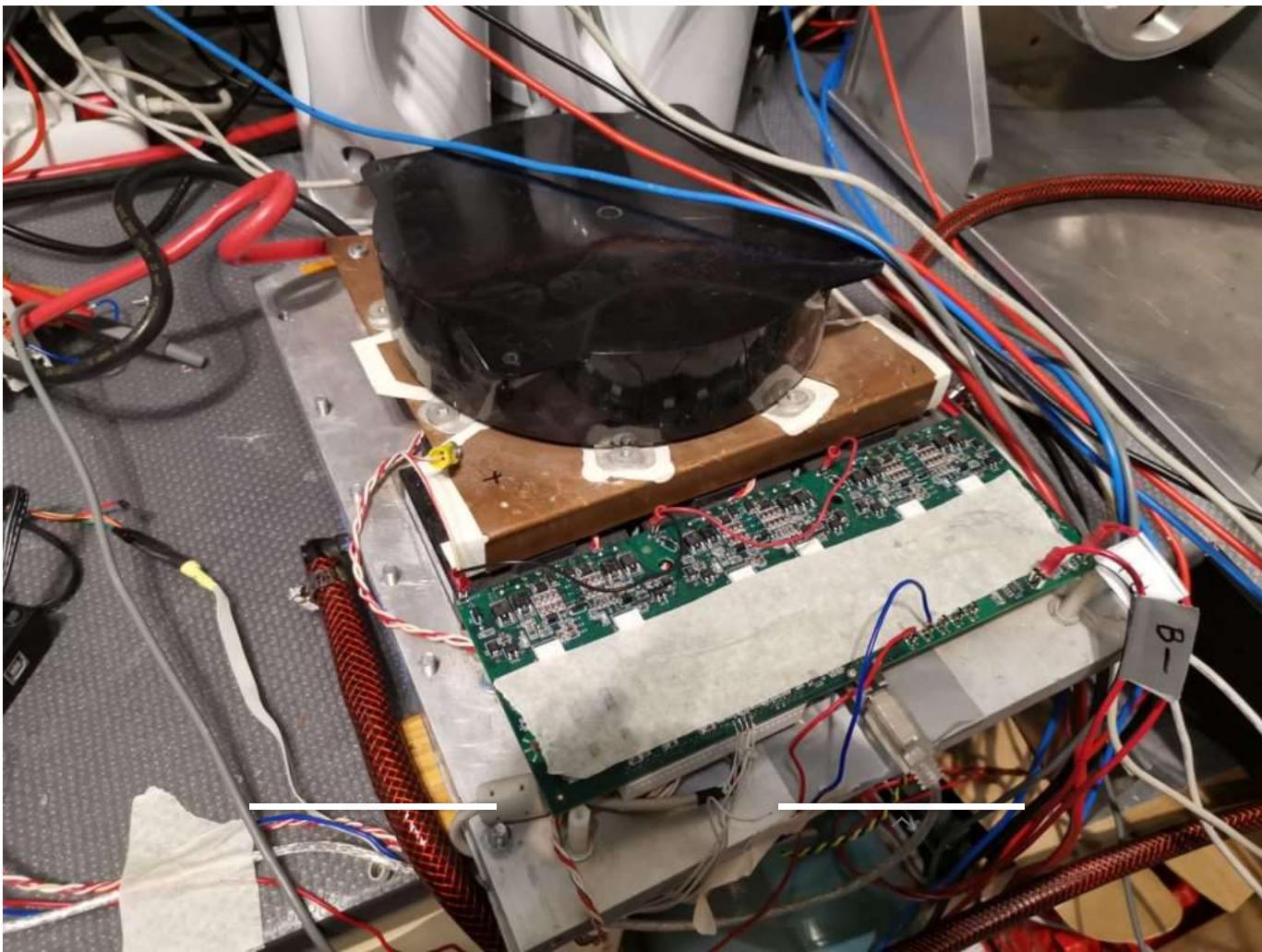




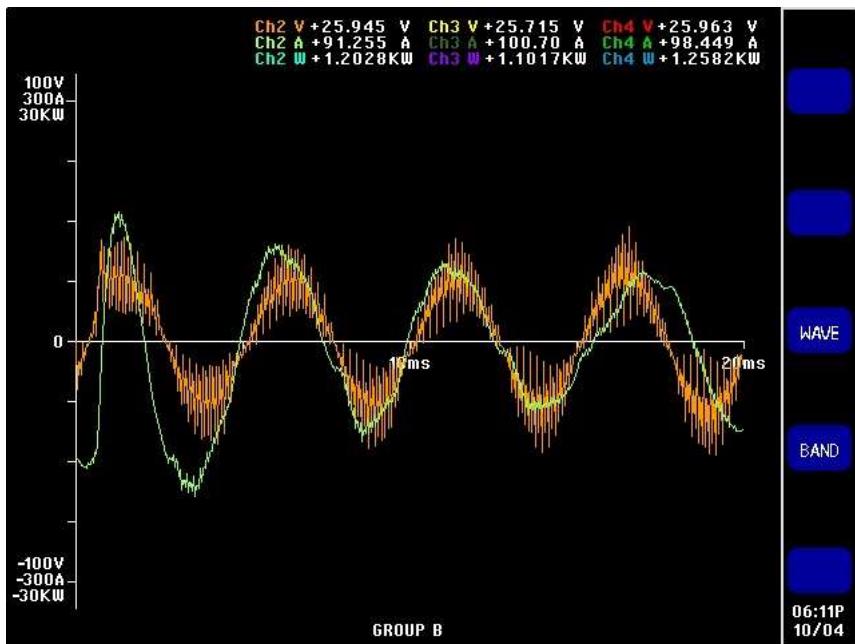
My small setup for testing motors and controllers, I have a very precise torque sensor with 0,1% accuracy  
Range is 0-500Nm and up to 8000 RPM



▼



Paul Open source controller test results, using water cooling plate and 400AMP IGBT modules. Encoder angle was fine tuned until i got the best performance. Advanced angle Id current was adjusted but without any practical improvement.



Orange wave is phase voltage, and green line represents the current.  
Using Paul Open source controller

DC BUS Ch1		GROUP B Ch2 Ch3 Ch4				Result 7578		
Vdc	45.463	Vrms	15.447	V	15.462	V	15.683	V
Adc	29.389	Arms	41.293	A	40.212	A	39.186	A
Watt	1.3360	kW	440.59	W	430.96	W	420.79	W
		VA	637.86	VA	621.75	VA	614.53	VA
		Freq	110.38	Hz	110.38	Hz	110.38	Hz
		PF	0.6907		0.6931		0.6847	
		Vrmn	11.034	V	11.048	V	11.203	V
		Vf	10.555	V	10.601	V	10.600	V
INV EFF	96.731	%	SPEED	1.3245	KRPM			
SHAFT PWR	1.2446	kW	TORQUE	8.9703	NM			
MOT EFF	96.306	%	TORQ/AMP	222.97	m	09:10P	10/06	

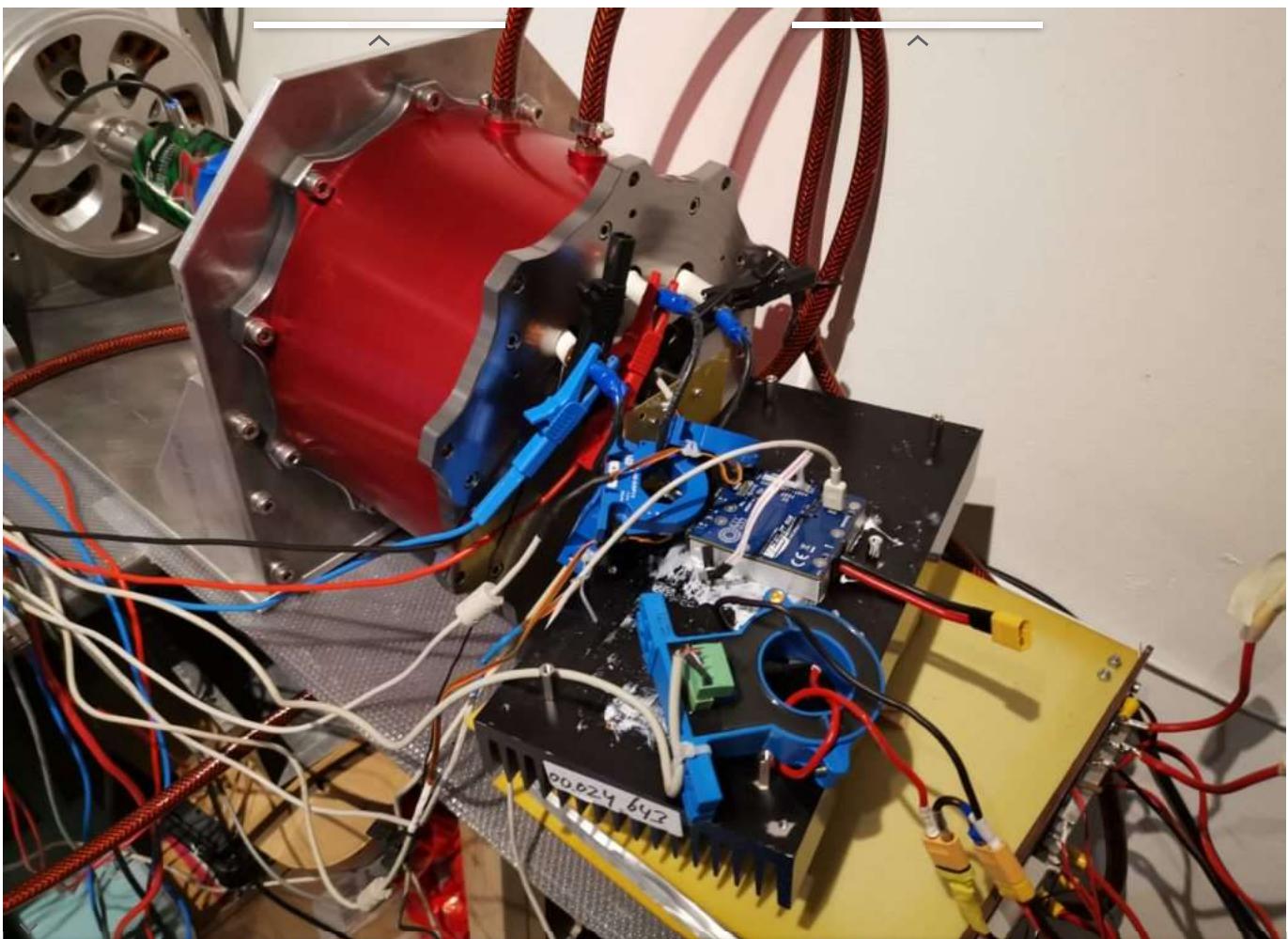
Even at low loads the efficiency parameters matched the simulation parameters in FEMM the software model.

Results using Vesc controller using in sensorless mode.

As we can observe the inverter efficiency is much better since is using mosfets instead if the IGBT's. The drive PWM frequency was set to 20khz.

The motor efficiency is also better because the sinewave generated is much closer to a pure sinewave. The torque per amp constant in this case was better from 0,14Nm/Arms to 0,22Nm/Arms

#### Open source controller used to run my RED 130 motor in sensorless mode



## Using Tektronix PA3000 to measure Vesc motor controller efficiency



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**MOTOR**

Sensor Mode	Sensorless
Motor Resistance (R)	5.4 mΩ
Motor Inductance (L)	21.69 μH
Motor Flux Linkage (λ)	20.967 mWb
Current Kp	0.0217
Current Ki	5.40
Observer Gain (x1M)	2.27

**APP**

**Detect and Calculate Parameters**

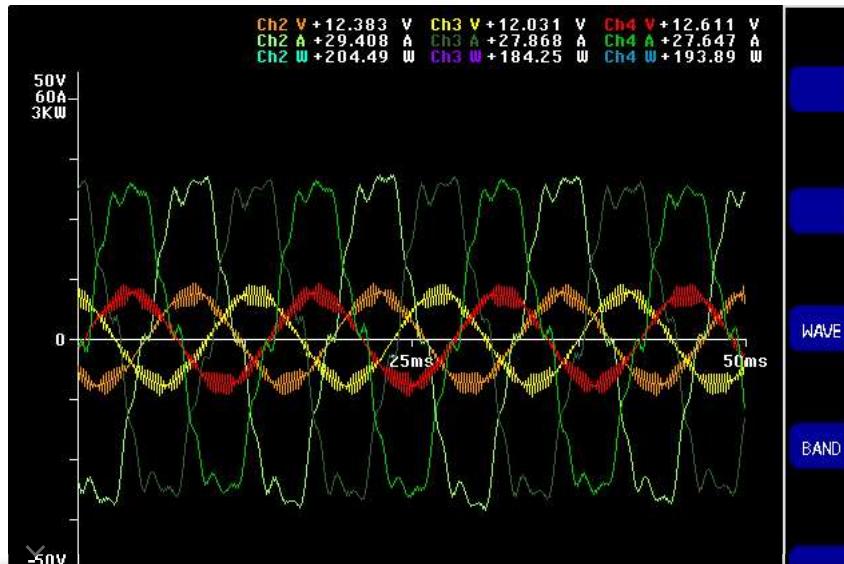
I: 60.00 A	D: 0.30	ω: 1000.0 ERPM/s
R: 5.40 mΩ	L: 21.69 μH	λ: 20.967 mWb
T: 1000.0 μS	Kp: 0.0217	Ki: 5.40
Observer Gain (x1M): 2.27	Calc Apply Old	

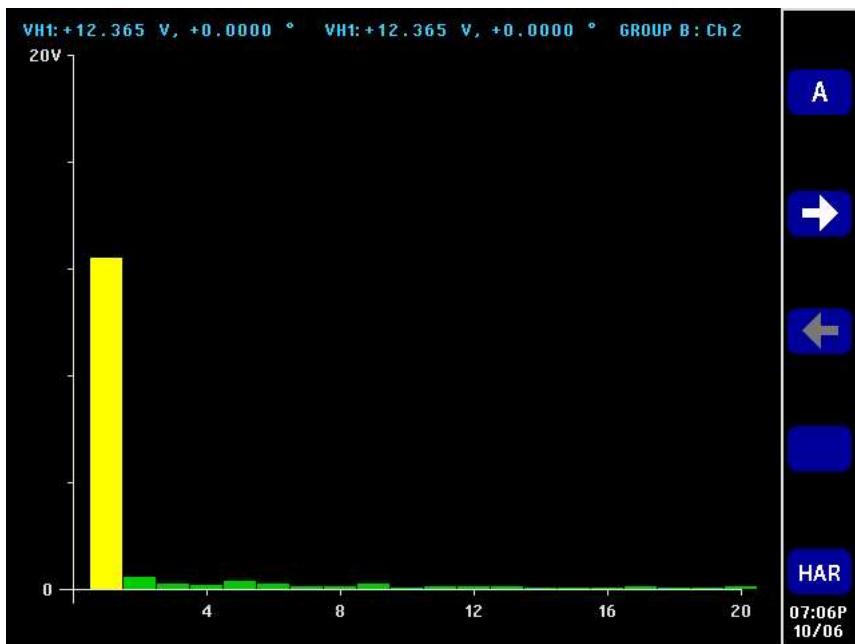
Duty: -0.1 %

Current: 0.00 A

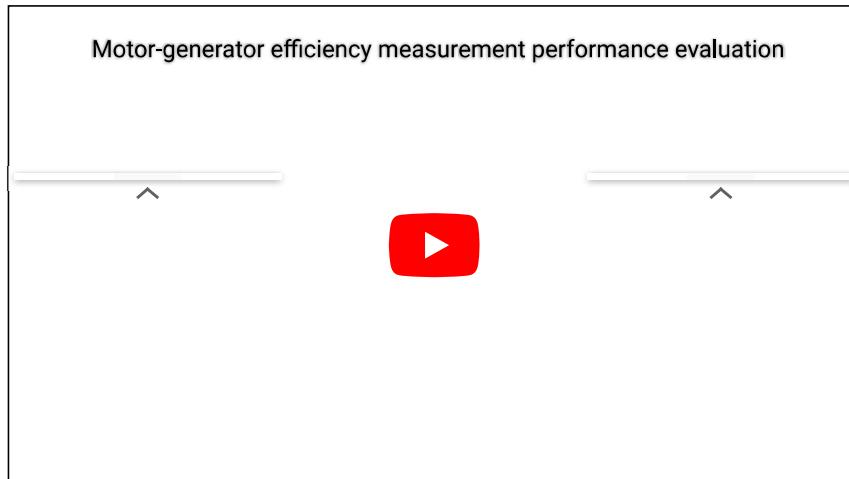
Not connected

The controller has automatically detected all the motor parameters including: Internal resistance, Inductance, Flux Linkage, Kp, Ki and observer gain





Voltage Harmonics. As we can observe due to very careful slot pole design the third and the fifth harmonic are very small, negligible



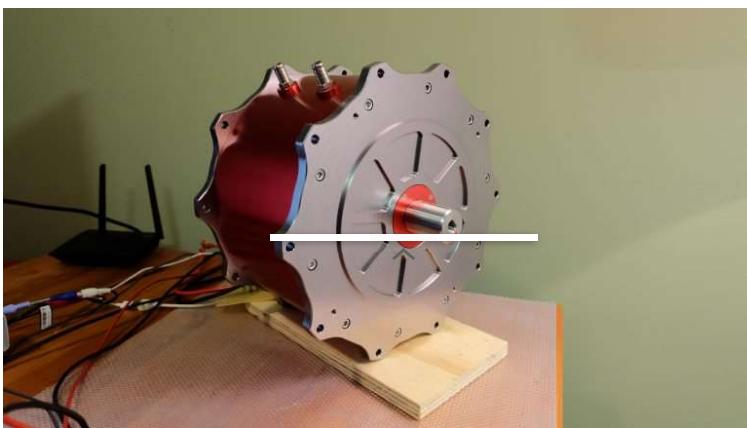
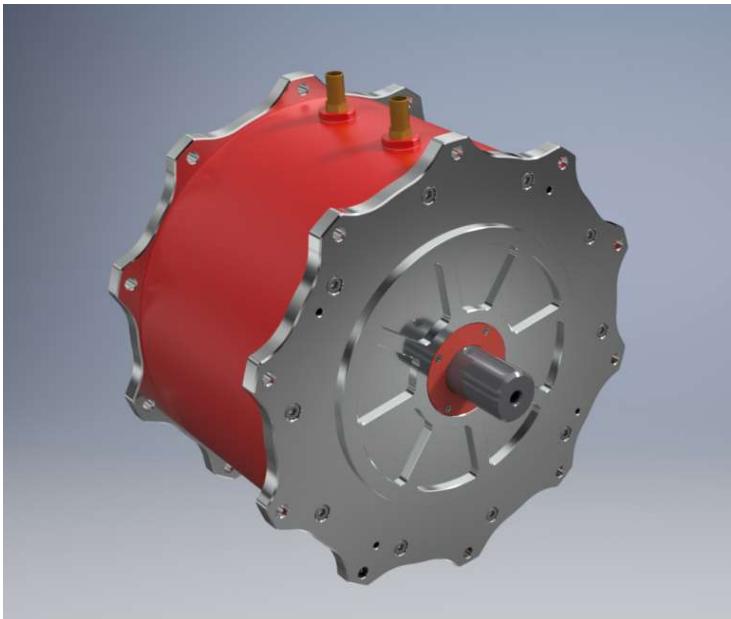
To be continued...

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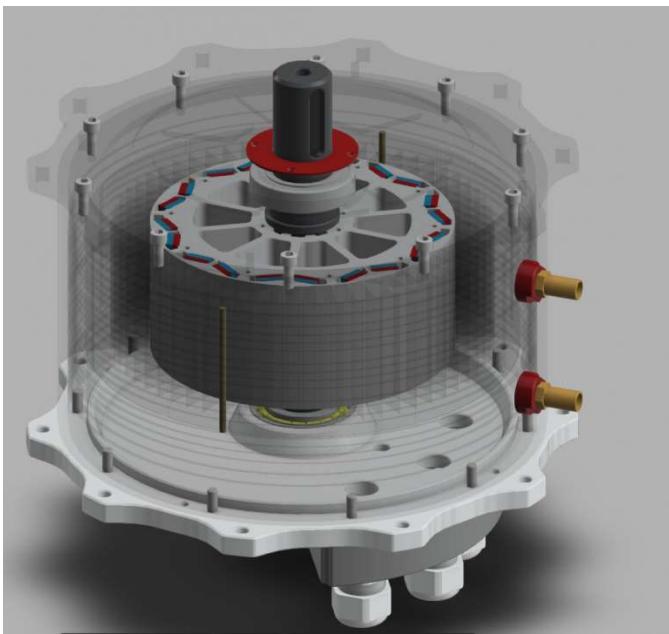
## [High Efficiency Brushless Motor PMSM 130kW 2018, 97% efficiency.](#)

Posted: 8th April 2018 by [iulian207](#) in [Projects](#)  
 Tags: [Open source BLDC motor](#), [open source brushless motor](#), [open source PMSM motor](#)  
 89

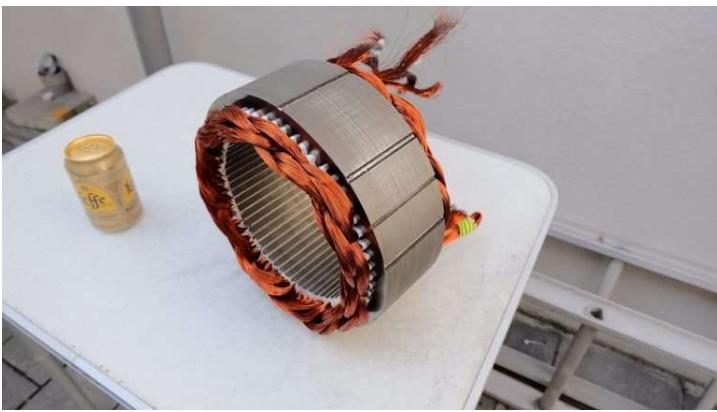
Prototype production phase done. Testing phase starting. Production phase started.



^



v



## 1. Introduction:

First BLDC motor was developed in 1962 by T.G. Wilson and P.H. Trickey unveiled what they called “a DC machine with solid state commutation”, it was basically because solid state thyristors appeared on the scene.

A permanent-magnet synchronous motor (PMSM) uses [permanent magnets](#) embedded in the steel rotor to create a constant magnetic field. The stator carries windings connected to an AC supply to produce a rotating magnetic field. At synchronous speed the rotor poles lock to the rotating magnetic field. Permanent magnet synchronous motors are similar to [brushless DC motors](#).

The difference between BLDC and PMSM is that BLDC is driven by square waveform and PMSM is drive with a sine wave current.  
These motors require a variable-frequency motor inverter, ESC, Frequency inverter.

With a strong background from Polytechnic University, Faculty of Electrical Engineering, electrical machines discipline, I started 10 years ago to modify some BLDC motors, and then designing my own.

After making my own electric car conversion Opel Agila: <https://youtu.be/y1orjHGmjeo>, i got enough experience to understand what are the motor needs for an electric car.

A motor should have high range of RPM and high peak capacity to be able to keep the car in one (gear) second or third, and have 120km/h at 7000RPM for example.

Many simulations were done until the final version (after about 50 simulations with various magnets sizes, angles, dimensions, stator tooth sizes and shape, and different winding configuration).

To be able to have a very small cogging torque and a sinewave back EMF the motor is consisted in three rotor slices with 3 degrees skewing; this also reduces harmonics and eddy currents losses in the magnets. The motor will work with sensorless sinewave controller or using resolver/ encoder.

## 2. Motor sizing, slot pole count:

To decide what is the best motor for your application you need to take in account many of aspects.

- You start from the power, speed and torque needed.
- There are many type of motors good for a car conversion ( DC motors, AC motor, Permanent magnets motors (PMSM), Hybrid PMSM ( less magnet material and increased use of reluctance torque), Switch reluctance motor (no magnets, no cogging torque high efficiency, some disadvantages not high start torque, noise and vibrations )
- For example BMW i3 has a hybrid design between PMSM and switch reluctance motor, by still using some magnets, they were able to create one of the best electric motors on the market.
- Higher RPM motors native offer higher power density, because for higher speed you do not need to increase the size of the motor, only for the torque and since the power of a motor is torque multiplied by speed you gain power easy in this way. Mechanical power at the motor shaft equation : **P[W]=Speed [RPM] \* Torque [Nm] / 9.55**
- ✓ If a motor has 65kw of mechanical output power the electrical power needed is bigger because of the efficiency. For a car a 60-65kW is enough to have

- Motor Frequency calculator :  $f[\text{hz}] = \text{Pole count (magnets nr.)} \times \text{RPM} / 120$  Example: this motor has 10 magnets and 7000rpm then  $f=10 \times 7000 / 120 = 583\text{Hz}$ .
- For this frequency is imperative to use high frequency lamination in the stator like NO-20 or equivalent. Thinner lamination will offer lower eddy currents losses. The thickness for this frequency should not be higher than 0.3mm

### 3. Choosing the Materials:

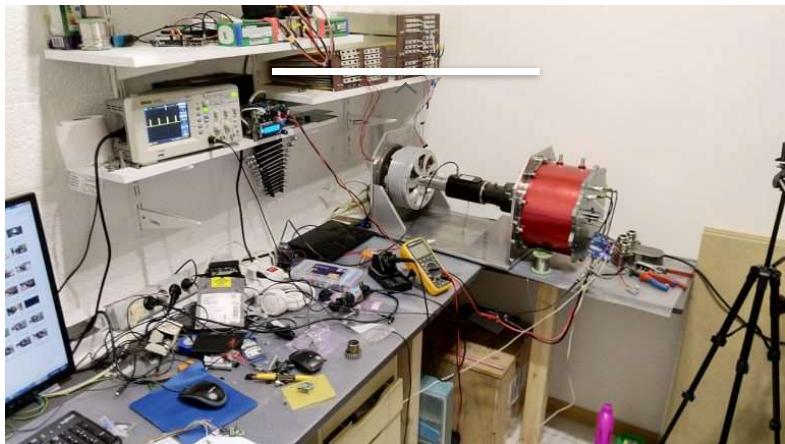
- Even if in the last 50 years the PMSM motor theory remained basically unchanged, the big improvement was in the materials area, especially in the magnets. This days the magnets are very powerful, and this allowed for very high motor densities to become a reality. Also with the use of thinner and thinner silicon steel laminations the frequency of the motors was increased 10 times versus of a conventional AC motor that runs at 60 or 50hz, this resulting in very high RPM and many pole pairs. For example regarding this motor a NO-02 material in the stator (0.2mm thickness) compared with M235-35A materials which is 0.35mm thick, will give an improvement of 0.5% in efficiency, which is quite a lot can mean 400w of less power dissipation.
- **Neodymium Iron Boron** is an alloy made mainly from a combination of Neodymium, Iron, Boron, Cobalt and of other transition metals and with varying levels of Dysprosium and Praseodymium. The exact chemical composition within NdFeB depends on the grade of the NdFeB magnet. Dysprosium and Praseodymium are added to improve the Hci (Intrinsic coercivity) of the “Neo” magnets.
- Stronger magnets produce more torque so more power in the end. I decided to have N42UH magnets with phosphate coating. Stronger than N42 can not sustain high temperature operation so i stop at N42.

### 4. Motor topology, calculations and simulations:

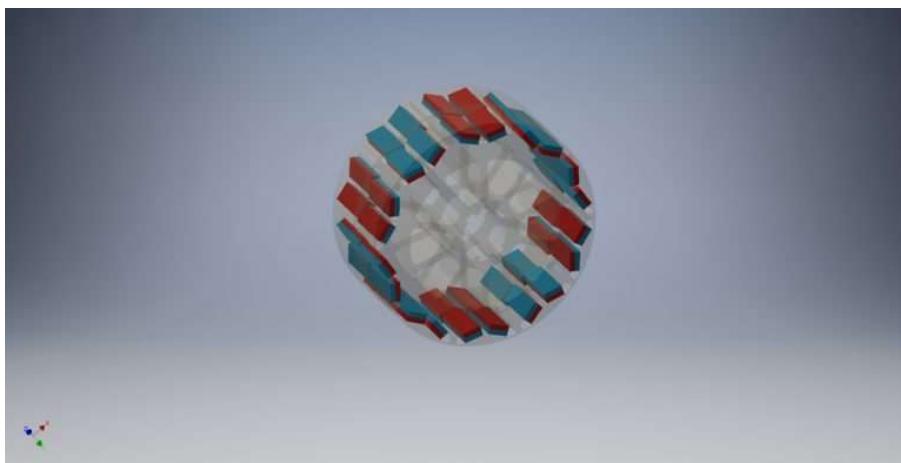
- To be able to calculate and determine all the parameters of the motor a simulation software is needed, they are extremely expensive unfortunately.
- Even with a tool to simulate the motor, you still need to know and understand all the parameters, to determine best motor topology for certain application, so is almost pointless for somebody to try the software if there is no university background for electric motor like me for example.
- The software can determine many parameters, but it can determine them well if you feed with the correct data, correct materials and correct assumptions.
- Underneath is the 60 slot and 10 pole model with transparent core to be able to visualize the coils and magnets.
- Mechanical analysis is done in a separate program to evaluate the centrifugal forces that tend to throw the magnets outside the rotor.

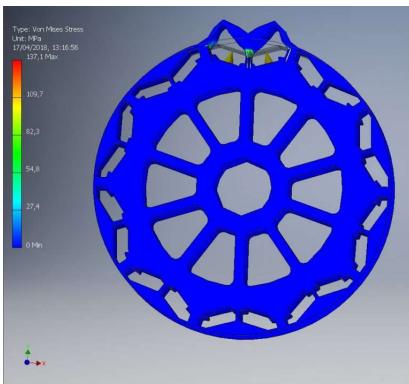
This test is very important because you want to make sure the magnets will not fly outside, but you also want to have them as close to the exterior with bracket as thin as possible.

My test bench setup



Skewed magnet representation:





Stress analysis. A force of 2700N was applied to each magnet in outer direction to simulate the centrifugal force produced by the rotor rotation at 7000 RPM. The limit force for the material is about 250 Mega Pascal. So we are way under the limit.

	<b>150A 4000 RPM</b>	<b>195A 7000 RPM</b>	<b>300A 8000 RPM</b>
<b>Torque (N·m)</b>	62.6	89.3	155
<b>Input power (kW)</b>	27.3	67.7	134
<b>Output power (kW)</b>	26.2	65.5	130
<b>Efficiency (%)</b>	96.1	96.7	96.9
<b>RMS voltage (V)</b>	106	203	277
<b>RMS current (A)</b>	150	195	300
<b>RMS current density (A/mm²)</b>	7.69	9.99	15.4
<b>Power factor</b>	0.987	0.981	0.924
<b>Torque per unit volume (N·m/mm³)</b>	3.14E-05	4.49E-05	7.79E-05
<b>Airgap stress (N/mm²)</b>	0.0157	0.0224	0.039
<b>Loss - Total (kW)</b>	1.08	2.22	4.1
<b>Loss - Winding (kW)</b>	0.457	0.772	1.83
<b>Loss - Stator back iron hysteresis (kW)</b>	0.126 ↗	0.274	0.466
<b>Loss - Stator back iron eddy current (kW)</b>	0.0311	0.118	0.231
<b>Loss - Stator teeth hysteresis (kW)</b>	0.222	0.484	0.68
<b>Loss - Stator teeth eddy current (kW)</b>	0.0741	0.277	0.475
<b>Loss - Rotor back iron hysteresis (kW)</b>	0.0104	0.0276	0.0513
<b>Loss - Rotor back iron eddy current (kW)</b>	0.0132	0.0545	0.114
<b>Loss - Magnets eddy current (kW)</b>	0.00149	0.00499	0.0162
<b>Loss - Friction (kW)</b>	0.1	0.13	0.145
<b>Loss - Windage (kW)</b>	0.04	0.08	0.09

Motor simulated at different parameters and loads

#### Main Electrical characteristics:

- Peak Power ..... 130Kw, 1min.
- Nominal Power(S1 continuous) : ..... 65 kW
- Max speed: ..... 8000 RPM
- Supply DC bus Voltage: ..... 280Vdc
- Supply DC bus Voltage: ..... 280Vdc
- Peak Torque : ..... 200 Nm
- Very low cogging torque (Zero current)..... 2,5Nm (lower than 1% of the motor torque)
- Ac Supply..... 151 VAC
- Turns: 2, parallel paths: 4

#### Mechanical Characteristics:

External diameter 292mm

length : 190mm

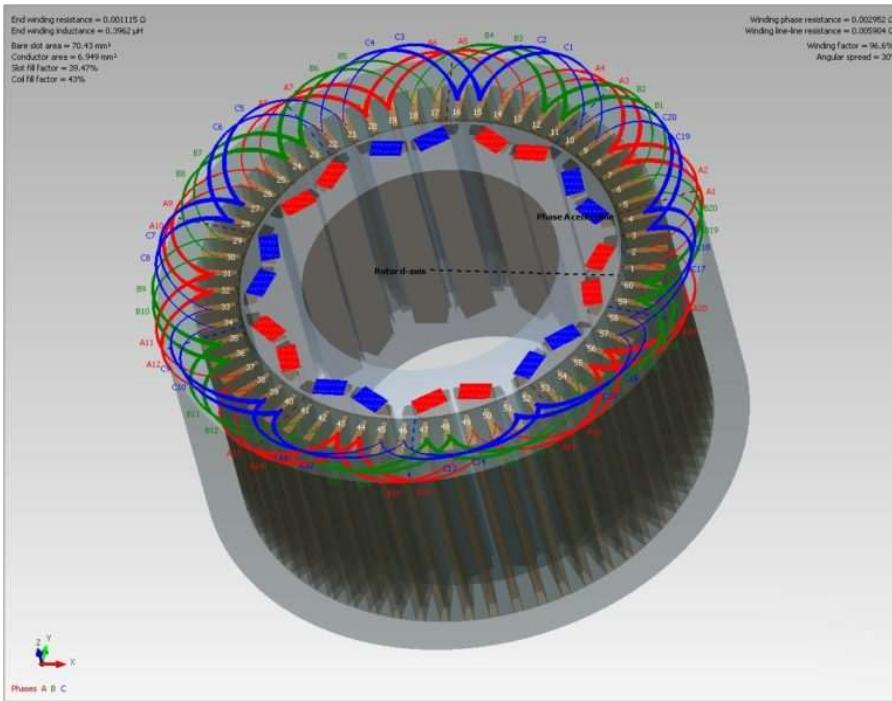
Weight ~ 36kg

The video represents the flux polarity, is displayed radially. As we can see the field is moving from coil to coil in front of the magnets.

The motor has big peak capability ~2x and enough iron not to saturate the core. At continuous operation the flux in the tooth and back stator is not higher than 1,5T

Winding distribution and parameters:

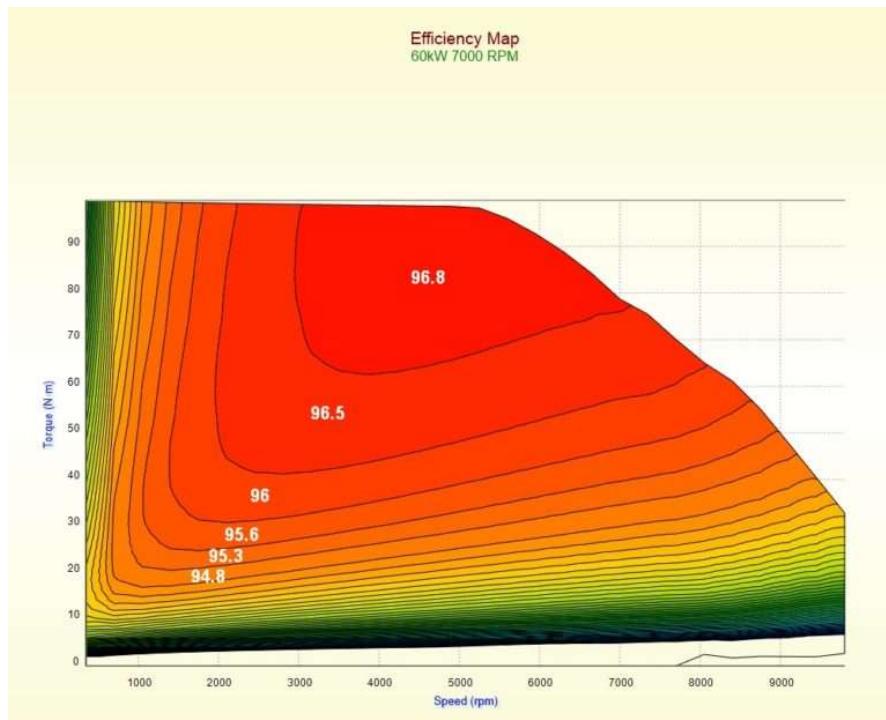




65kW Permanent magnet synchronous motor

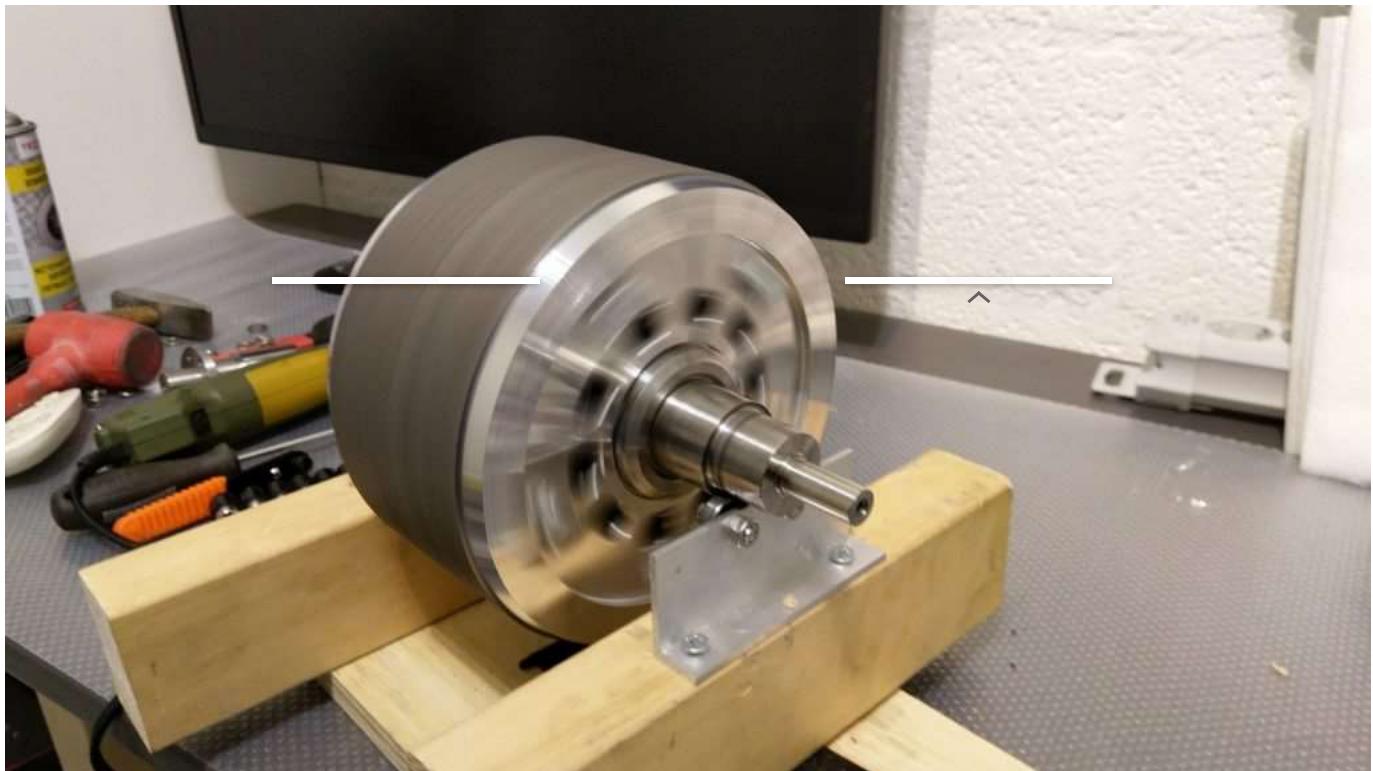
- Connection Type : Star
- Number of parallel paths: 4
- Number of turns: 4
- Wire diameter 0.61mm
- Number of strands in hank 25

Calculations: Conductor area is 6.75mm<sup>2</sup> per each path. Using 0.61mm diameter wire we end up with 23wires in parallel. Since we have 4 parallel paths total conductor area will be  $6.752 \times 4 = 27\text{mm}^2$  resulting a current density of 9,63Amp RMS/mm<sup>2</sup>. With a water jacket around the motor to cool down, the motor temperature will be about 90 degree celsius from the simulations.



The efficiency map in Non linear mode took 5 hours to simulate with core i7 7700 processor. In practice we expect some variations due to the materials, and motor construction tolerances.

💡 Dynamometer test bench for 130kW motor



Rotor balance evaluation



Rotor insertion

Iulian Berca

00:17

Our new motor uses in-house made parts, with best materials.  
 With the help of magnetic simulation software we achieved best efficiency for this class.  
 High temperature N35SH magnets where used.

Key points:

- It can be used on electric airplanes, paraglide, large drones. It can be also used on cars or boats but with limited power, because of air not blowing on the windings.
- Aviation grade Aluminum 7075
- Dual winding for 120v and 240V (end user can change the coil connections).
- Both delta and star connection possibility.
- Tapered roller Koyo ball bearing for high radial-axial load (4800Kg), long life and stability.
- High temperature high, coercively magnets with 150 degree Celsius maximum
- Triple coil insulation with lacquer impregnation and oven curing.
- Motor voltage can be made in accordance with customer request.
- 35 degree inclination spokes to push air into the coils.

Requests for quotations can be sent to iulian.berca@gmail.com.

Motor Specifications:

Power: 35Kw continuous, 60kw peak.

Nominal RPM: 2600.

Torque at 1200RPM 120Nm

Torque at 2670RPM 110Nm

Diameter 230mm

Total length: 95mm

Voltage 120 (parallel connection) or 240V series connection

Kv constant is 13 for 240v or 23 for 120v connection.

We can make any custom flange for the propeller or other mechanical connection.

Both bearings are whether sealed.

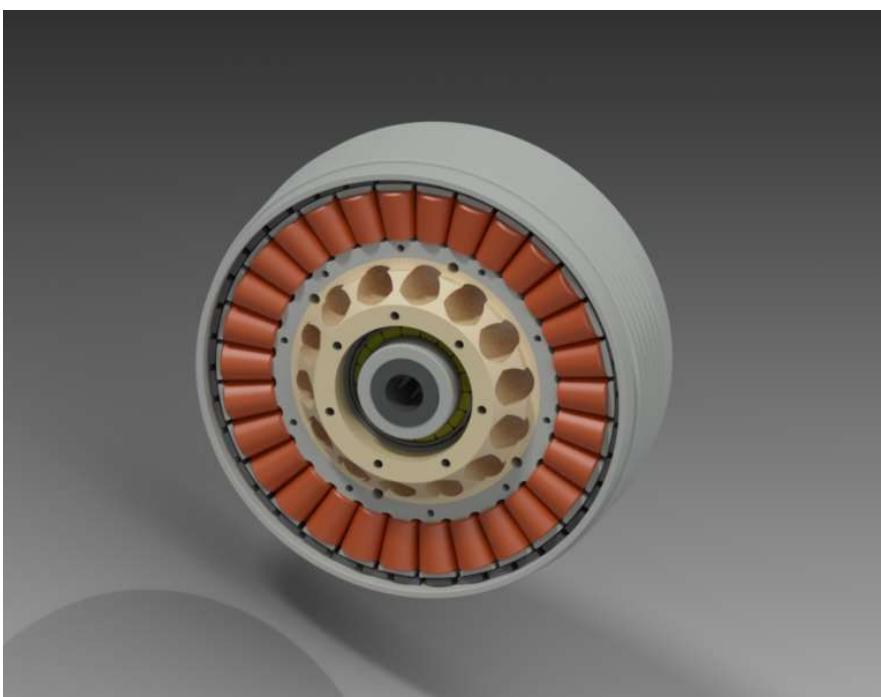
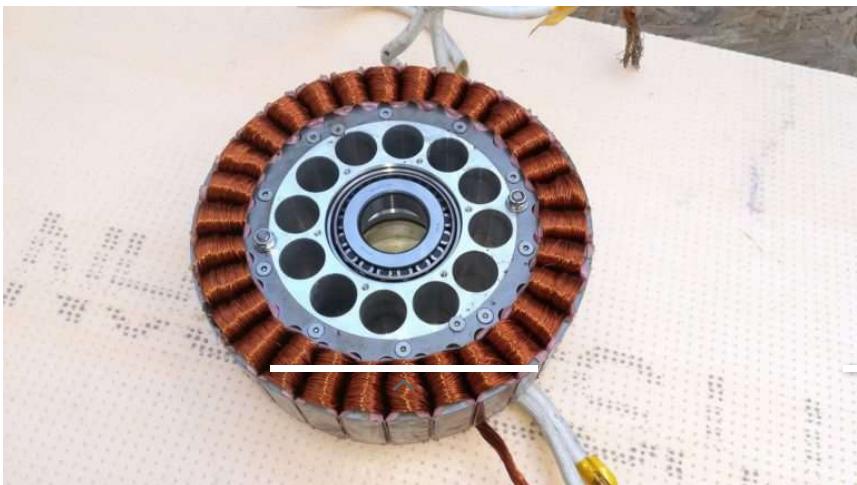
Bottom flange is 6xM6 on Ø82

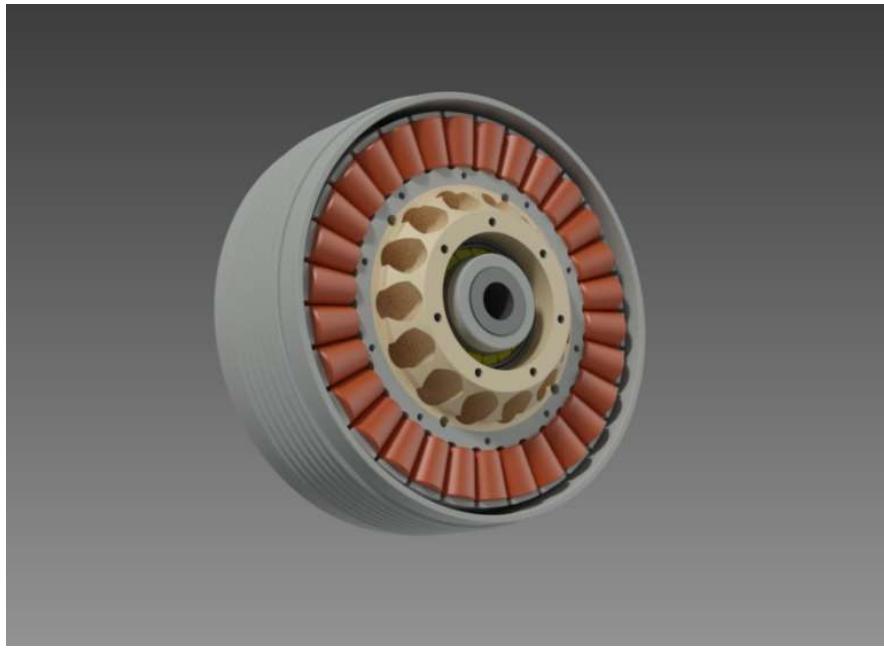
Underneath you find the efficiency map of the motor.



Motor rendering:

Some pictures of the motor





### Video Test with the motor :

High efficiency 35kW Brushless motor.



### [EmDrive independent test](#)

Posted: 12th May 2015 by [iulian207](#) in [Projects](#)

Tags: [em drive](#), [em drive test](#), [Emdrive](#), [emdrive dimensions](#), [emdrive frustum size](#), [emdrive independent test](#), [emdrive propulsion](#), [emdrive size](#), [emdrive thrust](#), [real antigravity](#)

I'm willing to collaborate for different projects or working in a creative company, to develop and test new technologies.

My youtube channel is on <http://www.youtube.com/user/julian207?feature=mhee>

### An Explanation of the EmDrive and Cannae Drive Part II



### An Explanation of the EmDrive and Cannae Drive Part I



If you have any question about this constructions you can send me an e-mail at [julian.berca@gmail.com](mailto:julian.berca@gmail.com)



I'm designing in Eagle-cad : brushless controller circuits, dc motor controller circuits, protection circuits, etc.

I'm constantly improving my design for a robust brushless controller, with all protection needed (overcurrent, over-temperature, fault output, torque control) also ramp acceleration and deceleration and adapting timing angle.

Now i also work on a new design of a large 48 pole 60kW direct drive brushless motor.

I'm now building new laboratory and i would appreciate any small donations for new shop materials or equipment to improve designs, and make new interesting and innovative projects.

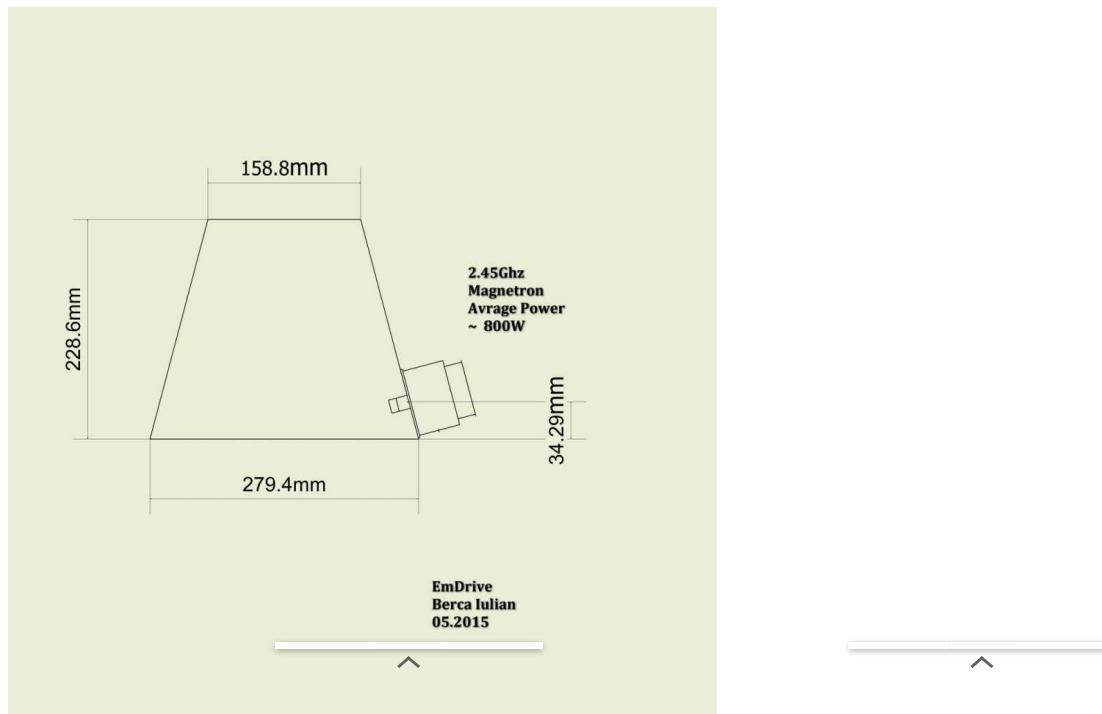
Experiment replication to observe the thrust in EmDrive device.

The device uses a [magnetron](#) to produce [microwaves](#) which are directed into a metallic, fully enclosed conically tapered [high Q resonant cavity](#) with a greater area at one end of the device, and a [dielectric resonator](#) in front of the narrower end. The inventor claims that the device generates a directional [thrust](#) toward the narrow end of the tapered cavity. The device (engine) requires an electrical power source to produce its reflecting internal microwaves but does not have any moving parts or require any [reaction mass](#) as fuel. If proven to work as claimed, this technology could be used to propel vehicles intended for all forms of travel including ground travel, marine travel, sub-marine travel, [airflight](#) and [spaceflight](#).

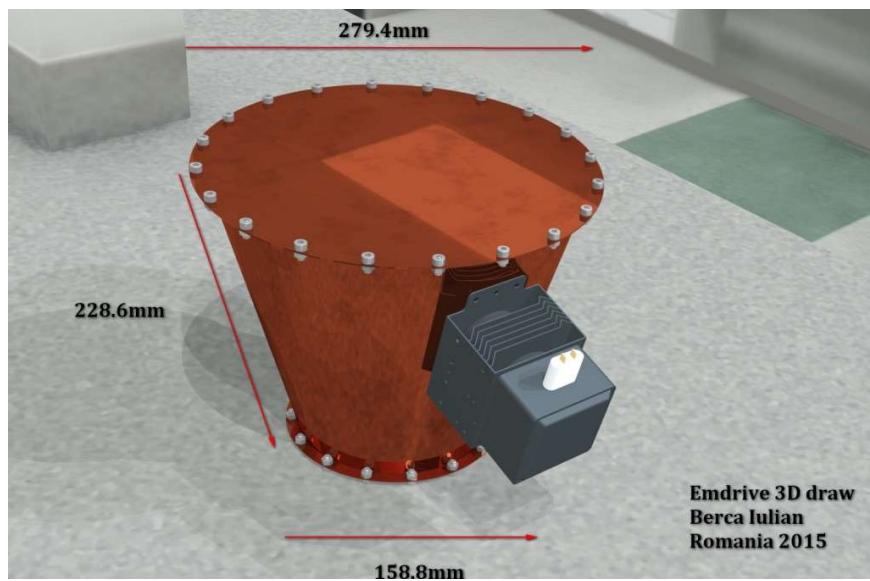
EmDrive is a device invented by Roger Shawyer in 1999 and replicated with success by a team from China led by Yang Juan and a team from Nasa this year. The

- transformer from a microwave oven (power ~ 800-1200W)
- Magnetron from microwave oven: Anode cathode voltage ~ 4kV , and 3-4V @13A for the filament. Frequency 2,45Ghz.
- solder, 4mm screws, pcb

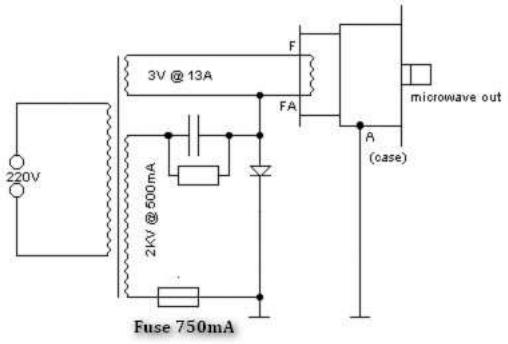
Test equipment: Current measurement, voltage measurement, temperature measurement, micro-gram scale.



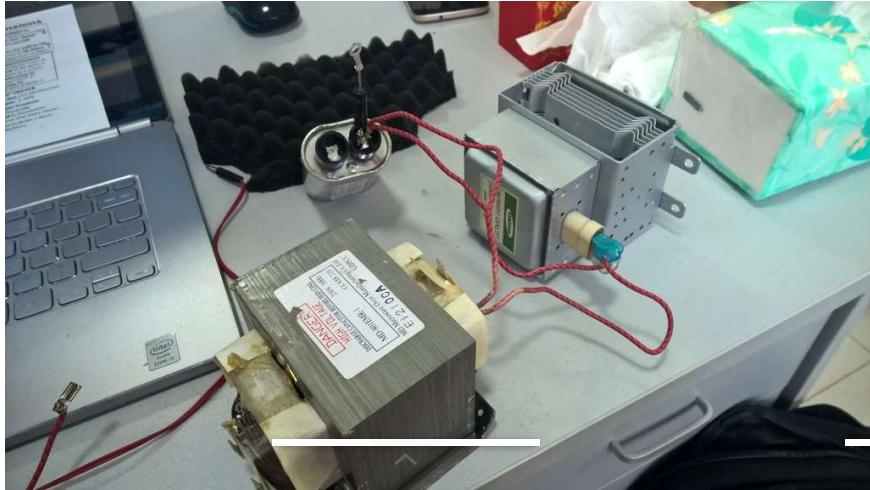
Frustum 3D modeling in Autocad Inventor

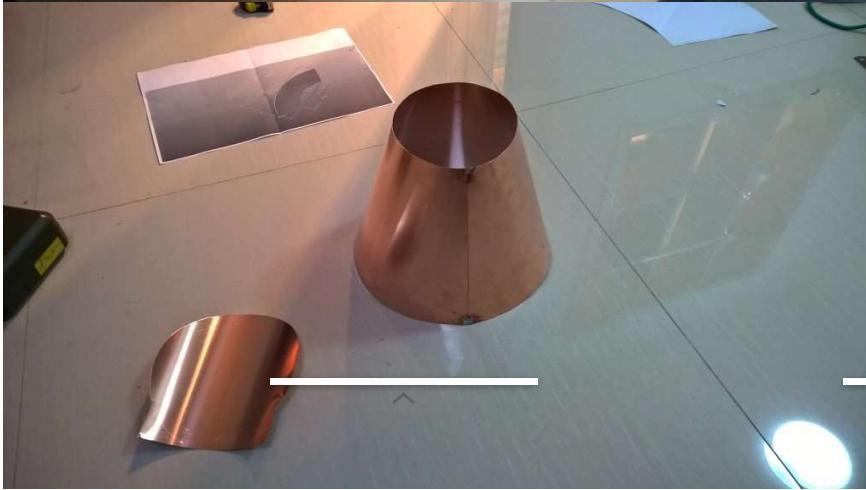


Connection diagram for the magnetron. **Warning:** charged capacitors can kill very easy. Always discharge the capacitor by putting 100kOhm resistor on the ends and also to the external case for your safety. After discharge put the ends in short circuit and wait for a couple of seconds to be absolute sure that there is no voltage.



Almost all the materials arrived today:





Today i will make the frustum to see the results.



After i power on the temperature of the magnetron increased to 60 degree celsius.(140F) in around 5-6 seconds. If the magnetron does not have any load the temperature should rapidly increase i think, even though the microwave oven will not burn (overheat) if left on without any food inside.

In this paper : <http://www.emdrive.com/IAC-08-C4-4-7.pdf> they say that the thrust comes after 20 seconds you power the magnetron. But in 20 seconds the magnetron will be very hot without proper cooling ( or maybe because the magnetron has no load)

I still do not know if the waveguide in a microwave oven plays other role than just feed the microwave in to the cavity.

Another think i want to test is to try to reduce the current on the filament with a separate power supply in hope that i will decrease the power in “search” of some thrust.

In a few days i will receive 2 plates of PCB single-sided, and i will try them instead of cooper ends.

---

Today i did the first test with the setup suspended in a pendulum. The power was applied for 40 seconds. No thrust resulted 😞

i will post the video soon.

Tomorrow i will move the magnetron above the middle of the frustum, to the small end. After power on for 40 sec temperature was 85 degree Celsius.

Next step will be adjusting the current of the filament and maybe the frequency by adding 2 separate coils over the magnets with adjustable current, to try to change the oscillating frequency.



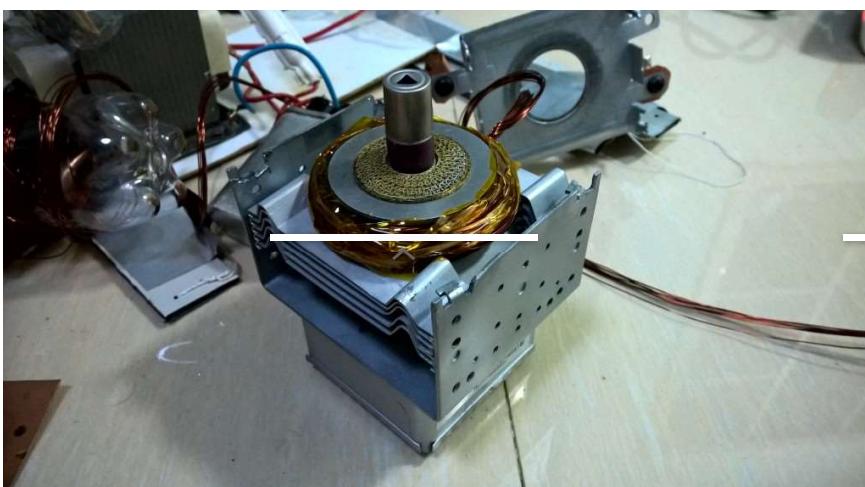


Test No. 2

Modifications: magnetron moved to the smaller side

Coil added to the magnetron to try to adjust the frequency.

Still no visible thrust in the pendulum







EmDrive Test No.03 Success, I have thrust !!!



change enough to find the resonance frequency of the frustum and hope for higher thrust. The other method of finding the resonance is to adjust the length of the cavity. This can be done with a movable plate and a screw . I can make that out of PCB.



---

Hi, guys, i'm still alive . Sorry if i did not post anything this days. I noticed some guys think i died , relay strange. I do not have a Tweeter account by the way.

I flipped the cone in the original setup and i have the thrust downwards (scale goes positive). Unfortunately the thrust downwards is around 7 times smaller. difference on the scale is only 0.20 grams and is consistent with the power on and off .

**Emdrive Test No 3.1 Frustum upside down position**



Temperature discussion: during the tests the temperature of the frustum does not change to much maybe 1-2 degree.

---

Andy P. said something interesting : "When comparing the different thrusts, you will also have to take into account that in test 3.1 the thruster has to fight against the upward force of the spring onto which it is attached. This will lower the observed weight change on the scale, but does not necessarily mean the thrust is lower."

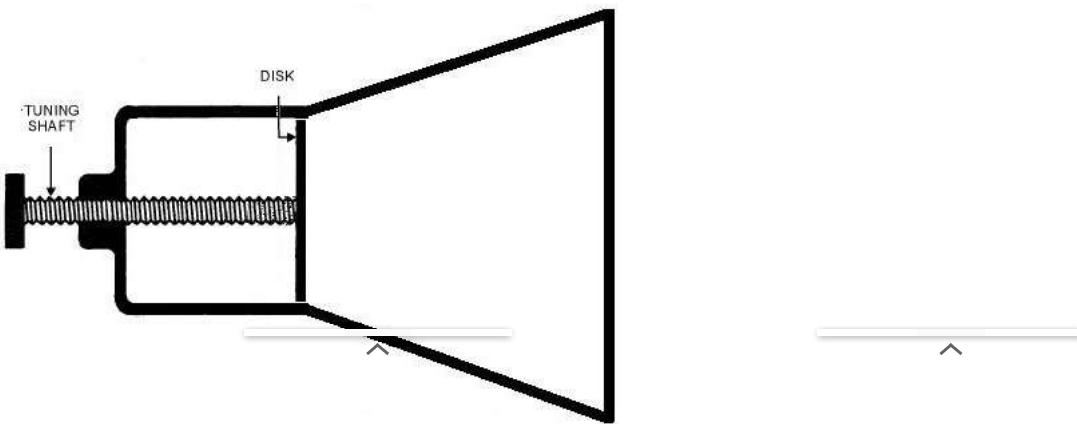
This fight "against the spring" is real or not ? Lets assume this: you put 1kg on a scale and push TARE button. When you remove the weight it should not indicate -1Kg if the fight against the spring was real. Inside the scale is also a "spring" to keep the weight the the test should be the same.

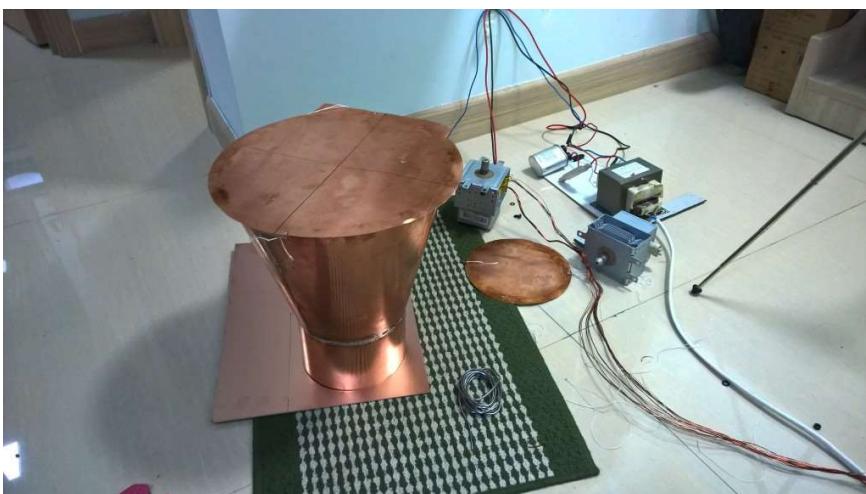
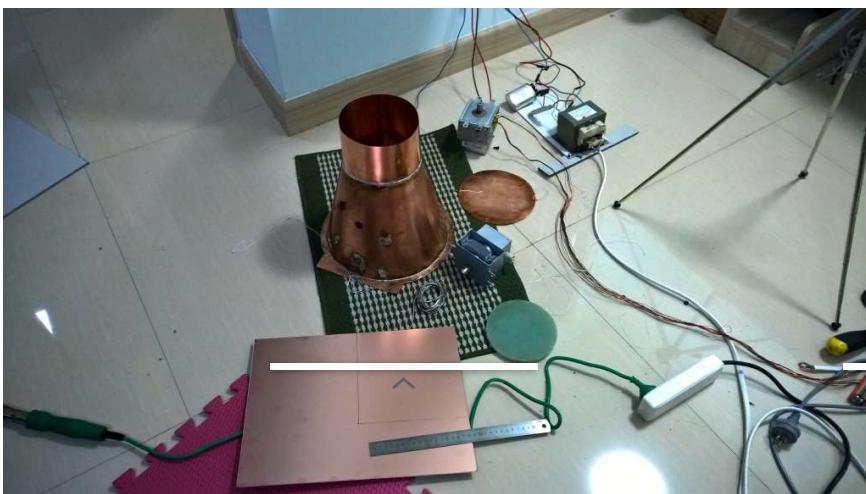
I'm working the modify the cone now. Test No 4 will be with new setup.

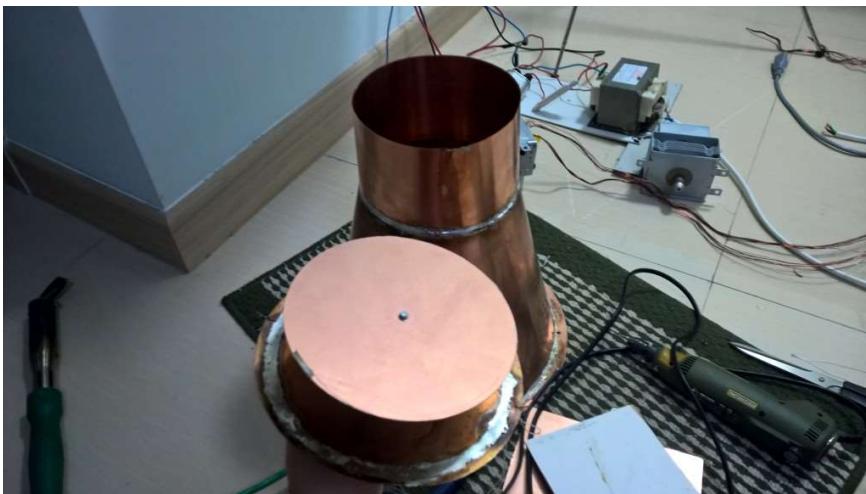
Because i do not have a cooler to the magnetron i can not put a servomotor to continuously adjust the cavity length because magnetron will heat fast. So i will need to manually adjust the length for each test to observe the scale and then let the magnetron to cool ant test again.

First i will adjust in bigger steps 1cm smaller for each test. then i will see witch one has the most thrust. After i will go around that value from mm to mm with the screw.

I do not own the picture with the adjustable setup. is from NasaSpaceflight forum.



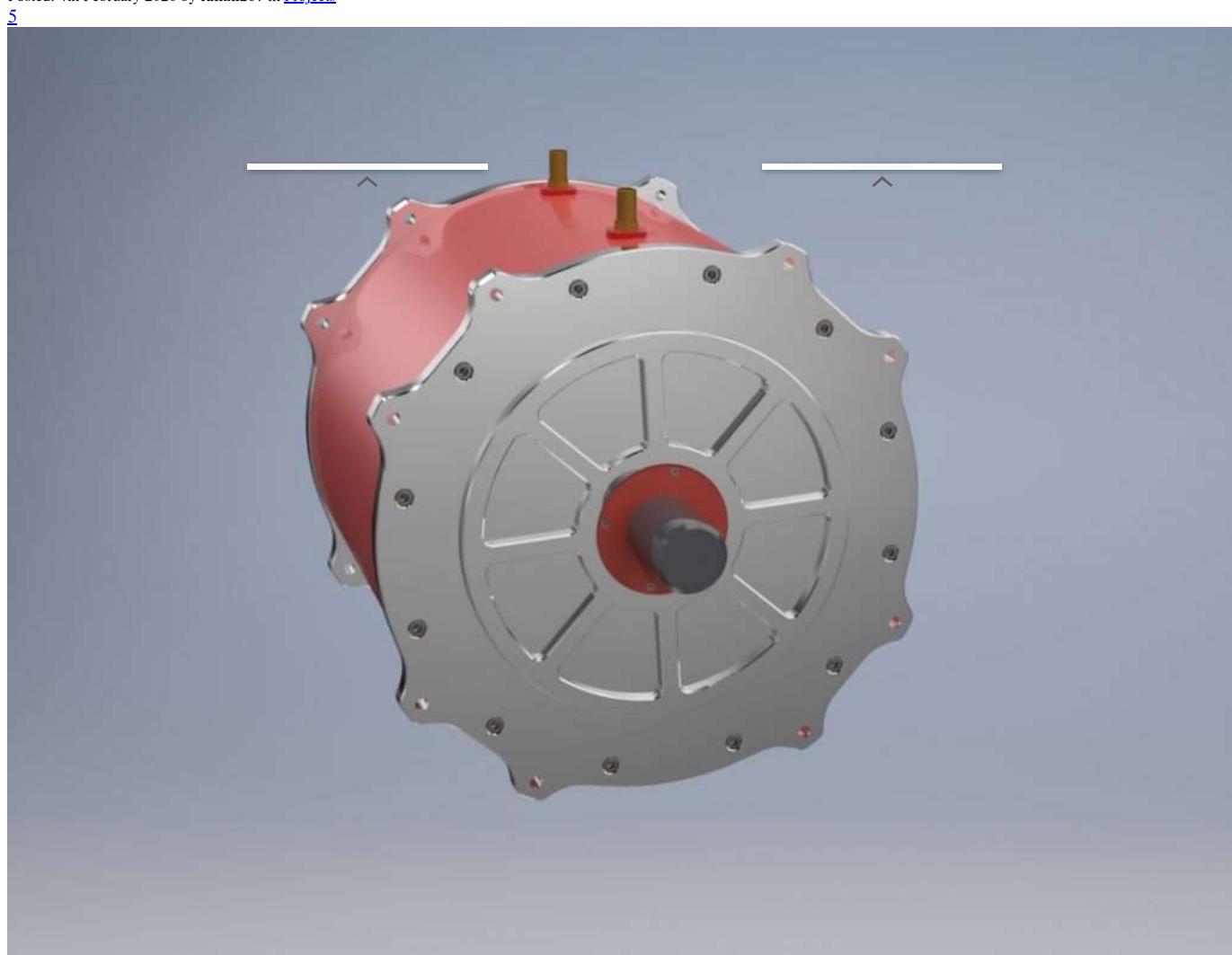




I hope others can make their own EmDrive and have positive results.

## [\*\*250Kw electric motor for electric BUS\*\*](#)

Posted: 4th February 2020 by iulian207 in [Projects](#)



3000 RPM/ 900Nm 650Vdc ~130kg

## [\*\*How Neodymium magnets are made\*\*](#)



The method for manufacturing Neodymium Iron Boron magnets is as follows:

- **MELTING OF THE ALLOY UNDER VACUUM** The Neodymium metal element is initially separated from refined Rare Earth oxides in an electrolytic furnace. The Neodymium, Iron and Boron and the other chemical elements are weighted and put in a vacuum induction furnace to form an alloy. The mixture is melted due to the high frequency heating.
- **CASTING AND MILLING** In simplified terms, the “Neo” alloy is like a cake mixture with each factory having its own recipe for each grade. The resultant melted alloy is ~~then cooled to form ingots of alloy or cooled down by the so called “strip casting process”~~. The alloy ingots or the strip cast faces respectively are then broken down by hydrogen depreciation (HD) and then jet milled down in a nitrogen and argon atmosphere to a micron sized powder (about 3 microns or less in size). This Neodymium powder is then fed into a container, and oriented by a magnetic field in order to allow the pressing of the so called green compacts of the magnets.
- **PRESSING AND ALIGNMENT IN MAGNETIC FIELD** There are three main methods of pressing the powder: isostatical pressing, axial pressing and transverse pressing. Die pressing requires tooling to make a cavity that is slightly larger than the required shape (because sintering causes shrinkage of the magnet). The Neodymium powder enters in the die cavity from the container and is then compacted in the presence of an externally applied magnetic field. The external field is either applied parallel to the compacting force or perpendicular to the direction of compaction. Transverse pressing gives higher magnetic properties for the NdFeB sintered magnets. The other method of pressing is isostatic pressing. The NdFeB powder is put into a rubber mold and is put into a large fluid filled container which then has the pressure of the fluid increased. Again an external magnetizing field is present but the NdFeB powder is compacted from all sides. Isostatic pressing gives the best possible magnetic performance for Neodymium Iron Boron. The methods employed vary depending on the grade of “Neo” required and are decided by the manufacturer. The external magnetizing field is created by a solenoid coil set either side of the compacting powder.

The individual powder particles of the NdFeB powder align with the magnetizing field that is applied – the more homogeneous the applied field, the more homogeneous the magnetic performance of the Neodymium magnet. As the Neodymium powder is pressed by the die, the direction of magnetization is locked in place – the Neodymium magnet has been given a preferred direction of magnetization and is called an-isotropic Rare Earth magnets exhibit uni-axial magneto-crystalline anisotropy i.e. they have a unique axis crystal structure corresponding with the easy axis of magnetization. In the case of Nd<sub>2</sub>Fe<sub>14</sub>B, the easy axis of magnetization is the c-axis of the complex tetragonal structure. In the presence of an external magnetizing field, it aligns along the c-axis, becoming capable of being fully magnetized to saturation with a very high coercivity.

- **SINTERING AND ANNEALING** The Neodymium magnet is then sintered to give it its final magnetic properties. The sintering process is carefully monitored (a strict temperature and time profile has to be applied) and occurs in an inert (oxygen free) atmosphere (e.g. argon). If oxygen is present, the resultant oxides destroy the magnetic performance of the NdFeB. The sintering process also causes shrinkage of the magnet as the powder fuses together. The shrinkage gives a magnet close to the required shape but the shrinkage is usually uneven (e.g. a ring may shrink to become an oval). At the end of the sintering process a final rapid quench is applied to rapidly cool the magnet. This is to minimize the unwanted production of other magnetic phases. A rapid quench maximizes the magnetic performance of NdFeB. Because the sintering process causes an uneven shrinkage, the shape of the Neodymium magnet will not be to the required dimensions.
- **MACHINING** The next stage is to machine the magnets to the required tolerances. Because machining is required, the Neodymium magnets are made slightly larger when being pressed e.g. larger outer diameter, smaller inner diameter and taller for a ring magnet. Standard magnet dimensional tolerances are +/-0.1mm although +/-0.05mm is achievable at extra cost. The possibility of even tighter tolerances depends on the shape and size of the magnet and may not be achievable. For note, the Neodymium magnet is very hard. Trying to cut holes in NdFeB with a standard drill or carbide tip will blunt the drill bit. Diamond cutting tools (CNC diamond grinding wheels, diamond drills, etc.) and wire cutting machines (EDM) have to be used. The NdFeB swarf powder produced during machining needs to be cooled by liquid otherwise it may spontaneously combust. For Neodymium block magnets, there may be

## How SuperMagnets are made



## neodymium magnet manufacturer



## 45KW Bushless motor design

Posted: 17th May 2014 by iulian207 in [Projects](#)

Tags: ["3 phase brushless dc motor"](#) ["3 phase brushless dc motor controller"](#) ["brushless esc"](#), ["Homemade electric Go Kart"](#) ["go kart"](#) ["electric go kart"](#) ["homemade brushless controller"](#) ["brushless dc controller"](#) ["tachometer"](#) ["Brushless Regler"](#) "Бесщеточный контроллер" "безщетковый контроллер" "sin e." ["自家製のポンツーンのボート"](#), ["10kw brushless motor"](#), ["20kw brushless motor"](#), ["30kw brushless motor"](#), ["40kw brushless motor"](#), ["50kw brushless motor"](#), ["60kw brushless motor"](#), ["blde motor esc"](#) ["초음향"](#), ["brushless motor build"](#), ["brushless motor construction"](#), ["brushless motor homemade"](#), ["Brushless motor design"](#), ["PMSM design"](#)

[108](#)

After successfully building a brushless controller i have decided also to build my own brushless motor. Autocad Inventor 3D cad software was used by me to design 3D model of the motor.

Before start to design something you need to know the RPM needed, torque needed, running voltage, max Amps. This formula is for calculating the torque if you

$$\text{Torque[Nm]} = \frac{9.55 \times P[W]}{\text{Speed [rpm]}}$$

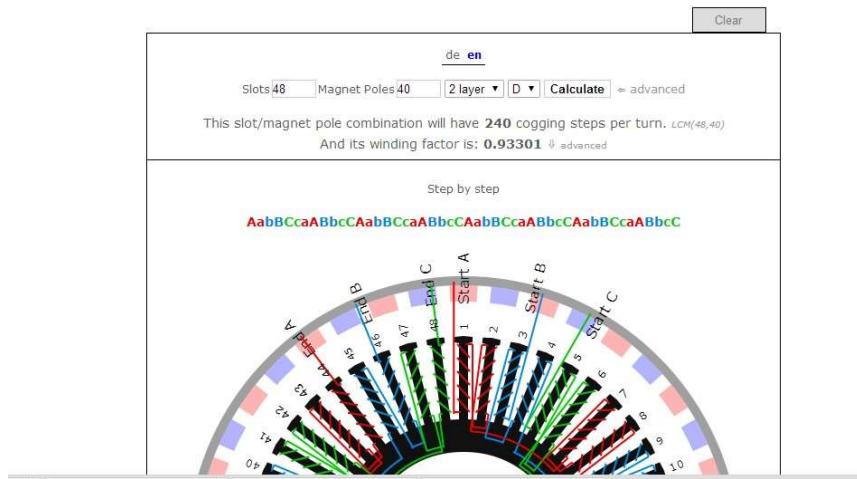
have the power and the speed.

After reading the gear ratio on Opel Agila it resulted that i need a speed of 4000RPM to reach 73.4Km/h in 3-rd gear ( i do not need more than that in a city). I decided to make a reverse outrunner motor design because is more easy to cool the stator if is outside. If the sator is inside is more difficult to cool down. The draw-back is that you loose torque because you have a smaller diameter of the rotor. I opted for a 48 slots (teeth) and 40 magnets design and i will have 142.5Nm of torque.

- Nominal Power: 45KW
- Nominal Voltage: 230v
- Nominal current: 200A
- Winding configuration : Delta
- RPM: 2600
- Torque 165Nm
- Construction : 48 slot, 40 Magnets Neodymium
- Lamination Grade M330
- Cooling : Glycol
- ✓ Weight: 17Kg

- STAR will give you lower RPM, higher torque (1,73 more than dela), higher the voltage, lower the current.
- Can be: concentrated or fractional slot type.
- If there are concentrated can be: LRK, distributed LRK etc.
- A good winding scheme calculator can be found [here](#).

### Winding Scheme Calculator

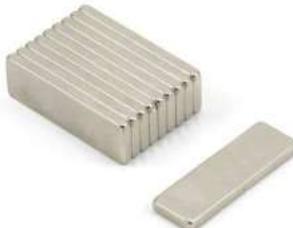


### How to choose the magnets ?

After you choose the slots and poles, you need to choose the magnets. This is not very easy task because you need hi temperature magnets and are not so cheap and easy to find for your needed size. I bought custom made magnets from a Chinese website.

- The temperature rating for Neodymium is only a guide value. The actual temperature where magnet start to lose strength is size, shape and magnetic circuit dependent. If you have a magnet attached to a piece of steel it will demagnetize at higher magnetic flux than in a free space. On the other hand demagnetization temperature will be lower if you subject the magnet to a strong opposite magnetic field such in a motor.
- If the thickness of the magnet is bigger and you will need a bigger magnetic field to start demagnetize it.

Temperature classification for neodymium magnets, N stays from Neo from .



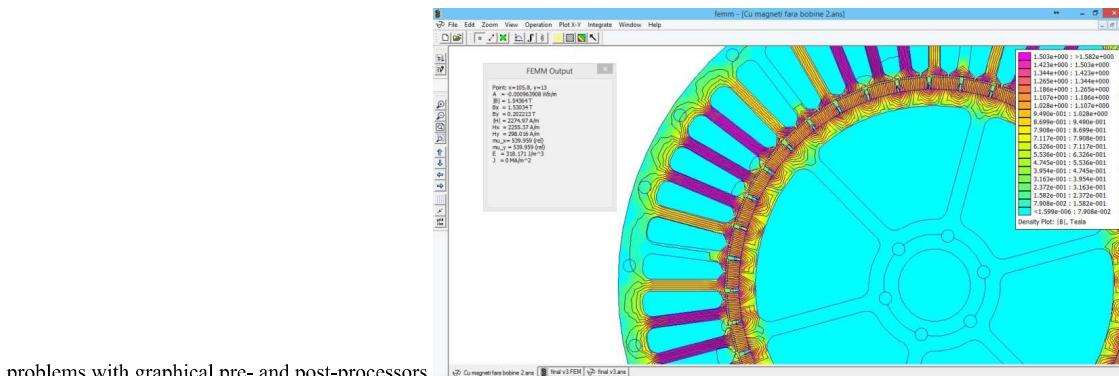
- N42       $\leq 80^\circ\text{C}$
- N42M     $\leq 100^\circ\text{C}$
- N42H     $\leq 120^\circ\text{C}$
- N42SH    $\leq 150^\circ\text{C}$
- N42UH    $\leq 180^\circ\text{C}$
- N42EH    $\leq 200^\circ\text{C}$
- N42VH    $\leq 230^\circ\text{C}$

Neodymium magnets needs a coating otherwise they will rust in contact with air.

The coatings can be: Nickel, Zinc, Phosphate, Epoxy, Gold and others.

### Mathematical magnetic flux density analysis.

Next step is to run mathematical analysis in magnetic field to see if i have some areas with saturated magnetic field. We want to avoid core saturation. For this i used **Finite Element Method Magnetics Tool** A Windows finite element solver for 2D and axisymmetric magnetic, electrostatic, heat flow, and current flow



problems with graphical pre- and post-processors



We can observe that i have areas in pink color with to much magnetic flux, above 2Teslas, so i need to increase the thickness of the tooth to stay under 2 Teslas because the saturation of the lamination.

Hall sensor spacing mechanical angle calculation formula:

$$360^{\circ} / 6^{\circ} \text{ pole pairs number}$$

6 is coming from 6 step commutation in the controller.

Example. 14 magnets motor means 7 pole pairs

So the resulting angle between two hall sensors is  $\Delta\alpha = 360^{\circ} / 6^{\circ} 7 = 11,14$  degree.

For 20 magnets motor mean 10 pole pairs.

resulting  $\Delta\alpha = 360^{\circ} / 6^{\circ} 10 = 12$  degrees

Multiple factors can have a huge difference in motor performance and efficiency:

This factors can be:

1. Maximum working frequency (depending on RPM and no. of poles). The frequency is calculated by next formula:

$f = \text{rps (motor rotation per second)} \times (\text{nr. of poles}/2)$ . no.of poles is equal with no. of magnets.

or:  $f[\text{Hz}] = \text{Magnets nr.} \times \text{rpm} / 120$

Example for 1000Rpm: the rps will be  $1000\text{rpm}/60\text{s} = 16,66$  then  $f = 16,22 \times 40\text{poles}/2$  will result in:  $f = 333.2\text{Hz}$

Because the losses in the core lamination are increasing with increase (non linear) in frequency we want to have a frequency as low as possible for max motor RPM. For example for lamination grade M330-50 the losses at 50Hz and 1Tesla are 1,29W/kg but 132W/kg at 1000Hz.

2. Proper combination between slots and poles count.

3. Material properties and thickness of stator and rotor laminations.

4. Air gap thickness.

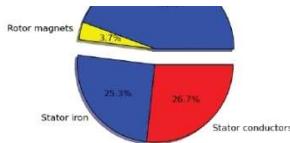
5. Magnets grade.

6. Current density.

7. Slot fill factor.

8. Cogging torque. A summary of techniques used for reducing cogging torque:

- Skewing stator stack or magnets
- Using fractional slots per pole
- ✓ Modulating drive current waveform



### Results from load simulation

BLDC with DC current source, delta connected

#### Current

DC current	150A
Current density	7586000A/m <sup>2</sup>
Phase resistance	0.0302Ω
Conductor losses	453W
Slot area	263.6mm <sup>2</sup>
Conductor area	6.591mm <sup>2</sup>
Advance angle	0deg
Phase end-winding leakage inductance	2.409e-07H

#### Torque

Mean airgap torque (by Maxwell stress tensor)	214.8mN
Torque reduction due to iron losses	3.165mN
Torque ripple	14.15%

#### Flux density

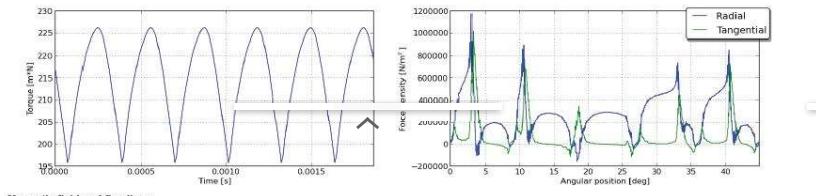
Maximum stator iron flux density:	3.557T
Maximum rotor iron flux density:	2.544T
Minimum permanent magnet flux density:	0.6423T
Fundamental airgap flux density:	1.11T
Iron losses:	533.6W

#### Voltage, Power, Flux linkage

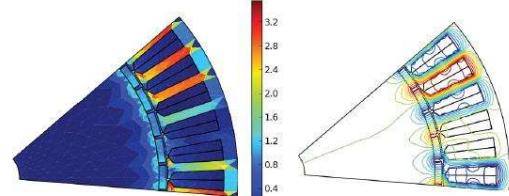
Average DC voltage	242.9V
Electrical input power	36440W
Mechanical output power	35450W
Efficiency	97.29%
Peak flux linkage, q-axis	0.09305Wb
Peak flux linkage, d-axis	0.07495Wb

#### Torque and force waveforms

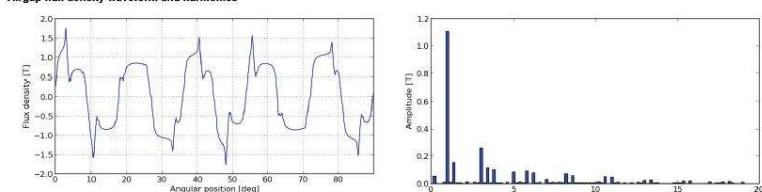
Brushless motor simulation 45KW 0.2mm lamination and 0.75mm airgap



#### Magnetic field and flux lines



#### Airgap flux density waveform and harmonics



#### Simulation statistics

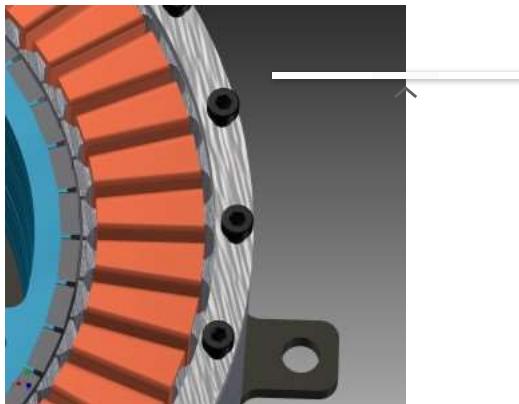
Load simulation number of mesh elements:	3420
Load simulation core type:	1x c2
Runtime of simulation:	00:01:26
Credits cost:	0.0481

Brushless motor simulation results

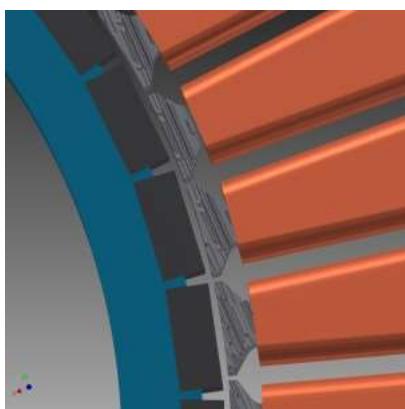
- The final motor without the caps.
- The motor with the caps and without the ball bearings.
- I have also an [YouTube](#) video to present the 3D model.



Design stage



Close up view of the coils, magnets.



Air gap 0.75mm

Custom Magnets received and tested by heating up to 120 Celsius to see if any drop in magnetic field occurs.





Custom cut, N35UH grade, Phosphate coating magnets

Because the laser shop in Romania only had M330-50 grade, i was forced to use this material with higher losses for high RPM (frequency)



Motor laminations Grade M330-50 and other elements.

Drill to embed the screw in material.



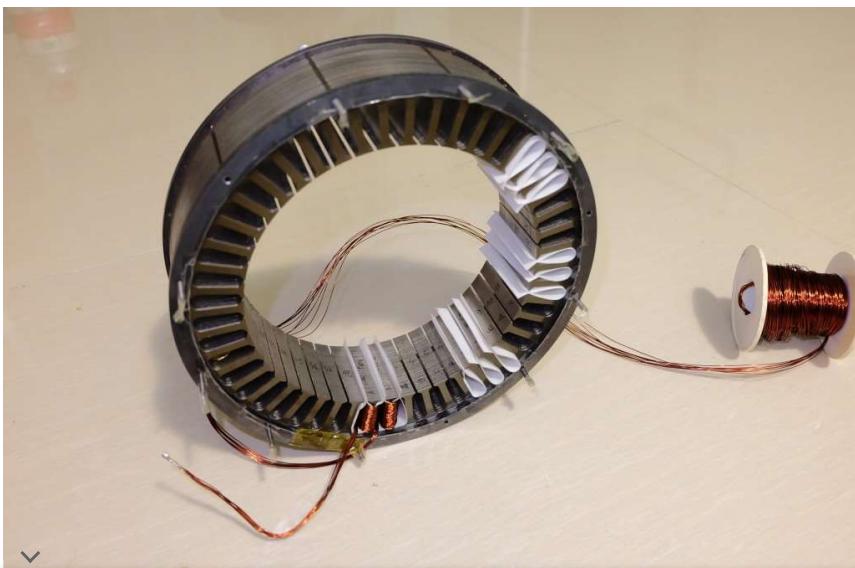


Rotor assembly



Rotor with epoxy and kevlar to keep the magnets from flying from centrifugal force.

I used 14 strands of cooper 0.5mm in parallel.





Winding the motor



Motor finished, ready for testing.



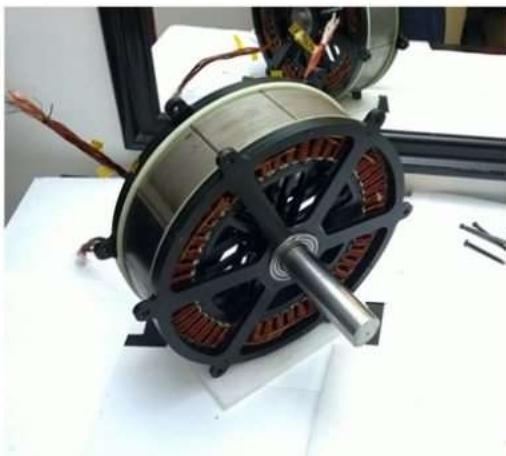
Motor Water Cooling



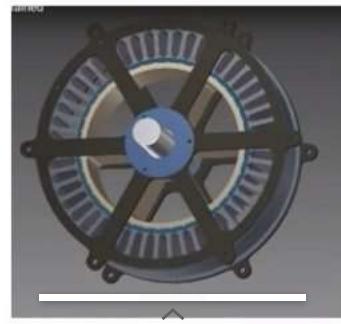
Brushless motor ready for testing under load.

How to make a brushless dc motor at home





Siemens motor released in 2018



My motor released in 2014 (youtube)

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Siemens released a similar motor in 2018 while i put mine in 2014 on youtube

## Electric car conversion, Opel agila electric

Posted: 19th July 2013 by [iulian207](#) in [Projects](#)

Tags: [car de motor](#), [electric car controller](#), [Electric car conversion](#), [opel agila conversion to electric](#), [opel agila electric](#), [opel corsa conversion](#), [opel corsa electric](#)



In the last 6 months i'm working at my electric car conversion. The car was in good shape and was a good candidate for the conversion.  
After i found a place where to work to the car (a garage) i took out the ICE and i begun to measure all the interior spaces to see if the motor will fit.  
Today (27.07.2013) i did the first test on the electric motor to see if the motor fits ok inside and join well the gear box.





After i removed the engine i start taking measurements for creating the mounting plates.

The car with the engine removed. Now i have space to work.



This motor is a Permanent Magnet Dc motor with 100Nm and 1900RPM total power will be limited to ~ 20Kw depending on how i will manage to remove the heat from the motor.

I also received the Batteries:

Model : A123 Systems.

Capacity : 20Ah

Nominal voltage 3.3v

Continuos Discharge 600A

Puls discharge 1200A

Total number of cells in series : 72

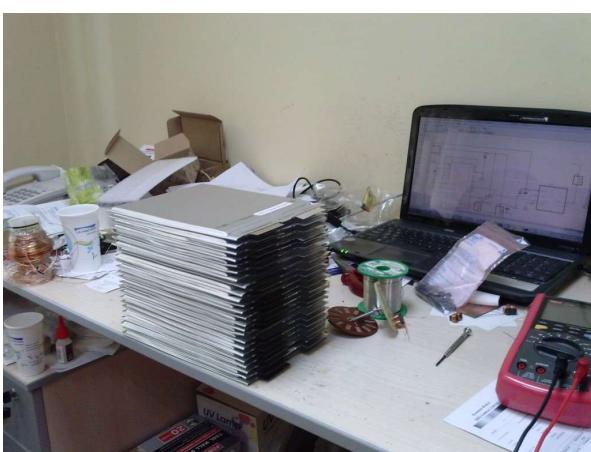
As the DC motor Controllers are to expensive for my budget, and i have experience in electronics i designed my own pwm controller using Eagle Cad program.

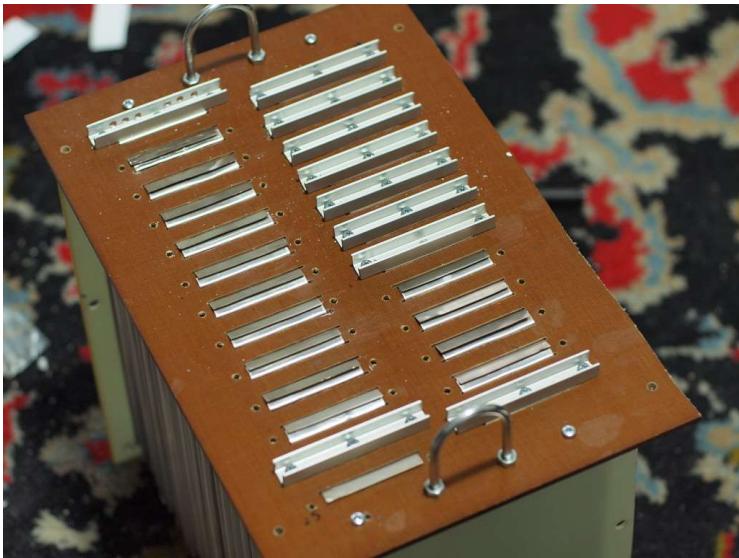
Specs: Max voltage: 400V

Max current: 200A

Adjustable current limitation feature,

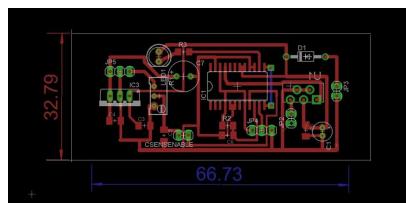
Undervoltage lockout, thermal shut down, current sense.



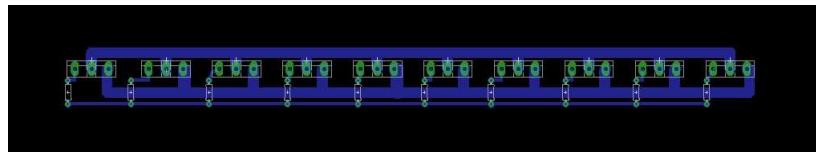
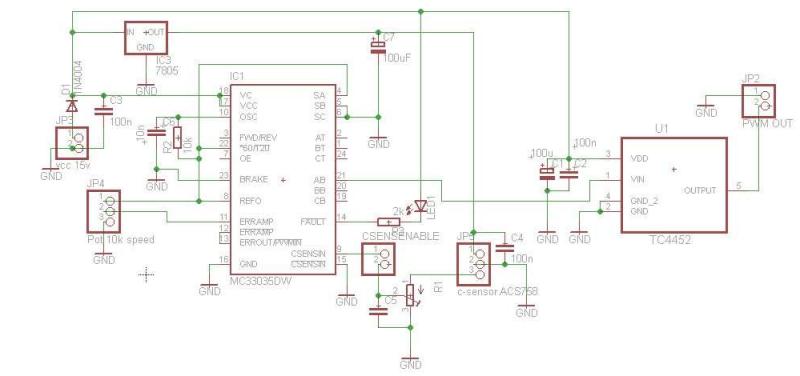


New 30 A123 Systems cells arrived.

The mounting plate for 24 cells in series

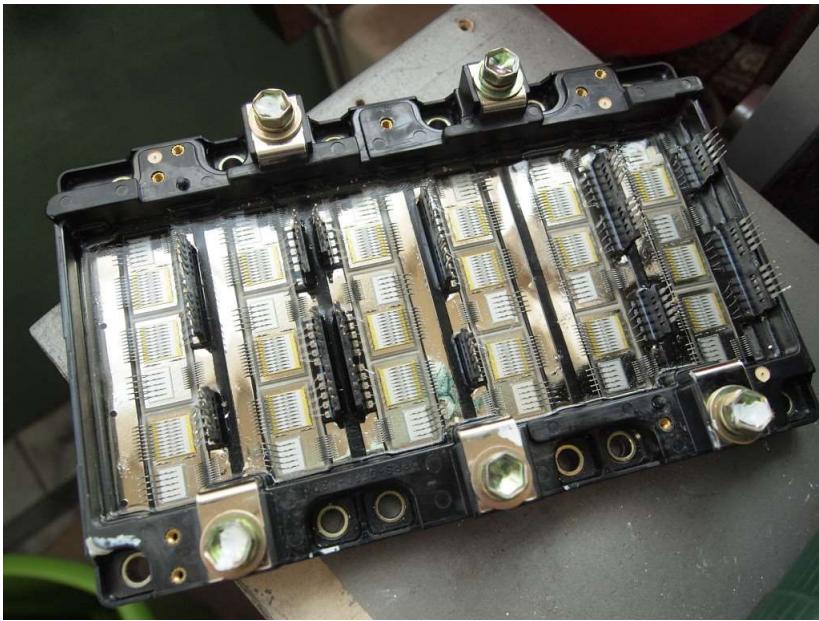


The schematic and PCB were designed in Eagle cad.

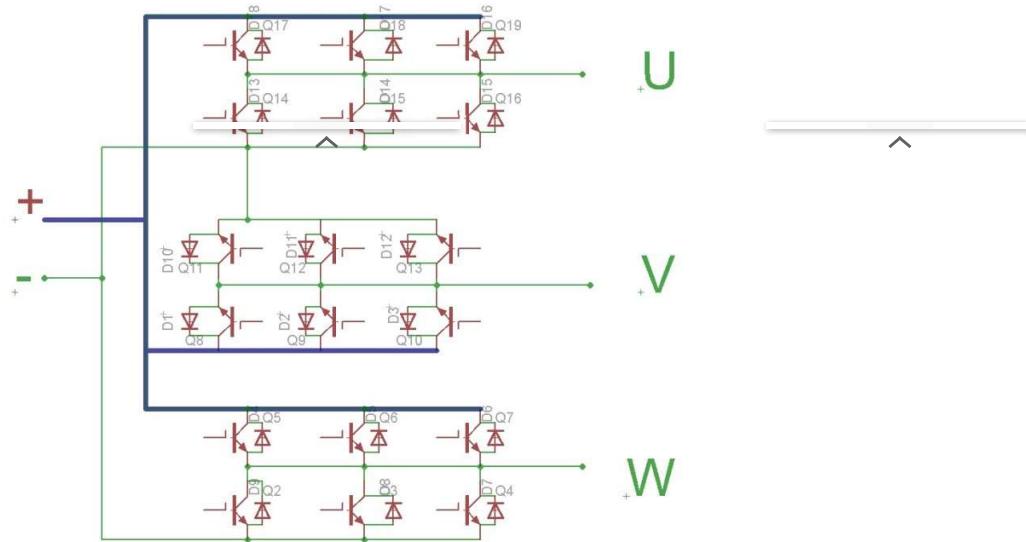


This is the power module formed by 10 mosfets 500V 32A in parallel.

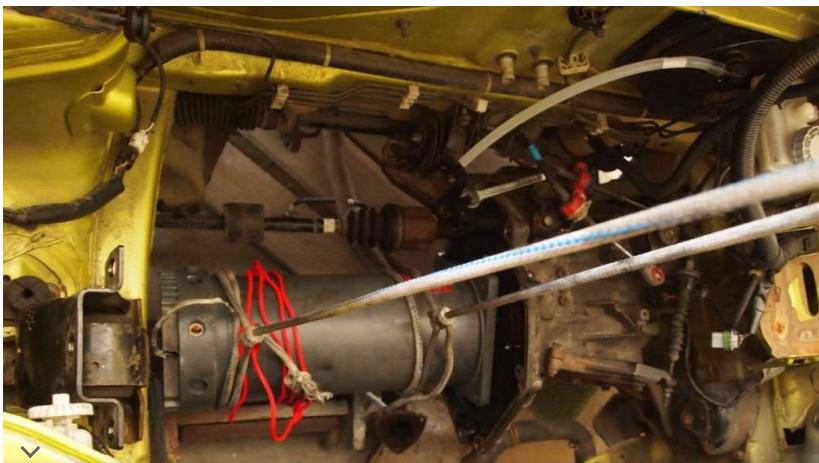
Now new Toyota Prius first generation IGBT module arrived from ebay.



The internal schematic, Each igbt inside has 27nF gate capacitance.



[Opel agila electric](#)





[CAM00645](#)

[CAM00649](#)

Opel Agila electric conversion 20Kw Part1



Opel Agila Electric Conversion 20Kw part 2 Homemade controller



Opel Agila Electric Part 3 (speed testing)



## Homemade Qadrocopter

Posted: 13th November 2012 by [iulian207](#) in [Projects](#)

Tags: [Homemade Qaudrocopter](#), [Smart Quadrocopter](#)

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Motors: 12v Dc Brushless 1000Kv, 20Amp

ESC Mystery 30A Blue series

Prop.9x4.5

Batt: 3 cell 2300mAh Li-Fe A123 Systems.

Home Made Contoller Board Atmega48, 3xMurata Gyroscops.

Homemade Quadrocopter Aluminium Frame



## Brushless motor controller Schematic

I just finished a new controller for my electric scooter project.

The current is limited to 50Amp but it can sustain at least 80Amp.

It is also based on Mc33035 IC and it is capable to decode signal from the hall sensors.

[Donate](#)



5300w RC Brushless motor on a Go Kart 42km/h



How a brushless motor works ( animation)



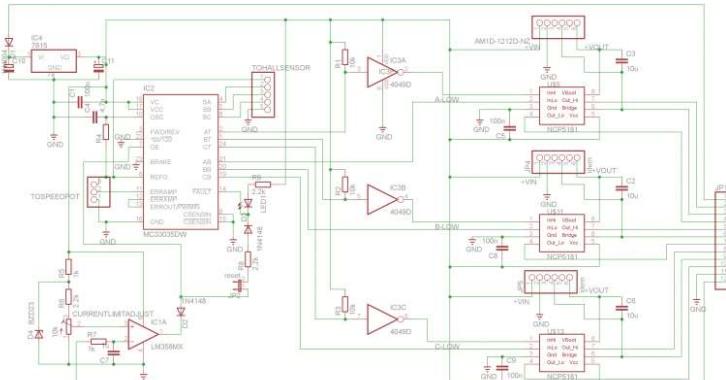
Brushless controller schematic Hall sensor



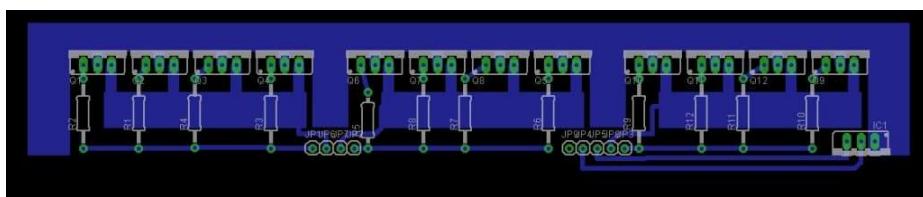
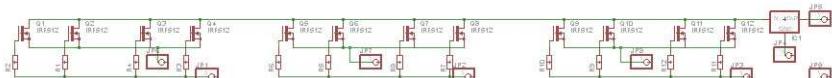
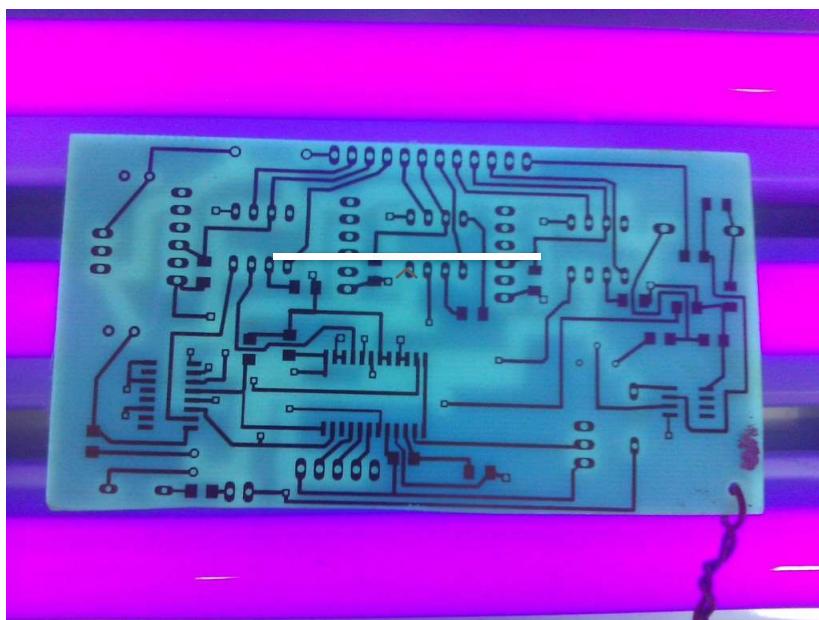
Now i'm also building a new Go kart controller. I will use the same logic board as the scooter.

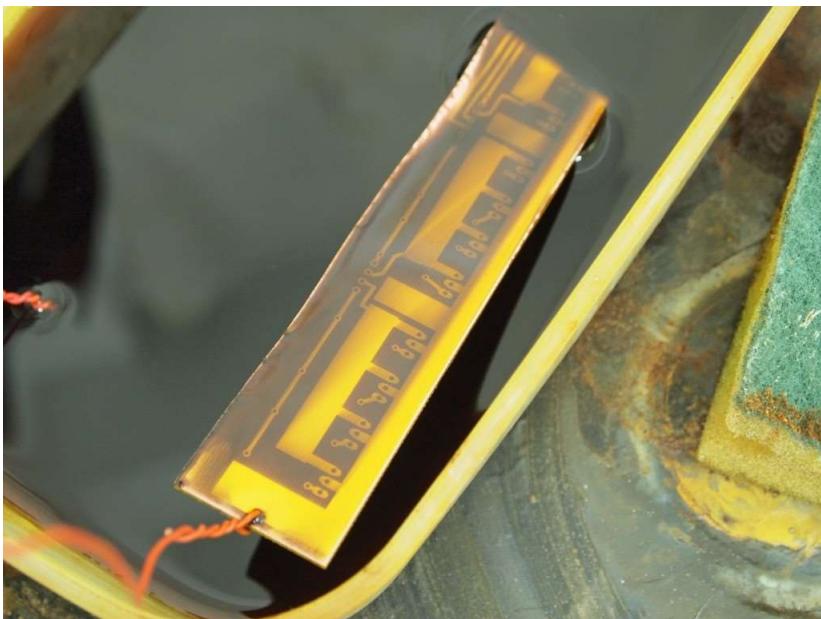
The board was designed in Eagle Cad:

The Brushless motor Controller Schematic: (command module)



IC3A 40460  
IC3B 40460  
IC3C 40460





I used photo resist PCB to make the circuit :

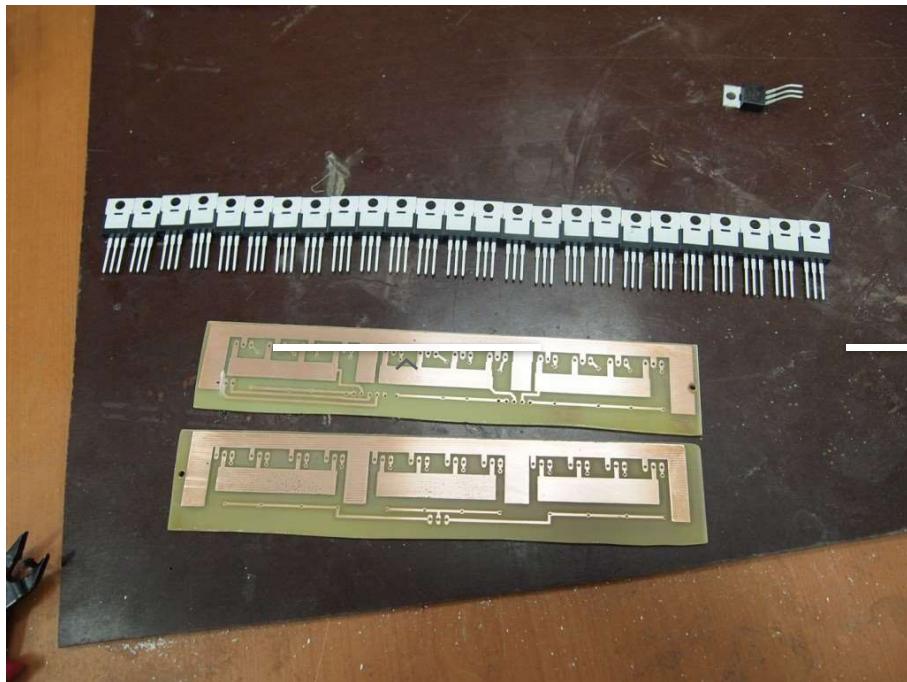


After that i used Ferric Chloride etchant to remove all the unnecessary cooper.



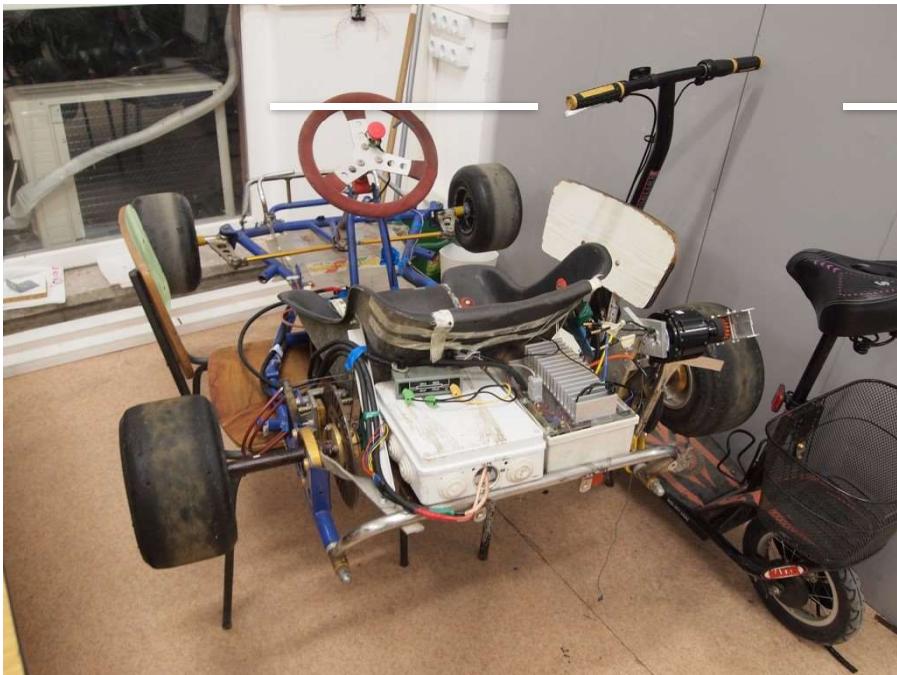


And the result was :





Here is a picture of my Homemade Electric Go Kart:



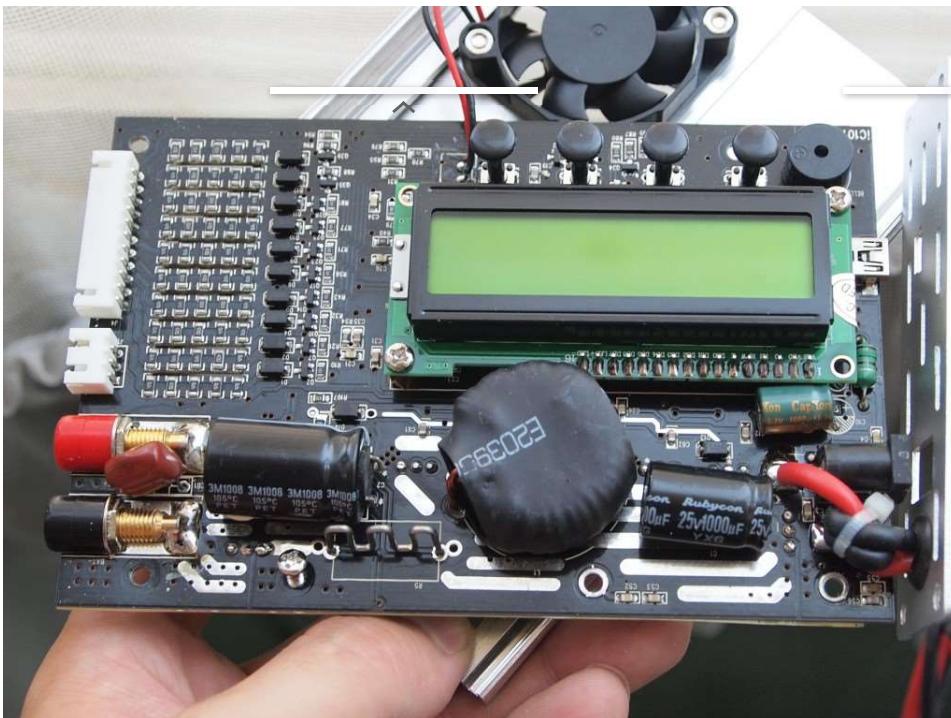
## ICharger 1010B+ testing

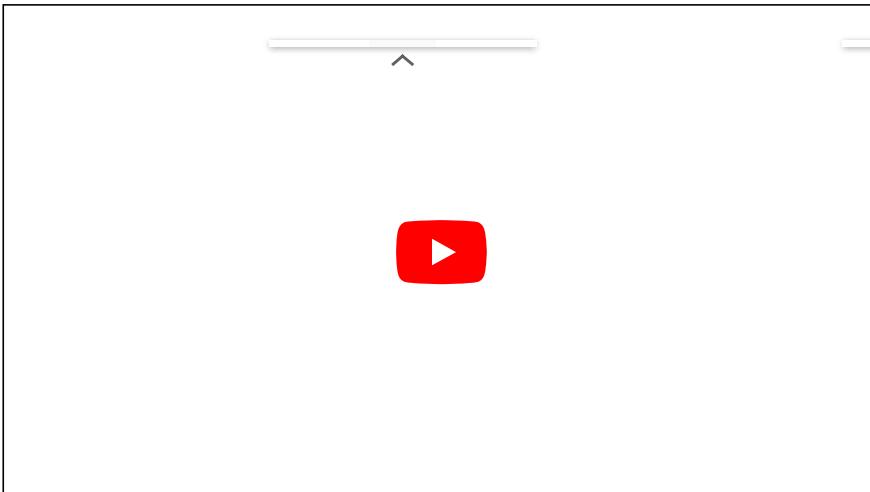
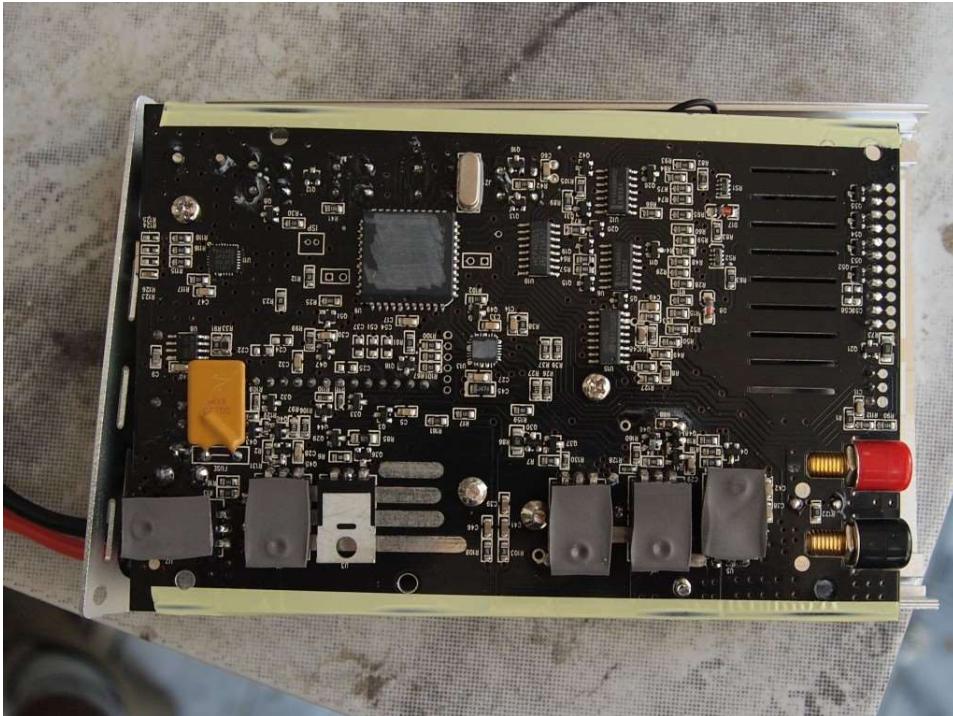
Posted: 9th August 2012 by iulian207 in [Projects](#)  
Tags: ["Best charger"](#), ["ICharger 1010B insite\\_pcb"](#), ["ICharger 1010B" testing external discharge resistor](#)  
[2](#)

Today i received the charger and i put it to the test.  
I am very satisfied about the capability of the charger.



iCharger 1010B+ insite board





## [DC Motor Speed Controller PWM 0-100% Overcurrent protection \(second circuit\)](#)

Posted: 30th May 2012 by [iulian207](#) in [Projects](#)

Tags: "[dc motor controller](#)", "[DC motor speed controller](#)" "[dc motor pwm](#)" "[Speed controller](#)" "[Best DC motor spped controller](#)" "[speed controller circuit](#)" "[pwm speed controller circuit](#)", "["Electric Car controller pwm"](#)", "[pwm controlle](#)", "[dc motor controller current sense](#), [hot to make a dc motor controller](#)"

[116](#)

This is the best cheapest DC motor speed controller circuit that you can find on internet.

In the past i tried wit NE555 and other circuits but the results were every time in shorted mosfet's 😊 and not stoppable GO KART (not very good thing when you do not have a **big red kill switch**).

In the following I will present my DC Motor speed controller capable of adjusting speed (PWM) from 0 – 100% and the frequency from ~ 400Hz to 3kHz, based on LM339 comparator.

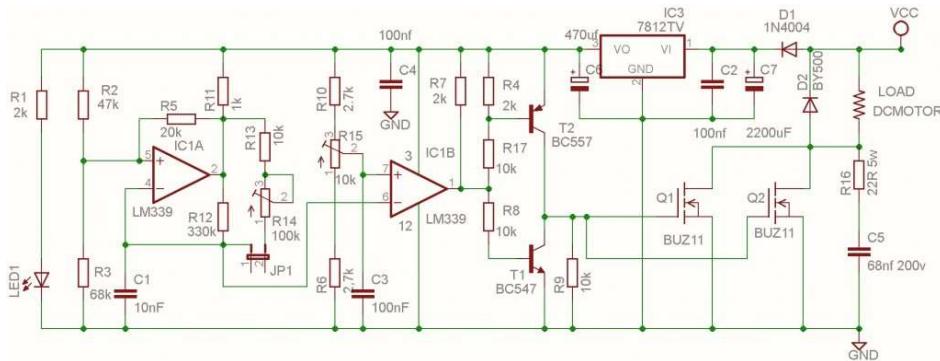
The power supply is from 14-30volts, expandable to practical any value with little modification.

From R15 VR 10k you can adjust the speed from 0 -100%

From R14 VR 100k you can adjust the frequency.

If the jumper JP1 is shorted you can adjust the PWM frequency from 400Hz to 3kHz. If jumper is open Freq is fixed at 100Hz.

The circuit is designed in Eagle cad 6.2



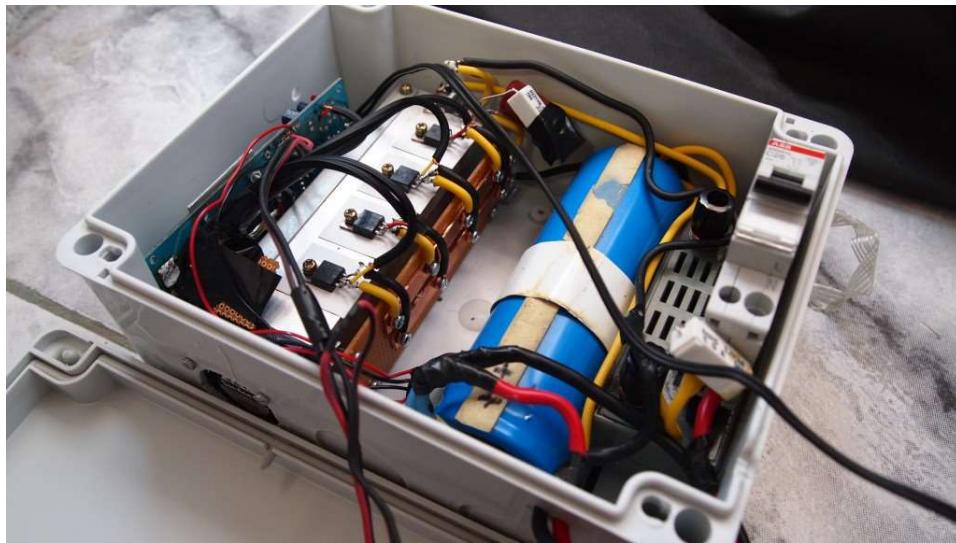
[Donate](#)



You can use almost any Channel N mosfet's you want. The fets will be mounted on a radiator if the current is higher than 2-5 amp.

It is possible to increase the voltage supply to any value if you separate the power to the logic circuit from power to the load and mosfet's

This is my second prototype of dc motor controller

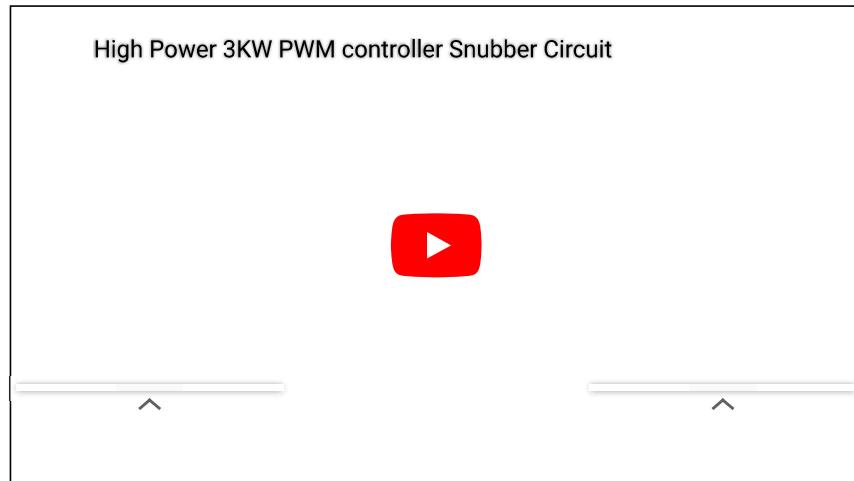


Depending on Rds ON value of the mosfet's you will need a smaller or bigger radiator.

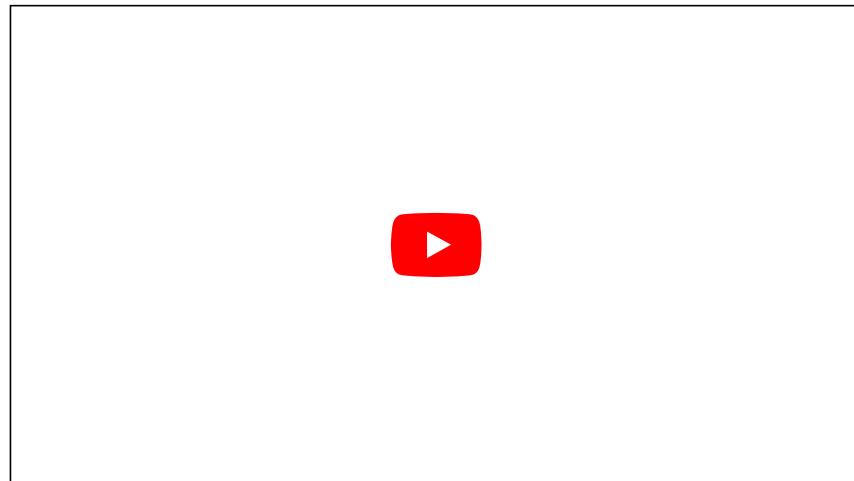
The wires will be at least 12 AWG for a 30 -35 amp load.

For any questions you can ask me any time via my e-mail found on about menu.

Success with the circuit.



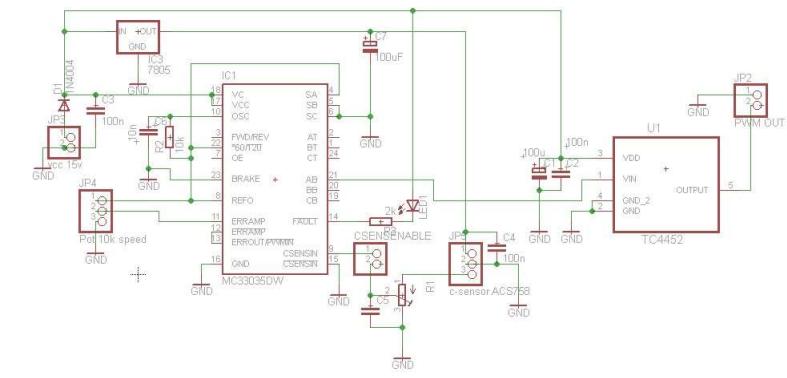
This is the real life testing of the circuit.



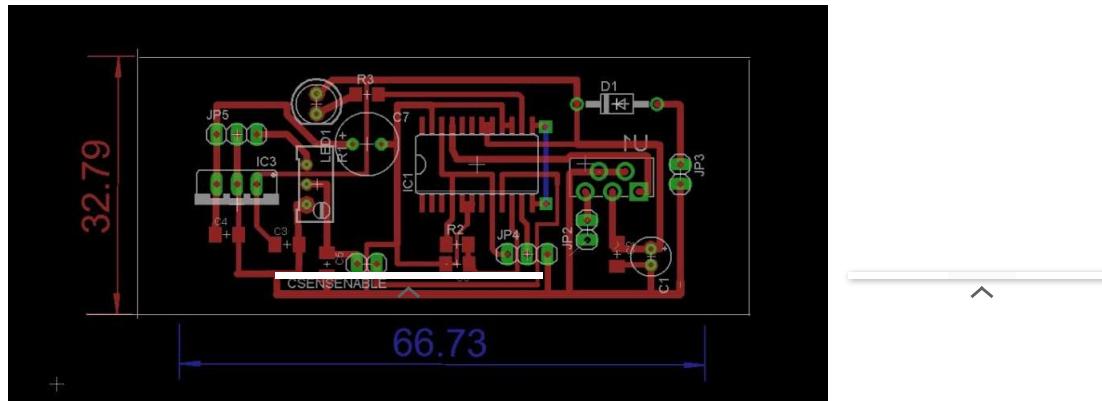
High quality improved PWM controller based on MC33035 IC.

Schematic :





Eagle 6.1 design.



I used as mosfet driver the TC4452 IC with is capable of 12Amp output.

I used this schematic in conjunction with power stage formed by 10 mosfets in parralel with all gate connected via 10pcs 20 ohm resistor to the output of the IC driver.

For current sense circuit i used allegro sensor [ACS758](#)

50-200A current sensor IC

The Allegro CA and CB package current sensor ICs are fully integrated current sensor solutions. They contain the primary conductor, concentrating ferromagnetic core and the analog output Hall-effect linear in a single IC package. The conductor resistance is a typical of  $100 \mu\Omega$  for ultra low power loss when sensing current up to 200 A. These sensors are automotive grade devices that can take the heat and deliver highly accurate open loop current sensing in the most harsh applications environments.

The Allegro medium current devices are much smaller than bulky current transformers and have the added advantage of sensing both AC and DC currents. The package design also provides galvanic isolation to 3000 VRMS and can be used in many line side applications.

## Updated Brushless controller schematic 2015

Posted: 24th May 2012 by iulian207 in [Projects](#)

Tags: ["3 phase brushless dc motor"](#) ["3 phase brushless dc motor controller"](#) ["brushless esc"](#), ["Brushless controller schematic"](#) ["brushless motor"](#) ["Brushless controller"](#) ["brushless controller circuit"](#), ["Homemade electric Go Kart"](#) ["go kart"](#) ["electric go kart"](#) ["homemade brushless controller"](#) ["brushless dc controller"](#) ["tachometer"](#) ["Brushless Regler"](#) ["Бесщеточный контроллер"](#) ["безщетковый контроллер"](#) ["sin e, blde motor esc 를 초음향"](#), [brushless controller schematic](#), [hall sensor controller](#), [motor controller schematic](#), [speed brushless](#)

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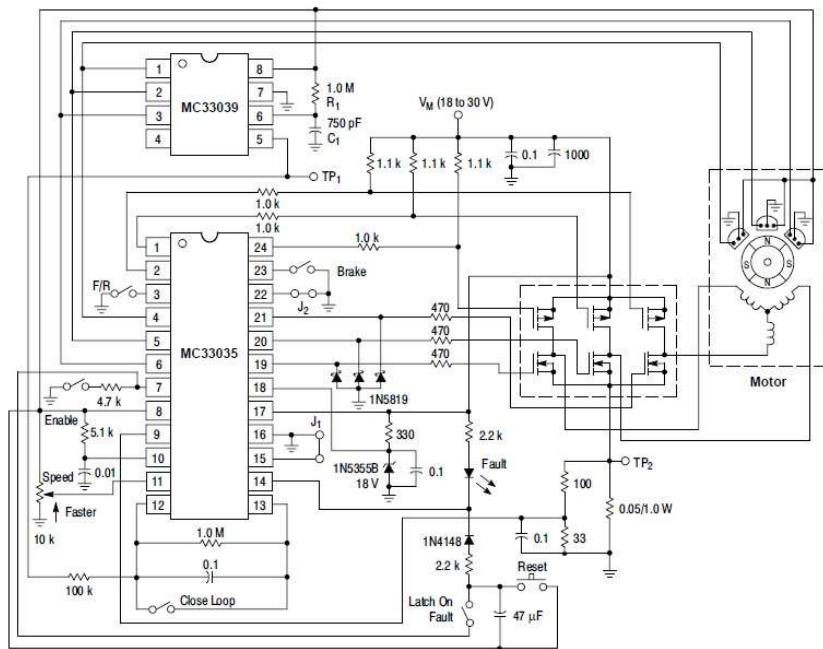
Controller Parameters and Features

### 1. Features:

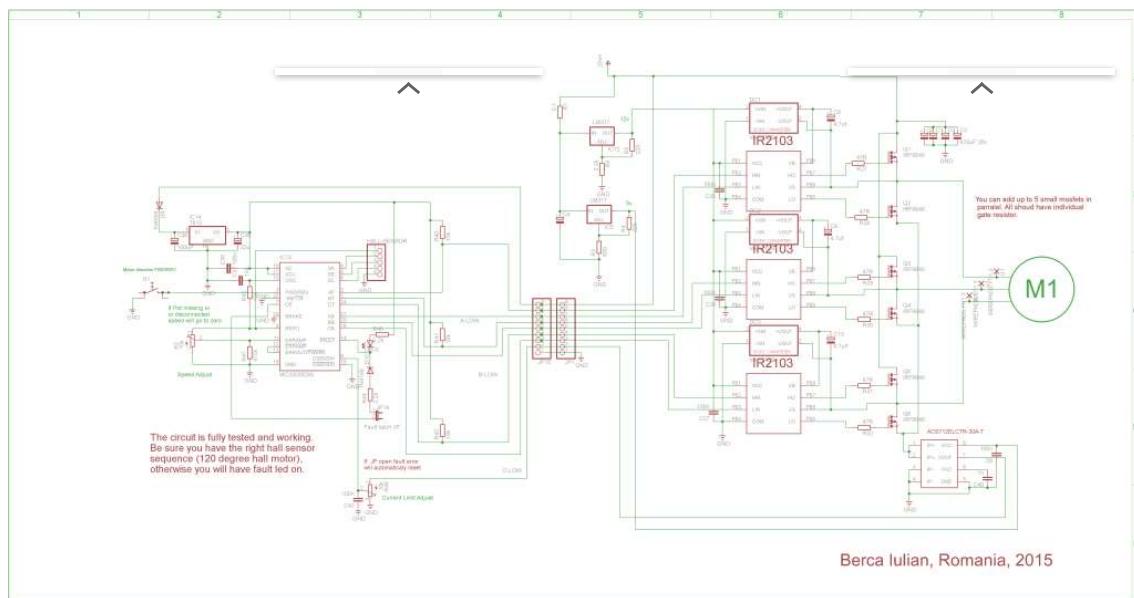
- – Uses analogical chip with no software inside.
- – Work only with sensored brushless motors.
- – Speed adjustable via a potentiometer
- – Adjustable acceleration deceleration
- – Loop Control
- – Fackword / Forward
- – Dynamic breaking
- – Over-current sense from external shunt resistor 100mV threshold level.
- – Overheat protection.
- – Undervoltage protection.
- – Fully Accessible Error Amplifier for Closed Loop Servo Applications
- – Adjustable PWM frequency
- – 6.25 V Reference Capable of Supplying Hall Sensor Power

I used Eagle Cad to make the schematic and the board.

▼



New version of the schematic and it is easy to understand :



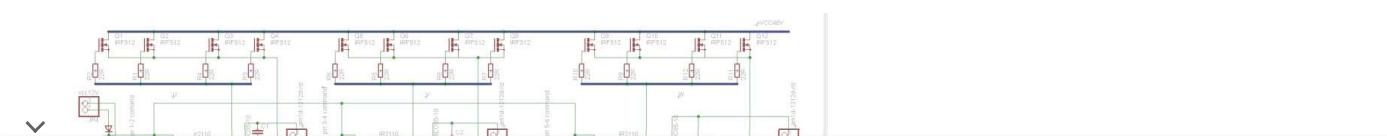
**Very important !!! without the DC-DC converter(above to IR2103) the controller can not work.**

This is another version with more mosfets in parallel and different drivers.

I used only N-channel mosfets in junction with IR2110 half bridge mosfet driver.

You also need a inverstor gate for the Top Drives (4049)

This is the newest version with 4 mosfet in parallel per switch IR4110 but doesn't matter what mosfet you use as the volgate and current is good for you.



The pin 1 JP8 goes to Board 2 JP20 pin 2

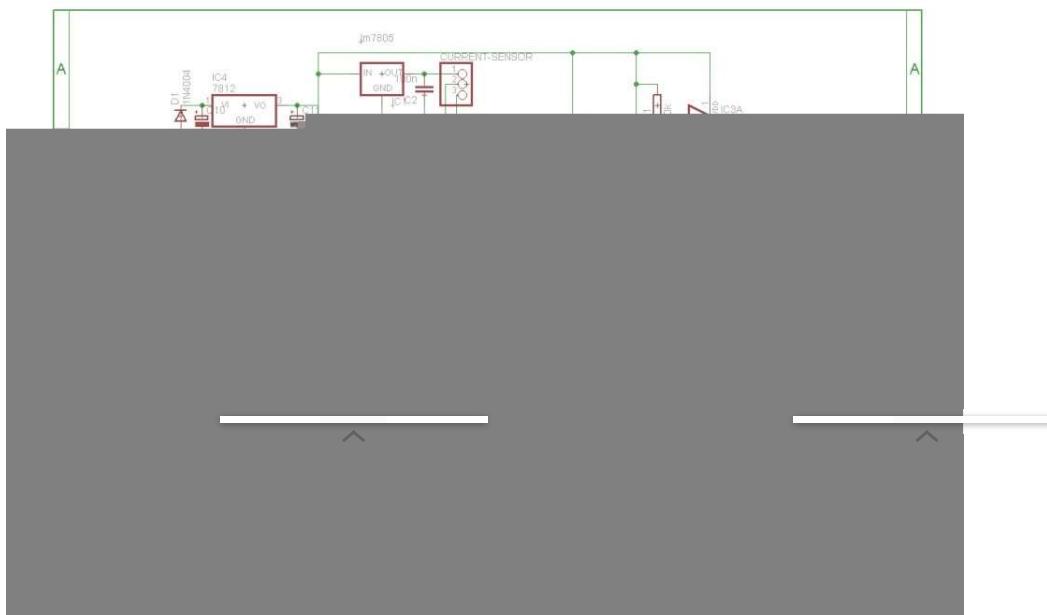
The Pin 3 JP9 goes to the Board 2 at JP20 pin 3

The pin1 JP9 goes to the Board 2 JP20 at pin 4

The pin 3 JP10 goes to the Board 2 JP20 pin 5

The pin 1 JP10 goes to the Board 2 jp20 pin 6

The pin 1 gnd of the JP5, JP6, JP7 can be left in the air. because gnd is common.



In the upper part you can see the current sensor Allegro ACS758 200A.

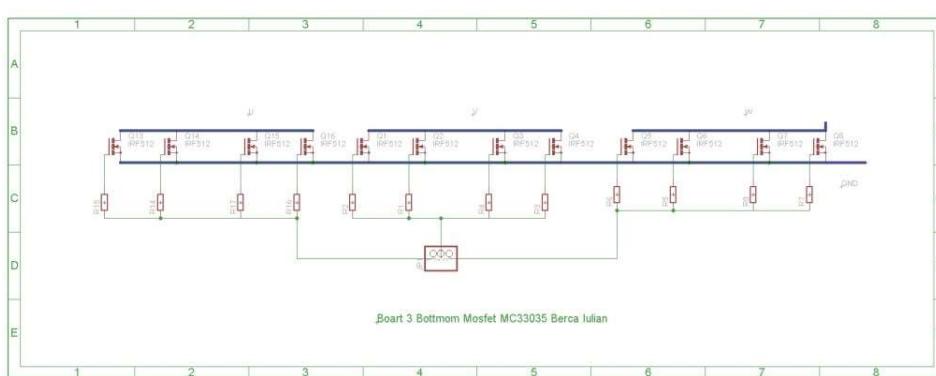
You can also see in the left the bottom module next to it the driver+ top module.

In the right upper corner command module and in lower corner a dc-dc converter module from ebay.

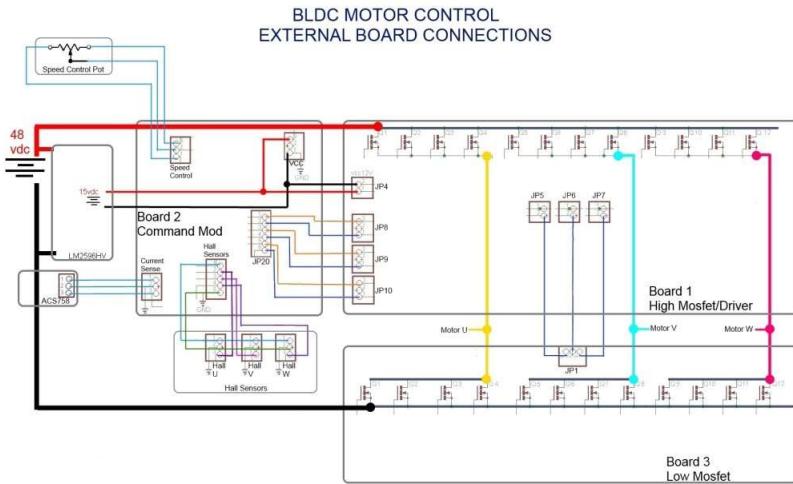
#### [DC-DC Step-Down Converter 4.5-60V to 3-35V LM2596HV Power Supply Module](#)

to power the command module and the driver board.

You can put any mosfets channel N you need. Best are with internal resistance as low as possible and higher current.



This diagram was draw by a website visitor by name "Bill Catalena" from my specification.



This is the 3-rd Board with bottom mosfets.

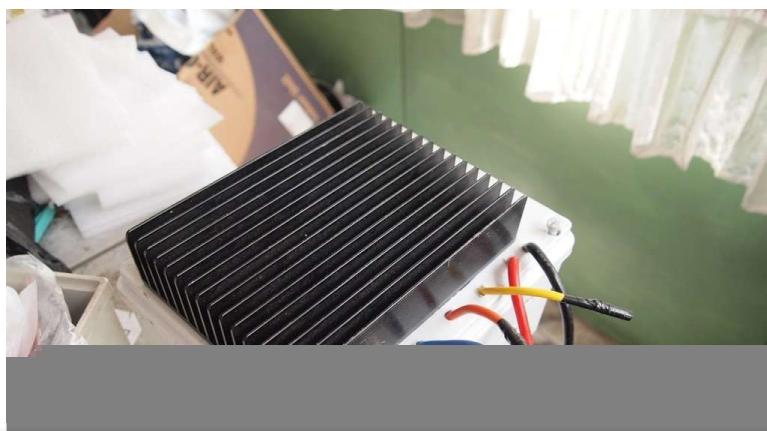
The U,V,W need to be connected to the U,V,W to the top part of the mosfets.

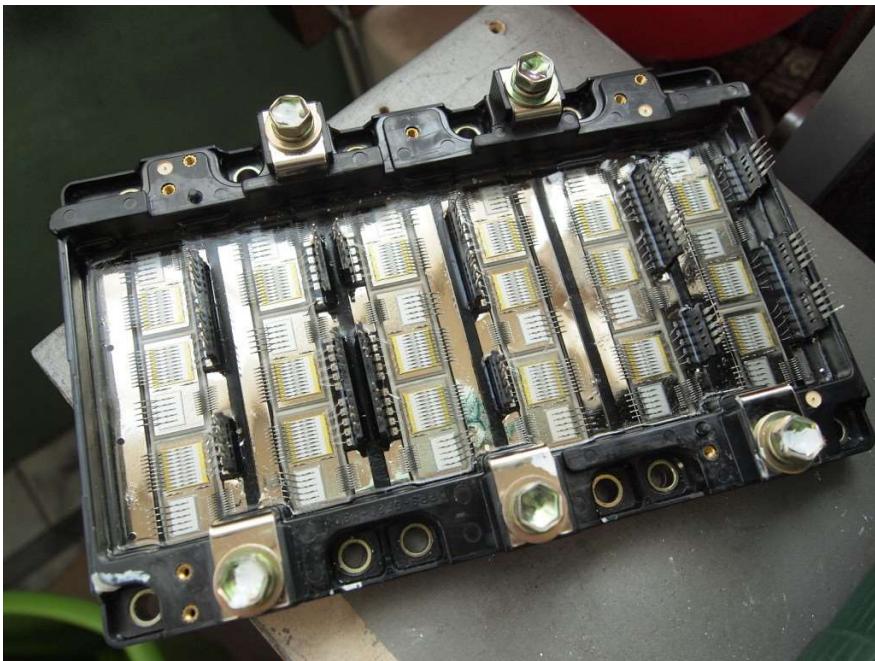
The Pin 1 of the JP1 goes to the pin 2 JP5 from the Board 1

The pin 2 of the JP1 goes to the pin 2 JP6 from the Board 1

The pin 3 of the JP1 goes to the pin 2 JP7 from the Board 1

Ground pin is connected from the power supply or 48v



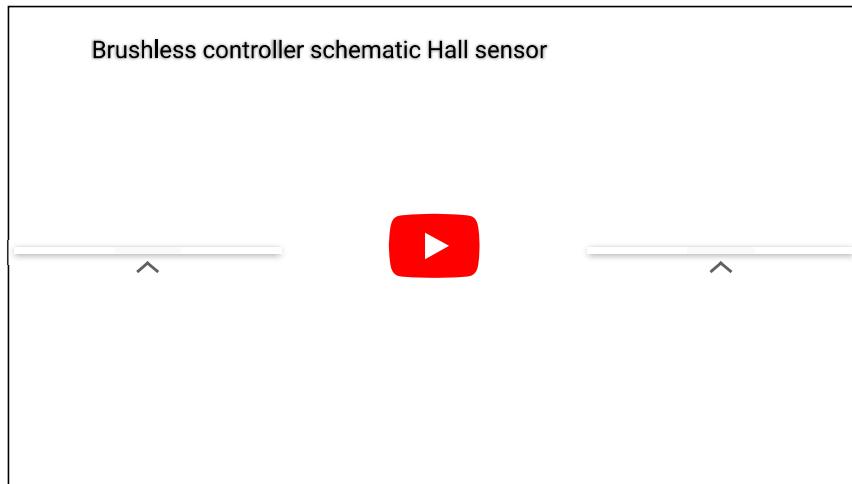


This toyota prius igbt module inverter was from a scrap and i took apart the driver board. now i'm building my own diver board.

In order to be able so start the motor you need to put a floating dc power supply to the IR2110

I have tested some of 1200V 600A FZ600R12KE3 IGBT module and the input capacitance was ~ 55nF. The time rise obtained with avago IC ACPL-P343 was 1,2uS at 12Khz, not so good if you want the switching losses low. the datasheet: <http://www.farnell.com/datasheets/1676975.pdf>

In this video i used 12 Mosfets irf3205z and IR2110 driver



Now the big thing : Testing the Electric Go Kart to measure the results

5300w RC Brushless motor on a Go Kart 42km/h



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